Physics
Direct

INCENTIVES FURNISH THE DRIVE

THE COMING YEAR, 1946, and the years to follow can bring unprecedented prosperity to the people of the United States if the incentives to secure it are provided.

We have the advantage of starting with an economy which has demonstrated a capacity for expansion unequalled in any other country in the world. Our economy has demonstrated, also, one grave weakness—a recurring interruption of the upward trend of production and living standards by wasteful and paralyzing periods of recession. Recovery from each depression always has carried us to new heights of economic welfare, but the toll of the years of blight has been harmful to everyone.

The job ahead of us is a dual one. We must maintain the vitality of an economy which, over the years, has yielded an enormous increase in the American standard of living, and we also must improve its stability.

The Dynamics of American Production

In the last prewar year, 1940, the population of the United States was 3⅓ times as large as it was in 1870. But the national production, measured in dollars of constant purchasing power, was 10 times as large at the end of the period, and industrial output had increased 20-fold.

In the meantime, the average number of hours of factory workers had been reduced from about 63 per week in 1870 to less than 40 in 1940, while average hourly earnings had more than trebled in dollars of constant purchasing power. Thus "real" weekly or annual wages in manufacturing had doubled over the 70-year period, even though the work-week was cut by 35 per cent. This was made possible chiefly by a tremendous increase in the quantity and quality of the mechanical facilities which were provided in American manufacturing industry. Manufacturing capital investment per worker was multiplied by 6 times over the period in question. But the return per dollar invested, while it has fluctuated widely between good years and bad, showed no general upward trend over that portion of the period for which measurement is practicable.

Incentives in American Manufacturing

There has been, historically, a remarkably consistent pattern in the division of the realized income from the expanding manufacturing output of America. Reliable statistics are not available for as far back as 1870, but from 1899 through 1939 the average share of wages and salaries has been 82½ per cent against 17½ per cent as the share to investors (including dividends, interest, rents, royalties, and non-corporate profits). There have been, from year to year, relatively minor divergences from this pattern of distribution, but there is no discernible trend during the period away from the averages cited.

It is suggested that the persistence of the average 17½ per cent share of realized income from manufacturing that was maintained for the 40 years preceding World War II may represent the proportion that is needed to produce the dividends, interest, rents, royalties, and non-corporate return that will provide for the continuing investment upon which an expanding productivity such as we have had in the past depends. At any rate, it would seem reckless to depart too radically from such an established pattern at a time when unprecedentedly large private capital investment is counted on to make up for the drastic curtailment of such investment during the war years, and to carry us to the new high levels of civilian production set as our postwar goals.

The Distribution of Manufacturing Income in War

At the beginning of the war, the Government adopted controls and a tax program designed to prevent wartime activity from resulting in un-

duly swollen private returns. Due primarily to huge volumes, the profits *before* taxes of manufacturing industry were very high, but throughout the war its profits *after* taxes averaged returns no larger than they had been in good prewar years. Relative to volume, they were considerably lower than in prosperous years in the past. Again, there can be no complaint at results that generally were in accord with a national wartime policy.

But it is fair to note that the wages of manufacturing labor were allowed to increase substantially during the war. Between January 1, 1941 and April, 1945, average weekly earnings per worker increased by 77 per cent. This was, in considerable part, a result of increased working hours and a shift from low- to high-paid industries, but straight-time hourly earnings on the same jobs increased about 40 per cent against a cost-of-living rise of about 30 per cent.

The net result was to alter drastically the 40-year relationship of the 17½-82½ per cent division of Realized Income from Manufacturing. The share of wages and salaries increased to over 90 per cent, and the investment share shrunk to less than 10 per cent.

Its Postwar Distribution

This wartime shift in the proportion of distributive shares has an important bearing upon current wage controveries. With union demands for wage increases ranging up to 30 per cent, and the economists of the Office of War Mobilization and Reconversion asserting that an average increase of 24 per cent is feasible without raising prices, it is pertinent to inquire how such increases would affect the prewar ratios that governed realized income distribution in manufacturing.

Forecasting is always hazardous, but if we assume (1) that in 1946 we shall reach the $160 billion level of national output which the Government proponents of general wage increases expect, and (2) that there will be little increase in productivity because of the continuing process of reconversion, and (3) that the Government will succeed in carrying its announced purpose to maintain present price ceilings, it appears that a 24 per cent general wage increase would reduce the share going to capital from 17½ per cent to 11 per cent even allowing for its increased return resulting from the repeal of the excess profits tax. The prewar ratios would be about maintained if wages remained at present levels.

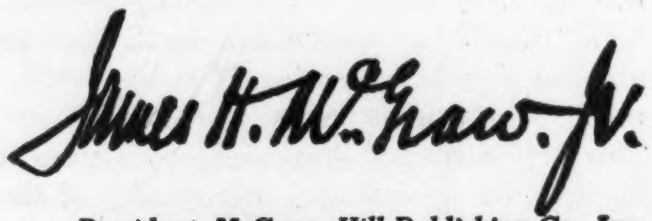
Conclusion

Since the maintenance of these prewar ratios was accompanied by an unparalleled rise in the "real wage" of American workers, there is a powerful prima facie case for not tinkering with them. It should be noted, however, that some economists think that the size of the investment share of manufacturing income tends to provide more capital than can be absorbed by a mature economy, and thus contributes to those breaks in the expansion of the economy which, as stated at the outset, have been its principal blight.

Regardless of what may ultimately prove to be the validity of this view, no one can responsibly contend that at this early but crucial stage in the reconversion process is the time to test it. Now, no one knows whether, or what dimension of, additional wage increases can be supported without forcing up prices or reducing profits to a point that will discourage vitally needed private capital investment.

We want high and increasing wages in American manufacturing. We need them to provide an active incentive to workers to support expanding productivity, as well as to continue the trend of rising living standards in America. Equally, we need a continuing profit incentive of sufficient attractiveness to call forth the new investment upon which expanding productivity depends.

We can never attain our dual objective if we push one of these aims so far and so fast that it defeats the other.



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THIS IS THE 43rd OF A SERIES

CROSS TALK

►**PROGRAM . . .** Most impressive is the program planned for the new Research Laboratory of Electronics at M.I.T. This new research center, a successor to the Radiation Laboratory, will be sponsored jointly by Tech's Departments of Electrical Engineering and Physics. Experimental researches will be directed toward these objectives, among others: klystrons for 0.25 to 0.75 cm waves; magnetrons for 10 megawatts power output at 10 cm wavelength; single-anode magnetrons for use as reactance tubes; waveguides as elements of measuring devices, particularly waveguide bridges. Among the studies in physics are: absorption of 0.25 to 1 cm waves by gases; low pressure gas discharges at microwave frequencies; magnetic and electrical properties of materials near zero absolute temperature and at microwave frequencies. Not concerned with microwaves are plans for a super-voltage accelerator, supersonics at 30 mc, timing devices with pulse durations of 0.01 microsecond. Similar work will be carried out in microwave communication and electronic aids to computation. Coming when there is a tendency to let down on fundamental research, news of this high-powered attack on the unknown is heartening.

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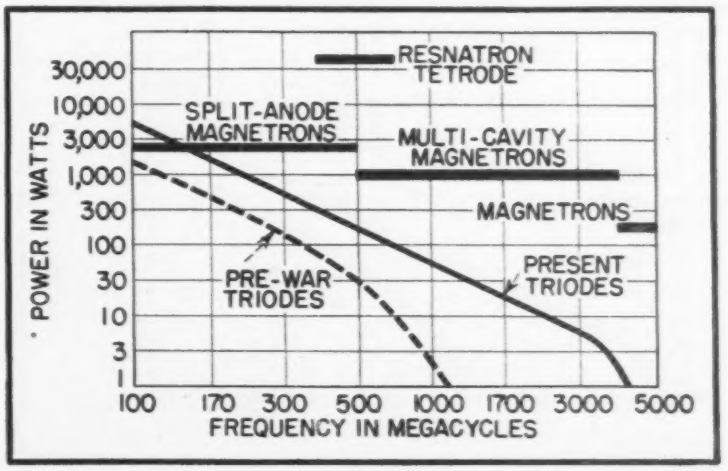
This is progress indeed. Further study is required to show how these oscillators may be frequency-



controlled and modulated for broadcast services. Up to now they have been modulated with noise and other non-critical waveforms. But our guess is that most of them are applicable to post-war needs. It has been argued, in fact, that the power output of the resnatron is *too* high for any peace-time purpose.

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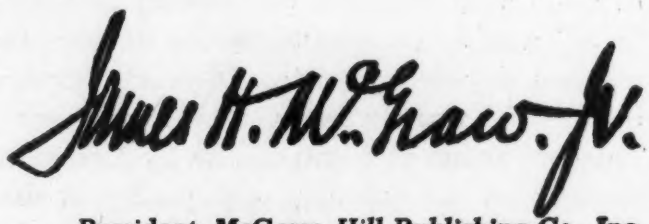

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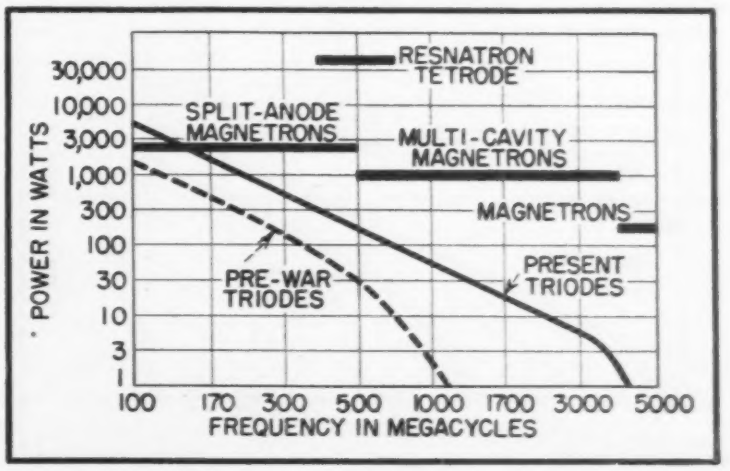

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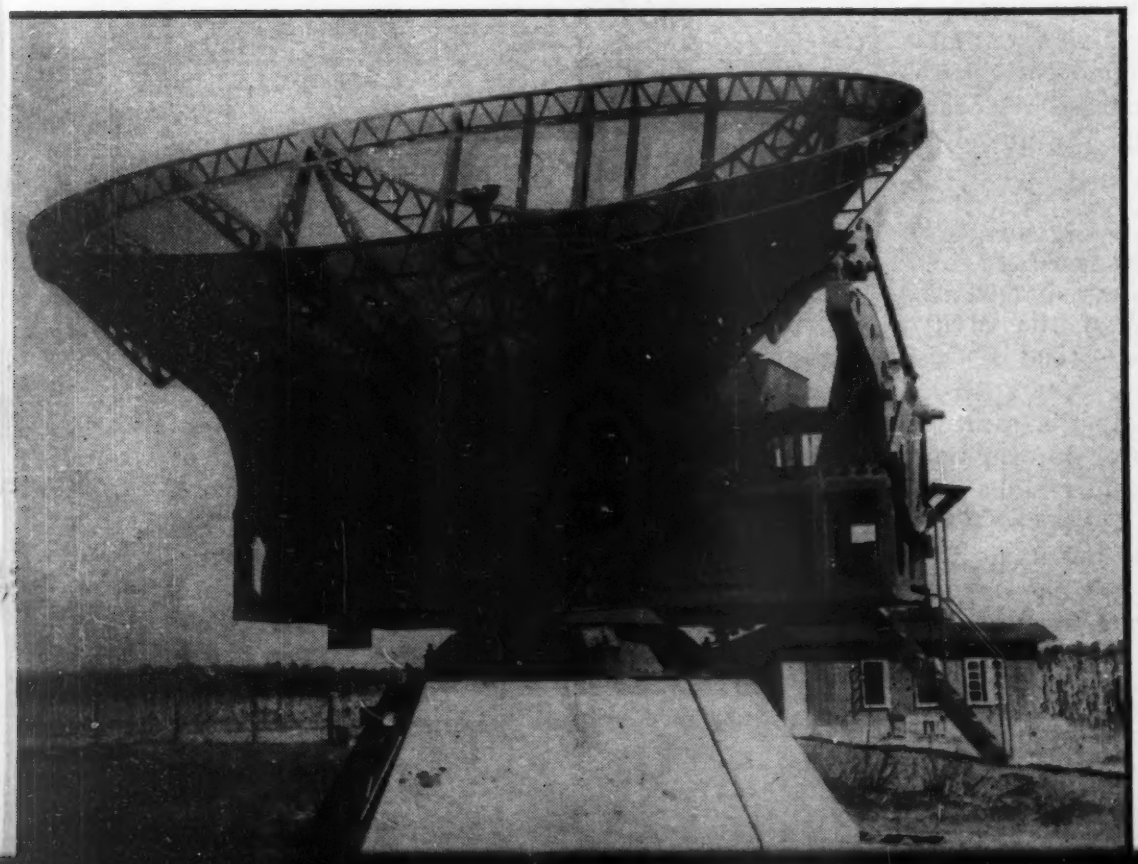
Radar COUNTERMEASURES

Equipment for detecting enemy radars, determining their location, analyzing their characteristics and then jamming them electronically or by means of chaff is described. Many of the devices used, such as shf direction-finders, wideband radiators, and the resnatron tube, have postwar applications

IN THE RADAR WARFARE just concluded there were two major campaigns. The first was to use radar against the enemy; the second to hinder the enemy in his use of radar against our forces. The second campaign was conducted by techniques known as radar countermeasures, (rcm) by which enemy radars were detected and put out of action.

In many ways rcm activity was the most fascinating aspect of the electronic war, since it involved direct contact with the enemy and required all the competitive strategy and inspiration of a campaign in the field. Technically, also, rcm commands attention, not only because the methods used were unique and different from those of radar, but also because many rcm devices have post-war uses which may outrank those of radar.

Radar countermeasures were developed and employed by the British in the early stages of the war. Prior to America's entry, rcm activity was also carried out in our own Army and Navy Laboratories, on a long-range basis. The need for coordinated activity led in 1942 to the formation, under Division 15 of the National Defense Research Committee, of an rcm laboratory at Harvard University. This laboratory was known as the Radio Research Laboratory. Of the $300,000,000 spent by the U. S. on rcm equipment and materials, approximately two-thirds was for equipment developed in this NDRC activity. The remainder was spent developing methods originating with the British, and in the Naval Research Laboratory, Signal Corps, and Air Technical Service Command Laboratories.



The principal targets of Allied countermeasures in Europe were Wurzburg antiaircraft radars such as this unit. The Germans had 4000, representing a billion-dollar investment. Jamming reduced their effectiveness to 25 percent of normal

Search and Jamming Functions

To knock out enemy radars, two basic functions are involved. The first is a search, conducted with receivers and direction finders, to determine where the enemy radar is located and as many as possible of its technical characteristics. The second is jamming, accomplished by means of aluminum-foil chaff or rope sowed in the sky by airplanes or by rockets fired from the ground and/or the transmission of signals which will interfere with the operation of the enemy equipment. In the initial stages of the program the two functions were separate. The search was conducted to obtain technical specifications and these specifications were sent home for use in the design of a suitable jammer. The jammer was then produced, on the fastest possible basis, and put into action.

Ten million pounds of aluminum foil (chaff and rope) were dumped on Europe by Allied bombers to produce radar smoke screens within which German radars could not detect Allied bombers. Each package of chaff, one of which is shown being dispensed, contains several thousand dipole reflectors

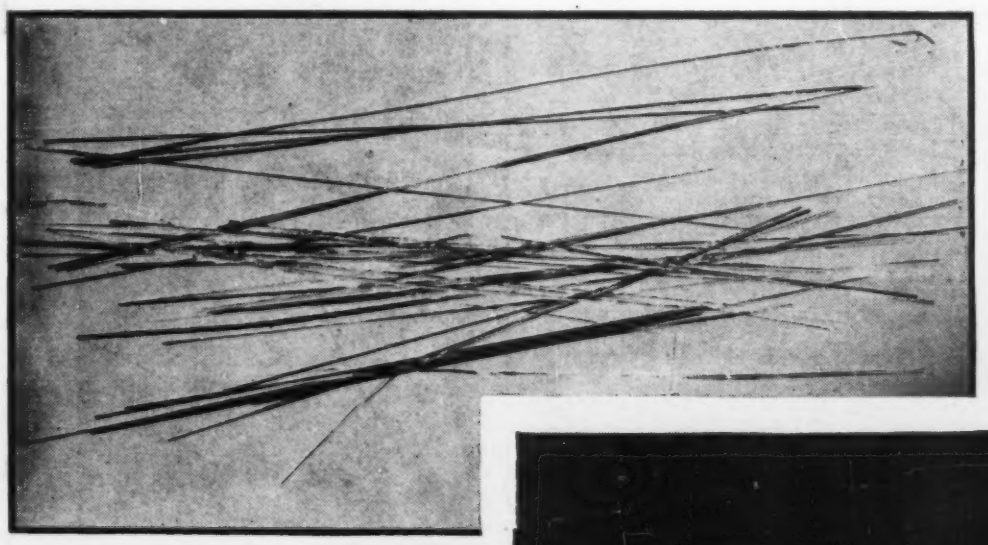
Later, the two functions were combined in equipment of such flexibility that virtually all types of enemy radar could be detected and jammed in a single operation.

Some idea of the equipment flexibility required may be obtained from the accompanying table, which lists jamming transmitters, search receivers and direction finders. Since the enemy had, within reason, a free choice of frequency, it was necessary to build a group of continuously tunable search receivers and direction finders to cover virtually the entire radar spectrum, from 25 to 6000 mc. Moreover, since the choice of pulse width lay with the enemy, bandwidths wide enough to accept a variety of pulses were required.

In the jamming transmitters the same continuously tunable frequency range was required. Moreover the transmitters were of necessity the continuous-wave variety, since they had to block out echo signals which might occur at any time, depending on the timing of the enemy radar and the distance to its target. Moreover, the highest possible power was required, continuously, to blot out the enemy indicators at great distances. In the interest of conserving power, suppressed-carrier transmitters were often used, modulated with random noise over a bandwidth of several mc.

Underlying the design of these search and jamming equipments are several basic relationships which put the enemy radar at a fundamental disadvantage. In the first place, a radar must transmit at high power to detect targets at useful distances.

The radar operates by reflection of its signals, which introduces attenuation of the radar signal proportional to the fourth power of the distance to the target. The search receiver detects the radar signal by one-way transmission, which is at-



Chaff, aluminum halfwave dipoles of various lengths, tuned to the Wurzburg radar frequencies. Each strip is embossed and crimped along its length to assure wide dispersal and reasonable rigidity after being thrown from the aircraft

Effect of chaff on Wurzburg type-J indicator. At left, normal indication of Allied bombers (at right of circle). Right, a trail of chaff some 16 miles long totally obscures the bombers. These shots were taken from a German training film

EQUIPMENT FOR RADAR COUNTERMEASURES

Jamming Transmitters

Frequency Range (mc)	Designation and Code Name	Bandwidth (mc)	Output Power (watts)		Description and Radars Jammed
			Carrier	Sidebands	
25–100	ARQ-8 Dina	0.15	0	40–20	Suppressed carrier, single sideband. German EW, Jap GL and SLC
85–150	APT-3 Mandrel	3	12–9	3–2	Grid-modulated MOPA. German EW
90–220	APT-1 Dina	6	0	15–8	Suppressed carrier, single sideband. German EW, Jap GL and SLC
200–550	APQ-2 Rug	7	20–5.5	5–1.25	Line oscillator, doorknob tubes. German coastal, Jap torpedo planes
450–720	APT-2 Carpet	7	8–3	1.6–0.6	Same as APQ-2. German Wurzburg GL
475–585	APQ-9 Carpet III	7	20	5	Parallel plate, using 8012's. German Wurzburg GL
350–1200	APT-5 Carpet IV	2.5–3.0	30–5	——	Lighthouse-tube cavity oscillator. German Wurzburg GL
150–780	APT-4 Broadloom	7–10	150	——	Current-modulated c-w magnetron. German Wurzburg GL
300–2500	APT-9	2–8	25–10	10–3	Cavity oscillator
2230–4030	APT-10	——	25–50	——	Tunable c-w magnetron, four heads to cover range

Search Receivers

Frequency range (mc)	Designation	Input power (watts)	Description
25–100	ARQ-8	75 a-c	Dinamate, used with Dina and tuned electronically to transmitter frequency. Superheterodyne
40–3000	APR-4	90 a-c, 9 d-c	Bandwidth 4 mc or 0.5 mc. Single-dial tuning. Four r-f heads cover range Motor-driven sector sweep. Superheterodyne
1000–3100	APR-5	150 a-c, 25 d-c	Coaxial-antenna input, cavity oscillator. Crystal mixer. Superheterodyne
3000–6000	APR-8	150 a-c, 25 d-c	Same as APR-5, but mixer operates on local oscillator harmonics. Waveguide input. Superheterodyne

Direction Finders

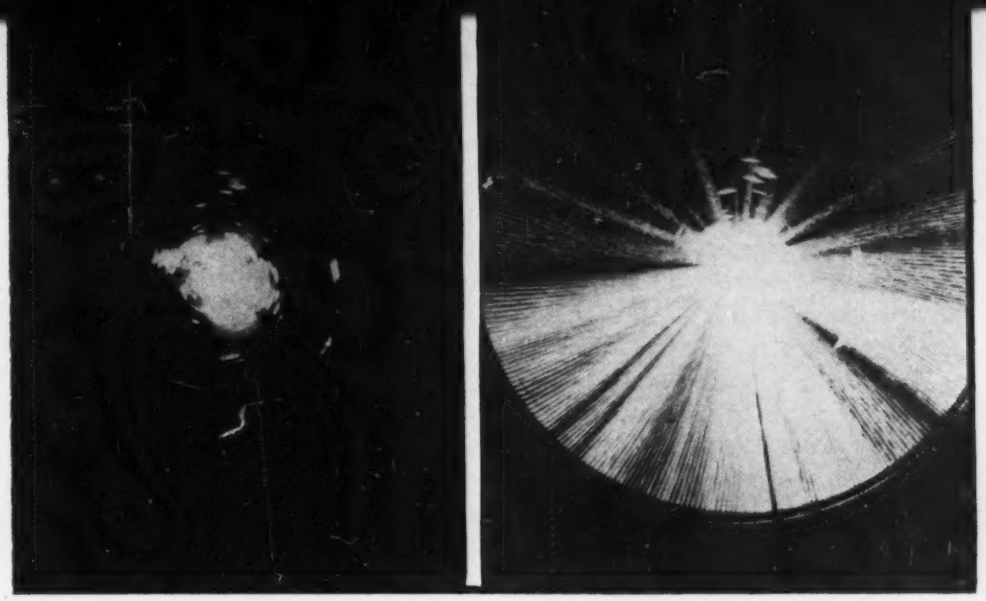
100–450	APA-24	——	Vertical Adcock plus horizontal dipole, manual remote control. Null indication
300-1000	APA-17	125 a-c, 50 d-c	Whirling radiator for use with any search receiver. C-r indication on maximum of antenna pattern

tenuated only as the square of the distance. Consequently, a radar signal can be detected at distances far greater than the maximum range at which the radar can see a target. In the second place, the direction to the source of the radar signals may be observed by the use of a directional antenna on the search receiver. Two or more bearings so obtained reveal the location of the radar. Radar reconnaissance can thus be carried out, with airborne search receivers for example, without fear of detection by the radar itself.

The same favorable discrepancy exists between radar range and jamming range, since the jamming signal competes only with a weak reflected echo on the radar screen. Consequently, a jamming power level in the tens of watts is sufficient to compete with a radar peak power in the hundreds of kilowatts. This advantage is reduced, however, by the necessity of jamming with a c-w signal, and so the average power of the jammer is often nearly as great as the average power of the radar.

The accompanying table illustrates the frequencies, power levels and bandwidths of the jamming transmitters. Power in the tens of watts, modulated over bands up to 10 mc, is obtained at frequencies up to about 700 mc using triode tubes, notably the door-knob types. Cavity resonators, using disk-seal (lighthouse) tubes, give about the same performance at frequencies up to 2000 mc. For higher power levels, especially at the highest frequency ranges, c-w magnetrons are used. One important example is the Broadloom jammer, which produces 150 watts

Electronic jamming, caused by noise-modulated c-w transmissions tuned to the radar frequency. Above, normal scope. Below, echo pulses obscured by artificial noise produced by jamming signal



Effect of electronic jamming on plan position indicator (ppi). Left, normal indication. Right, radial traces, brightened by jamming signal, obscure echo signals

of cw carrier up to 780 mc.

Modulation of the jamming signal is, of course, essential to achieve the maximum blanketing effect. Experience has shown that random noise, such as may be obtained conveniently from the space current of a gaseous vacuum tube, provides the most effective modulation waveform. Noise modulation, received by the radar, has the effect of multiplying enormously the normal noise level present in the radar receiver.

Search Techniques

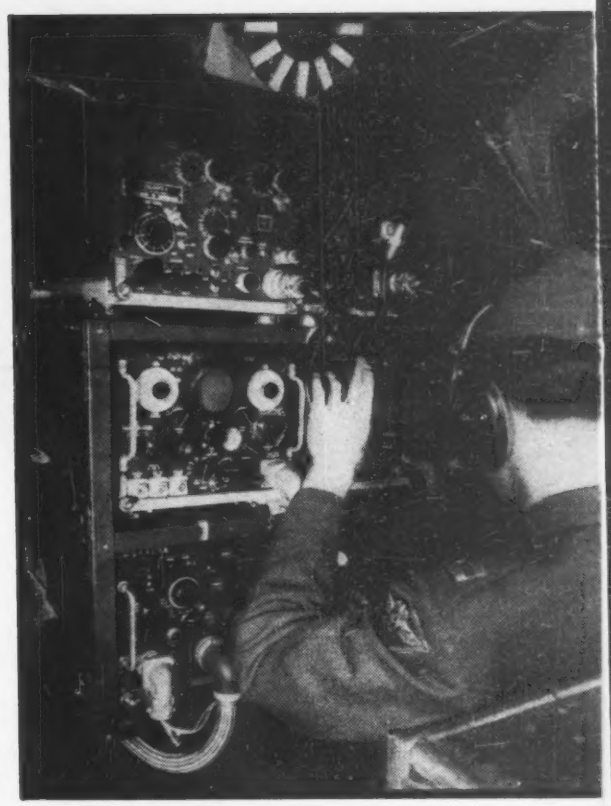
The technique of searching for enemy radar signals, as a preliminary to jamming them, consists simply in tuning the search receiver repeatedly over the radar spectrum. This is not only difficult technically, but physically tiring. The technical difficulties reside in the great width of the spectrum to be covered. One excellent example of how the problem is solved is the AN/APR-4, which covers the range from 40 to 3000 mc, using four r-f heads. The tuning is motor driven over a frequency sector which can be selected by the operator, thus relieving him of a considerable physical burden. An automatic tape-recording system is available to record the frequency at which signals are detected as the spectrum is swept, thus further reducing the attention demanded.

The simplest method of observing the radar signals is by an aural indication. Radar pulses are transmitted at repetition rates which lie within the audible spectrum. Moreover, the pulse represents, in effect, a high degree of overmodulation on a c-w carrier, and this modulation can be recovered in a conventional second detector, amplified at audio frequencies, and fed to headphones. When a radar signal is intercepted a whine (repetition frequency plus harmonics) is heard in the headphones. The strength of the signal varies periodically as the radar beam sweeps past the search plane. So long as this variation continues, the radar is searching. But if the signal becomes steady, at maximum volume, the chances are that the radar has detected the search plane and is tracking it. Appropriate action is then taken to avoid enemy gunfire and aircraft.

While aural or tape-recording methods serve to identify the presence and carrier frequency of the enemy radar, they give little indication of the pulse characteristics. A cathode-ray pulse analyzer (oscilloscope) is available to determine the pulse repetition rate, the pulse width, the pulse shape and relative amplitude. Such an analyzer gives important clues to the type of radar under observation, since it reveals the radar's maximum range, minimum range, and range accuracy.

Wideband Radiators

Implicit in the wide frequency ranges covered by search and jamming equipment is the necessity for radiators which will cover these ranges without excessive tuning adjustments. The Radio Research Laboratory undertook to develop antennas which would cover frequency ranges of several thousand megacycles without any adjustment whatever. One of the most spectacular of these antennas is an approximately cylindrical structure which covers the range from 950



Search and jamming equipment on a Navy rcm plane: from top to bottom: antenna selector switch, APT-1 electronic jammer, panoramic adapter (to sweep receiver frequency) pulse analyzer, cathode-ray d-f indicator (in use by operator), second jamming transmitter, and search receiver (hidden by operator)

to 2900 mc, a frequency ratio of 3-to-1, matching the transmission line throughout this range. In general, the wideband antennas make use of the principle that a thick, stubby radiator has low stored energy and hence responds well over a wide band. Several of the wideband radiators are of the turnstile type, two dipoles at right angles, extending through massive collars.

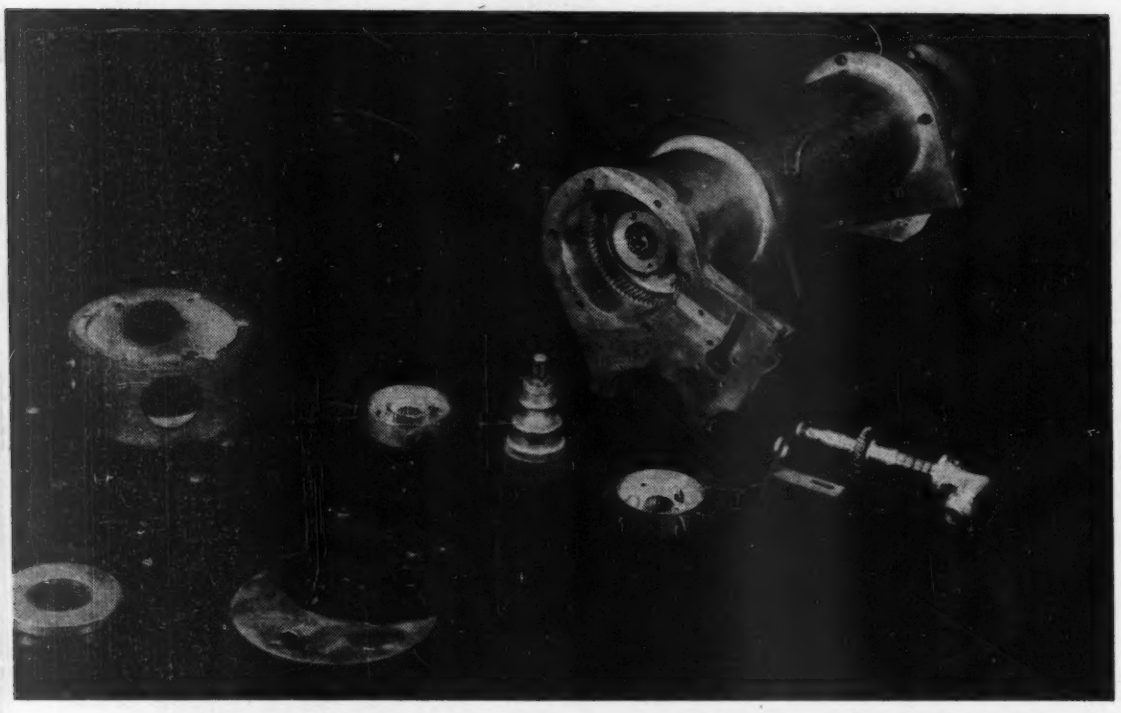
Closely allied with the wideband antennas are suitable direction-finding structures. The direction-finding problem is complicated by the fact that the enemy may choose vertical or horizontal polarization at will. In the AN/APA-24, which operates in the range from 100 to 450 mc, a four-element Adcock system is used to receive vertical polarization, and a single horizontal dipole is used for horizontally-polarized signals. The system operates on the null of the pattern.

For higher frequencies (300 to 1000 mc), an automatic direction finder was produced, using a continuously rotating radiator. A cathode-ray oscilloscope, with polar sweep, indicates the strength of the received signal and plots a polar

diagram on the c-r screen, the maximum of which indicates the direction of the radar under observation. This equipment, when observing a point source, plots the polar radiation diagram of the antenna in use. It has found much use in measuring the polar diagram characteristics of developmental antennas.

Chaff Dipoles and Rope

An effective way to confuse enemy radar operators is to simulate targets by dispersing large quantities of reflecting material in the sky. The most efficient material for this purpose, from the standpoint of echo area per unit weight, is aluminum foil cut in strips one half wavelength long at the enemy radar frequency. Such dipole strips, when used by the British, were called window. The American version is known as chaff. Three quarters of the entire wartime production of aluminum foil, some 20,000 tons in all, was devoted to the manufacture of chaff. Allied aircraft dispensed hundreds of packages of foil strips, each containing several thousand dipoles, on every flight over enemy territory. The material was designed to disperse widely and to remain aloft as long as possible, thus providing a radar smoke screen within which following aircraft could avoid detection by gunfire-control radars below.



Oscillator of APT-9 jamming transmitter, using disk-seal tube. This oscillator covers the enormous range of 300 to 2500 mc with a power output of 10 to 25 watts, continuous wave

While apparently a simple device, chaff presented many interesting technical problems. The primary objective was to obtain as large as possible a reflecting area from a given weight of aluminum foil. This implies thin, pliable foil which tends to bend when thrown into the slip stream of the aircraft. Such bending causes interweaving of the strips in tangled bird's nests which present little area and fall rapidly. Moreover, adjacent strips of foil tend to adhere to one another, preventing rapid dispersal. These problems were solved by embossing the foil and crimping each strip along its length to give it rigidity. Chaff thus manufactured is highly dispersive and falls at the slow rate of 150 feet per minute. The material most widely used over Europe was tuned to the region 450 to 600 mc, which covered the operating frequencies of the German Wurzburg fighter-direction and gunfire-control radars.

Since the resonance of the foil strips extends over a band only 8 percent of the center frequency (at 3 db down), it was necessary to provide two lengths in each package, roughly 10 and 11.5 inches long. About 1000 such dipoles, dispersed at an average separation of about four inches, were found to equal the echo area of a heavy bomber. The weight of these 1000 dipoles in the latest version of the material was only 2 ounces. Each heavy bomber carried with it sufficient chaff to simulate 700 bombers, and dispensed it at regular intervals over areas known to be protected by radar-controlled gunfire. Large areas of the German countryside thus became littered with aluminum strips, which were used by the natives to decorate Christmas trees.

On the indicator of a gun-fire control radar, the chaff-dispensing aircraft appears as if it were reproducing itself. As the pulse representing the aircraft moves across the indicator screen, additional pulses appear behind it and remain stationary. As the chaff disperses, the pulses assume an amorphous shape in which succeeding aircraft are nearly invisible. Aircraft outside the cloud of dipoles are not hidden. Aircraft behind (but not within) the cloud are detected by signals which pass through the dipoles.

The most effective protection against German radar-controlled flak was a combination of electronic jamming by transmitters tuned to the radar frequencies, and chaff dipoles. In addition to adding to the general confusion on enemy indicators, electronic jamming protected the first plane in a flight as well as succeeding planes. This combination reduced the effectiveness of anti-aircraft fire to about 25 percent of normal, which saved the U. S. forces an estimated 450 aircraft and 4500 casualties. The value of these aircraft alone more than equalled the cost of the countermeasures program directed against German flak. During the height of the campaign, 20 billion dipoles were scattered on Germany and France each month.

In the Pacific, chaff was not used to any great extent because the Japanese radars used many widely different frequencies, which would have required as many different sizes of chaff to combat them. Instead, very long strips of aluminum foil, about one half inch wide and 400 feet long, were dropped, sometimes supported from small parachutes. This device, known as rope, was effective over a very wide range, covering all of the many frequencies employed by the

Typical wideband antenna structure, used to cover wide frequency range of jamming transmitter. Known as a fish-hook, this radiator produces circularly polarized signals to jam regardless of the polarization of an enemy antenna

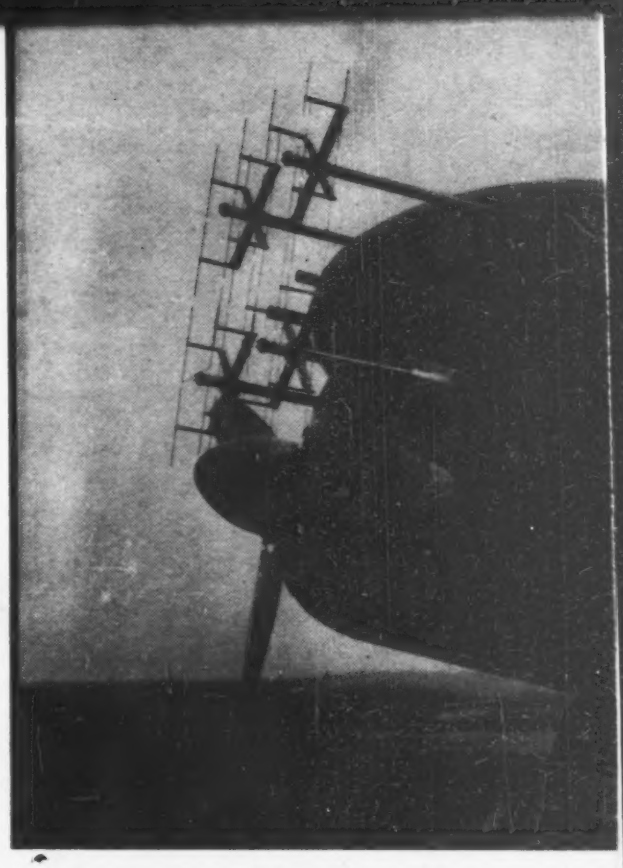
Jap equipment. Eventually, each B-29 carried 600 pounds of this material on every mission. This weight necessarily subtracted from the bomb load, but was well worth it.

Tuba and the Resnatron

Perhaps the most spectacular of all rcm developments was the high-powered ground-based jammer known as tuba. Beginning in 1942 the Germans installed airborne Lichtenstein radars in their night-fighters, operating at about 500 mc, and designed to detect night-flying British bombers. So successful was this German equipment that consideration was given to an airborne jamming equipment to be carried in the British planes, but this plan was abandoned when it became clear that the Germans could locate the bombers from the ground by accurate d-f equipment. So it was decided to install a super-power jammer on British soil, aimed at the coast of France and intended to blot out the screens in the night fighters as they chased the bombers home. To achieve this result, the highest possible power was required. Calculations indicated that tens of kilowatts, continuous wave, tunable from 375 mc to about 600 mc would be required to cover the possible tuning range of the German equipment with a sufficiently strong signal. The highest c-w power achieved up to that time at that frequency was measured in the tens of watts.

The answer was found in the resnatron, a water-cooled, continuously-pumped tetrode which overcomes the effects of electron transit time by a 90-degree phase shift introduced between grid and plate circuits. This tube achieved the astounding power output of 30 kw, continuously, at a frequency of 500 mc, and was pushed to an output of 100 kw for short periods. The upper frequency limit of the tubes developed for tuba was found to be about 700 mc.

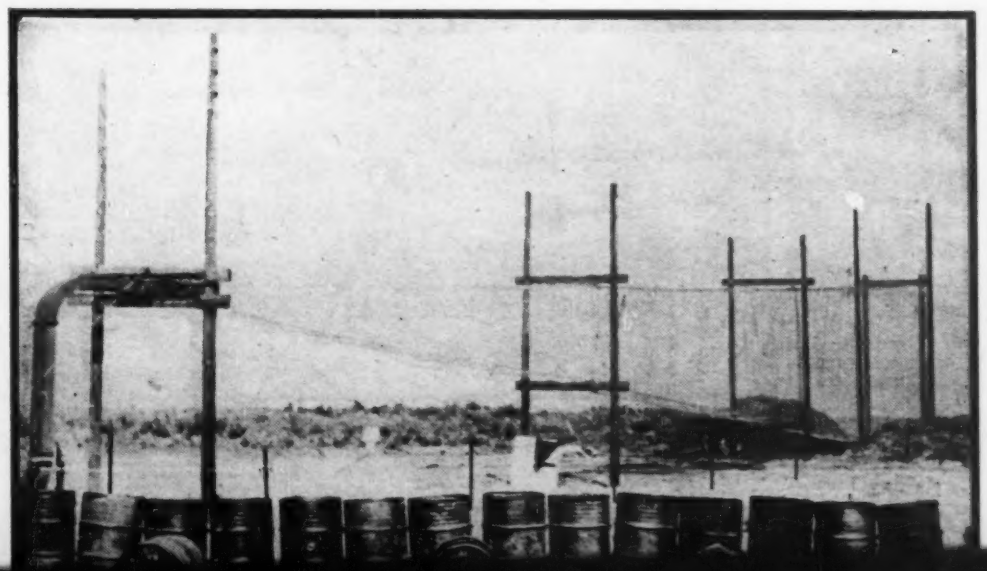
The original design was modified to permit tuning it through the range (about 200 mc) over which the Lichtenstein airborne radars could be shifted by simple modifications. The tuba equipment consisted of two oscillators, noise modulators, power supplies, etc, mounted in trucks. Two oscillators were required to cover the frequency range, not only for convenience in tuning the resnatrons themselves, but also to permit the use of waveguides. Since a waveguide of given crossection can transmit power over but a limited range of frequency, two systems were used to cover the band. The waveguide cross-sections were respectively 16 by 6 inches and 22 by 6 inches, which is easily the largest waveguide system ever used. The radiator itself was a sectoral horn, constructed of chicken wire supported on telegraph poles, 150 feet long from neck to mouth. Work on tuba began in America early in 1943. The first operation against the enemy began from the south coast of England June 1944. In that month the Germans changed to a radically different type of airborne interception radar.



Antenna of German Lichtenstein airborne radar, operating at about 500 mc and used on night fighters to combat Allied bombers. This successful equipment led to the development of a super-high-power jammer known as tuba

Aside from its military success, the resnatron is a major milestone in the history of electron tube development. It can develop more power than, perhaps, can be used economically for any peace-time purpose at frequencies above 100 mc. While its modulation capabilities have not been thoroughly investigated except for random noise waveforms, the advance indications are that it can be modulated in frequency, phase and amplitude over very wide sidebands. Its importance in the future of f-m and television broadcasting can scarcely be doubted.—D.G.F.

Radiator of the tuba jammer, a chicken-wire horn 150 feet long. Based on the south coast of England, and fed with 30 kw of c-w power at 500 mc, this equipment successfully jammed the Lichtenstein radars over Europe. Note 22-by-6-inch waveguide feeding horn at left



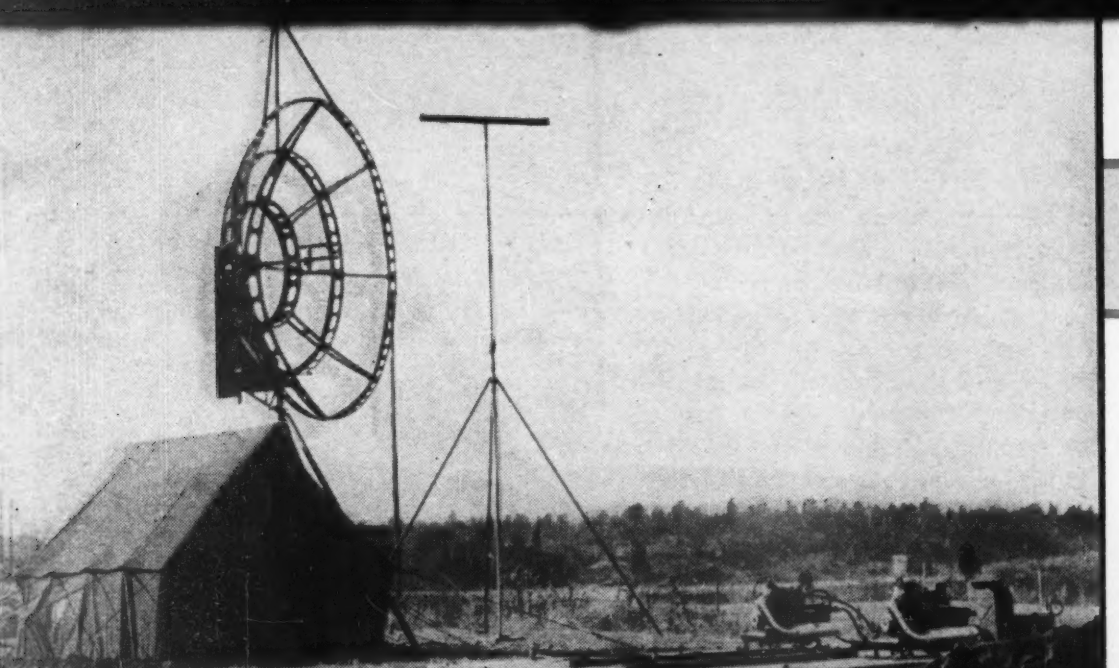


FIG. 1—Typical setup of the TPS-3 radar

RADAR

By

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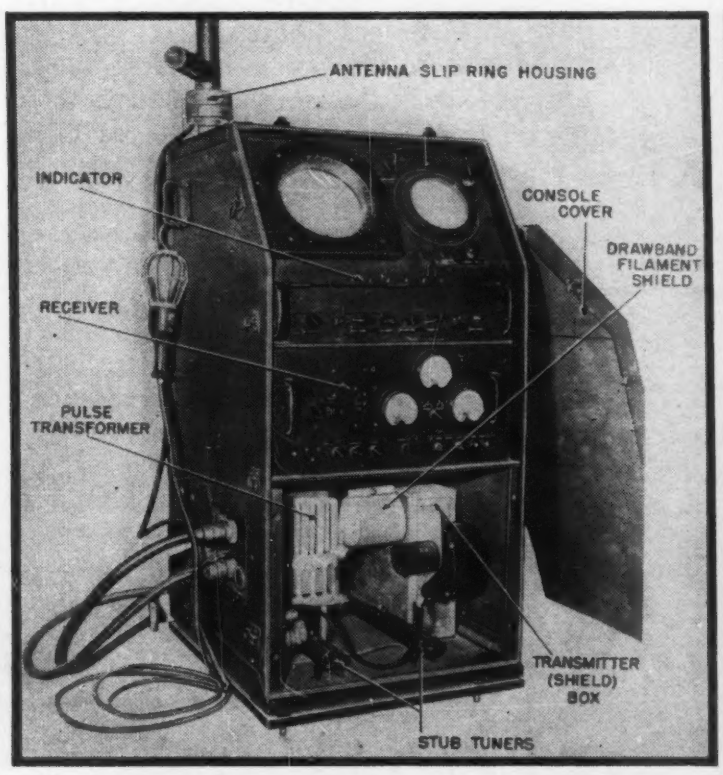
PRIOR TO AMERICA'S entry in the war, available radar equipment operated on 100 and 200 mc. This equipment had been set up at the Panama Canal, but it was felt that additional measures were necessary to avoid surprise raids by low-flying aircraft against which the existing radars were least effective.

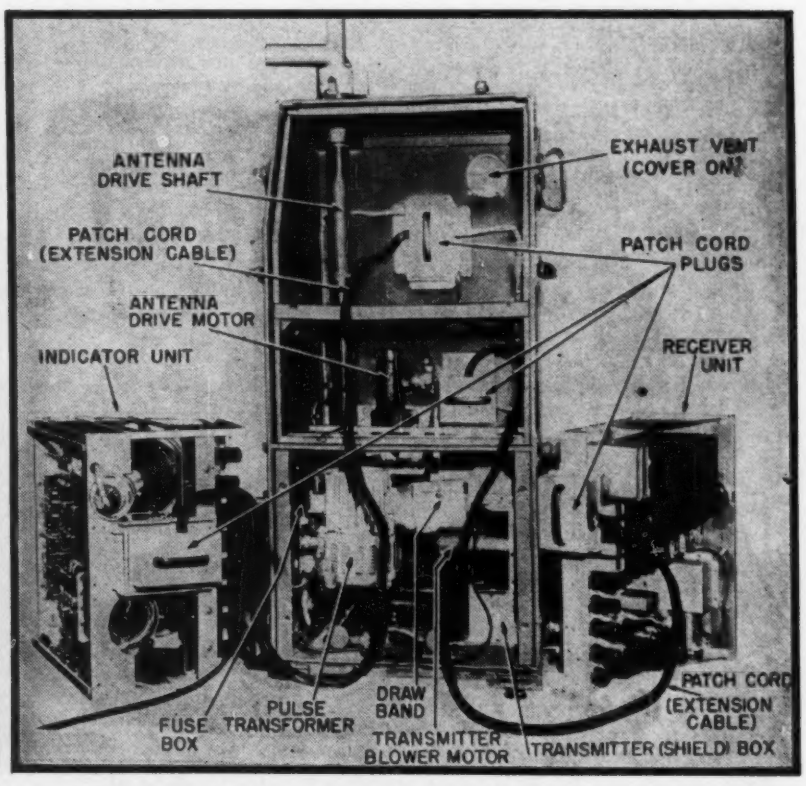
A plan was set up to construct a small number of radar sets which were to be mounted on small boats anchored in the vicinity of the Canal entrances. It was felt that in these advance locations they would provide a radar screen making surprise raids impossible. After considerable experimentation, it was decided to construct these sets to operate on a frequency of 600 mc and, accordingly, a model was built up of components already developed by the Signal Corps and an installation was made on the motor vessel "Nordic."

Tests of this equipment were so successful that it was immediately apparent that extremely long ranges and low angle coverage were obtainable in this frequency band even from a set only 15 feet above sea level, and that with the components available a very lightweight medium-warning radar could be constructed. At the request of Col. William Cody and others of the AAF, the Signal Corps was asked to repackage this equipment into a lightweight assault-type radar that could be both air transportable and hand-carried and have a range of well over 100 miles on bombardment aircraft.

To prove that the first laboratory model was air transportable it was flown from Newark Airport to Florida on February 27, 1943, in a B-18 and was set up and operating

FIG. 2—A view of the console and, at the right, of the console with the indicator and receiver units removed, showing the maintenance cable system





ON 50 CENTIMETERS

Combining high power with light weight, the TPS-3 radar detects approaching bombers at 120 miles, yet can be carried by hand and set up by a four-man crew. This first of two articles on the equipment furnishes an important example of 600-mc technique

at the test site on March 1, 1943. The next two weeks were spent in calibrating and determining operational performance of this equipment in comparison with three other lightweight equipments. At the conclusion of the tests it was definitely determined that the AN/TPS-3 (then known as the 602-T8) had a range in excess of 110 miles and could be mass-produced. This performance was sufficient to indicate an immediate combat requirement. The model was flown back to Camp Evans, Belmar, N. J., on March 18, 1943. The engineers responsible for the design of the equipment took all of the information available to a manufacturer, along with the model, so that production might start as soon as possible. Nine hundred sets were ordered, and the first started coming off the production line about a year later.

To cover the interim period it was necessary to produce on a crash basis a small number of sets for immediate air shipment to critical theater areas. Accordingly, it was decided to construct 12 models within the Camp Evans Signal Laboratory. These 12 models were completed in three months with the aid of GI crews who later formed the operating teams for the equipment and were flown directly to the theaters.

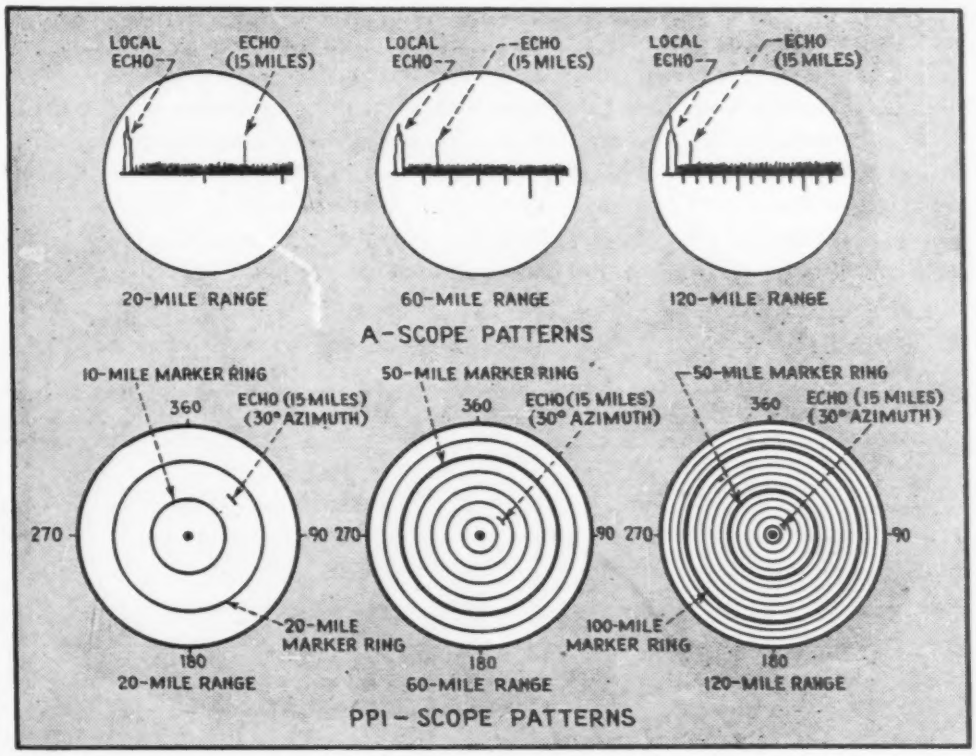


FIG. 3—A-scope and PPI-scope patterns

The production models of this set found their way to all the theaters of the war. The first 25 were produced in time to take an active part in the Normandy invasion.

General Description

The AN/TPS-3 was designed chiefly for medium-long-range early warning against aircraft. It is composed of units which are small and light enough to be either transported by air or hand-carried. Its total weight (including spare parts and power units) when packed for air transport is 1200 lbs. Maximum weight of any single component is 200 lbs. The set can be completely assembled and put on the air by a crew of four men within thirty minutes of arrival at a site. A typical installation is shown in Fig. 1.

The major component is a console which houses the receiver, the transmitter, the indicator, part of the modulating system and part of the r-f system. This unit is shown in Fig. 2. The console is normally housed in a tent which is provided with the set and acts as both a light-proof covering and a shelter. A section of transmission line with very heavy steel walls plugs into the top of the console and forms a pedestal upon which a 10-ft parabolic-reflector antenna system is mounted. This section of transmission line is braced by two wooden struts whose ends are buried in the ground. The top of the parabolic reflector is further secured by three guy wires and the entire structure is so made that the antenna can rotate continuously in either direction or be inched

electronics WAR REPORT

slowly for accurate azimuth orientation.

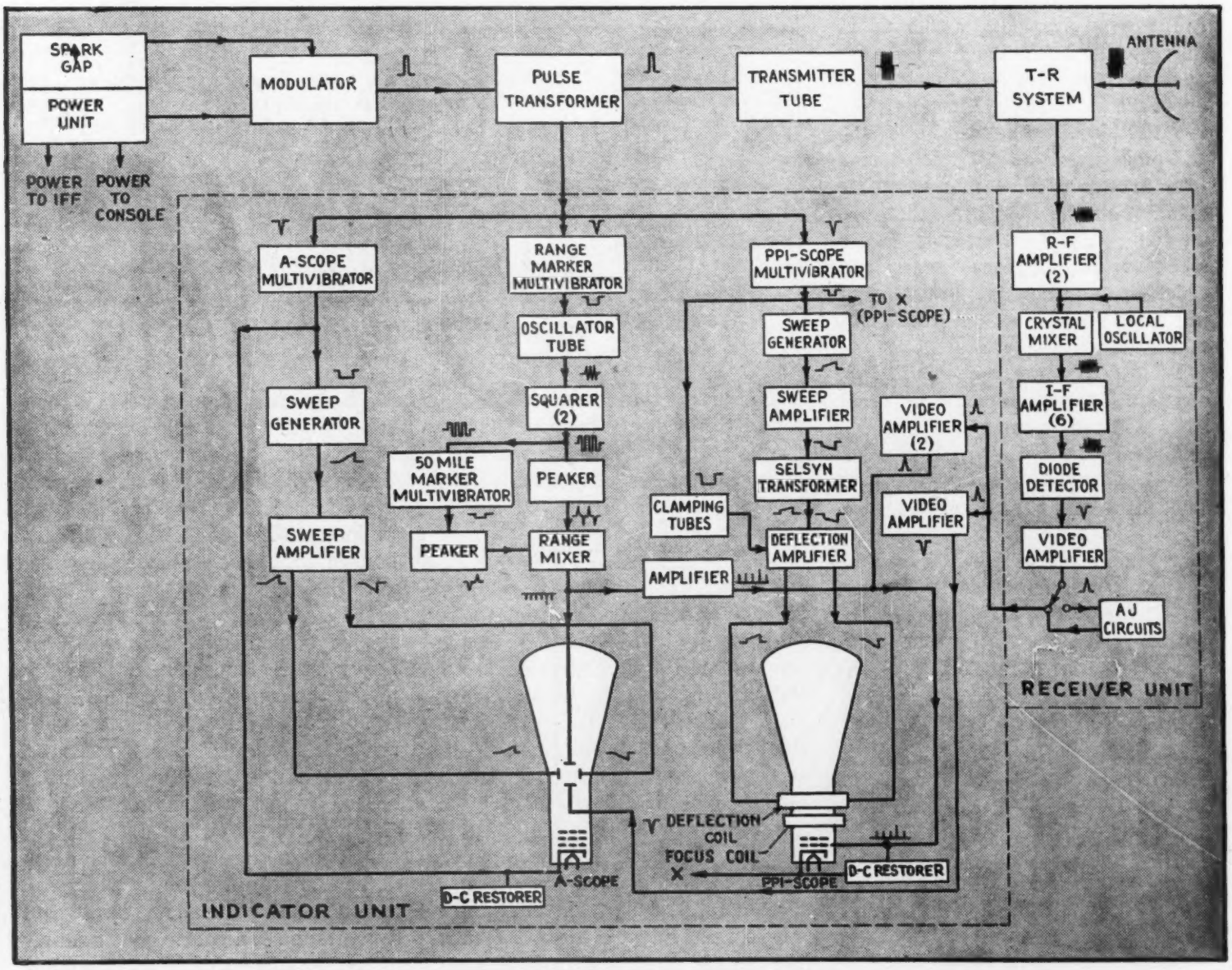
The power unit and modulator are kept 50 ft from the tent and are connected to the console by means of cables. The power unit is a single-cylinder gasoline engine driving a 400-cycle alternator and a d-c generator mounted on the same shaft. The radar components of the set use 400-cycle power, thus effecting a great saving in weight and size. The d-c generator produces 28 volts that is used to drive fan motors and the antenna turning motor.

Unlike the SCR-268 previously described in ELECTRONICS, the AN/TPS-3 provides only range and azimuth information. This information is displayed on the indicator on two cathode-ray tubes used in an A-scope and in a PPI-scope. The A-scope resembles an ordinary test oscilloscope and presents a horizontal sweep line with the signal appearing as vertical deflections along this line. The PPI presentation employs a sweep line which starts at the center of the tube and sweeps outward toward the edge. This sweep line is made to rotate about the center of the tube in synchronism with the rotation of the antenna and therefore indicates the position of the antenna. Signals are applied to this tube by intensity modulation so that they appear as bright dots on a dark background. Both of these tubes indicate the range of the target by the distance of the signal from the beginning of the sweep line. The PPI tube measures azimuth by noting the position of the sweep line when a signal appears. The two forms of presentation are shown in Fig. 3.

The pulse rate of the set is 200 pulses per second. The interval between pulses is 5000 microseconds. Using a sub-multiple of the alternator frequency for timing the pulses provides a clear and steady picture on the indicators with a minimum of filter weight. The transmitted radio-frequency power and the sensitivity of the receiver are such that the radar set will "see" a medium-size aircraft at 120 miles. The total time necessary to cover this 120 miles is approximately 1300 microseconds, and this amount of time is all that is used on each sweep of the cathode-ray tube. The remainder is dead time.

Figure 4 is a block diagram showing the flow of signals through the equipment. The modulator, in conjunction with a rotary spark wheel, mounted on the power unit, produces high voltage d-c pulses which are synchronous with the 400-cycle power supply at half its frequency.

FIG. 4—Block diagram of the complete TPS-3 radar



These pulses are applied to the transmitter, which converts them to pulses of 600-mc energy. They are then radiated from the antenna, which is coupled to the transmitter by means of a coaxial line. When one of these pulses strikes an object some of the energy is reflected. Measurement of range is facilitated by injecting into both cathode-ray tubes a series of markers spaced 107 microseconds apart, and also synchronized with the transmitted pulses. These markers represent 10-mile intervals and provide the scale by means of which the range is measured.

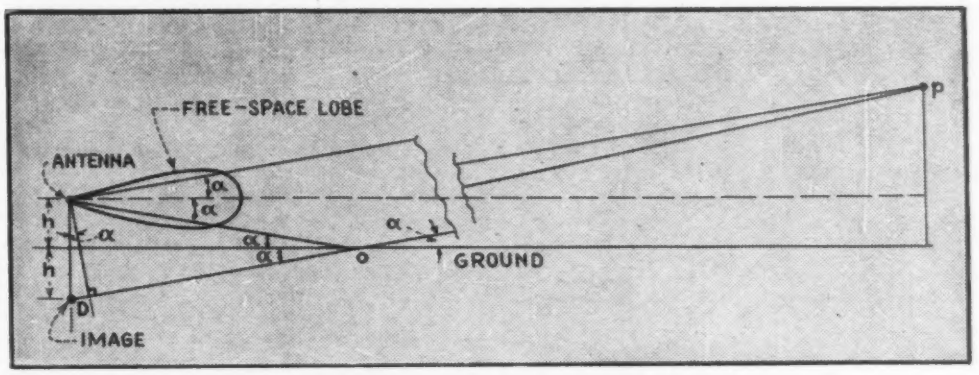


FIG. 5—Determination of reflected energy maxima and minima

Antenna and Propagation

The antenna used with the AN/TPS-3 is a 10-ft parabolic reflector with a radiator at its focus. This produces a free-space beam which is about 10-deg wide at its half-power points. Consider such an antenna situated at height h above a plane earth as shown in Fig. 5. Assume a reflecting target at point p at a great distance from the antenna. The antenna will appear as a point source as seen from p. The radiation pattern shows the free-space pattern of the antenna. It is obvious that energy can reach the point p from the antenna by traveling two paths, one directly from the antenna and the second reflected from the ground. These will be called the direct ray and the reflected ray. The angle made by the direct ray with the horizontal is almost exactly equal to the angle made by the reflected ray and the horizontal. Therefore, the amount of energy reaching point p along each of these two paths will be almost equal. However, because the distances along the two paths are not equal, the phase of the direct ray and the reflected ray at point p will in general not be the same. Therefore, the total energy at point p is the vector sum of the energies reaching it along the direct path and reflected path.

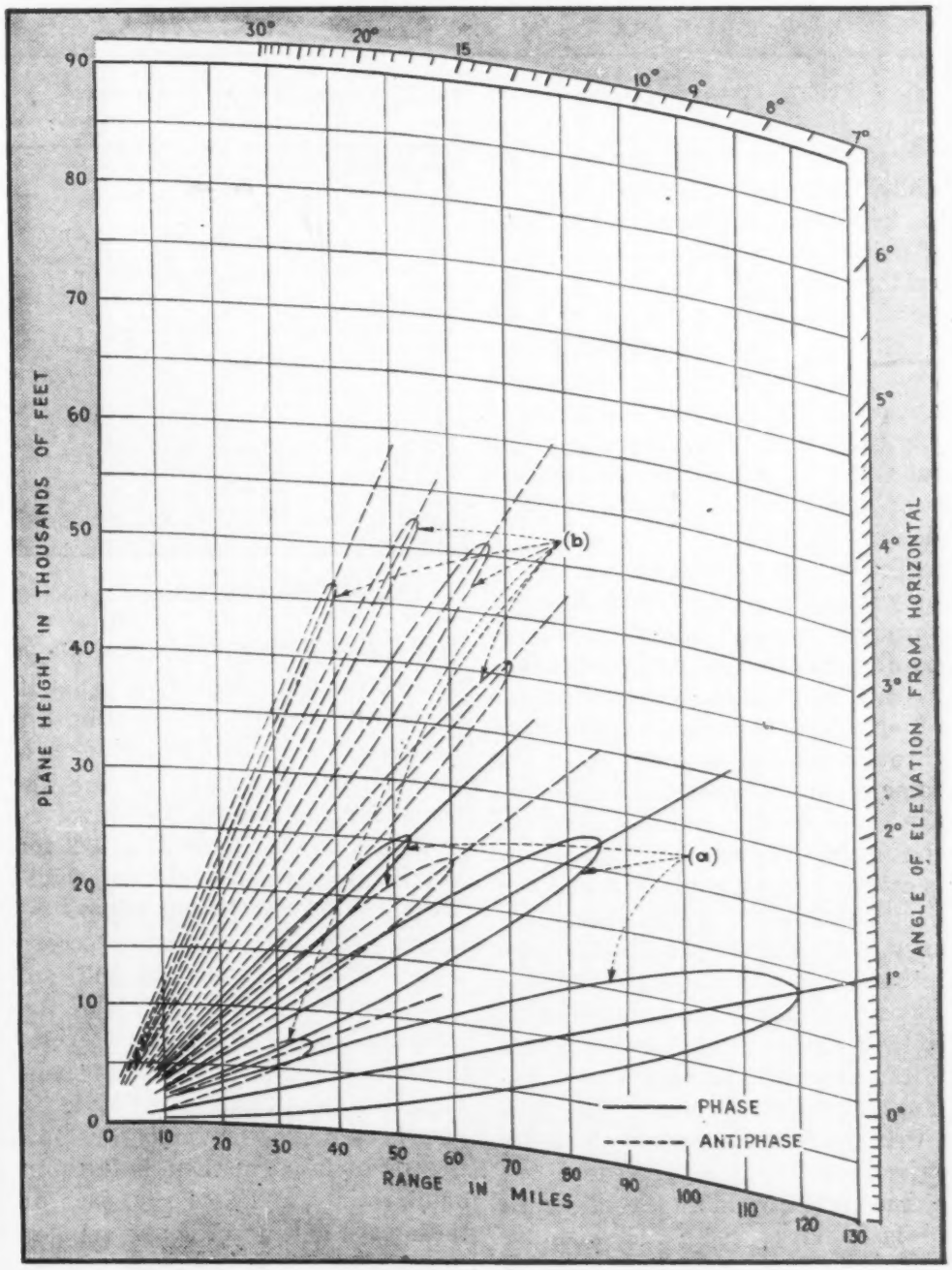


FIG. 6—Complete coverage pattern, with the antenna horizontally polarized and having a 24-ft effective height

In order to determine this total energy it is not necessary to know the length of each path but merely the difference between the path lengths, which will determine the difference between the phases of the direct and reflected rays at point p. It may be assumed that the earth is a perfect reflector. At the point of reflection there is a 180-deg change of phase. Because radio energy travels with the speed of light, 180-deg phase differences occur every half-wavelength in space. Therefore, in order for the direct ray and reflected ray to arrive in

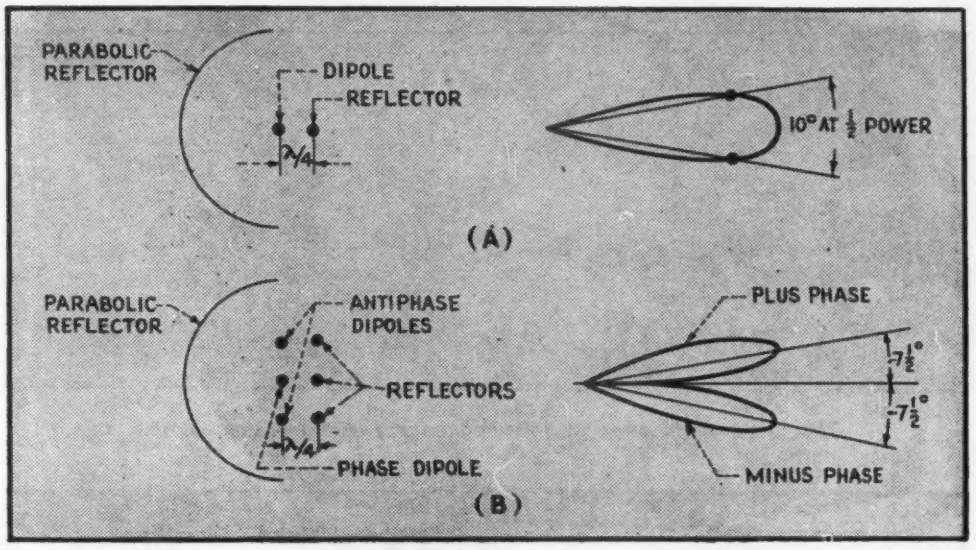


FIG. 7—Free-space patterns for the two TPS-3 radar antenna connections

phase at point p and add algebraically the path difference must be a half wave. Whenever the path differs from a half wave or some odd multiple of a half wave, the total energy at point p will be slightly less than the sum of the direct and reflected ray. Whenever the path difference is an even multiple of a half wave, the energy at point p will be zero, because the direct and reflected wave arrive 180 deg out of phase.

To find this path difference it is only necessary to extend line op through the point of reflection to a point directly under the antenna. It can be seen that the extension of op will intersect a vertical line through the antenna at a point the same distance below the surface of the earth as the antenna is above the surface of the earth. This point is called the image antenna and can be considered to be radiating energy toward point p with the same intensity as the true antenna but opposite in phase. A perpendicular is then dropped from the antenna to the extension of op and the distance D then represents the path difference. While there may be some objections to this procedure on geometrical grounds the distance op is so large compared with $2h$ that for all practical purposes the results are correct.

If point p is moved up and down the vertical line it will be seen that the distance D varies greatly, passing through several odd and even multiples of half-wave lengths. This means that the distribution of energy along the vertical line through p will pass through maxima and minima. The first maximum will occur when D is equal to a half-wavelength. The relationship between the angle of this first maximum and the height of the antenna for the small angles involved is then $\alpha = \lambda/4h$ since $D = \lambda/2$ for maximizing, where λ is the wavelength and h the height of the antenna. The first minimum occurs where $D = 1$ wavelength and will therefore be twice the angle of the first maximum. The resulting complete radiation pattern is shown in Fig. 6 at (a). The succeeding lobes get shorter and shorter because although each one of them is the sum of direct and reflected ray, each of these rays gets smaller and smaller in energy value as the angle α is increased and finally reach zero when the direct and reflected paths are tangent to the free-space radiation pattern. This means that the envelope of the complete radiation pattern will be the same as the free-space radiation pattern, although the length of the longest lobe will be twice the length of the free-space lobe because it is the sum of two rays, each of which is nearly equal to the maximum energy in the free-space radiation pattern.

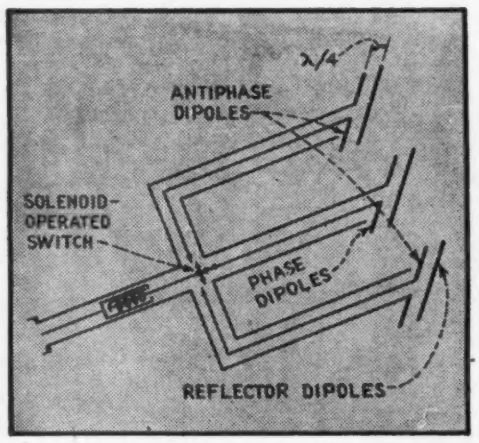


FIG. 8—Triple-head antenna system

From the complete radiation pattern it can be seen that there are large areas where no radio energy is present. This would allow enemy aircraft to fly into these areas without being detected. Many schemes have been used to overcome this deficiency in early warning radar sets. The one adopted for the AN/TPS-3 is known as the phase-antiphase system. Looking again at Fig. 5 it can be seen that if point p is at a minimum and by some means the reflected ray could be moved 180 deg out of phase with the direct ray, point p would immediately become a maximum. This phase shift is the basis of the phase-antiphase system. The antenna is arranged so that by means of switching it can produce either the free-space pattern shown in Fig. 7A or the free-space pattern shown in Fig. 7B; in the latter case the upper lobe is 180 deg out of phase with the lower lobe.

The antiphase radiation pattern is shown in Fig. 6 at (b). Here the lobes are small at the low angles, pass through a maximum, and are small again at the high angles. This occurs because the energy in the free-space pattern is low at small angles, reaches a maximum at about 7½ deg, and then decreases again. Superimposing the phase and antiphase coverage patterns gives a pattern of the total coverage of the radar set. Contours can be drawn representing the loci of point where the energy is just sufficient so that the amount reflected from a medium-sized aircraft will be detectable by the receiver. Such contours represent the maximum range.

From the discussion on formation of maxima and minima it can be seen that the coverage diagram can be greatly altered by changing either the frequency or the height of the antenna. The height of the antenna is usually restricted by mechanical considerations and features of the terrain. Therefore, to get low-angle coverage as high a frequency as possible is used. That is why 600 mc gives more efficient low-angle coverage than the lower-frequency radar sets. Higher frequencies would give still better low angle coverage.

Radio-Frequency System

As stated previously, the AN/TPS-3 employs a 10-ft parabolic reflector. It is also provided with two different antenna feeds that can be used interchangeably. One is a simple halfwave dipole at the focus of the parabolic reflector, with a parastically excited halfwave reflector a quarter wave in front of it. This combination produces a single free-space lobe such as that shown in Fig. 7A. The purpose of the reflector is to prevent direct radiation from the dipole, and results in an increase in gain.

The second type of antenna feed is the one used to produce phase and antiphase patterns. This array consists of three dipoles spaced vertically a quarter wave apart in the plane of the focus, with center dipole at the focal point. Each one of these dipoles has its associated half-wave parasitic reflector a quarter wave in front of it. These three dipoles are so arranged that either the center dipole alone or the two outside dipoles may be driven. When the center dipole alone is driven the result is a single free space lobe such as shown in Fig. 7A. The outside dipoles are so connected that they are fed 180 deg out of phase, which is very simply arranged by connecting the left side of the upper dipole to the center conductor and the right side of the lower dipole to the center conductor and then feeding the pair in parallel from a common point. When these two dipoles are driven the result is a split pattern such as that shown in Fig. 7B, with the upper lobe 180 deg out of phase with the lower lobe.

The switching between the phase dipole and the antiphase dipoles is done by means of a solenoid-operated plunger which is controlled from a switch on the panel of the radar set. The plunger merely connects the center conductor of the transmission line to either the center dipole or the outside dipoles as shown in Fig. 8.

The antenna radiators are connected to the transmission line by means of a 50-ohm rigid coaxial line. In order that this line be flat, or without appreciable standing wave, every precaution is taken to match the antenna radiator to this line. This is done by means of quarter-wave transformers consisting of sections of inner conductor of different diameter from the diameter of the inner conductor of the transmission line itself. By properly choosing the diameter and position of such a quarter-wave transformer, the antenna radiator can be matched to the transmission line.

The requirement of a flat line precludes the use of insulating beads as supports for the inner conductor, since many beads in the line would produce reflections and, therefore, standing waves. Instead of beads, quarter-wave stubs are used to support the inner conductor. A section of transmission line, short-circuited at one end and a quarter-wave long, has an extremely high impedance looking into the open end. When such a section is shunted across the line it does not produce any appreciable reflection. Such quarter-wave stubs, spaced at intervals along the transmission lines, can be used to support the inner conductor. However, since a stub is a sharply tuned resonant circuit it can be used only at one frequency. The AN/TPS-3 operates over a band from 590 to 610 mc, so that provision must be made for allowing the transmission line to pass this band of frequencies without appreciable reflections. This is done by making the characteristic impedance of the support stubs 100 ohms as compared to 50 ohms for the transmission line proper.

Another interesting feature of this transmision line is the rotary joint. The antenna must be able to rotate continuously, so the transmission line must be broken at some point and means provided to pass energy from the stationary to the rotating side. This breaking is done

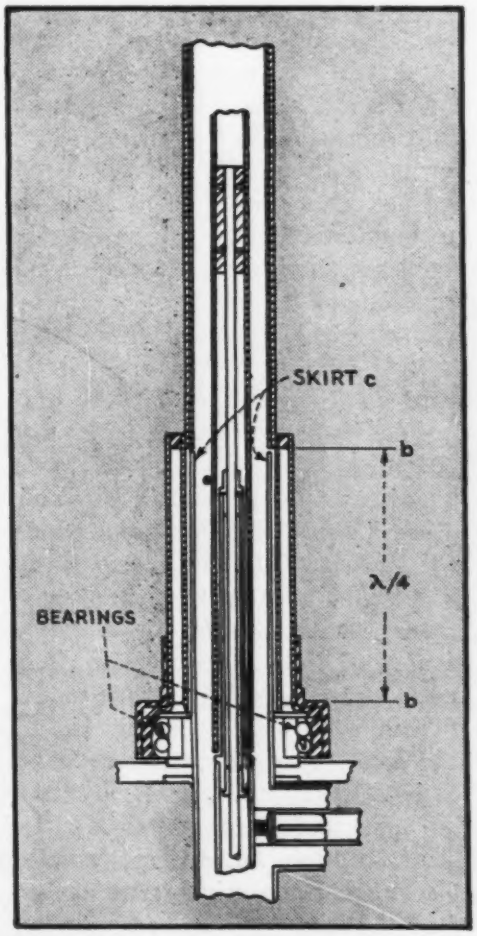


FIG. 9—Transmission-line capacitance joint

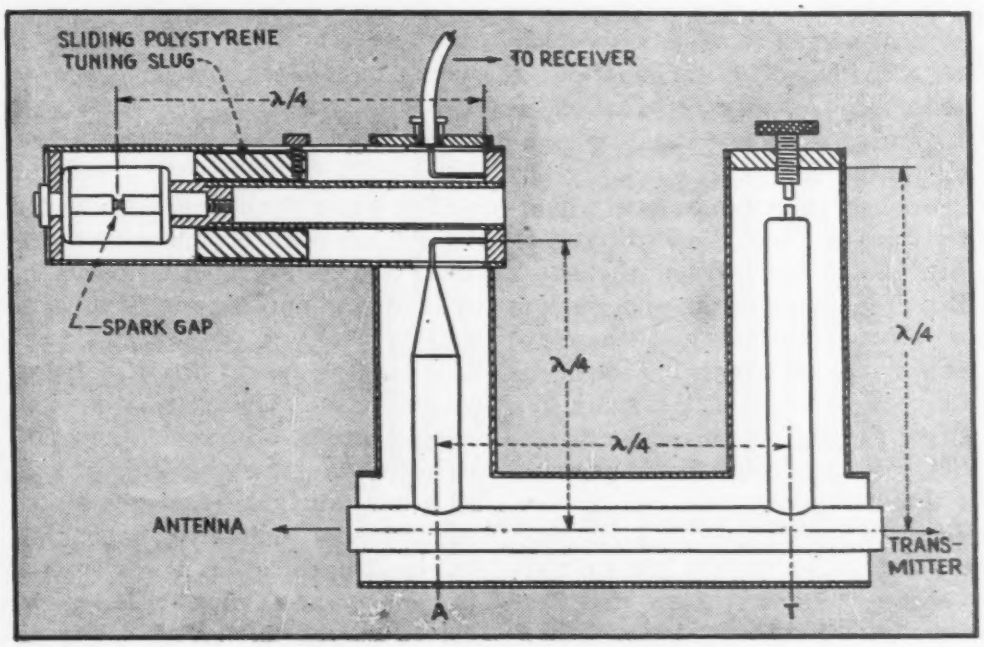


FIG. 10—Sketch of the t-r switching system

by means of a capacitance joint. A sketch is shown in Fig. 9.

The sections *bb* can be considered as forming a transmission line of very low characteristic impedance by themselves. Because this line is open circuited at the antenna end and is a quarter wave long, it is effectively short circuited at the transmitter end. The transmitter energy therefore passes from the stationary to the moving part. The function of the skirt *c* is for broad banding and also allows the bearing to be placed at a zero-current point. This bearing carries the total thrust load of the super-structure.

The inner conductor functions in essentially the same manner except that it is unnecessary to fold the line back upon itself in a skirt as in the case of the outer conductor. In the inner conductor, the ¼-inch rod extending beyond the rotary joint itself performs the same function as the skirt in the outer conductor. Oilite bearings are used on the inner conductor.

Moving down the transmission line toward the transmitter, the next interesting feature is the t-r or transmit-receive system. As shown in the block diagram (Fig. 4) the transmitter and receiver are connected to the same antenna. Some means must be provided to protect the input of the receiver from the high power present in the line when the transmitter is operating and to prevent the received signal from being dissipated in the transmitter instead of flowing into the receiver. This is accomplished by a combination called a receiver-disconnect switch and a transmitter-disconnect switch. The combination of these two components is called the t-r system. A functional diagram is shown in Fig. 10. The transmitter-disconnect switch consists of a quarter-wave section of line with a spark gap in the end. When the spark gap fires, this quarter-wave section is short circuited and therefore presents a very high impedance at its open end. When the spark gap is not firing, the quarter-wave section is open circuited and presents a short circuit at the transmission line. A quarter wavelength toward the antenna from the transmitter-disconnect switch is the receiver-disconnect switch. This consists of a high-Q cavity a quarter wavelength long, short-circuited at one end with a spark gap between the center conductor and ground at the open end. The transmission line and the receiver are both coupled into this cavity by means of loops. When the spark in this cavity is firing, the cavity is detuned and the receiver is decoupled from the transmission line. When the spark gap is not firing the cavity is tuned and energy passes freely from the transmission line to the receiver.

The operation is as follows:

When the transmitter is operating, a high voltage is present in the line. This fires the spark gap in the transmitter-disconnect switch, which then acts as a quarter-wavelength short-circuited stub and has no effect on the line. A high voltage is also built up in the cavity of the receiver-disconnect switch, firing its spark gap as well. This effectively decouples the receiver from the transmission line. When a received signal comes down the transmission line from the antenna, the voltage is extremely small. This voltage is far too small to fire either the spark gap in the transmitter-disconnect switch or the spark gap in the receiver-disconnect switch. The transmitter disconnect switch will therefore be an open-circuited wavelength and will appear to be a short circuit at the transmission line end. A quarter wavelength toward the antenna, at the point where the receiver-disconnect switch ties in, this will be reflected as an open circuit. Therefore, no energy will flow toward the transmitter. On the other hand, the receiver-disconnect switch is now a high-Q tuned cavity. The received energy will therefore flow freely from the transmission line cavity into the receiver.

The transmission line is connected to the transmitter through a network which provides a means of matching impedances. This network consists of two stubs of variable lengths spaced 3/5 of a wavelength apart. These are known as tuning stubs. Since a short-circuited stub of variable lengths acts as a pure variable reactance either capactive or inductive, depending on its length, such a stub when properly placed may be used to match a transmitter of arbitrary impedance to a transmission line. However to avoid the mechanical difficulty of properly placing this stub on the transmission line two stubs are used and spaced 3/5 of a wavelength apart. By properly adjusting the length of these stubs, any impedance from infinity to one half the impedance of the line can be matched to the line.

The next and concluding installment on the AN/TPS-3 will describe the transmitter, modulating system, receiving and indicating system.

Acknowledgments

Responsibility for the construction of the AN/TPS-3 was directly assigned to J. W. Marchetti, who was assisted by William P. Goldberg as civilian engineer in charge. Many individuals of the Camp Evans Signal Laboratory contributed in the design of various features of this equipment. The modulator, indicators, and transmitter were turned over to a group headed by Dr. John E. Gorham. Within this group H. P. Pacini was responsible for the design of the indicators and I. Sager for the modulator. The transmitting tube, which was designed by the first named author, required very little additional development, since it had been in semi-production for some time and required only minor changes to adapt it for use in a lightweight radar set being styled for mass production. Later, when the set went into production, Dr. Gorham handled the manufacturing problems incidental to the construction of the transmitting tube. The physical arrangement of the set was due to William J. Smith and he further played a large part in the overall coordination. The major part of the mechanical design of this set was due to Arthur H. Hood.

The VT-158 vacuum tube, which is the heart of the set, is unique in that it includes all of the transmitter under vacuum in a single envelope, thus minimizing seal losses and arcing. The first model of this tube was built by H. A. Zahl in 1939 at the suggestion of Major General Roger B. Colton. The personal interest shown by General Colton, and his many valuable suggestions during the design of all elements of this radar, in no small way contributed to its success.

Vehicular-Mounted MINE DETECTOR

Detection of a mine, or failure of the electronic apparatus, automatically sets the brakes of a jeep before the front wheels reach the explosive. False signals are discriminated against by phase selection. Variometer tube minimizes circuit drift

By

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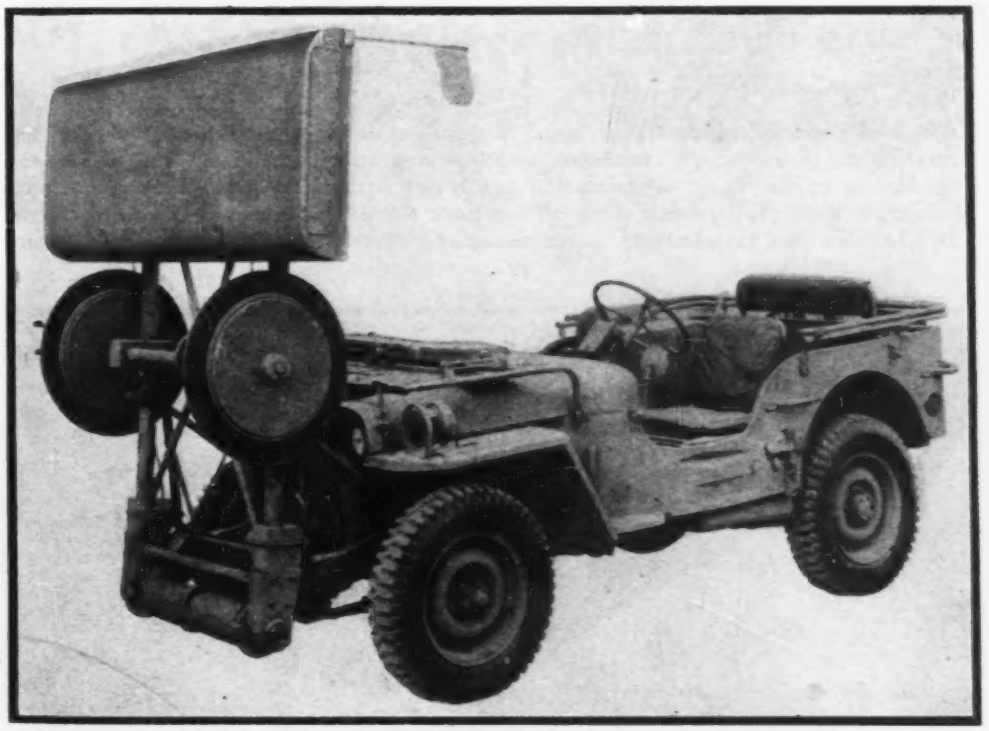

The mine detector mounted on a jeep, showing how articulation on the front bumper permits it to be raised to a vertical position when not in use

IN ORDER to speed up the clearance of mines in roads and fields a vehicular-mounted detector was developed during the war.

The problem was to devise a metallic-mine detector, the sensitive elements of which could be carried in front of a vehicle running at a reasonable speed, the system being so arranged that the vehicle would be automatically stopped before rolling over the mine and exploding it.

The circuit adopted is similar to those used in portable mine detectors, comprising a transmitting coil energized by an oscillator, and a receiving coil connected to an amplifier, the mutual impedance between the two coils being nullified in the absence of mines. Vehicular operation calls, however, for a number of special features which are described in the following paragraphs.

The Boom

In a vehicular device the coils obviously have to be carried ahead of the vehicle, and therefore must be supported by some kind of boom. If a rigid attachment were used, the elevation of the coils above ground would vary too much in rough terrain, therefore an articulation must be provided which makes it possible

* All three co-authors on leave of absence, at Electro-Mechanical Research, Inc., Houston, Texas

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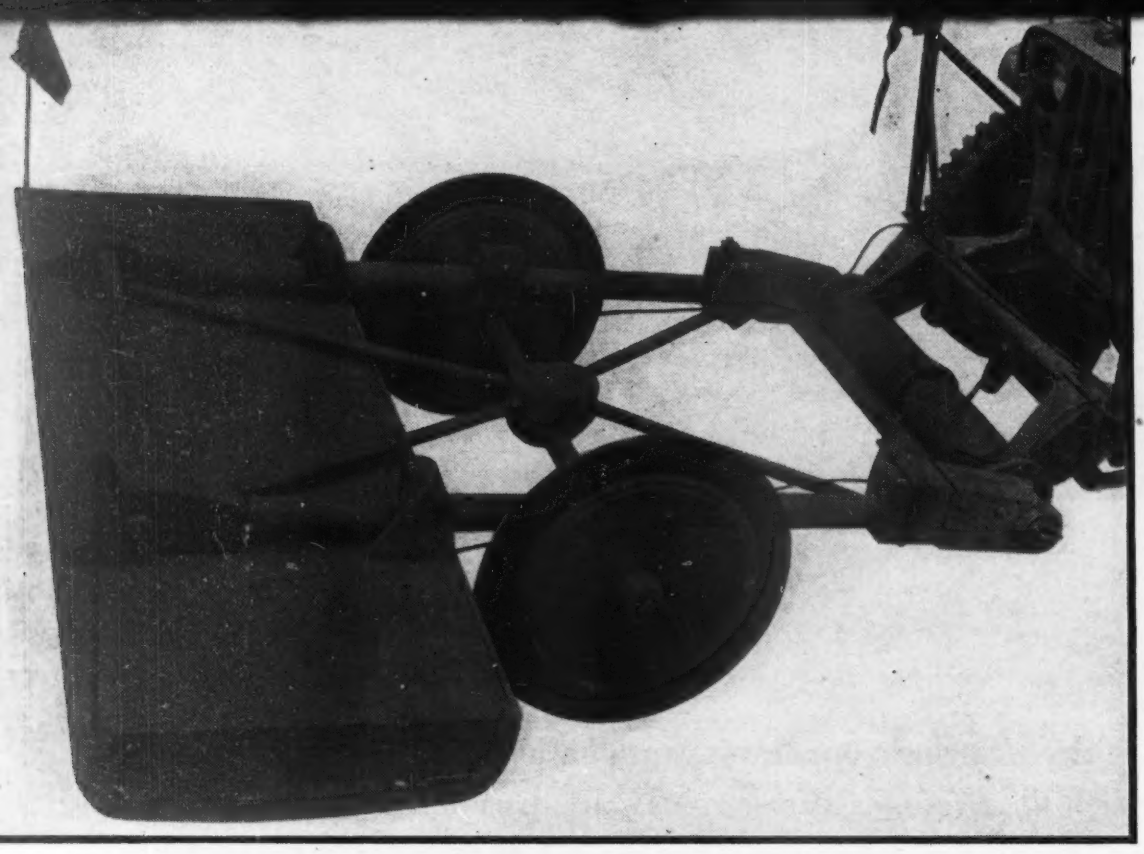


FIG. 1—The vehicular-mounted mine detector with boom lowered

to keep the coils at a reasonably short distance from the ground.

As a manual operation of the boom is impossible when the vehicle moves at a speed of 5 to 8 miles an hour, as required, and as a system of servo control would be complicated and bulky, the solution was to balance most of the boom weight by means of an adjustable torsion-spring system, and to rely on wheels supporting the boom exerting a low pressure on the ground. The boom structure, in other words, was made sufficiently sensitive so that its weight, working against the torsion system, keeps it at a reasonably uniform height above ground at the normal driving speed. The structure is shown in Fig. 1.

To prevent electromagnetic action on the coil system, the boom is built of plywood tubing and wooden blocks. Due to the articulation on the front bumper, the entire boom assembly can be set in a vertical position when not in use. This is the traveling position which provides adequate visibility for the driver through the boom frame.

The Flexible Wheels

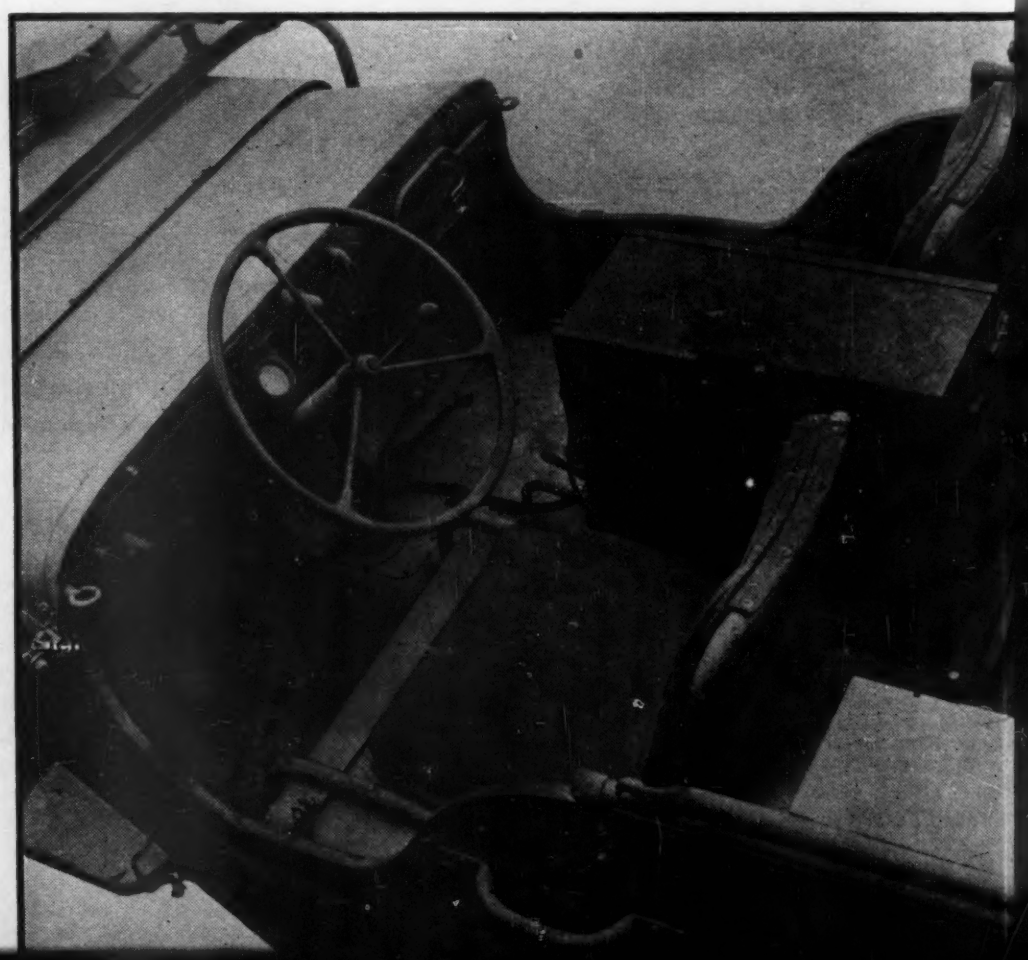
The use of wheels to support the boom, even though they carried only a little weight, created a difficult problem of steering. On sharp turns, and especially in rough terrain, ordinary wheels would catch on the surface irregularities of the ground, and this would result in tremendous stresses on their bearings and on the boom. The seriousness of this difficulty was further increased by the fact that the wheels, being necessarily in the vicinity of the coils, had to be of a nonmetallic nature in order not to give false signals.

The problem was solved by the use of special wheels. In these wheels the tire is replaced by a flexible rubber disc clamped between two wooden flanges of substantially smaller diameter. The rubber disc is rigid enough to support the small pressure on the wheel when the motion is forward, but when a turn is taken a lateral effort is exerted on the disc, which bends and slides over surface irregularities.

Brake and Declutching Mechanism

A fair driving speed is desirable to achieve efficient clearance of mined areas, and also to avoid staying in dangerous battle zones longer than necessary. If the car had to be manually stopped by the driver when a mine signal was received, a human time lag would be involved, and the coils would have to be carried far ahead to avoid accidents, while the operator would be constantly under strain. For these reasons, it was considered necessary to provide the equipment with an automatic stop system. This automatic stop system employs a combination of springs and trigger mechanisms working almost instantaneously on both the foot brake and the clutch,

FIG. 2—Equipment installed in a jeep. To the right of the driver's seat is a case that contains the power supply, oscillator, amplifier and balance controls. Visible between the spokes of the wheel is the control box which includes a meter indicating circuit balance and the presence of a mine, an automatic stop-control switch and warning lights. Two handles used for recocking the automatic stop system may be seen near the bottom of the steering column



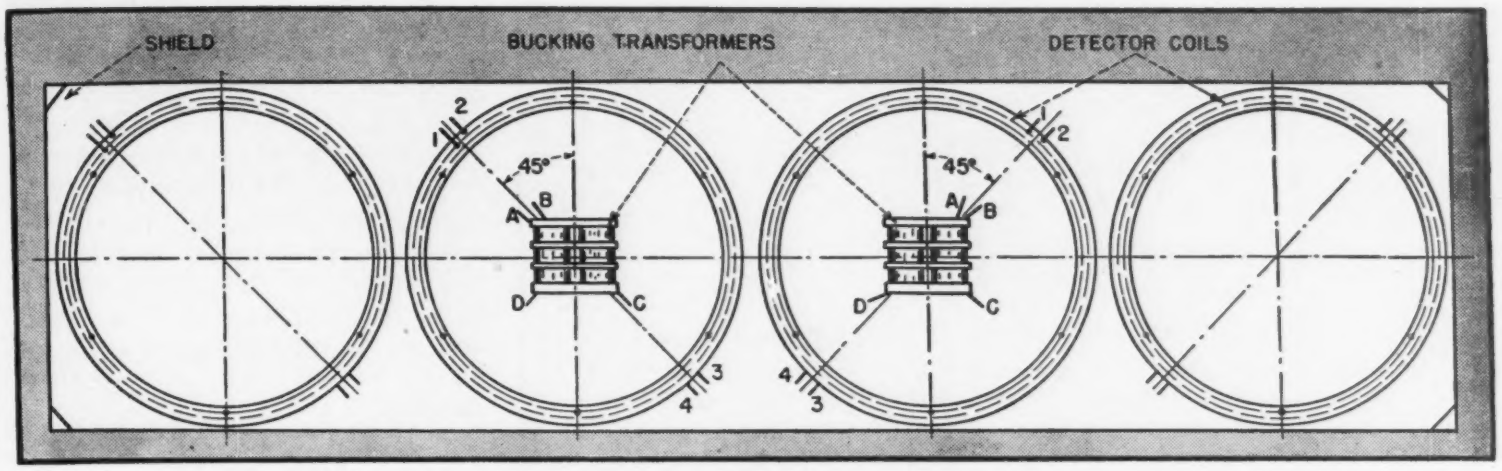


FIG. 3—The arrangement inside the coil box

and is actuated by the mine signal itself.

The automatic brake and clutch mechanism is fast and powerful; it applies 500 lbs on the brake master cylinder in a fraction of a second. It is so arranged that manual operation of the foot pedals by the driver is not impaired in any way. Guard plates are provided to prevent the driver's feet from being trapped by the pedals, as well as handles for recocking both pedals, as shown in Fig. 2.

Coil Box

The coil assembly must be wide enough to cover the whole width of the vehicle, which is about 6 ft for a jeep, and cannot be made rigid without adding unduly to the weight. In rough terrain the coil box is badly shaken, and when bumps are encountered which lift one boom-supporting wheel more than the other or when turns are made, substantial twisting effects take place. Were it not for the special construction used, this would result in substantial variations of the mutual inductance between the transmitter and the receiver coil, and therefore in false signals.

To minimize this difficulty, the coils are mounted in a hermetically-sealed and relatively thin rectangular wooden box which is fastened in the front housing through a three-point suspension. With this arrangement, motions or deformations of the front housing have very little tendency to produce any twisting effect on the coil assembly and practically no false signals result from mechanical shocks. Such false signals as might otherwise occur from that source are further discriminated against by phase selection, as will be explained later.

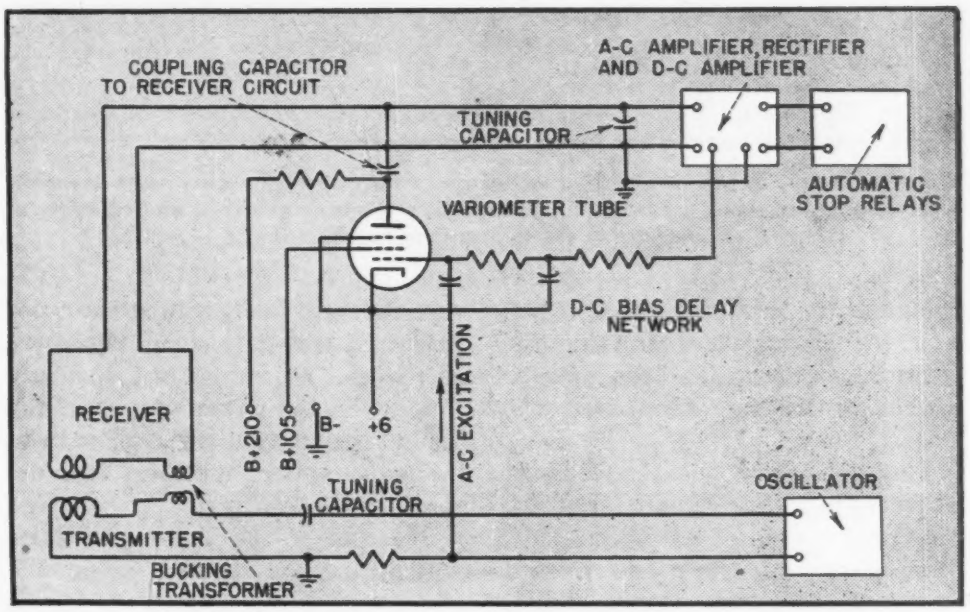


FIG. 4—Simplified schematic of the mine detector circuit, showing in particular the automatic drift stabilizing system of which the variometer tube is an essential part. A complete schematic is given in Fig. 6

Four circular coils are laid side by side as shown in Fig. 3 and connected in a series. These four coils are used as a transmitter. A second series, also of four coils, mounted flat on the first ones, is used as a receiver. The mutual inductance between transmitter and receiver coils is nullified by a group of air-core transformers giving an equivalent mutual inductance of opposite polarity.

Electrostatic shielding, to avoid capacitance effects when the coils approach the ground, is provided by a Faraday shield of thin copper wires imbedded in the bottom part of the coil box.

The front housing serves as a sunshade for the coil system and is provided with ventilating holes. These holes have the additional advantage of letting water fill the front housing when driving in streams or ponds, thus preventing the housing from floating on the surface, and allowing it to follow the ground.

Automatic Drift Control

In conventional detectors the mutual impedance between transmitter and receiver coils has to be manually nullified time and again, generally by means of two variometers corresponding respectively to the inductive and resistive components of the signal. This is primarily because changes in temperature of the coil assembly produce uneven thermal expansion, which results in a progressive change of the mutual impedance, with a corresponding drift of the output signal. In the vehicular detector drift is particularly undesirable. Rebalancing takes

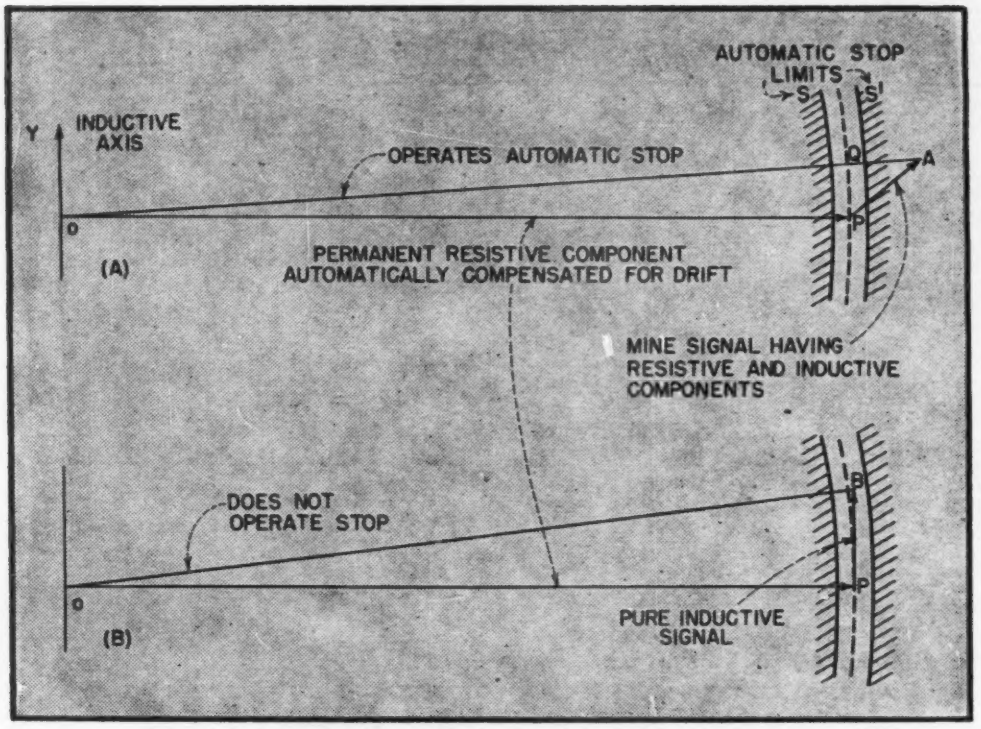


FIG. 5—Principle of phase relation which causes a mine signal to operate the automatic stop system (A), while a moderate inductive signal due to magnetic ground effect or mechanical deformation has no effect (B)

time, and in this case speed is the goal. For that reason, an automatic electronic system has been devised which practically eliminates the drift.

The drift compensator consists essentially of a variometer tube. This tube is a variable—mu pentode, the grid of which is a-c modulated by the transmitter current, while the plate is coupled to the receiver circuit, as shown in Fig. 4. A signal is therefore injected into the receiver circuit, and the magnitude of this signal depends on the d-c bias impressed on the control grid of the variometer tube. The signal at the output of the main receiver-amplifier is rectified and the d-c voltage obtained, after one stage of amplification required to get enough power to operate a relay, is fed back through a resistance-capacitance delaying network to the grid of the variometer tube.

It will be seen that with this arrangement the system is self-stabilizing, and tends to keep the output signal constant, or at least insensitive to slow drift. The action is somewhat similar to that of an automatic volume control, except for the fact that the use of a variometer tube makes it possible to predetermine the phase of the signal injected by proper choice of the excitation phase and coupling to the receiver circuit. The automatic control takes into account not only drift due to thermal expansion of the coil assembly, but also any changes in the power output of the oscillator and of the overall gain in the a-c amplier, rectifier, and d-c amplifier network.

Because of the automatic drift control, manual rebalancing of the system is very seldom required. While conventional detectors frequently have to be readjusted every five minutes, the vehicular detector will usually work without touching the manual controls for at least a whole day, whatever the weather conditions.

The drift control does not adversely affect the equipment's sensitivity to mines. When the detector coils pass over a mine, the signal is suddenly modified and the change is too quick for the automatic control to balance it. The output signal level changes and operates a differential relay which is normally in its neutral position. This trips a latch-type relay, which operates the brake and clutch mechanism.

The above system, based on the use of an automatically controlled permanent signal, has two additional advantages. One of them is to give some phase selection. The second is to provide a warning in the event of breakdown of the circuit. If conventional operation at true balance were used, there would be no output signal under normal conditions, but something could deteriorate in the circuit with the result that the output would remain at zero even though a mine was encountered. In the vehicular device, every breakdown of the circuit automatically produces a substantial change of the output signal, either up or down, so that the brake and clutch mechanism is immediately tripped and the vehicle stopped.

Phase Selection

The a-c excitation of the variometer tube, together with its coupling to the receiver circuit, can be so chosen as to give to the injected signal any desired phase with respect to the transmitter current. In the vehicular-mounted detector the phase of the injected signal has been chosen at 90 deg with respect to inductive signals, such as those produced by a change in the mutual inductance of the coils. The result is that the detector has been made substantially more sensitive to resistive components, which are always an important part of a mine signal, than to purely inductive components which might result from changes of elevation above magnetic ground, or sudden twists of the coil assembly when passing over rough ground.

The theory of the system of phase selection will be better understood by referring to the vectorial diagrams of Fig. 5. In these figures the vectors represent a-c potential differences applied to the input of the amplifier. These potential differences are characterized by both their amplitude and their phase. They are considered as inductive, and the corresponding vectors are plotted parallel to the inductive axis (*OY*), when they have the same phase as that which would be produced by a mutual inductance between transmitter and receiver coils. Conversely, they are considered as resistive, and the corresponding vector is plotted parallel to the *OQ* axis, when they are 90 deg out of phase with respect to the inductive voltage. A mine signal is neither purely inductive nor purely resistive, and is therefore represented by a vector which makes an angle with both axes.

The permanent resistive compo-

nent, which is injected into the receiving circuit by the automatic variometer tube, is much larger than the signals from mines or other sources. Its amplitude is kept substantially constant by means of the d-c feedback bias system already described. (Strictly speaking, it is not the input potential difference that is kept constant by the automatic control but the output current of the whole amplifier; the two things are equivalent only if the gain is constant, which was assumed here for simplicity.) As Fig. 5A shows, a mine signal, such as *PA*, combining with the permanent resistive component *OP*, will give a resulting amplitude *OA* which is outside of the narrow limits within which the automatic stop system is not operated (the automatic stop limits are represented by the circular arcs *S* and *S'*). If a pure inductive signal such as *PB* on Fig. 5B, which might be due to magnetic ground, or mechanical distortion of the coils, combines with the permanent resistive component *OP*, the resulting amplitude *OB* is little different from *OP*, and stays within the limits where the automatic stop is not operated.

Field Operations

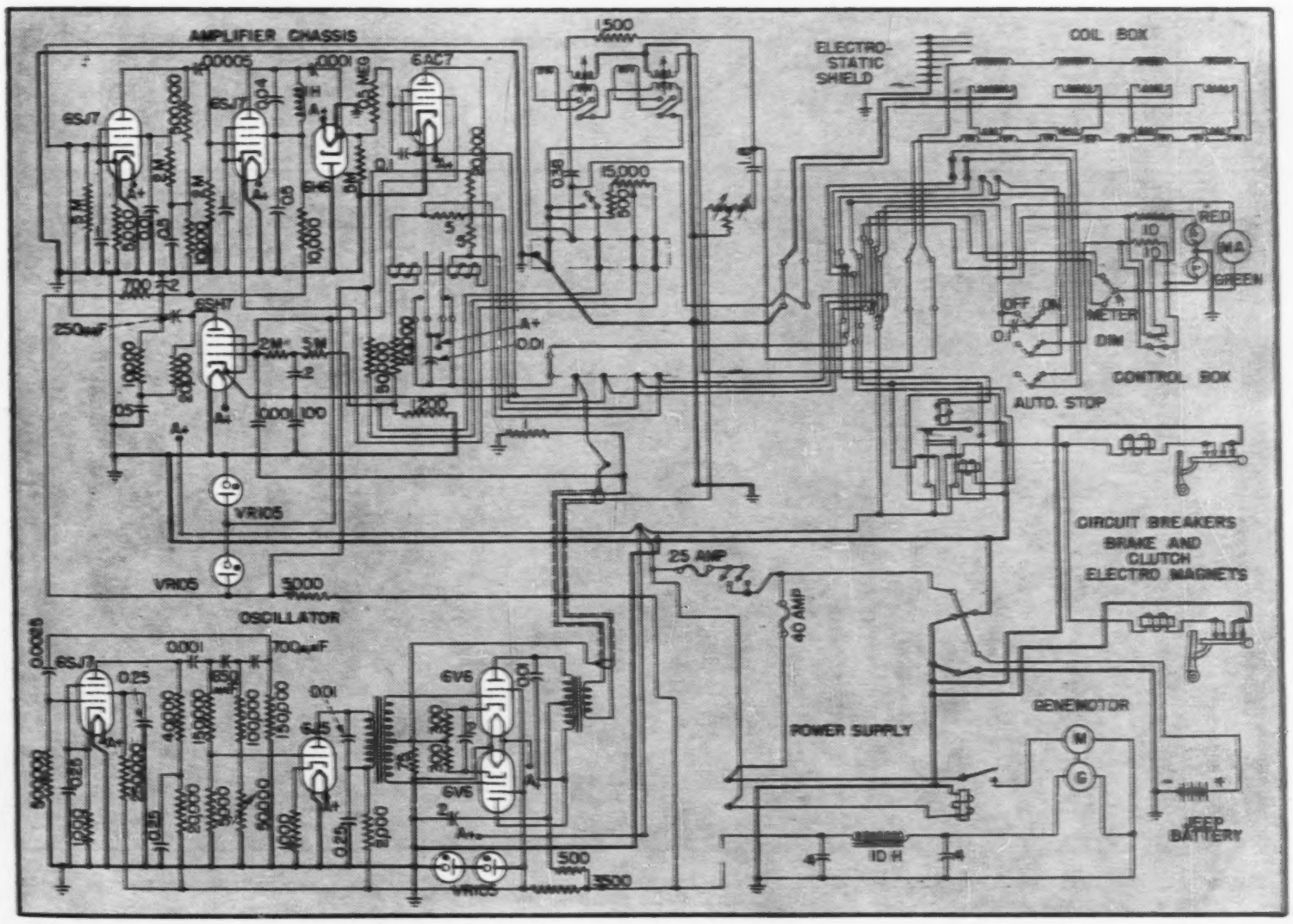
Before starting actual detecting operation, the circuit is balanced by means of manual controls which consist essentially of two variometers giving respectively inductive and resistive components of compensating voltage. While this adjustment is being made, the automatic control is put out of action by means of a switch. A waterproof cover is then set over the control panel of the oscillator-amplifier box, and from then on the operator uses a very simple control box mounted on the steering column. This control box contains a meter which enables the operator to know at all times if the circuit is still in good balance, and also indicates the presence of mines. In addition, it contains a switch to disable the automatic stopping equipment when automatic stopping is not desired. A pilot light shows when the equipment is turned on, and another warning lamp tells whether all the switches and equipment are in operating condition for the detection of mines.

A complete circuit diagram of the vehicular-mounted mine detector is shown in Fig. 6.

Acknowledgment

The responsibility for the development of mine detection equipment is assigned to the Engineer Board, the development agency for the Corps of Engineers, U. S. Army, and the device described above was developed for that agency. Brigadier General John W. N. Schulz is president of the Engineer Board, with Capt. G. S. Slaughter of the Applied Electronics Branch now assigned the responsibility of directing the work as project officer.

FIG. 6—Complete schematic of the vehicular mine detector



The MPG-1 Radar

The transmitting, r-f, receiver and antenna-positioning systems of the 3-cm coastal defense radar are described in this second article of a series. Details of the scanner which swings the radiated beam over a 10-degree arc at a rate of 160 degrees per second are included

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IN THE FIRST installment of this series the history, characteristics and a general description of the AN/MPG-1 radar were presented. In this installment the transmitting system, r-f system, receiving system and antenna-positioning system are described.

Transmitting System

Transmitted radar pulses must be narrow in width to give good range resolution, and a steep leading edge is required to give good range accuracy. The peak power must be high in order to produce strong echoes from distant targets, since the intensity of received signals is inversely proportional to the fourth power of the target range. To obtain good definition a high repetition rate must be used. The AN/MPG-1 transmitting system is designed to produce r-f pulses at fixed intervals. The modulator unit generates a high-voltage keying pulse which fires the magnetron transmitting tube. During the keying pulse, the magnetron oscillates, generating r-f pulses at approximately 10,000-mc frequency and at least 35 kw peak power (35 watts average).

When using the ppi-scope the keying rate is 1024 cps and the transmitted pulse width is 1.0 microsecond. On the B-scope, the keying rate is 4094 cps and the

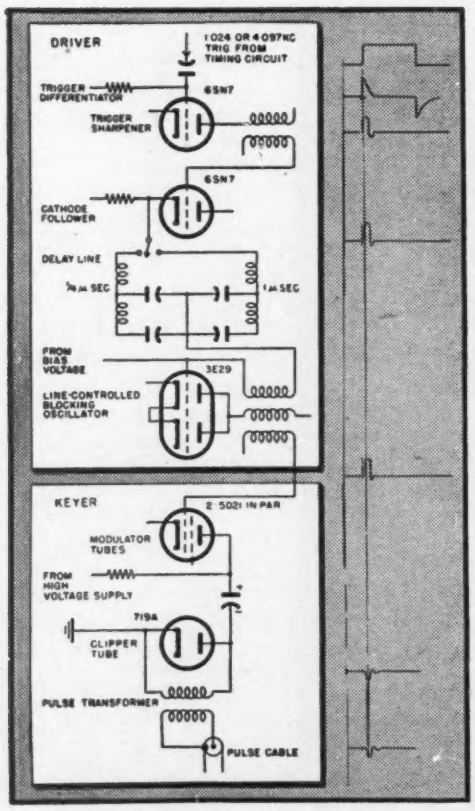

FIG. 1—Simplified schematic of the driver and keyer circuits

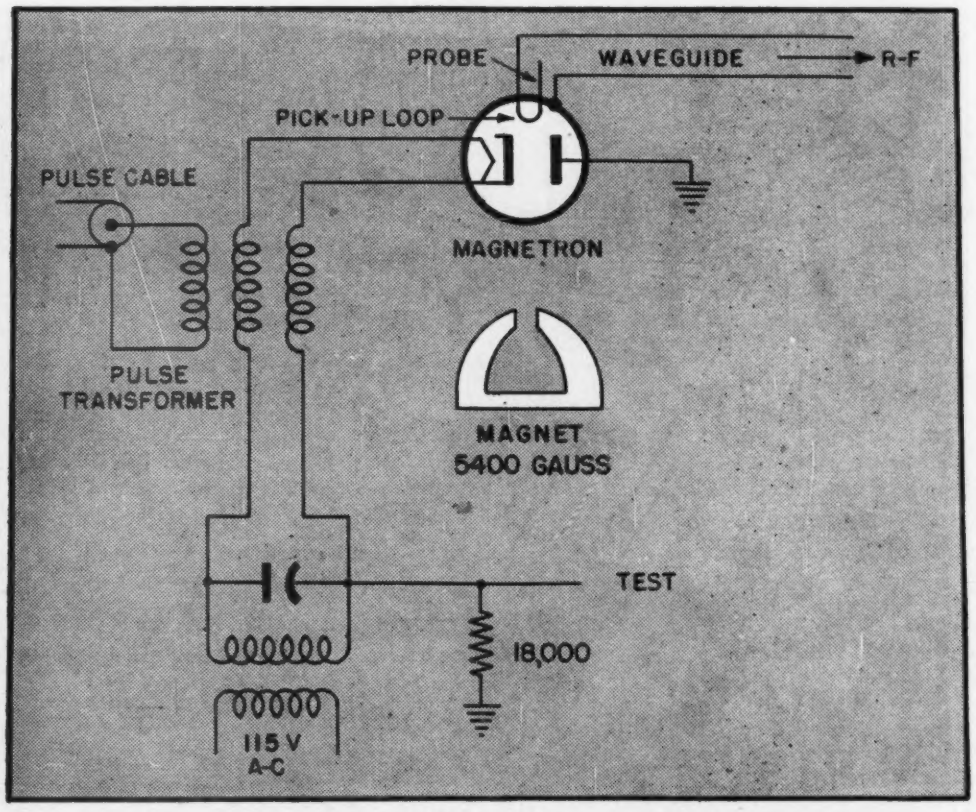

FIG. 2—Pulse transformer input to the magnetron. The magnetron operates at an efficiency of about 36 percent, producing 40-kw pulses in the output circuit

AN/MPG-1 trailer, tower and radiator. The "bath-tub" radiator is elevated within the tower to raise it to operating position

pulse width is 0.25 microsecond, to achieve better range resolution and definition on the B-scope than the ppi.

The modulator, shown schematically in Fig. 1, consists of two main circuits: driver and keyer. The driver circuit produces a synchronized pulse of the desired width and the keyer circuit generates a high-voltage, negative keying pulse which fires the magnetron. Basic timing is established by a 2-microsecond pulse from the timing system. In order to obtain stability and good range accuracy, the line-controlled blocking oscillator must be triggered by a narrow pulse which is accurately fixed with respect to the range circuit and has a very steep leading edge of high amplitude. Therefore the input trigger is differentiated, giving a positive and negative peak, and the positive peak is differentiated again and amplified before being used to trigger the blocking oscillator. The cathode follower matches the high impedance of the trigger sharpening transformer to the lower impedance of the delay line.

The positive trigger is applied to the blocking oscillator control grid through the pulse-forming delay line, thus raising the grid bias above cutoff and initiating plate current flow through the primary of the plate transformer. The voltages induced in the grid winding and output winding of the transformer are proportional to the rate of change of magnetizing current in the plate winding. At the first increase in plate current, an induced voltage of positive polarity appears at the grid, making the grid more positive, which increases the plate current still more. The action is cumulative, resulting in a high, rapidly-induced positive grid voltage. Because of the inductance of the plate winding, the plate magnetizing current increases exponentially. Meanwhile, since a positive voltage appears at the grid end of the secondary winding, a negative voltage appears at the delay line end, and the capacitors of the delay line begin to assume a negative charge at a rate determined by the L-C time-constant associated with the line. The time constant of the delay line and plate transformer are such that by the time the plate current has increased to some extent along the linear portion of the exponential the delay line has become fully charged and the grid voltage drops rather suddenly. This causes a sudden decrease in plate magnetizing current, which results in an induced negative voltage which lowers the plate current still more, inducing more negative grid voltage, etc. At the end of this degenerative process, which is very rapid, the grid is driven far below the fixed negative bias; but before the next trigger arrives it has returned to the fixed bias. The output of the blocking oscillator is a +1000-volt pulse having a steep leading edge, a relatively flat top and a duration of 0.25 or 1 microsecond, depending upon the pulse line used.

One plate of the pulse capacitor is connected to the plates of the modulator tubes and the other plate is grounded through the primary of the output pulse transformer. In the interval between synchronizing triggers, the pulse capacitor is a charge-storing reservoir, receiving a positive charge from the high-voltage power supply through the modulator plate resistor. When the synchronizing trigger arrives, the modulator grids are driven approximately 240 volts positive for the duration of the driver pulse, allowing the modulators to conduct, and thus providing a low-impedance-to-ground discharge path for the pulse capacitor. This is equivalent to shorting the capacitor to ground through the modulator plate resistance, causing the capacitor voltage to drop by 12.5 kv minus the 1.5 kv voltage drop across the modulator

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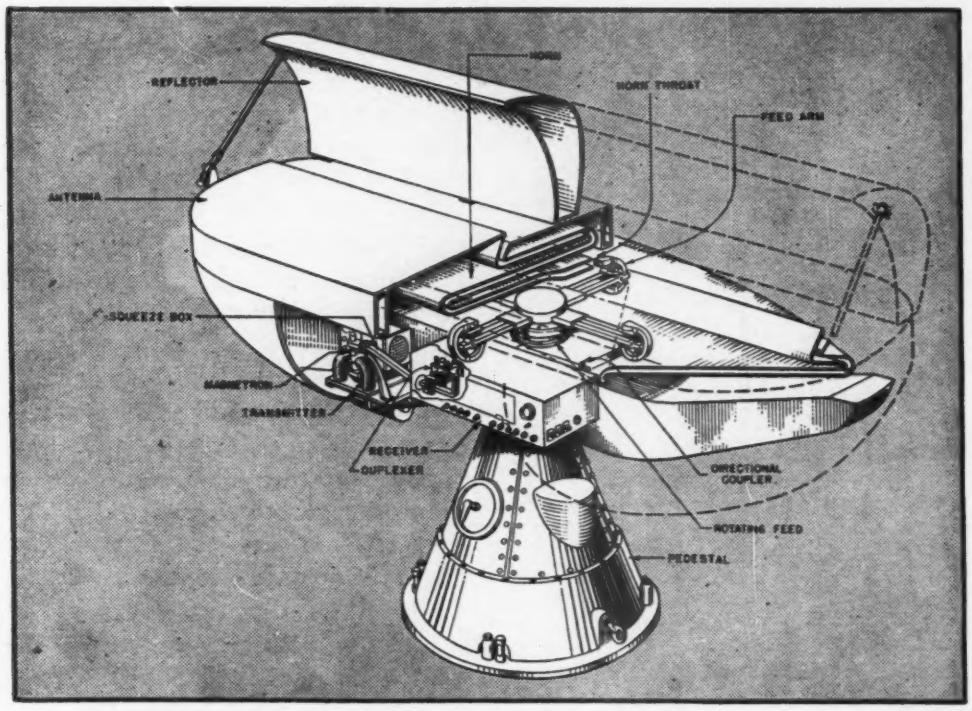


FIG. 3—Interior structure of the high-speed scanner and radiator. Waveguide feeds rotate past the input opening of a folded horn and reflector which produce a beam 0.6 degree wide

tube, ie, a net drop of 11 kv. The load side of the capacitor drops to −11 kv, resulting in current flow to ground through the pulse transformer winding. Since the capacitor-resistor time constant is comparatively large, discharge takes place only on the linear portion of the exponential characteristic before the modulator tubes are cut off by the trailing edge of the driver pulse. Thus, the current surge through the primary winding of the transformer is such as to produce an output voltage of rectangular waveform. A 4 to 1 stepdown pulse transformer matches the keyer circuit impedance to the 50-ohm coaxial cable which conducts the pulse to the magnetron. At the magnetron, shown in Fig. 2, a 4:1 step-up transformer is used for impedance matching and voltage-change. This arrangement makes it possible to conduct the pulse at relatively low voltage, thus minimizing the danger of high-voltage flashover and breakdown. Another function of the step-up transformer is to make possible operation of the magnetron filament transformer at ground potential, despite the fact that the magnetron filament is made highly negative during the pulse.

One of two identical secondary windings of the pulse transformer is placed in each line from the filament transformer to the magnetron filament. A potential of −11 kv is induced across each winding and applied to the magnetron cathode. Since the filament transformer is connected across the low-voltage ends of the two windings, special insulation is unnecessary and the capacitance effect is negligible.

The transmitting tube is a magnetron which oscillates at approximately 10,000 mc when the keying pulse is applied. The magnetron current for the duration of the pulse is approximately 10 amperes, so the peak power input is approximately 110 kw. With magnetron efficiency at about 36 percent, the peak power output is approximately 40 kw.

Radio-frequency energy is inductively coupled to a pick-up loop and transferred to a short section of coaxial line terminated by a probe which radiates the energy down the waveguide toward the antenna. This coaxial line and waveguide are integral parts of the magnetron. The rectangular waveguide propagates the energy in the $TE_{0,1}$ mode.

The R-f System

The r-f system is perhaps the most striking and original feature of the AN/MPG-1 radar. It consists of the "squeeze box," duplexer, directional coupler, rotating feed, horn and reflector. The r-f system is housed completely in a bathtub-like structure commonly referred to as the antenna. Pivoted on a pedestal which contains the azimuth drive motor and gear train, the antenna is capable of 360-deg horizontal rotation, with control centered at the console in the trailer. Electrical connections between system components mounted in the antenna and those located in the trailer are made through a system of slip rings in the pedestal. Figure 3 shows the antenna semi-schematically and some of the units which it houses.

The chief features are (1) a suitable transmission system for conducting the magnetron output to the antenna feed, (2) an antenna-feed mechanism which provides a 10-deg scan 16 times per second on B-presentation so that fast, maneuvering targets can be tracked smoothly and accurately and, (3) a radiator which produces a fan-shaped beam very narrow in the horizontal dimension to assure good azimuth accuracy. A block diagram of the r-f system is given in Fig. 4.

Radio-frequency energy generated by the magnetron is conducted towards the antenna feed via rectangular waveguide. The squeeze box, or line stretcher, is a section of rectangular waveguide with a slot cut in the longitudinal dimension in the center of each wide face of the guide. By adjusting the width of the squeeze box it is possible to change the guide wavelength in the box and adjust the standing wave at the magnetron to the point where the magnetron has the greatest frequency stability.

The transmission system includes the duplexer (t-r) assembly which is located in the receiver. By means of the t-r and anti-t-r switch tubes, it is possible to use the same antenna for transmission and reception. The t-r tube is a high-speed gaseous switch which permits received signals to enter the receiving system, but prevents the high-power transmitted pulses from doing likewise, since this would result in destruction of the receiver crystal detector. The anti-t-r tube prevents received energy from entering, and being dissipated in, the magnetron cavity.

The directional coupler allows a small part of the transmitted energy to enter a test set, but blocks

the test set from receiving energy traveling in the opposite direction, ie, from the antenna towards the receiver.

The radiator is a folded, sectoral horn. Four waveguide feed arms (see Fig. 3), spaced 90-deg apart, rotate in front of the horn throat, transferring r-f energy to the horn from one feed arm at a time. Energy is conducted to the feed arms via a cylindrical rotating joint and r-f switch, shown in Fig. 5. Before reaching the cylindrical waveguide the energy is propagated in a rectangular waveguide in the $TE_{0,1}$ mode. Symmetry considerations indicate that the lowest acceptable mode which may be used in a cylindrical waveguide is the $TM_{0,1}$ mode. The necessary mode transformation is accomplished by means of a matching diaphragm and resonant ring placed near the junction of the rectangular and cylindrical waveguides. Impedances introduced by the mode transformer result in a broad band unit. At the junction of the stationary and rotating sections of the rotating joint a choke joint is fitted to permit free rotation of the upper member without having excessive loss by radiation. Feed-arm tuning is assisted by the resonant ring in the rotating member of the joint. Note that it is necessary to effect a transformation from the $TM_{0,1}$ mode to the $TE_{0,1}$ mode at the junction between the rotating joint and feed arms due to the fact that the feed arms consist of rectangular waveguide.

The rotating feed is tuned to the magnetron by means of a plunger in a closed cavity located above the feed arms. The inside diameter of the cavity is so chosen that the guide wavelengths of the $TE_{0,1}$ and $TM_{0,1}$ modes differ sufficiently to result in suppression of the former mode and enhancement of the latter by resonance.

The radiator, shown in Fig. 6, is a parallel-plate folded horn with a parabolic reflector which is a parabolic cylinder. The separation of the plates is held within such tolerances that the only modes which are propagated are those with the electric field perpendicular to the surface of the plates and having a wavelength equal to the space wavelength. Matching to the reflector is obtained by flaring the horn mouth. To keep out moisture, a plastic member covers the horn mouth. Since the flare cover is built to a thickness of $\lambda/2$, reflections from the two surfaces of the cover cancel out and do not affect the match of horn to reflector. Any reflections resulting from local irregularities in the horn are scattered at random and do not reach the feed in appreciable quantity. The reflector, which reduces the vertical width of the beam, is illuminated by an equivalent line source of plane waves. The horn folds are so designed that when a feed arm is centered at the horn throat, the horn emits a plane wave propagated in a direction perpendicular to the length of the reflector. The principal sections of the horn folds are equivalent to the diametrical sections of the mirrors in an aplanatic optical system designed by J. G. Baker of Harvard.

The half-power beam-width on transmission is about 0.6 deg in the horizontal dimension and 3 deg vertically. The antenna system is relatively free from side lobes. Since the antenna is not frequency-sensitive, the system may be used with magnetrons covering a wide frequency range.

Scanning is accomplished by rotating the feed arms in front of the horn throat. Each of the four feed arms is a rectangular waveguide with a flared end which provides a good match between the guide and the horn. For each degree of displacement of a feed arm across the horn throat, the emitted beam is displaced $\frac{1}{8}$ of a degree in space. Thus a 10-degree scan is obtained by utilizing 80 degrees of feed arm rotation. This allows an adequate switching interval (10 deg) in which the next feed arm is brought into play. While one feed arm feeds the antenna through the opening in the fixed circular shorting ring, the other three arms are shorted by the ring. Each arm in turn feeds the antenna as the arm passes the opening in the ring.

When the B-scope is used for tracking targets the feed unit rotates steadily at approximately 4 rps. Since there are four feed arms, this results in 16 sweeps per second across a 10-deg sector in space. Smooth, accurate tracking of rapidly moving targets is greatly facilitated by a scan of this type.

When the ppi is used for surveillance one of the feed arms remains fixed at the center of the horn throat and the beam is always pointed along the antenna axis. In any case, whether on B-scope or ppi, the antenna axis may be rotated to any desired direction.

Receiving System

The receiving system shown in Fig. 7 converts high-frequency echo signals into video pulses for presentation on the ppi, B, and remote-B scopes. The receiver itself is a sensitive superheterodyne with automatic frequency control, duplexer, an A-scope and power sup-

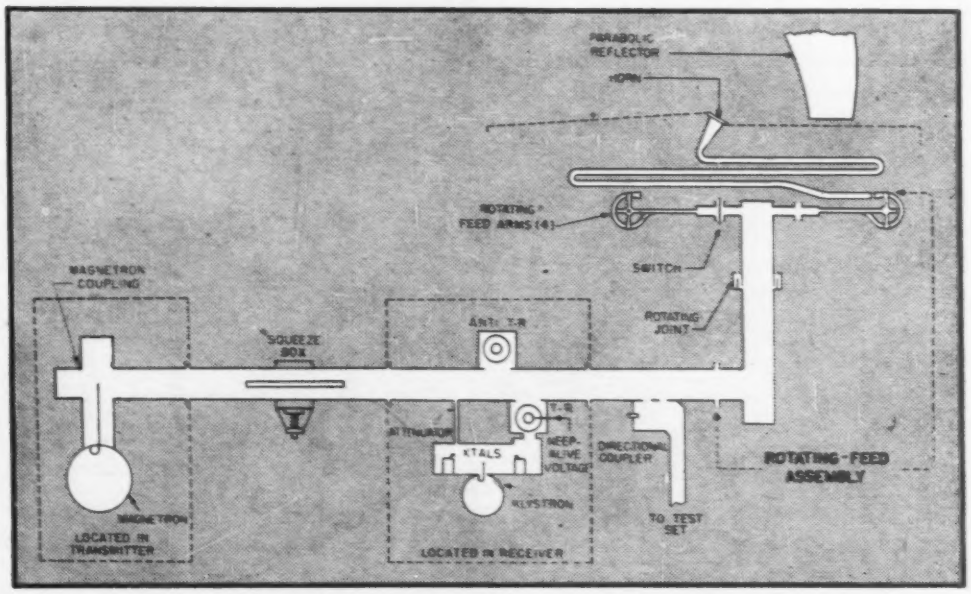


FIG. 4—The r-f system, showing the duplex switch (t-r and anti-t-r switch tubes), squeeze box, and rotating joints between transmitter and receiver and the radiator

ply. A sensitivity time control (stc) circuit is in the radar console in the trailer, but this circuit is properly considered part of the receiving system.

The mixer circuit consists of a klystron local oscillator and two crystals, one for a-f-c and the other for signal, as shown in Fig. 8. The afc crystal receives a small part of the transmitted energy through a 70-db attenuator, whereas the signal crystal receives echo signals from the r-f system. Since the transmitted energy and received signals are both of the same frequency, the same i-f signal (30 mc)

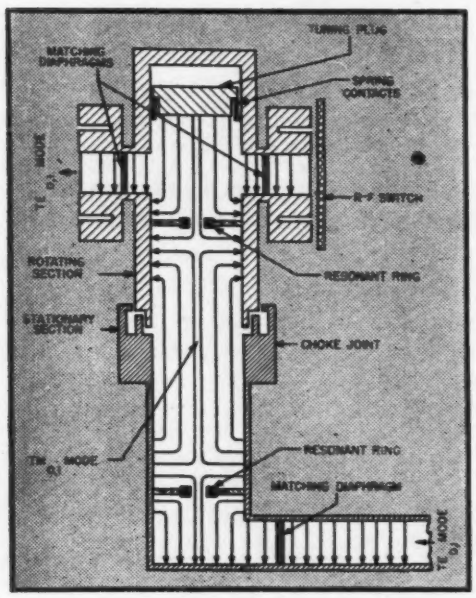


FIG. 5—Disposition of electric fields in the rotating joint and r-f switch, in which the wave is transferred from a rectangular to a cylindrical waveguide

is produced in each case as a result of mixing with the local oscillator output. It would have been possible to use a single crystal for generating 30-mc signals for the afc and i-f circuits, but the possibility of electronic jamming made it necessary to employ two crystals. If the signal-crystal output controlled the afc circuit, the enemy, using r-f energy of the proper frequency, could remotely tune the klystron frequency so far from the magnetron frequency that the resultant i-f signals would fall outside the frequency band of the i-f circuits, resulting in loss of normal signals.

The weak 30-mc signals from the signal crystal are amplified by the i-f circuits. In a radar receiver it is essential to preserve the steep leading edge of the echo pulse to permit accurate measurement of target range. If the receiver bandwidth is too narrow the echo pulse becomes distorted. If the bandwidth is too wide, the pulse waveform remains good, but there is an increase in noise originating in the crystal and first i-f stage. A bandwidth of 106 mc with ample overall gain is obtained by staggered tuning of the i-f amplifiers and use of high-gm tubes (6AC7).

There are nine i-f amplifiers divided into 3 groups of staggered triples. In each staggered triple one stage is tuned to 30 mc, another to 24.8 mc and the third to 36.3 mc. The i-f coils are pretuned. The amplification of each staggered triple is made uniform over the working range by adjustment of the bandwidth and gain of each stage in the triple. Figure 9 shows the response of the i-f amplifiers. In order to prevent undesirable feedback between stages, the first eight stages of amplification are decoupled from one another in pairs. Bias is applied to the control grids of the second and third stages. The i-f gain is more than 110 db. The last i-f stage is coupled to a diode detector having a very low value load resistor (1200 ohms). The negative video output from the detector is inverted and amplified by the following tube. The positive video output from the amplifier tube is used to drive the grid of a cathode follower. The cathode follower feeds a 70-ohm coaxial line which conducts the video signal to the video amplifier circuits in the range unit mounted in the operating console. Use of this low impedance coaxial line terminated at the console permits the video signal to be transmitted through slip rings in the pedestal and over a distance in excess of 100 feet with negligible distortion of the video pulse waveform.

An afc circuit keeps the klystron tuned to 30 mc above the transmitter frequency as shown in Fig. 10. When the afc circuit is used, the klystron frequency is regulated by automatic control of the repeller plate voltage which is governed by the voltage across capacitor C_2. By means of the action of C_1, R and the 2050 control tube, the voltage across C_2 is maintained at a little less than 100 volts.

The discriminator receives an amplified i-f signal and provides a negative output pulse if the transmitter has drifted below the proper

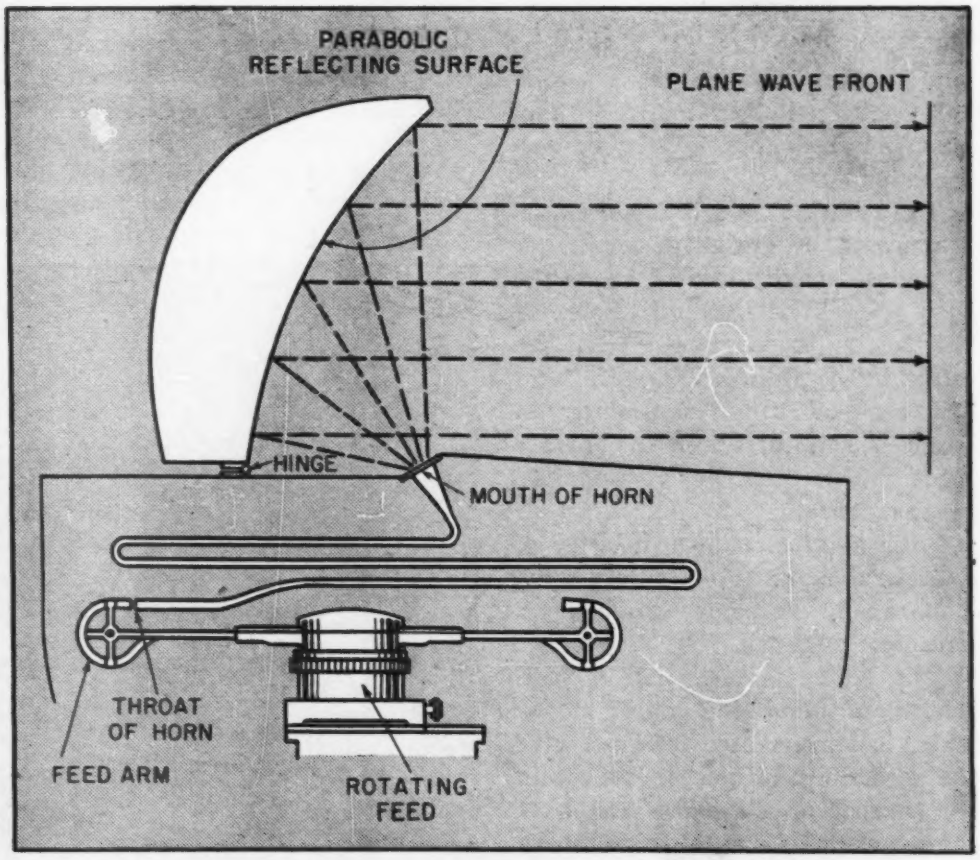


FIG. 6—Detail of waveguide feed, folded sectoral horn and reflector. As the feed rotates, the axis of the beam swings through a sector of 10 degrees

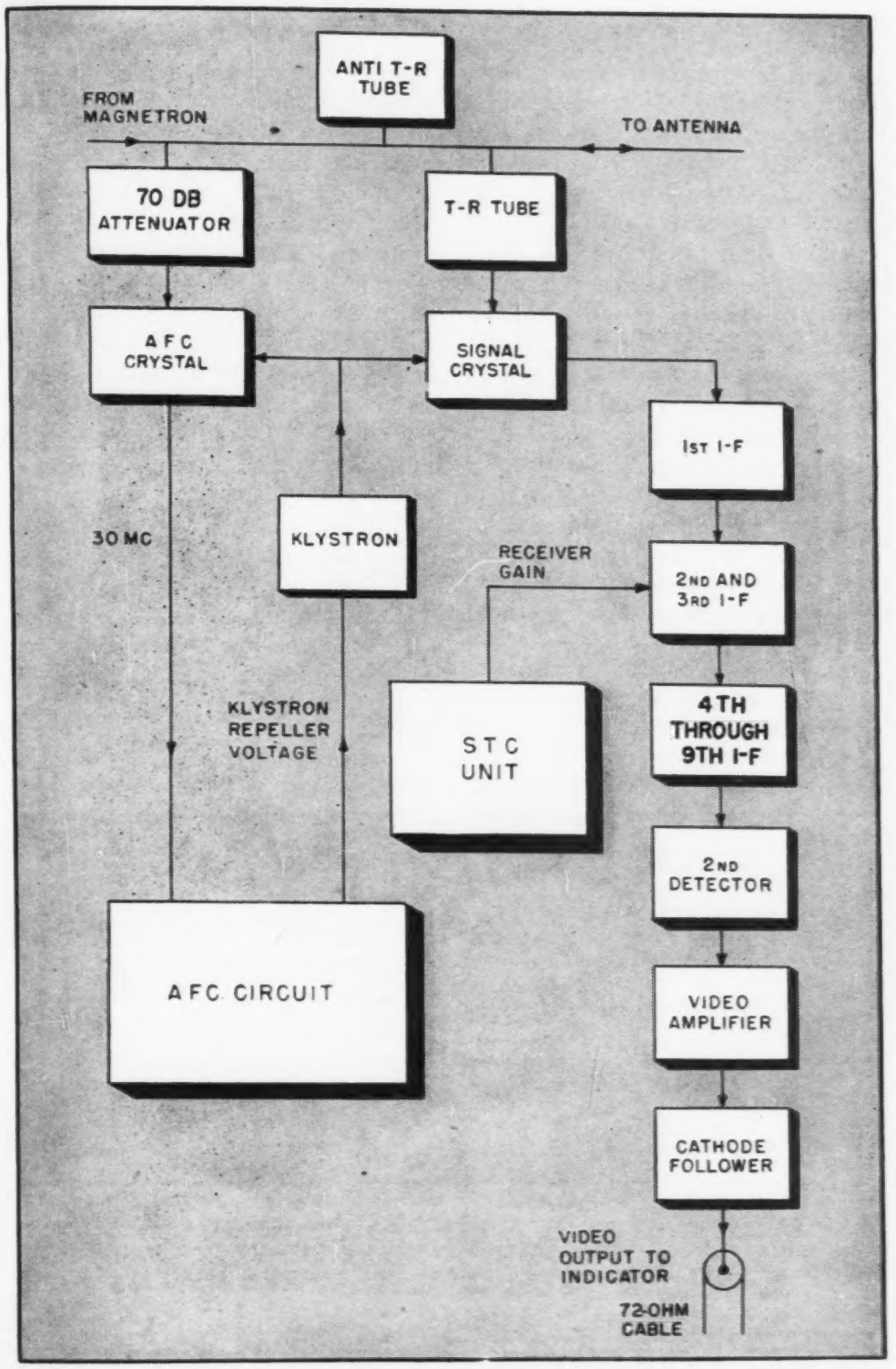


FIG. 7—Block diagram of the AN/MPG-1 radar receiver system. Two silicon crystals are employed, one for afc control, the other for the signal circuit

frequency, a positive pulse if the transmitter frequency is too high, and no pulse at all if the transmitter frequency is of the proper value to yield a 30-mc intermediate frequency. The discriminator output, if any, is applied to the pulse amplifier.

The sawtooth oscillator generates a saw tooth voltage waveform which causes the klystron frequency to vary over a considerable range. This pulse operates only when the transmitter frequency is very far from being 30 mc above the klystron. When the transmitter frequency is close to 30 mc above the klystron the control tube takes control of klystron frequency and the sawtooth oscillator is inoperative. The sawtooth oscillator output is a positive-going sawtooth of approximately 100 volts amplitude and a repetition rate of $\frac{1}{2}$ cps.

Once the sawtooth oscillator has set the klystron frequency to a value such that the intermediate frequency is slightly below 30 mc, the discriminator output will have become slightly negative. This triggers the control tube, resulting in a plate-voltage drop to approximately ground potential. The drop in plate voltage is applied to C_2, resulting in increased klystron frequency and an intermediate frequency which is slightly higher than 30 mc. The control tube triggers now become negative and the tube remains inoperative while C_2 charges, causing the i-f to drift toward 30 mc again. When the i-f passes 30 mc the cycle is repeated. The sawtooth oscillator remains inactive as long as its plate voltage is maintained at less than 100 volts.

If the receiver gain were constant, signals from targets of a given size would appear much brighter on the scopes at short range than at long range. "Blooming" of the scopes due to multiple close-in targets or sea-return might interfere with the operators' ability to distinguish targets at close range. The sensitivity control circuits make it possible to adjust the receiver gain automatically so that signals from the same target appear equally intense on the scope regardless of range. This is done by superimposing on the bias level set by the manual gain control a trapezoidal voltage which is synchronized with the transmitted pulse by the method shown in Fig. 11. The receiver gain is made low for short ranges and higher for longer ranges. The receiver and r-f system can be tuned by reference to the built-in A-scope.

Antenna Positioning System

The antenna pointing direction is controlled by a servo. Selsyn generators geared to the tracking

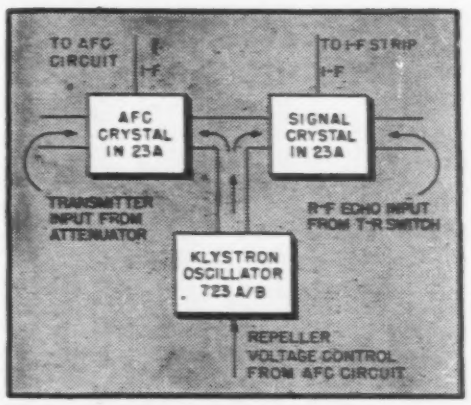


FIG. 8—The afc crystal employs a minute portion of the transmitter signal to operate the frequency control. A separate crystal develops the echo signal intermediate frequency

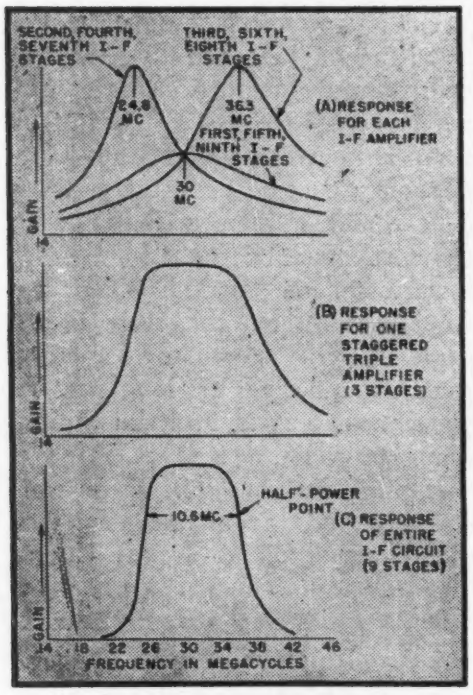

SECOND, FOURTH, SEVENTH I-F STAGES
THIRD, SIXTH, EIGHTH I-F STAGES
24.8 MC
36.3 MC
30 MC
FIRST, FIFTH, NINTH I-F STAGES
(A) RESPONSE FOR EACH I-F AMPLIFIER
GAIN
(B) RESPONSE FOR ONE STAGGERED TRIPLE AMPLIFIER (3 STAGES)
HALF-POWER POINT
10.6 MC
(C) RESPONSE OF ENTIRE I-F CIRCUIT (9 STAGES)
14 18 22 26 30 34 38 42 46
FREQUENCY IN MEGACYCLES

FIG. 9—Band-pass characteristics of the nine i-f stages: (A) Staggered tuning of the i-f stage groups produces a flat top response (B) for three stages. Nine stages produce the overall result shown in (C). This is one of the widest bands ever employed in radar equipment

output transmit electrical position data to selsyn control transformers geared to the antenna drive. The difference between the desired direction and the actual pointing direction is transmitted as an error voltage to an amplifying system, the output of which actuates the antenna drive motor to correct the pointing error as shown in Fig. 12.

The pointing error must be less than 0.05 deg. Due to inherent electrical error of selsyns, a gearing ratio of 1:1 does not provide sufficiently accurate pointing. Therefore a 36:1 system, which reduces the selsyn electrical error by a factor of 36, is normally used during the tracking operation. For pointing errors in excess of 4 deg, however, such as might occur when the operators switch from ppi to tracking or when the antenna is slewed rapidly back and forth, a 1:1 selsyn system takes control of the antenna. When the positioning error falls below 4 deg, the 36 to 1 selsyn resumes control and accurate positioning is reestablished. This is accomplished by applying the 1:1 rectified error voltage to the biased control grid of a 6AG7 which has a relay coil in its plate circuit. When the rectified 1:1 error voltage reaches an amplitude corresponding to 4 deg of pointing error, the tube draws enough plate current to operate the relay, which removes the 36:1 error input from the servo amplifier and applies the 1:1 error input instead. When the pointing error falls below 4 deg the relay is de-energized and the 36:1 system takes control. Use of the 36:1 selsyn system reduces the peak inherent electrical and mechanical error to less than 0.5 deg. On ppi, however, where high accuracy is not required, the 1:1 selsyn system is used exclusively.

The error voltage from the 36:1 or 1:1 selsyn system is introduced into a phase-sensitive rectifier circuit located in the antenna servo amplifier and shown in Fig. 13. The reference and error voltages are mixed by applying each voltage to separate primaries of transformers T_1 and T_2. In T_1 the reference and error voltages add, while in T_2 they subtract. Each transformer provides a 12:1 step-up ratio. Losses are kept very low as a result of the high impedance windings and silicon-steel cores. When the error input is zero, R_5 sets the d-c level of points A and B, and the ripple voltages due to the reference voltage cancel out due to the manner in which rectifier V_1 is connected. When an error voltage exists however, the outputs of the halves of V_1 are unequal and total cancellation is impossible. This causes an unbalance between points A and B

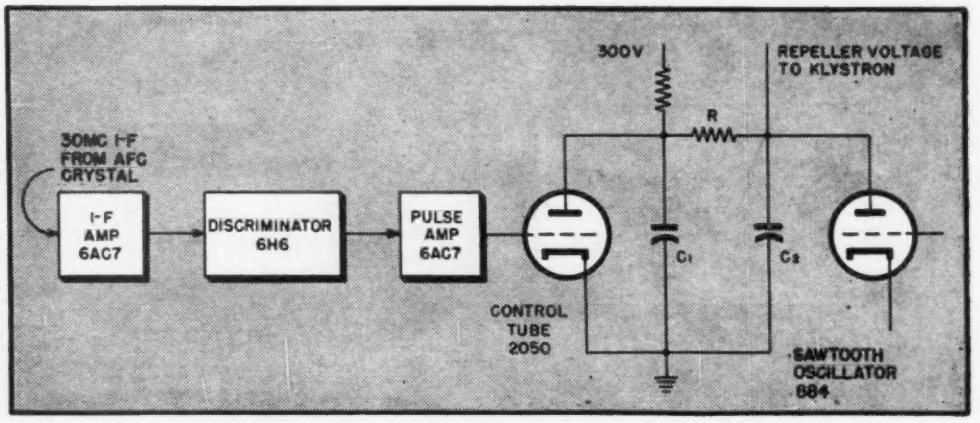


FIG. 10—Automatic frequency control system. When the transmitter frequency drifts excessively a sawtooth wave is applied to the tuning electrode (repeller) of the klystron local oscillator, causing it to hunt for the correct frequency

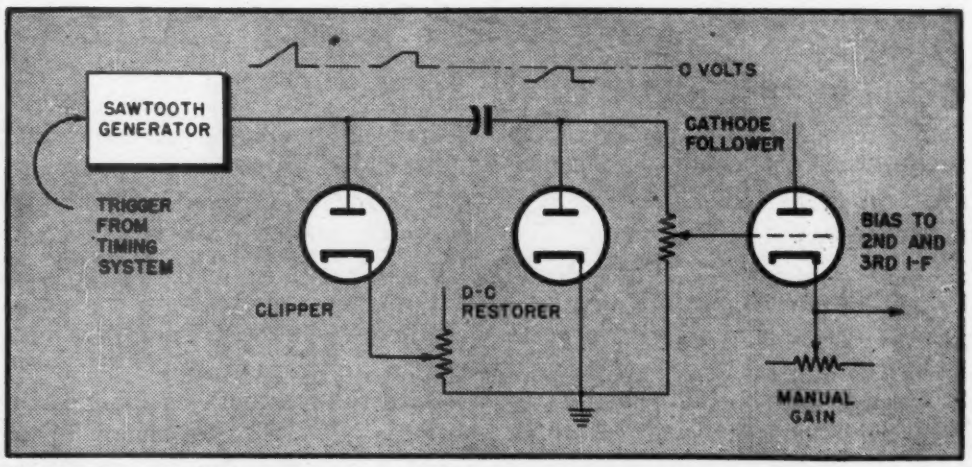


FIG. 11—Automatic brightness control scheme. The amplitude of echoes from nearby targets is automatically cut down to prevent "blooming" of the c-r tube screen

which appears on the grids of V_2. Tubes V_3 and V_4 therefore draw unequal plate current, resulting in an amplidyne output which has a maximum of approximately 500 watts. The amplidyne output is applied to a ½-hp motor which drives the antenna. The velocity error constant is 1613 deg/sec/degree error and the torque error constant is about 122,000 ft lb/degree error. At constant antenna velocities up to 4.5 deg/sec the pointing error is less than 0.015 deg. For a step increase in angular velocity of 3 deg/sec the pointing error is less than 0.03 deg in 2 seconds. Antenna inertia is 7 ton-ft.

Excessive time lag in the filters R_1C_1 and R_2C_2 is prevented by tolerating 10 percent ripple. Integral response to very small error voltages is provided by the R-C network R_5, R_6, C_5. R_3, R_4, C_3, C_4, is a flutter filter.

On ppi the antenna may be made to scan automatically back and forth over any assigned surveillance sector or, by means of a lever switch, the antenna may be slewed either to left or right at a velocity of 20 deg sec. The ppi deflection yoke is part of a separate servo system which makes the yoke follow the antenna pointing direction. Operating details were published in a previous issue of ELECTRONICS.[1]

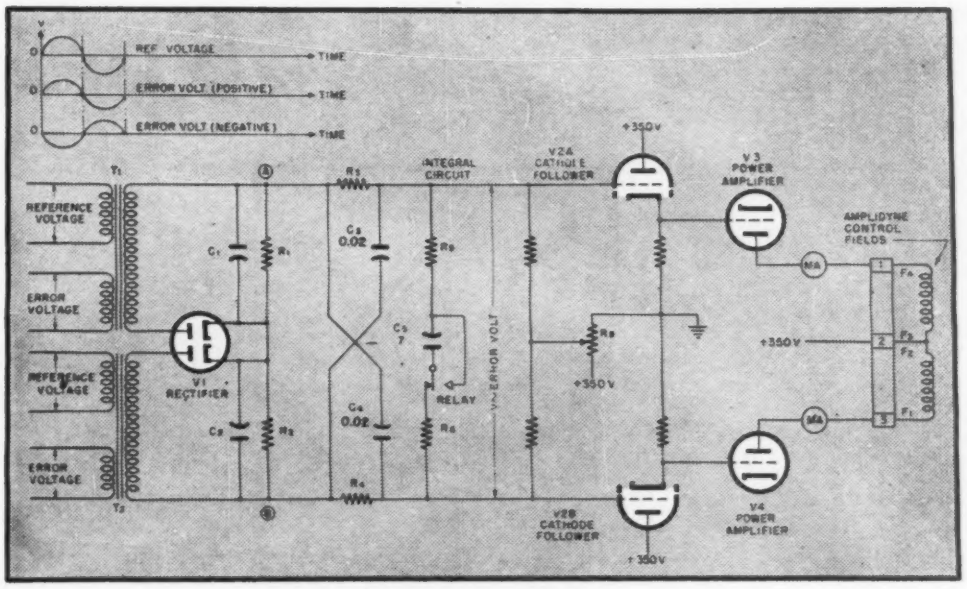


FIG. 13—Error voltage comparison circuit and amplidyne drive which controls the antenna-rotating motor

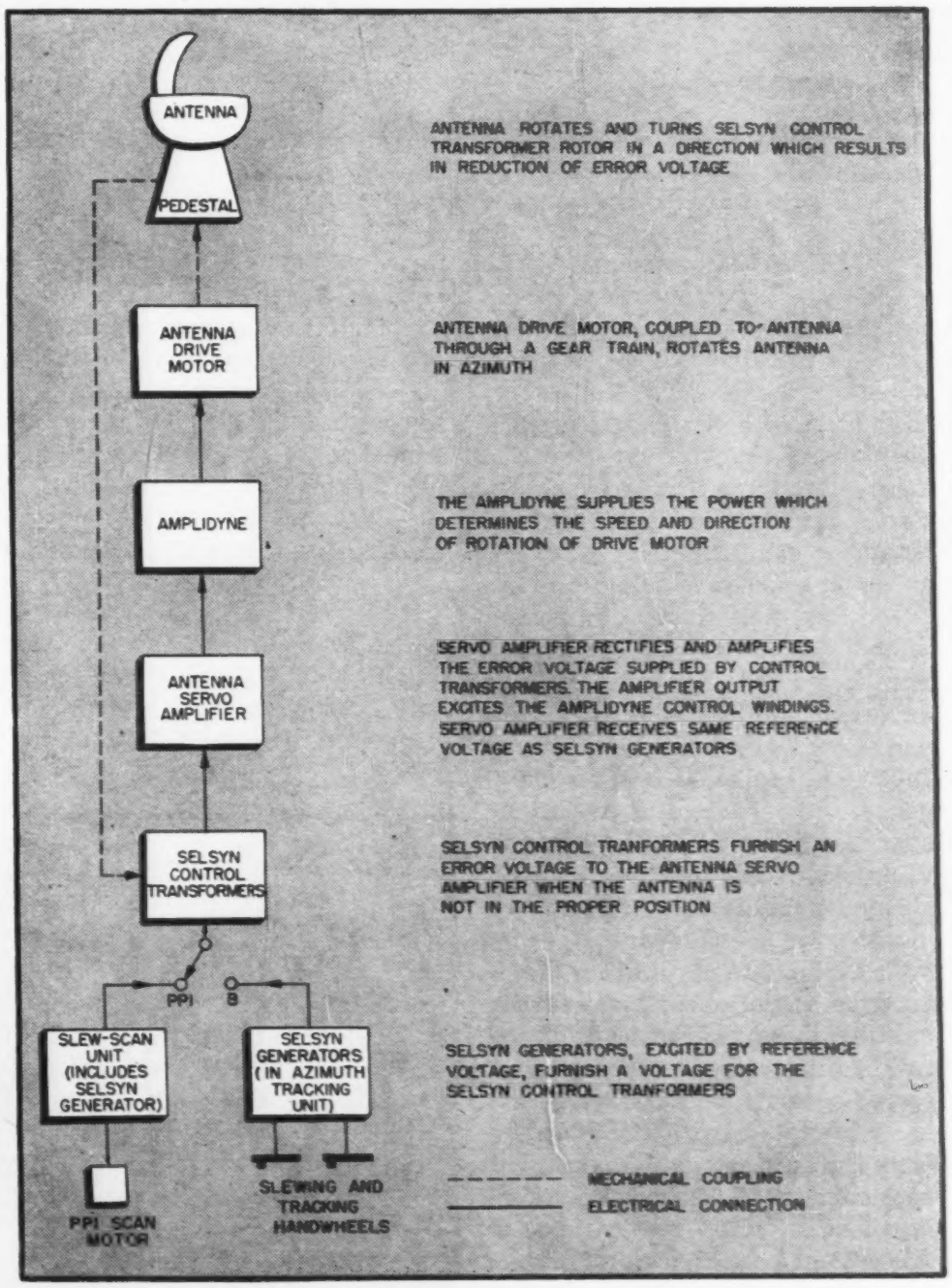

ANTENNA
PEDESTAL
ANTENNA DRIVE MOTOR
AMPLIDYNE
ANTENNA SERVO AMPLIFIER
SELSYN CONTROL TRANSFORMERS
PPI
B
SLEW-SCAN UNIT (INCLUDES SELSYN GENERATOR)
SELSYN GENERATORS (IN AZIMUTH TRACKING UNIT)
PPI SCAN MOTOR
SLEWING AND TRACKING HANDWHEELS
ANTENNA ROTATES AND TURNS SELSYN CONTROL TRANSFORMER ROTOR IN A DIRECTION WHICH RESULTS IN REDUCTION OF ERROR VOLTAGE
ANTENNA DRIVE MOTOR, COUPLED TO ANTENNA THROUGH A GEAR TRAIN, ROTATES ANTENNA IN AZIMUTH
THE AMPLIDYNE SUPPLIES THE POWER WHICH DETERMINES THE SPEED AND DIRECTION OF ROTATION OF DRIVE MOTOR
SERVO AMPLIFIER RECTIFIES AND AMPLIFIES THE ERROR VOLTAGE SUPPLIED BY CONTROL TRANSFORMERS. THE AMPLIFIER OUTPUT EXCITES THE AMPLIDYNE CONTROL WINDINGS. SERVO AMPLIFIER RECEIVES SAME REFERENCE VOLTAGE AS SELSYN GENERATORS
SELSYN CONTROL TRANFORMERS FURNISH AN ERROR VOLTAGE TO THE ANTENNA SERVO AMPLIFIER WHEN THE ANTENNA IS NOT IN THE PROPER POSITION
SELSYN GENERATORS, EXCITED BY REFERENCE VOLTAGE, FURNISH A VOLTAGE FOR THE SELSYN CONTROL TRANFORMERS
MECHANICAL COUPLING
ELECTRICAL CONNECTION

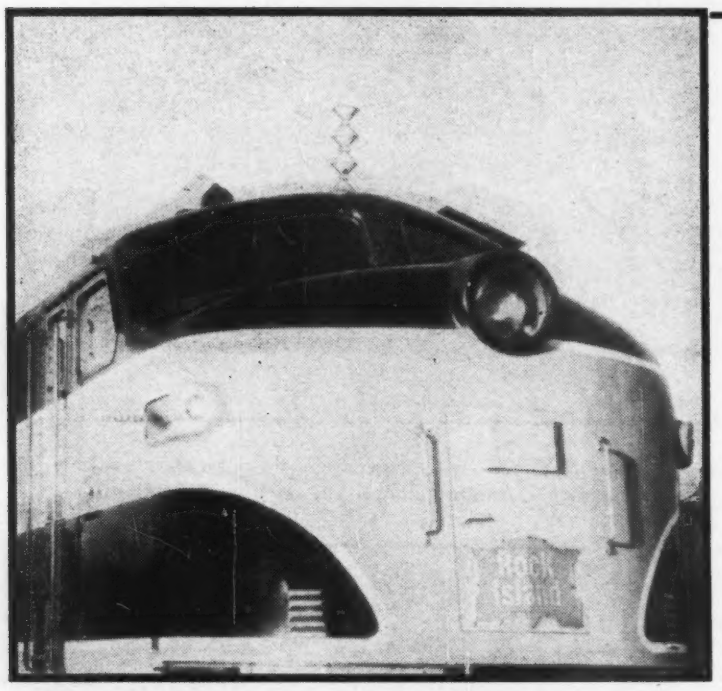
FIG. 12—Block diagram of the AN/MPG-1 radar antenna-positioning system. This electro-mechanical system must indicate the bearing of the beam correct to within 0.05 degree

REFERENCE

(1) McNaney, J. T., A Continuous Control Servo System; ELECTRONICS, p 118, Dec. 1944

electronics war report

2,660-Mc Train C



Reproduction of front cover of this issue, showing six-bank antenna array installed atop a locomotive

Details of complete two-way microwave f-m system to be installed on Rock Island Railroad. Transmitters and receivers are crystal-controlled and use klystrons as frequency multipliers and amplifiers. Antenna array has eight triple radial dipoles with biconical parabolic reflectors. Waveguide effect of train-filled tunnel is analyzed

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SOME of the prerequisites of a train communication system are dependability, ease of operation, and low maintenance cost comparable to corresponding features of present telephone and telegraph facilities. The equipment must also be constructed so that servicing can be done by railroad telegraph and telephone maintenance crews.

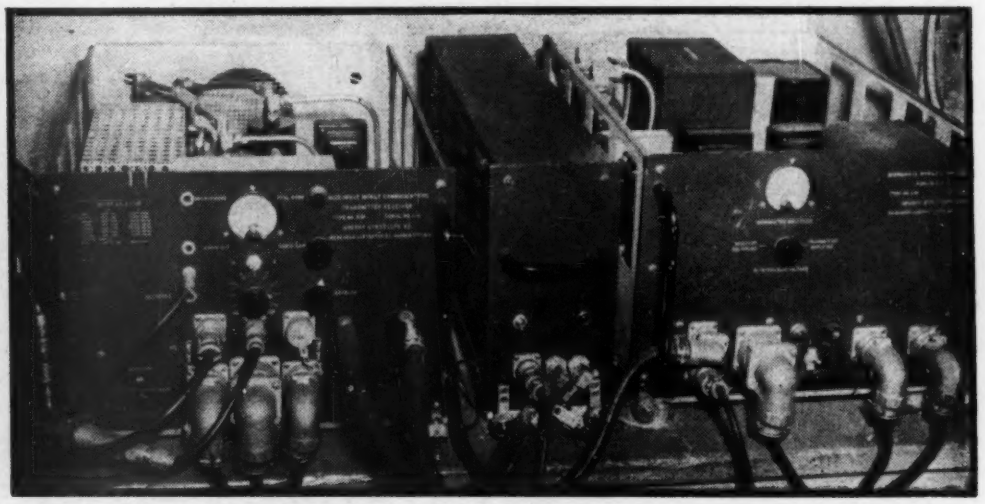
Radio communication from engine to and from the rear of moving trains and to wayside stations is now considered highly desirable as an extension of existing wire lines and carrier communication circuits. Likewise, the need is seen for emergency radio equipment to bridge stretches where wires are down and to provide auxiliary communication channels in storm or flood-damaged areas.

To explore the possibilities of microwave radio communication for meeting these requirements of the railroad industry, the Sperry Gyroscope Co. carried out extensive tests in collaboration with the Rock Island Railroad on a frequency of approximately 2,660 mc. Engineering evaluation of results showed this system to be highly practicable. In general, microwave equipment was found to possess excellent progagating characteristics under the conditions imposed in railroad operation. Except for interruptions lasting 15 to 20 seconds in unusual combinations of rolling hills and deep cuts, signals were consistently strong through all types of terrain, around curves, through gorges, and even in the long Moffat Tunnel.

Complete System Now Planned

Success of the operating tests resulted in the decision to lay out for the Rock Island Railroad a single radio communication system employing crystal control and frequency modulation in the region of 2,660 mc. The first complete system is to be installed from Chicago to Rock Island, Illinois, a distance of 180 miles. Steam and diesel freight trains operating in this territory will be complemented with two-way front-to-rear radio continuously operated, as will high-speed passenger engines and wayside stations.

The maintenance of all railroad equipment will be simplified by using only complete units that are replaceable easily and quickly. Whether the equipment is to be used in mobile or fixed installations,



Microwave mobile communications transmitter-receiver (at left) with liquid-cooling unit for klystrons (center) and power supply (right), all made by Sperry Gyroscope Co.

Communication System

it will be completely interchangeable since all units are designed for 115-volt, 60-cycle power. As failures occur, replacement of the unit will be the only servicing required in the field and the defective unit will be brought to a central point for repair. Frequency measurements and sensitivity tests will be continuously conducted by traveling field engineers to maintain the standards required of this important railroad service.

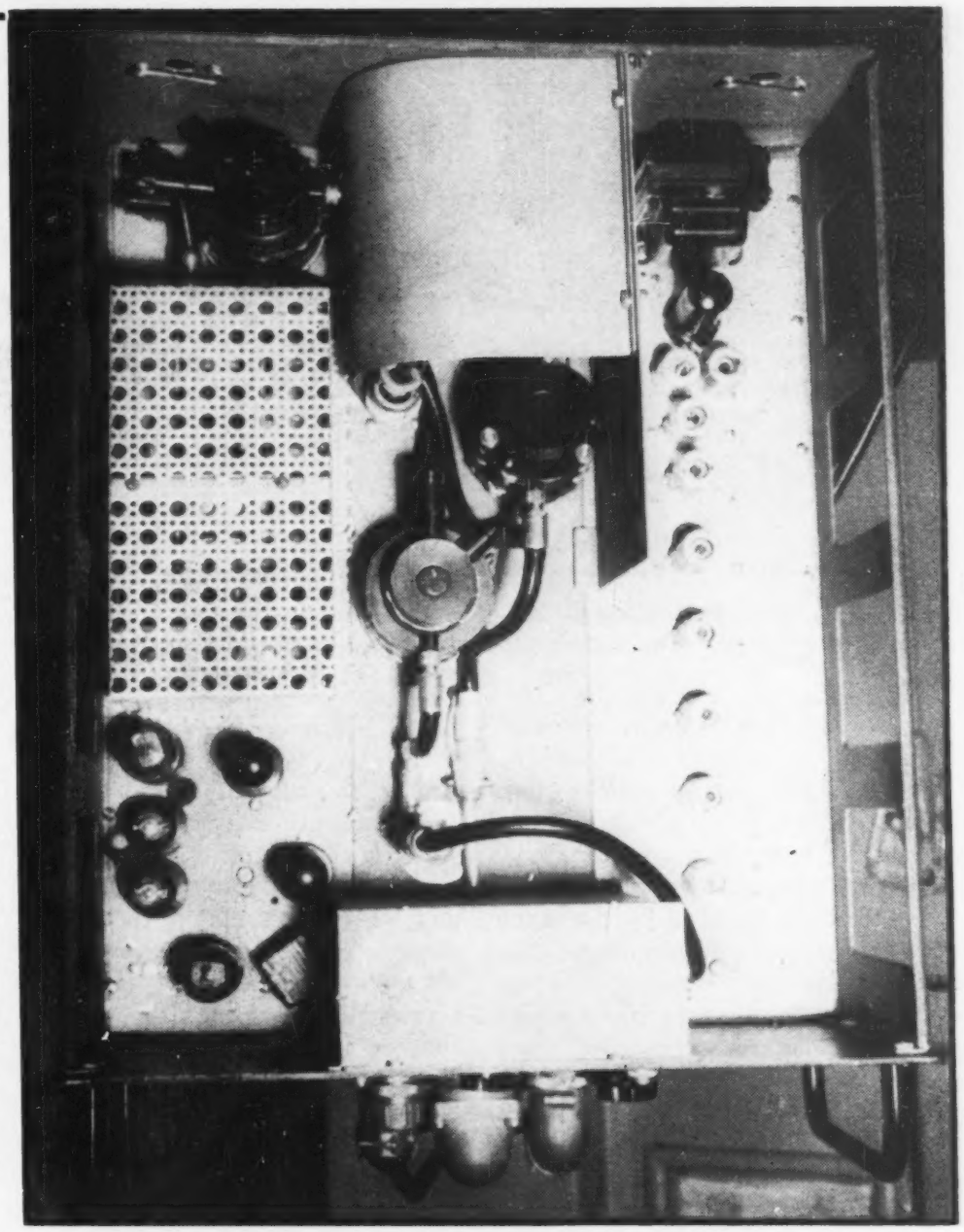
Antenna Gain Required

In railroad service, an antenna gain of about 20 (13 db better than an isotropic radiator) is the maximum amount desirable and at 2,660 mc this corresponds to an array about 60 inches high.

An isotropic point receiving antenna has an absorption cross-section area of $\lambda^2/2\pi$. However, only half the energy picked up by the antenna is available at the terminals of a receiver whose input impedance is equal to that of the antenna. Therefore the effective absorption cross-section area of an antenna is $\lambda^2/4\pi$. Gain then becomes the ratio of the absorption cross-section area of an actual antenna to the absorption cross-section area of an isotropic radiator.

Wave Propagation Problems

In free space or over reasonably flat terrain, a simple communication set will work well out to distances where the curvature of the earth has its effect. This is more than a sufficient range to satisfy the railroad service problem. The chief concern is with the effects of diffraction around an obstruction. This does not mean that the more normal propagation losses of free space are unimportant, but only that they are independent of frequency and can be overcome in almost any system. The modification of these free-space values as determined by tunnels, cuts, and curves in the railroad tracks is frequency-dependent and is the real heart of the problem of finding the best operating wavelength for a railroad radio system.

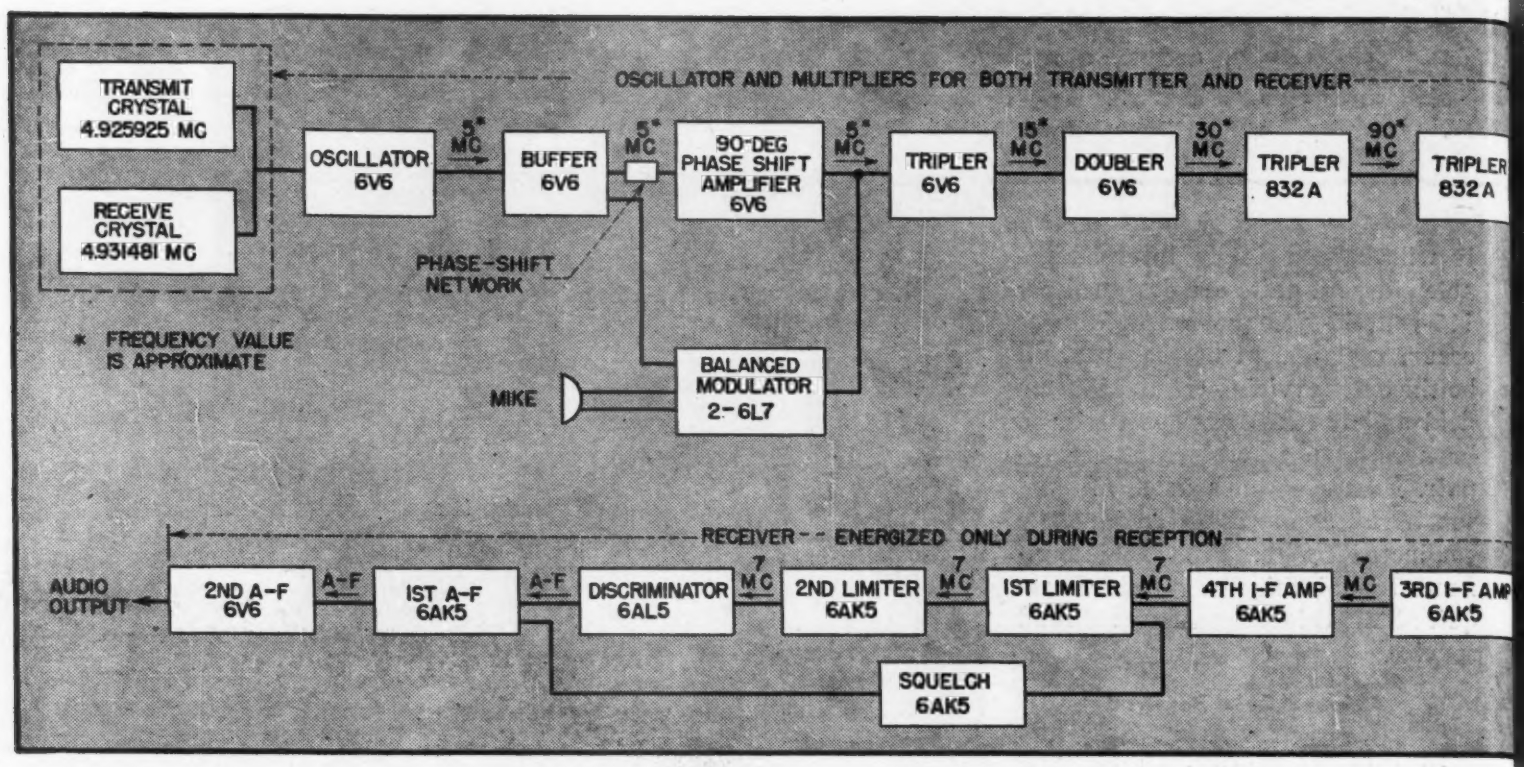


Complete 2,660-mc transmitter and receiver, with crystal multiplier and frequency-multiplier klystron at left, klystron power amplifier and klystron oscillator for receiver in center panel, and i-f stages on right-hand panel

Waveguide Effect of Tunnels

In tunnels or deep cuts which have an appreciable length, microwave frequencies are very advantageous. The tunnel problem is one of transmitting through the tunnel when a train is already blocking most of the available space. The problem of deep cuts is much the same; in both cases the space volume available for the transmission of radio energy is extremely restricted. The construction of tunnels and cuts is expensive and they are notoriously built with small clearance. This situation has been aggravated by the constantly increasing size of the railroad equipment, so that free spaces having at least one dimension as small as 6 inches are not uncommon.

The transmission of radio energy in a rectangular waveguide gives a rough idea of the size of an opening through a tunnel or cut which is necessary to complete a transmission. In one waveguide mode, the cutoff frequency is given by $f_c = c/2b$, where c is 3×10^4, b is the narrow dimension of the waveguide



Block diagram of 2,660-mc transmitter and receiver developed for train communication systems. Depressing push-to-talk switch on handset connects transmit crystal to oscillator stage, removes plate voltage from receiver stages and receiver klystron, and applies plate voltage to the two transmitter klystrons

measured in cm and f_c is the cutoff frequency. No energy at a lower frequency than f_c can be made to pass through such a waveguide and in a practical case, the opening must have a dimension at least 20 percent greater than specified by the cutoff frequency in order that attenuation be reasonable.

It is true that in an opening which has a cross-section 6 inches by 6 feet, other modes (the $TE_{0,1}$ mode for example) will be limited by the 6-foot dimension rather than by the 6-inch one. At best, modes of this sort will normally carry only half the energy so that restricting all transmission to them will reduce performance by a factor of 2.

In practice, the situation is even worse. Tunnels and cuts are not made up of plane surfaces. They are irregular and energy passing through them is, therefore, repeatedly coupled from one transmission mode to another and from that to a third and so on. It takes only one small dimension anywhere in the route to limit the frequency transmission characteristics of the path.

Diffraction Effects at Curves

When the railroad tracks curve around a hill or a similar obstruction, line-of-sight transmission from the transmitter to the receiver is not possible. Energy must then reach the receiver by diffraction around the edge of the obstruction or by reflection from a second object. If the obstruction has a sharp edge and is truly large compared to any wavelength that is to be used, the case must be compared to the diffraction of energy around a straight edge. Since the obstruction is assumed to be large, this case is entirely independent of frequency.

At least three factors modify this case for actual practice. Two of them favor using as short wavelengths as possible, while the third says lower frequencies are better. First, the train antenna does not move exactly along a line grazing the edge of the obstruction. Instead, it misses the edge by at least the width of the track. This means that, depending on how great that distance is in terms of wavelengths, the signal strength is reduced less than it would be in the optical case just referred to. Second, the edge of the obstruction is not sharp and, as measured in terms of microwaves, it may easily have irregularities great enough to raise the energy levels at points behind the obstruction. Third, the obstructions are not of infinite size and whenever they are only a few wavelengths high or wide, additional energy gets past. Larger obstacles satisfy this condition at longer wavelengths.

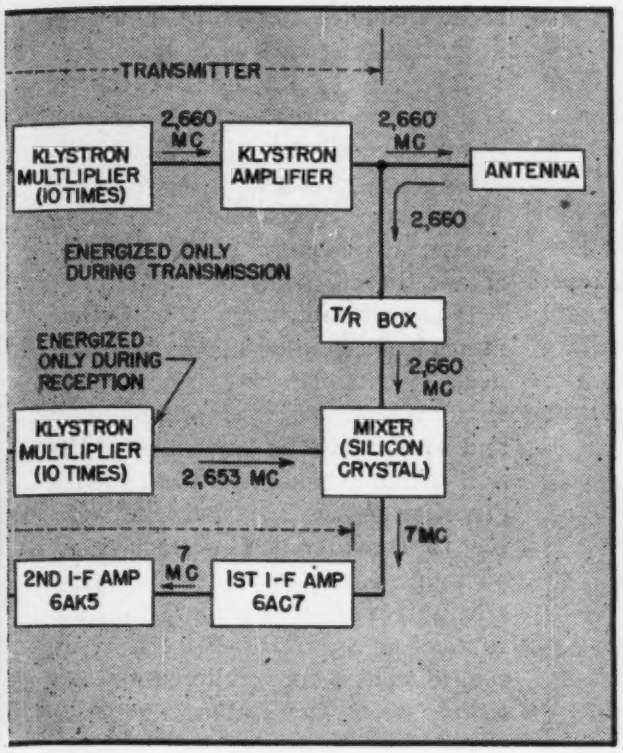
Because of the wide variations in the types of obstacles encountered, it is impossible to say from theory alone that a given wavelength is the best. This fact is the chief reason for the railroad tests which have been initiated and for additional tests scheduled for the future.

Transmitter Power Required

The transmitter power is limited chiefly by the bulk of the equipment and the power drain from the train's source that can be tolerated. Almost regardless of everything else, these considerations dictate a nominal 10 watts. It is true that this could be made 5 or 20 but since 20 watts adds only 3 db to a transmission link previously operating with 10 watts, it is not as great an advantage as might at first seem likely.

The cables connecting the antenna to the transceiver may be made to have very low loss if designed with care. The present microwave equipment uses stub-supported coaxial lines causing a loss of less than 3 db.

Receiver sensitivity, of course, depends on frequency bandwidth as well as on how well the receiver is



constructed. Because of variations in temperature, humidity, vibrations, and shocks that are encountered in actual service, a noise factor of about 15 db is present in the best receiver that can be built for railroad operation. This means that in the receivers used for the tests, at least a 15-db loss must be considered in conjunction with the 150-db-below-one-watt sensitivity which is theoretically possible.

The Equipment

The microwave transceiver is housed in a case measuring 16x11x 19 inches, containing all the microwave and audio circuits and a transmit-receive switch which automatically blocks the antenna connection to the receiver whenever power from the transmitter is fed to the antenna. The chassis also contains enough audio power to operate a loudspeaker in addition to the receiver of the handset. The power supply is housed in a case of the same size and converts the power from the train's supply system to produce all the necessary voltages.

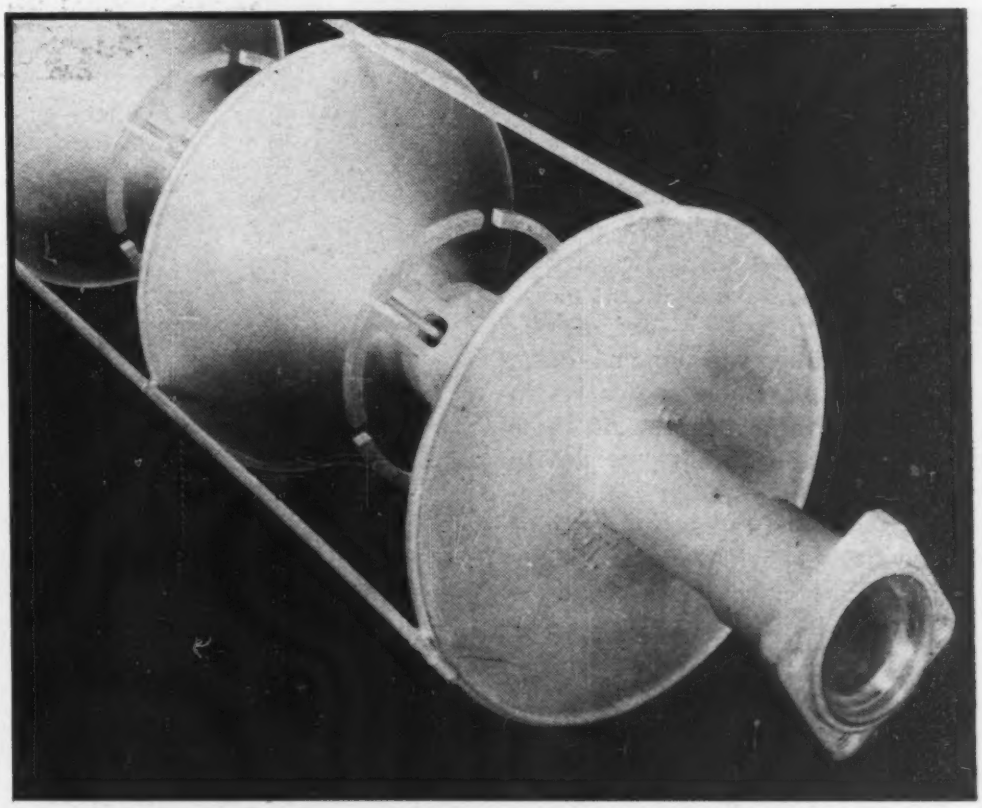
When a transmission is being made, a crystal-controlled oscillator produces a signal which is frequency-modulated, multiplied in frequency to the microwave level, amplified and fed to the antenna. During reception, the same frequency multiplier arrangement is used with a different crystal to produce a stable local oscillator signal at a frequency such that the receiver's intermediate frequency is produced as the difference. The local oscillator energy is fed into a microwave converter where it is mixed with the received microwave signal. The resulting intermediate - frequency signal is amplified, limited and demodulated by methods well known to the art.

During transmission a quartz crystal is used to control an oscillator whose output is modulated, amplified, and multiplied. The stability of the crystal oscillator is 5 parts in a million or better under normal operating conditions.

The frequency multiplier chain is quite conventional in design. As is usual in frequency multipliers, the resonant plate circuits of each stage are tuned to a new frequency two or three times higher than their grid circuits. The electrode voltages on the tubes are then so adjusted that an optimum amount of the proper harmonic of the grid frequency will be carried to the plate circuit and the oscillation there is, in consequence, excited so as to be locked in with the lower-frequency input to the grid.

Frequency multiplication is accomplished in four stages and raises the crystal oscillator frequency (with the modulation signal during transmission) up to a level where the klystron multiplier tube comes into operation. At high frequencies, conventional vacuum tubes are difficult or impossible to use because of the transit times of the electrons for which allowances must always be made. Fortunately, however, it is entirely feasible at 300 mc and higher to use resonant cavities in place of resonant circuits. This means that klystrons can be used for the final multiplication necessary to reach the desired microwave frequency. It is also fortunate that, because of the nature of velocity modulation, much higher multiplication factors can be used at frequencies where resonant cavities are of reasonable size.

During transmission, the microwave signal at 2,660 mc is obtained from the multiplier klystron and is amplified in a power-amplifier klystron. The amplifier tube is



Closeup of portion of antenna array, showing method of mounting curved dipoles and energizing inward-projecting stub. All parts shown here are precision-machined metal

relatively simple. Its cavities are all tuned to 2,660 mc and consequently straight amplification is obtained. The nominal rated power output during transmission is 10 watts.

Superheterodyne Receiver Used

The receiver portion of the transceiver is a superheterodyne. Except that the modulator is inoperative during reception and except that a different crystal frequency is being used, the production of a microwave frequency is accomplished, except for power amplification, just as it is during transmission. Now, however, the klystron multiplier is tuned to a lower frequency and produces a signal at 2,653 mc which is used to supply local oscillator energy. The 2,653-mc signal is mixed with the received signal in a resonant cavity type mixer.

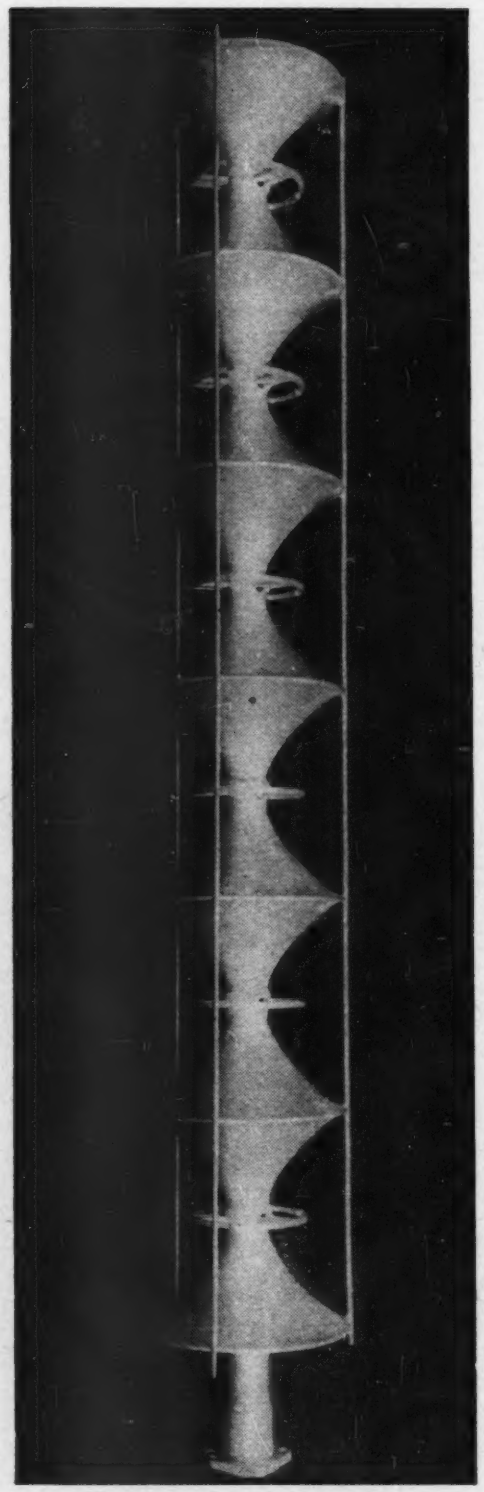
The purpose of the mixer is the same as it is in any heterodyne receiver. The 7-mc signal from the resonant cavity converter is fed to an i-f amplifier which has a band width of 300 kc at 6 db down and is conventionally constructed.

Antenna Design

One of the most interesting components in the equipment used for the 1945 test on the Rock Island was the antenna. The general problem encountered in the design of a microwave antenna is that of getting as much gain as possible while supplying coverage in the desired azimuthal directions. In the present equipment, it was decided that coverage over a full 360 degrees horizontally was advisable. Antennas must also have sufficient elevation coverage to take care of tilts as the train travels up and down grades. If the solid angles which must be covered because of these factors are added and compared to the 4π solid-angle radians of a completely isotropic radiator, a ratio of about 1 to 20 will be found. The gain which is desired for a fixed installation is, therefore, approximately 20 by definition.

If we can take all the energy from a transmitter and concentrate it in the desired directions alone, our signal strengths in those directions will be about 20 times as great as they would be if the energy were equally spread in all directions. Similar, such an antenna will absorb 20 times as much power from the desired direction as would be absorbed if it were designed to handle messages which approach from any arbitrary direction.

The approach to the problem consisted essentially of three steps. The first was to design a small structure which would radiate nearly equally in all directions. The second was to mount several of these elements, one above the other, at multiples of half-wave spacing so that much or all of the energy transmitted in the vertical direction would cancel because of phase differences introduced by the spacing. The third step consisted in adding biconical reflectors to further increase the efficiency.



Antenna array used on mobile units. Each of the six banks of the array has three radial dipoles at the focus of a biconical parabolic reflector. Overall height of this array is only 27 inches

It is well known that a half-wave dipole has a gain of 1.5. Such a unit used vertically does not have enough gain for this application and when mounted horizontally, the null which is found in the endwise direction makes it entirely impossible to consider the pattern near enough to isotropic to allow straightforward combination into a vertical array. Consequently, three such dipoles were combined and deformed so as to be able to produce a radiation pattern that is very close to being completely isotropic.

The general method of combining these triple units is to place them in a linear array with biconical parabolic cones acting as reflectors. For fixed installations in railroad yards or adjacent to the right of way, a stack of eight triple radial dipoles will be used with cones having a maximum diameter of about 8 inches. When mounted in a protective Plexiglass housing, the height is somewhat greater than the 54 inches occupied by the actual cones.

The antenna on a train (see front cover) is of similar construction but is smaller. It uses a stack of 6 radial dipoles, and the overall height of the antenna without the Plexiglass covering is 27 inches. This antenna has a gain of about 10.

Conclusion

While it is not believed that 2,660 megacycles is the optimum railroad communications frequency, it has been demonstrated that satisfactory operation at this frequency is now a reality. Further experimentation with this and other frequencies, perhaps higher, should reveal still more advantages of such a system. It is conceivable that railroads may find that those portions of the frequency spectrum which are devoid of man-made noise and static present the most practical solution for dependable communications.

Capacitor Charging Rectifier

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Experimental determination of design criterion for reactance-limited rectifier that will charge a large capacitor bank, which is used to supply power for industrial processes, in the shortest time and with the least power demand on the mains

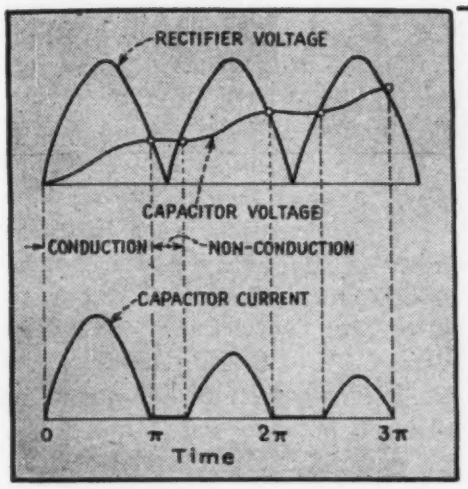


FIG. 1—Capacitor is charged by sinusoidal current pulses which raise capacitor voltage as shown

A NUMBER OF INDUSTRIAL processes depend upon storage of electrical energy in a capacitor. One process employing this principle is spot welding aluminum, magnesium and other high conductivity materials. In this case, a high voltage rectifier charges a large capacitor bank from an a-c supply. When the capacitor voltage reaches a predetermined value, (between 1500 and 2500 volts) the rectifier is blocked and the capacitor energy discharged into a welding transformer. After the weld is made the capacitor bank is recharged and maintained at the correct voltage until another weld is made.

Rectifier Design Problems

The rectifier used to charge the capacitor bank must be designed so that the initial charging current to the capacitor is limited. This can be accomplished by the reactance in the rectifier transformer. Also, the rectifier transformer should be designed to take into account the intermittent operation and the decrease in rectifier current as the capacitor is charged.

The problem is to design a rectifier which will charge a given capacitance to a particular voltage in as short a time as possible with a minimum demand on the a-c supply. To design this rectifier, the characteristics of capacitor voltage and capacitor current versus time must be determined. A recent paper[1] on waveshapes of capacitor current and voltage obtained from a single phase rectifier with resistance-capacitance load points out that capacitor current is delivered in discontinuous pulses as shown in Fig. 1.

To determine the capacitor voltage, a calculation must be made for each half cycle using the transient equation[2] for current delivered by a controlled rectifier to a complex load. Capacitor voltage at the end of each half cycle is used to determine starting voltage for the next half cycle.

While such calculations are possible when the instantaneous rectifier voltage is known and the current limiting resistance is constant, they become laborious if the time constant of the load is long. When transformer leakage reactance is used to limit the capacitor current, calculation of the capacitor voltage becomes very difficult, because the

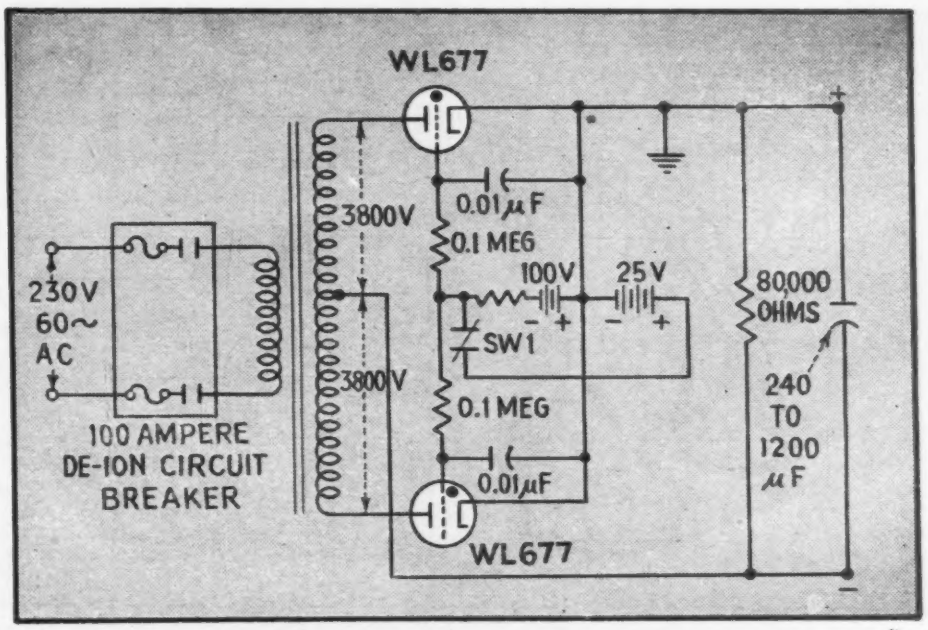


FIG. 2—The controlled rectifier used for empirical determination of capacitor charging characteristics

EDITORS' NOTE: The material for this paper has been taken from the author's thesis toward a Master of Science degree at the University of Pittsburgh.

rectifier output voltage is a function of capacitor current.

In view of the impracticability of calculating the capacitor voltage, an empirical equation for capacitor voltage versus time was evolved from experimental data in a manner illustrated in the following sections.

Experimental Rectifier

The rectifier used in this investigation is shown in Fig. 2. It is a single phase full-wave type using thyratrons as grid-controlled rectifying elements.

The rectifier transformer is a special reactance-limited transformer. The leakage reactance of the transformer was sufficient to limit secondary short-circuit current to 5.8 amperes rms. This reactance was obtained by wide physical spacing of the primary and secondary windings, but could have been accomplished more compactly by inserting a magnetic shunt between the two legs.

A negative voltage of 100 volts was applied to the grids of the rectifier tubes to prevent them from conducting. The rectifier tubes were made conductive by closing SW 1 which changed the bias on the rectifier tube grids from 100 volts negative to 25 volts positive (measured outside of the grid resistor). An 80,000 ohm resistor was connected across the output terminals of the rectifier to serve as a bleeder on the capacitor bank.

Capacitor charging tests were made on capacitor banks of 240, 600 and 1200 μf. Oscillograms showing capacitor voltage, capacitor current, and rectifier input current versus time were taken. From these oscillograms were derived the empirical equations for the capacitor voltage and current upon which rectifier design is based.

Oscillogram Observation

From oscillogram 1 of rectifier short-circuit current, a transient was observed during the first few cycles after the rectifier was started. This transient was observed in all capacitor-charging tests. It is of the same type that occurs in a-c circuits containing resistance and inductance when the circuit is closed at other than the natural-current zero instant.

In this rectifier the transient results in one tube carrying a higher peak current and conducting for a longer period (more than 180 degrees) than under steady state conditions. Because the second tube cannot begin to conduct until the first tube has ceased conducting, the second half cycle will be short and the tubes will carry unequal currents. Similar transients have been observed by the writer on other types of rectifiers such as the three-phase bridge type.

Such transients are undesirable because tube ratings may be exceeded. In practice the transient can be eliminated by starting with the rectifier tubes phased fully back and advancing the firing angle to full forward in several cycles.

Oscillogram 2 shows capacitor voltage and current versus time. By taking data from three such oscillograms, the curves shown in Fig. 3 were plotted. These curves do not contain the ripple which appeared on the oscillograms but are a plot of the maximum voltage attained during each half cycle. Consequently, the data presented here is not applicable to small capacitors being charged at high rates.

Development of Empirical Equation

From Fig. 3 it is apparent that the voltage curves follow a logarithmic law. Therefore, each of these curves was replotted on semi-logarithmic paper and the curves of Fig. 4, obtained. Each of these curves can be approximated by two straight lines with an error not exceeding five percent.

The first straight line (of slope M_1) is used as long as the rectifier current pulses join each other as shown on oscillogram 2. The second straight line (of slope M_2) approximates the observed curve after the current pulses become discrete. The change from one curve to the other occurred between $0.6E_M$ and $0.7E_M$ for each of the different capacitor banks.

Rectifier Design

From curves of slope M_1, the equation of the capacitor voltage versus time is $e_c = E_M(1 - \epsilon^{-M_1 t})$ where E_M is the maximum no-load transformer secondary a-c voltage and M_1 is the straight-line slope.

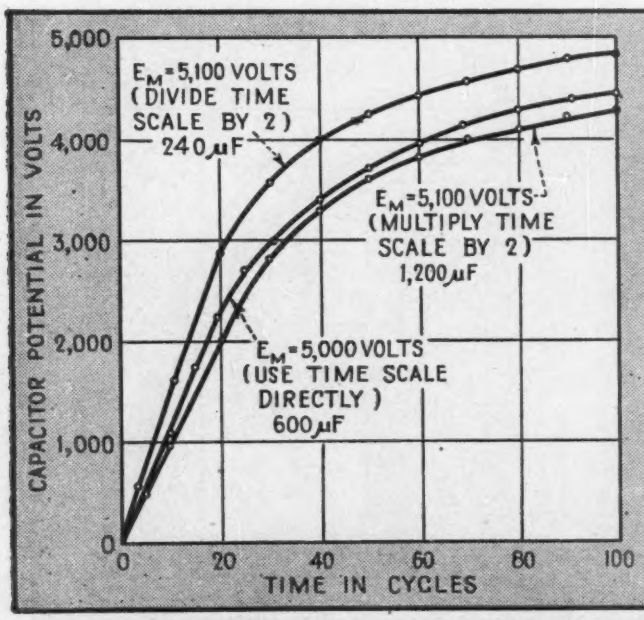


FIG. 3—Curves show the rate at which large capacitance banks charge when supplied from a full-wave rectifier

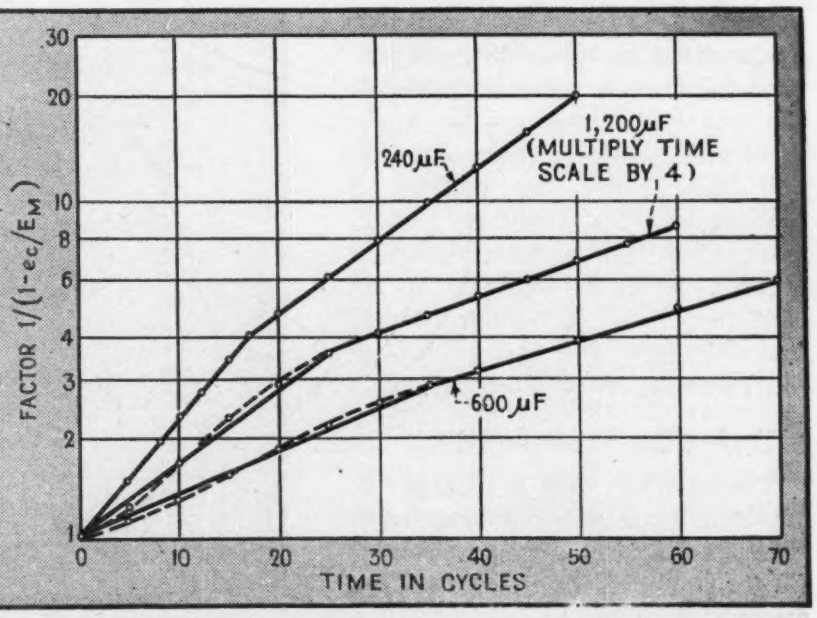


FIG. 4—Logarithmic plot shows that capacitors charge first at a fast rate, then change abruptly to a slow rate

Because the objective is to charge a capacitor in the shortest time with least demand on the power system, the no-load rectifier voltage should be selected so that the capacitor is charged to the desired voltage before the portion of the curve with slope M_2 is reached.

The slope of the capacitor charging curve is a function of transformer reactance. This inductance is in turn proportional to the maximum no-load rectifier voltage divided by the rectifier short-circuit current K_{sc}.

Because the only variable in the tests was load capacitance, the slopes of curves in Fig. 4 could be compared. They were found to vary as the 1.16 power of capacitance.

The inductance in the reactance-limited case is somewhat comparable to the resistance in the resistance case where the slope is $M = 1/RC$ when a steady d-c voltage is applied. Therefore the slope was assumed to be a function of the product of the inductance and the capacitance and was found to be $M_1 = 0.95(K_{sc}/E_M C)^{1.16}$. This assumption, although neglecting the reactance, is satisfactory for engineering design.

The equation for capacitor current can be found by differentiating the expression for capacitor voltage and multiplying by the capacitance, giving $i_c = CE_M M\epsilon^{-M_1 t}$.

The average capacitor current during the charging period is found by integrating the expression for the current during the charging period and dividing by the period T giving $I_{av} = CE_c/T$.

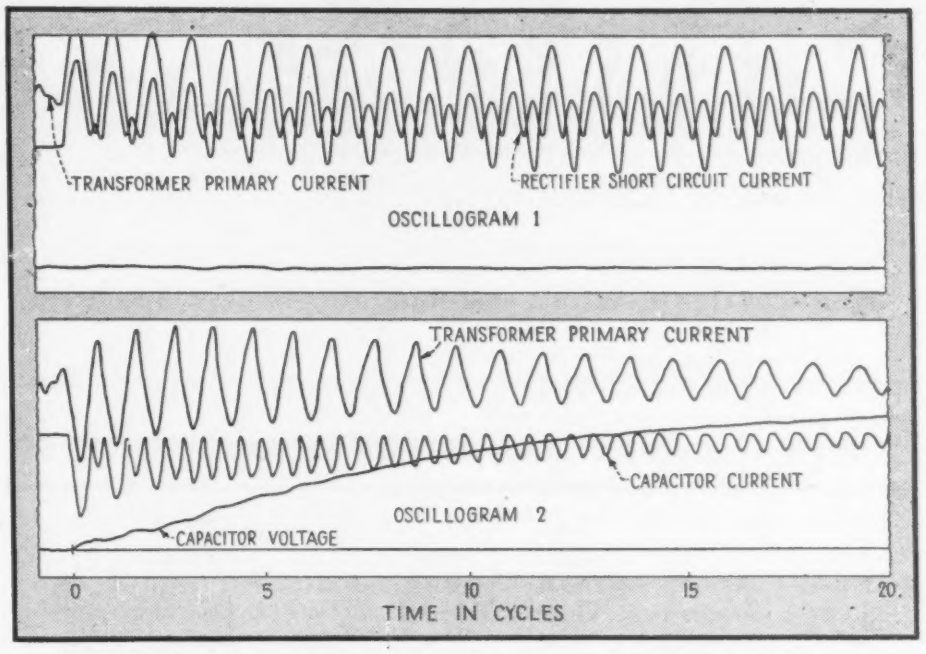
The expression for the rms capacitor current is slightly more complicated. Capacitor current consists of loops approximately sinusoidal in shape. Investigation on other types of rectifiers has shown that the form factor of these waves is within a few percent of being 1.11. Therefore the total rectifier rms current during the charging period can be expressed by

$$J = 1.11\, EC_M \sqrt{\frac{M_1}{2}\left(\frac{1-\epsilon^{-2M_1 t_1}}{t_1}\right)}$$

Because each secondary winding conducts current only one half of the time, the rms current rating of each secondary is $J_{sec} = 0.707\, J$.

The current rating of each secondary is also influenced by the time between charging periods. If the charging time is denoted as T_1 and the time between charges is T_2, the continuous secondary current rating is $J_{sec} = 0.707\, J\sqrt{T_1\,(T_1 + T_2)}$, if $T_1 + T_2$ is considerably less than the thermal time constant of the transformer.

When the rectifier is short circuited, the primary load current lags the applied voltage by about ninety degrees because of reactance limiting in the transformer. Therefore the magnetizing current of the transformer should not be neglected in calculating the kva of the primary winding. A satisfactory design can be obtained by adding the magnetizing kva to the lead kva arithmetically.



From oscillogram 1 an inequality in the current carried by two tubes was observed. From such curves as that of oscillogram 2 the charging conditions for various capacitances were determined

Illustrative Calculations

To illustrate a practical rectifier design, assume that it is desired to charge a 1200 μf capacitor bank to 2500 volts, 60 times per minute. About one-sixth of a second is required to discharge the capacitor for each operation. The kva demand on the power system should be as small as possible.

Let 2500 volts be about 0.6 of E_M or $E_M = 4150$ volts if the capacitor is to be charged before the charging rate decreases to the slower rate.

From the equation for capacitor voltage, the slope is $M_1 = (1/T_1) \times \log\,[1/(1 - E_c/E_M)]$. When T_1 is 5/6 second, $M_1 = 1.1$. Also, from the equation for slope in terms of rectifier impedance and this value of slope, we obtain $K_{sc} = E_M\, C$ antilog 0.132, or $K_{sc} = 4150 \times 1200 \times 10^{-6} \times 1.14 = 5.7$ amperes.

The rms rectified current during conduction is

$$J = 1.11\, E_M C \sqrt{\frac{M_1}{2}\left(\frac{1-\epsilon^{-2M_1 T_1}}{t_1}\right)}$$

or $1.11 \times 1200 \times 10^{-6} \times 4150 \times \sqrt{(1.1/2)\ (1 - \epsilon^{-2 \times 0.915})/(5/6)} = 4.1$ amperes during conduction.

The secondary rms current during the charging period is $J_{sec} = 0.707J = 2.9$ amperes. The total effective current in each secondary is $J_{sec} = 0.707J\ \sqrt{T_1/(T_1 + T_1)} = 2.9\ \sqrt{50/120} = 1.875$ amperes.

Calculation of the kva rating of each secondary can be made by $KVA_{sec} = E_{sec}J_{sec} \div 1000 = 2920 \times 1.875 \div 1000 = 5.5$ kva, and the total secondary kva is 11.0 kva. The primary kva due to the load is $KVA_{prr} = (E_{sec}J/1000)\ \sqrt{T_1/(T_1 + T_2)} = 0.645 \times 2920 \times 4.1 \div 1000 = 7.75$ kva.

The transformer can now be designed. It should produce a short-circuit a-c secondary current of $I_{sec} = 1.11K_{sc} = 1.11 \times 5.7 = 6.3$ amperes. The primary should be built to carry 7.75 kva plus the magnetizing kva continuously; while each secondary should be built to carry 5.5 kva.

References

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(2) Kurtz, E. B., Corcoran, C. F., "Introduction to Electrical Transients", John Wiley and Sons, Inc., New York, 1935, p. 163.

Cavity Magnetrons

First published information on microwave pulse generators capable of producing four million watts peak power at 3000 mc. Designed originally for radar but equally applicable to pulse communication systems, the cavity "maggie" is a milestone in electronics

TYPICAL CAVITY MAGNETRON CHARACTERISTICS

Type No. or Series	Frequency Range (mc)	Max. Peak R-f Power Output (kw)	Nom. Peak Anode Rating	Maximum Duty Cycle	Max.. Pulse Duration (μsecs)	Nom. Average Input Power (watts)
		L-Band (25–50 cm)				
700A-D	680–710 (fixed)	100	12 kv 10 amp	1/400	2	120
728AY-Gy	920–970 (fixed)	400	21.5 kv 35 amp	1/1000	2	600
5J21–5J25	1060–1110 (fixed)	600	22 kv 50 amp	1/1000	2	750
4J21–4J30	1220–1350 (fixed)	800	28 kv 60 amp	1/500	6	1500
4J42	660–730 (tunable)	200	12 kv 10 amp	1/200	2	500
4J51	900–970 (tunable)	400	21.5 kv 20 amp	1/1000	2.1	500
5J26	1220–1350 (tunable)	800	27.5 kv 60 amp	1/500	6	1500
		S-Band (8–11 cm)				
720Ay-Ey	2720–2890 (fixed)	1000	25 kv 70 amp	1/1000	2	1500
718Ay-Ey	2720–2890 (fixed)	200	11 kv 22 amp	1/500	5	600
4J45–4J47	2785–2890 (fixed)	600	25 kv 45 amp	1/1000	6.6	900
4J36–4J41	3400–3700 (fixed)	850	24 kv 43 amp	1/1000	2.5	1000
2J38–2J39	3249–3333 (fixed)	12.5	5.4 kv 5 amp	1/500	2	27
		X-Band (3 cm)				
725A	9345–9405 (fixed)	60	12 kv 12 amp	1/1000	2.1	120
2J51	8500–9600 (tunable)	60	14 kv 10 amp	1/800	2	200
2J55–2J56	9215–9405 (fixed)	70	12.8 kv 12 amp	1/1000	2.5	150
4J52	9345–9405 (fixed)	60	15 kv 20 amp	1/500	6	300

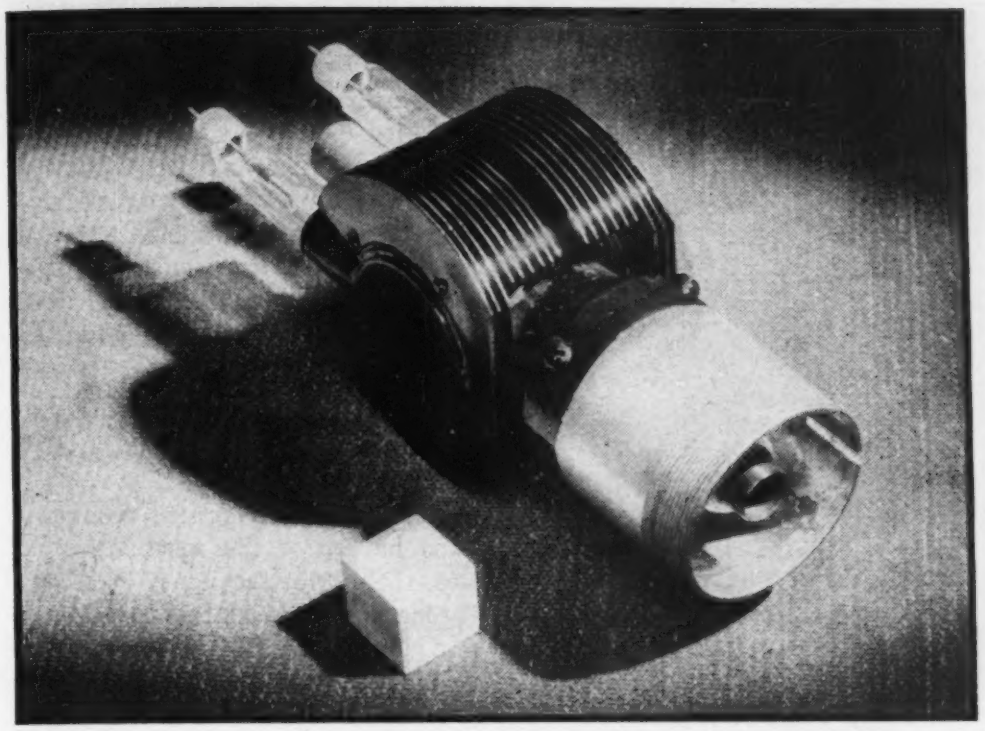
ONE OF THE outstanding electron tube types developed during the war is the cavity magnetron, a device which combines the principles of the magnetron and the resonant cavity. This type of magnetron, the basic transmitting tube of all American and British microwave radar sets, was developed in England in 1940 at the University of Birmingham. In September of that year the British Technical Mission brought one of the early models to the United States and revealed its construction and operating principles to scientists of the NDRC, Army and Navy.

It was evident that this tube made microwave radar practical for the first time, so a British-American agreement was concluded whereby the development of microwave radar would be undertaken in the United States. Within five weeks, copies of the British model were made at the Bell Laboratories. In November, 1940 the Radiation Laboratory was set up at M.I.T. to exploit the microwave radar field, using the cavity magnetron as the central element.

Since that time some 50 varieties of the tube have been produced, varying in peak output power from 2500 watts to 1,500,000 watts, at frequencies throughout the microwave region (50 cm waves and shorter). Experimental tubes, operating at 10 cm, have developed a peak power level of 4,000,000 watts.

General Characteristics

The cavity magnetron, like the conventional magnetron, is a diode. It consists of a cylindrical unipotential-heater cathode surrounded by a massive copper anode. The anode has, cut into it, several resonant cavities. In a typical tube, the cross-section of each cavity has the shape of a keyhole. The cavity



One million watts, peak r-f, is produced by this type 720 10-cm cavity magnetron. The heater leads, insulated to withstand the negative modulating pulse, are at the rear, coaxial r-f output fitting at front. The scale is indicated by the one-inch cube

proper is a cylindrical hole cut into the anode, which connects with the central cathode chamber through a thin slot. The cavities are excited by electrons emitted by the cathode and whirled in the cathode chamber, past the slots, by the action of an axial magnetic field.

The dimensions of the cavities are so chosen that their natural resonant frequency is the desired output radio frequency. The distance between slots is selected so that the adjacent cavities oscillate in reverse phase, that is, the voltage across one slot is opposite to the voltage across the adjacent slots. An even number of cavities is employed to permit maintaining this sequence of polarities throughout the structure. A coupling loop, inserted in one of the cavities, abstracts the power.

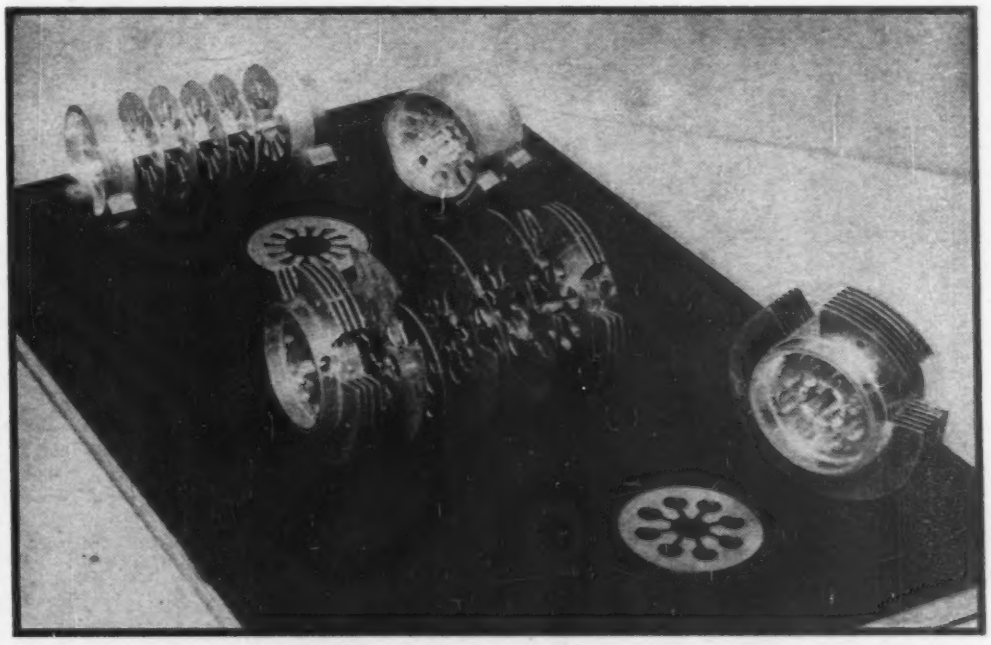
The cavity magnetron circuit is simple. For convenience and safety the anode is grounded, and negative modulating pulses are applied to the cathode. The shape of the modulating pulse must be as nearly rectangular as possible. When the pulse is applied, a current of several amperes (nearly 100 amperes in the high power types) flows between cathode and anode. Coincidentally, the cavities are excited and an r-f pulse appears across the coupling loop and its associated coaxial line. The plate efficiency of the tubes varies from 20 to 60 percent. Thus from a modulating pulse input of 1150 kw (24 kv at 43 amperes), an r-f peak power of 490 kw is developed by the type 4J36-4J41 series of tubes, an efficiency of 47.5 percent.

Two power ratings must be considered in the use of the cavity magnetron; the peak power and the average power. The peak power input is the product of the peak value of modulating voltage times the peak value of the resulting anode current. The peak power output is equal to this value multiplied by the plate efficiency of the tube. The peak power is limited fundamentally by the available emission from the cathode surface during the pulse, and also by the ability of the internal structure and output termination to withstand arcing.

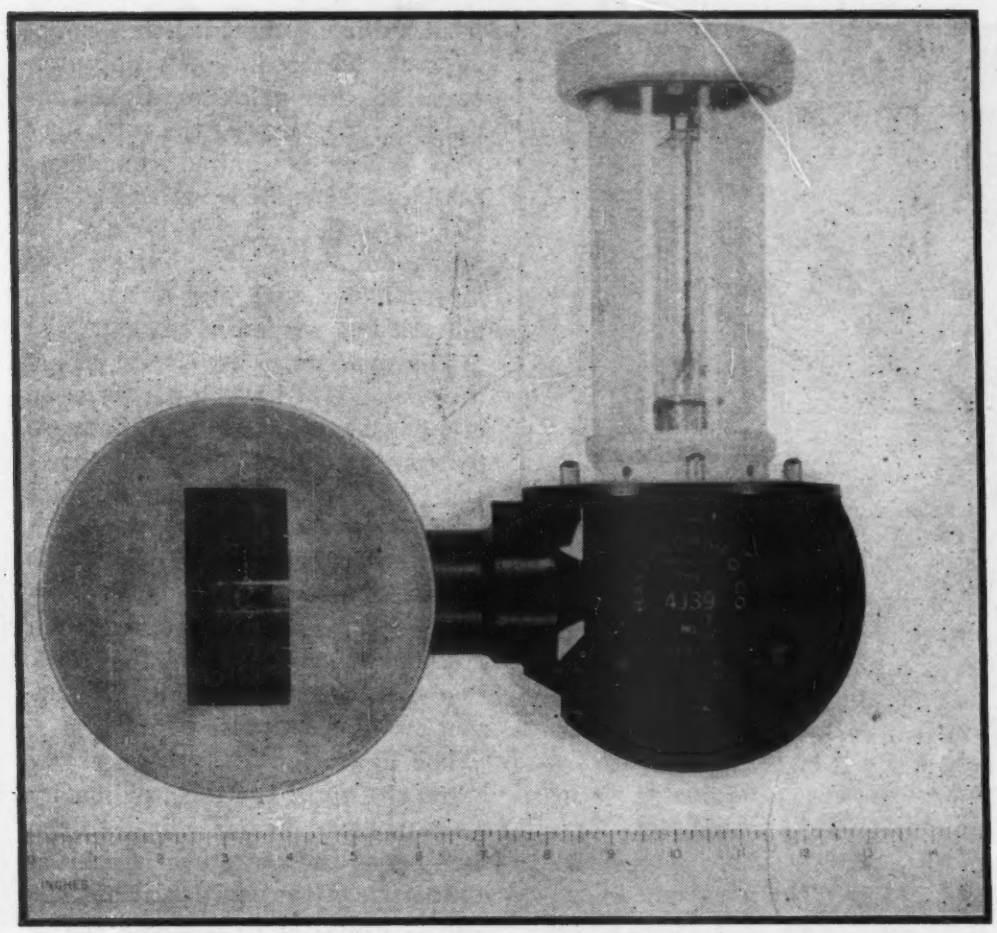
The average power is, in effect, the peak power in each pulse spread evenly over the interval between pulses. Consider a one microsecond pulse, transmitted at a pulse repetition rate of 1000 pps. Then the interval between pulses is 1000 microseconds and the tube is operative for only 1 microsecond in 1000, or 0.1 percent of the time. The average power is then 0.1 percent of the peak pulse power.

The limitation of average power is the ability of the tube, primarily the cathode, to dissipate heat. Since the size of the structure is limited by the size of the cavities, it has not proved practical to dissipate more than a few hundred, or at most a few thousand, watts continuously even by forced air cooling. The peak power outputs are generally about 500 to 1000 times as great. The maximum allowable anode temperature is generally 100 deg centigrade.

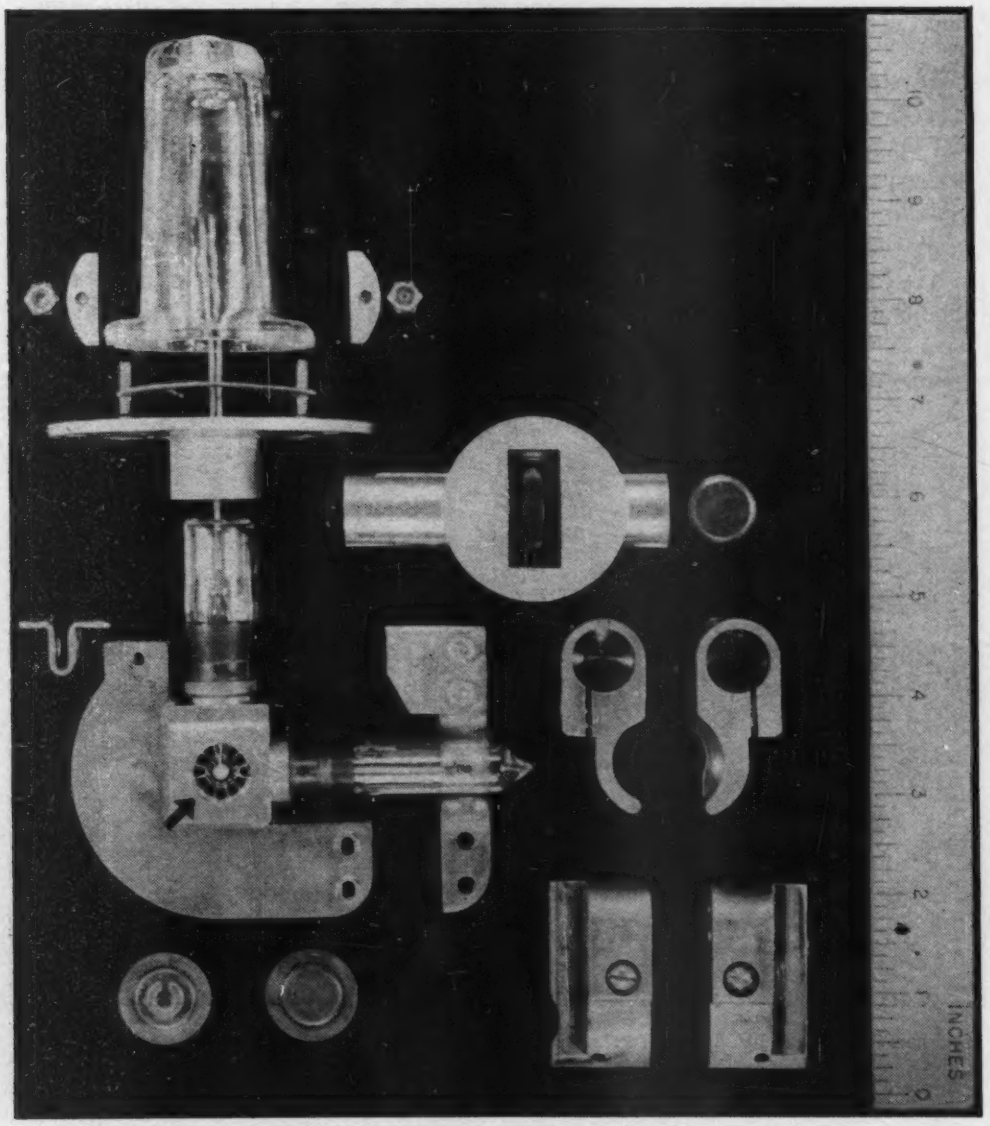
In practice the ratio between average power and peak power is stated in terms of the duty cycle that is, the pulse length in seconds



Construction of cavity magnetron anodes. Accurately punched laminations are built up to form a structure containing a number of cylindrical cavities surrounding a central chamber, which contains the cathode



Another form of megawatt magnetron, type 4J39. The power is taken from the cavities by a coupling loop and coaxial line which in turn is matched to the waveguide fitting



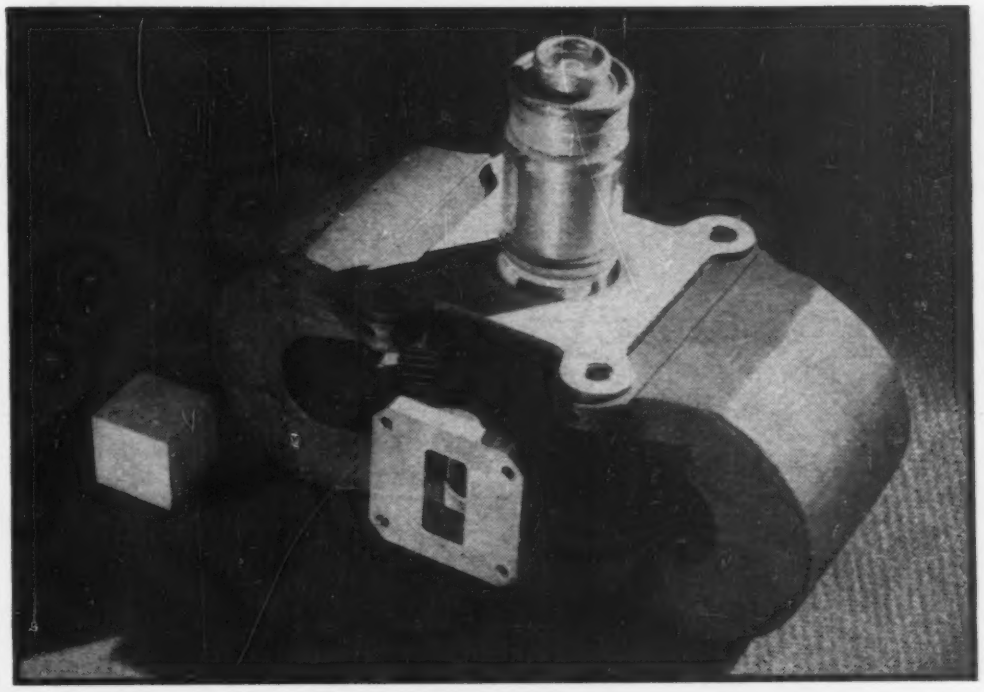
Exploded view of the 2J48 3-cm magnetron. The arrow marks the resonant cavity structure and cathode. The sealed-off coaxial output lead feeds the waveguide fitting. Power output, 50 kw

multiplied by the pulse repetition frequency in pulses per second. The duty cycles of typical production-type cavity magnetrons vary from 0.001 to 0.005. Within reasonable limits the average power capability may be used to produce long, widely-spaced pulses or short, closely-spaced ones. Pulses longer than a few microseconds are generally not recommended. Long operating life is achieved by operating at a low value of duty cycle.

Output Frequency and Undesired Modes

As might be imagined from the fixed nature of the resonant cavities, the output frequency of a cavity magnetron cannot be tuned over a wide range. The frequency can be shifted over a range of 1 to 2 percent of the carrier frequency, by adjustment of the load circuit connected to the r-f output. Tunable magnetrons are available which achieve wider variations in frequency (about 10 percent) by adjustment of an end plate which forms the top and bottom of the cavities and cathode chamber. When fixed magnetrons are used a series of tubes, having otherwise similar characteristics, is designed to cover overlapping regions of the spectrum. An example is the 4J36-4J41 series of six tubes which covers the range from 3400 to 3700 mc in six bands, each 50 mc wide.

Closely allied to the problem of obtaining the desired output frequency is the possibility of undesired frequencies due to improper operation of the tube. This effect, called moding, results from excitation of the cavities at modes different from the desired mode. Moding occurs when the anode current is lower than the normal value. Thus in the 2J38-2J39 series of tubes, operation at a peak anode current of less than 3 amperes is not recommended because the output frequency is then likely to be composed of many different modes, all of which require input power and thus lower the efficiency. More important is the fact that at low peak anode currents, the operation is



Type 4J52 2-cm matgnetron within integral permanent magnet. This is among the latest designs, employing a coaxial cathode terminal (top) and waveguide for r-f output

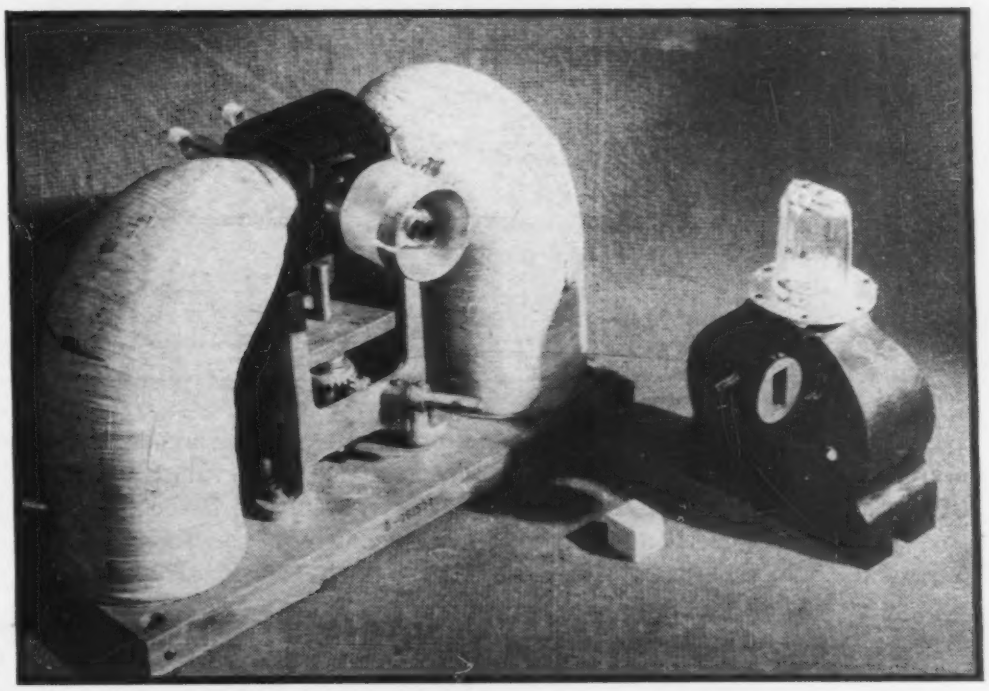
unstable, and the useful power output is apt to vary in amplitude and frequency. The same conditions may obtain at high space-currents with abnormally low peak anode voltage, which can occur if the magnetic field has too low a value.

The presence of off-frequency modes is minimized if the modulating pulse is closely rectangular in form, since the peak values of anode voltage and current are then attained almost instantaneously and the tube has no opportunity to oscillate at low values of current or voltage. The sideband spectrum of a rectangular pulse contains appreciable energy at frequencies far removed from the carrier, much more so than would occur if the modulating waveform were rounded. This outer sideband energy is generally not accepted by the radar receiver, and thus represents a waste of power as well as a source of interference to other receivers. But these disadvantages are outweighed by the advantage of stable magnetron operation which results from rectangular modulation. The time of rise of the pulse is generally about 0.1 to 0.2 microsecond and the time of fall about 0.4 microsecond.

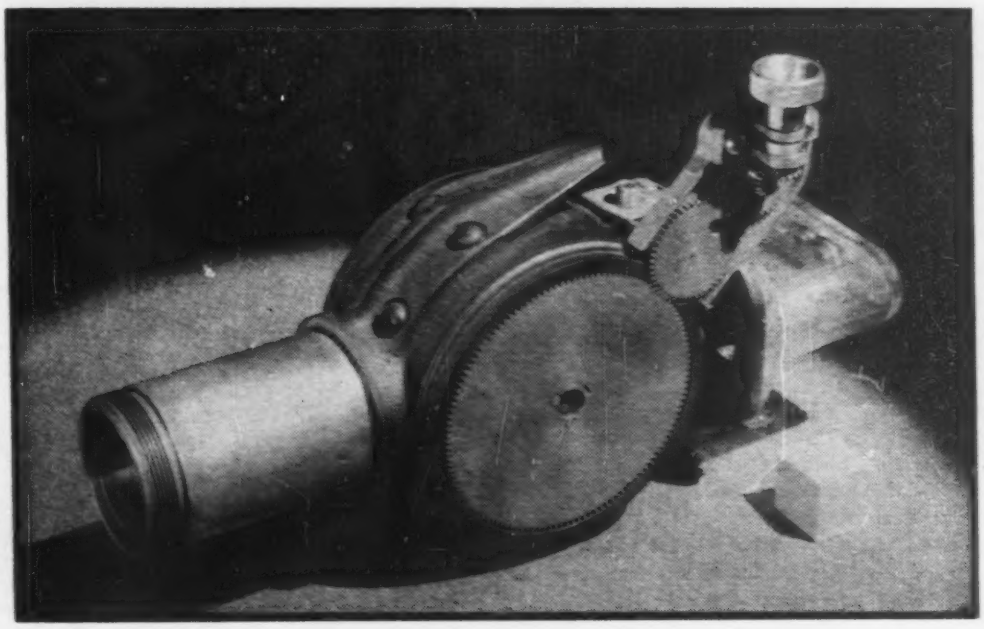
Magnetic Aspects

The magnetic field applied to a cavity magnetron serves to deflect the electrons as they pass from cathode to anode, causing them to follow curved paths of cycloidal form. Values of magnetic field from 1000 to 3000 gauss are usual in 10-cm magnetrons, while higher values, up to 6000 gauss, are used in the 3-cm types.

In early equipment, electromagnets were occasionally used, but these have now been supplanted by permanent magnets composed of Alnico or similar high-energy material. In most equipments the magnet is permanently installed in the transmitter, and the magnetron in-



Typical magnet structures. At left a 2900-gauss unit for the type 720 10-cm megawatt tube. At right, type 725, 3-cm tube with cast 5500-gauss magnet



Tunable cavity magnetron, type 4J51. This tube covers the range from 900 to 970 mc, with a peak power of 400 kw. Cathode input at right, coaxial output at left

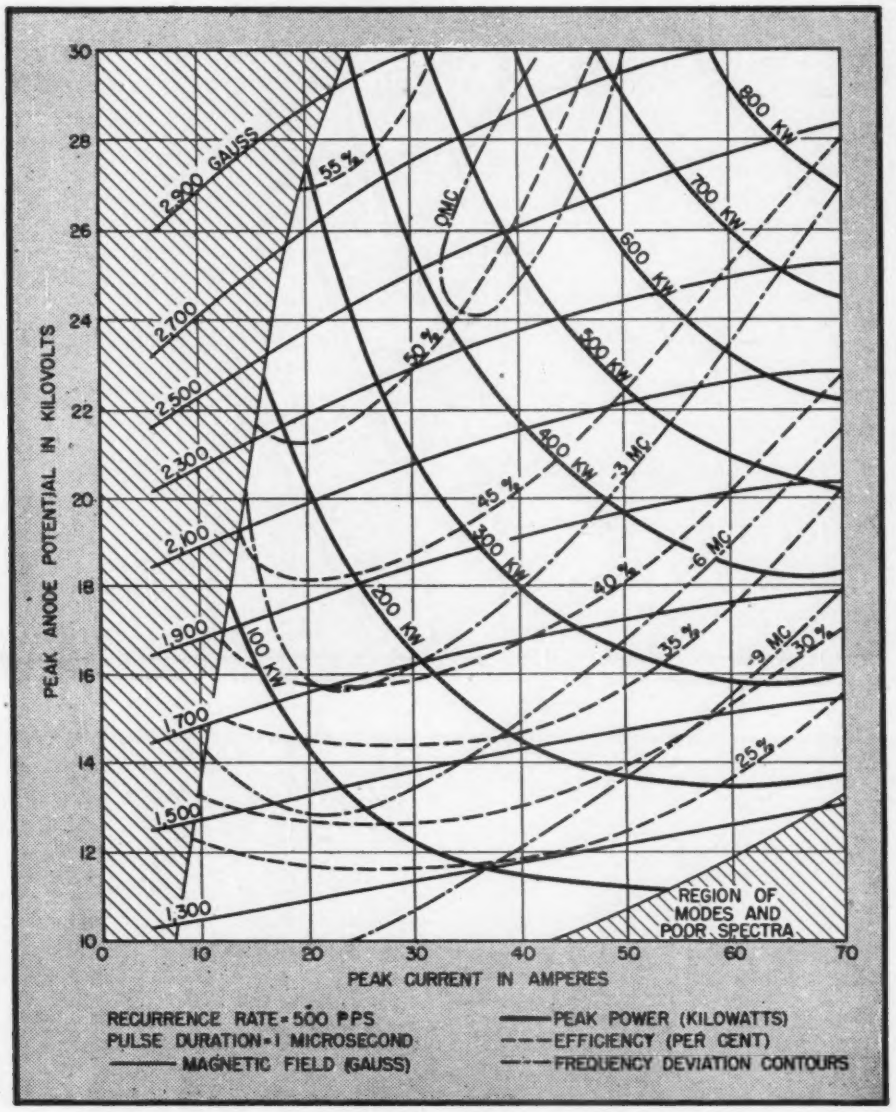


FIG. 1—Power output and efficiency of the 4J39 as functions of peak plate voltage, peak plate current and applied magnetic field. The shaded areas are regions within which unstable operation occurs due to the formation of multiple resonant modes within the cavities

serted between the pole pieces. More recently the "packaged" magnetron has made its appearance. In this type the magnet is an integral part of the tube structure. This design is economical in types requiring small amounts of magnetic material, such as the 3-cm tubes and low-power 10-cm tubes. In the high-power 10-cm and 50-cm tubes, the magnet is a bulky and heavy affair, so the package form is seldom used.

Figure 1 illustrates the effect of the magnetic field on tube operation. The light solid lines, marked with values of magnetic field, are static characteristics, that is, corresponding values of d-c peak modulating voltage and resulting d-c peak anode current. As the magnetic field is increased, the curvature of the electron paths in the cathode-anode space becomes more pronounced and the anode current drops. Thus high values of modulating voltage are associated with high values of magnetic field, to achieve the desired value of anode current. High efficiency of operation is also associated with high

Variations in the magnetic field are to be avoided since they introduce, simultaneously, variations in output power, operating efficiency and output frequency. Highly constant fields may be achieved, fortunately, from permanent magnets, provided only that the reluctance in the air gap remains constant, which implies merely a steady support for the magnetron itself.

Those aspects of magnetron operation which are independent of the magnetic field are illustrated in Fig. 2, which applies to an integral-magnet 3-cm type. As the peak modulating current is increased, with constant magnetic field, the anode voltage rises slowly, the power output rises almost linearly, the efficiency drops, and the bandwidth of the r-f spectrum rèmains substantially constant.

The Rieke Diagram

An important representation of cavity magnetron operation is the Rieke diagram which indicates the values of magnetic field.

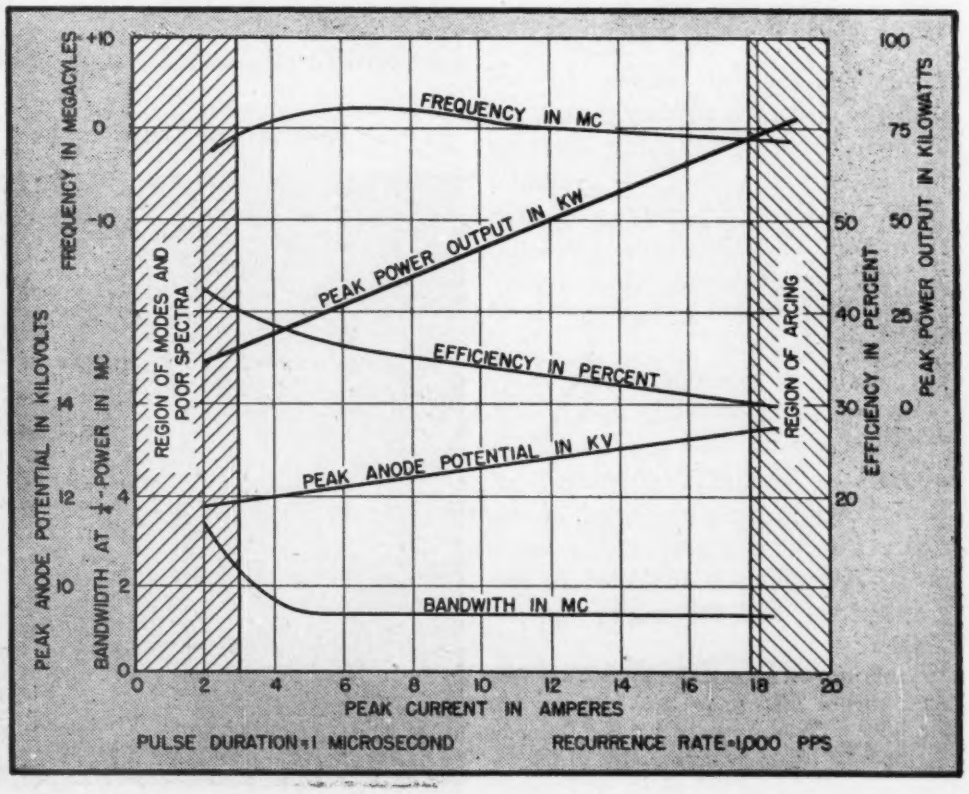


FIG. 2—Operating characteristics of the 2J55-2J56 3-cm tube, which operates at a fixed magnetic field supplied by the integral magnet. Note the increased bandwidth resulting from multiple modes at the lower values of anode current

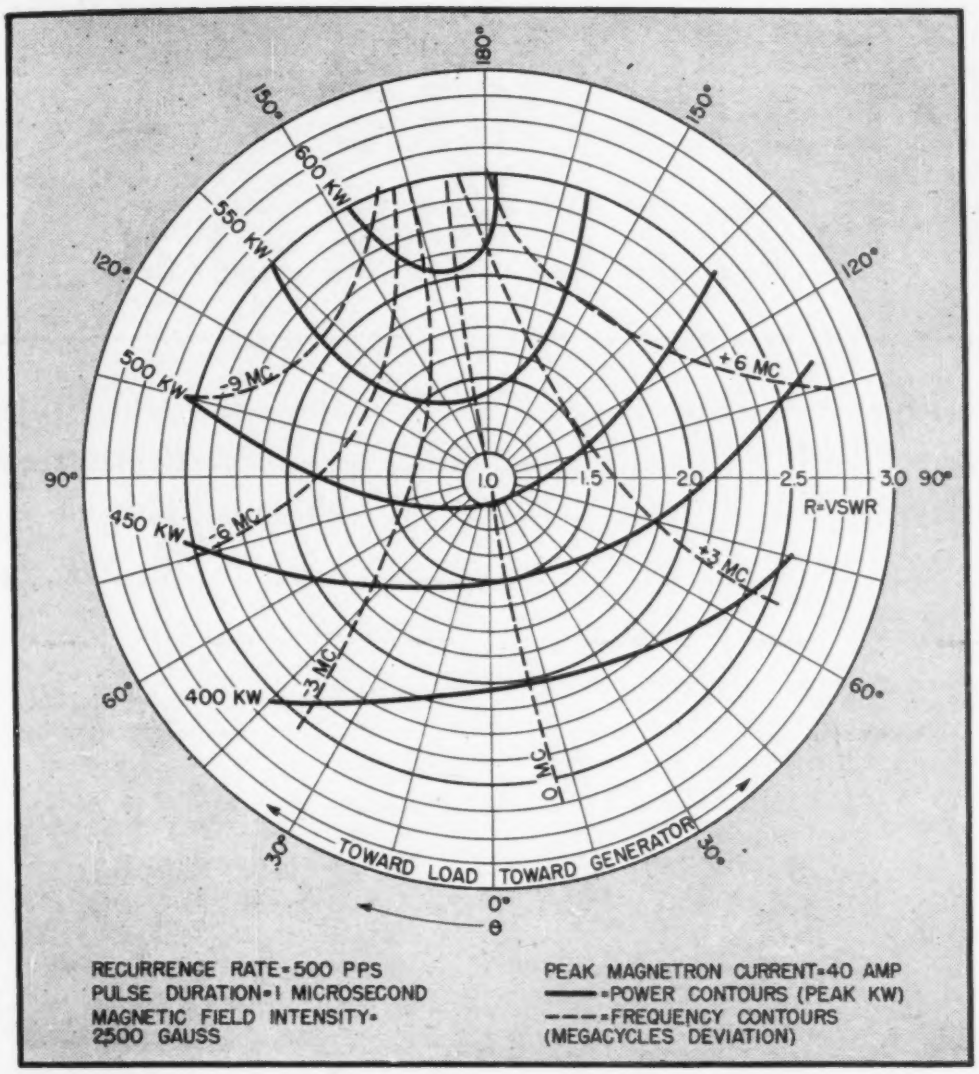


FIG. 3—Rieke diagram of the 4J39 megawatt tube, a contour diagram showing the output frequency and power level as functions of the standing wave ratio and position of the voltage minimum along the attached transmission line or waveguide

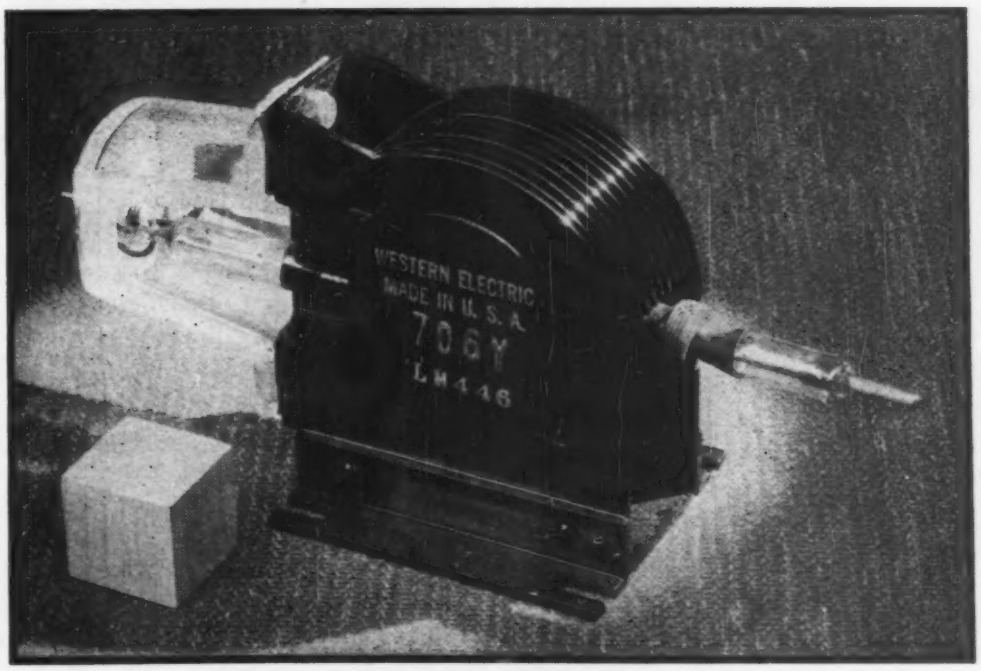
effect of the impedance of the connected load. The impedance may be varied intentionally to tune the magnetron or to vary its power output. The diagram indicates the value of standing-wave ratio and the position of the voltage minimum required to achieve a given frequency and power output. More generally, the diagram is used to indicate the operating condition of the magnetron from impedance measurements. The recommended operating condition corresponds to the center of the diagram (matched load).

A typical Rieke diagram is shown in Fig. 3. The standing-wave ratio (ratio of maximum voltage to minimum voltage in the standing wave on the connected transmission line or waveguide) is measured radially from the center. The position of the voltage minimum, measured in electrical degrees from the output flange of the tube, is represented as an angle from 0 to 180 deg. On these coordinates are plotted contours of peak power output and frequency deviation relative to the design value (the value at the center of the diagram corresponding to a matched load, with unity standing-wave ratio). The frequency contours tend to converge, in the case shown, along the angle 175 deg.

Mechanical Design

The mechanical features of typical cavity magnetrons are illustrated in the accompanying pictures. The cathode heater leads are insulated to withstand the high voltage modulating pulses. The r-f output is taken from the opposite side of the structure, in a terminal of the coaxial or waveguide form, depending on the r-f system to which the tube is connected. Since it has not proved practical to abstract the power except by means of a coupling loop, the output from the tube proper is always in the form of a coaxial line. When waveguide output termination is required, the coaxial segment is coupled to the waveguide by extending the inner coaxial conductor across the waveguide cross-section.

The editors wish to acknowledge the assistance of the following men, and their associates, in furnishing characteristics and photographs of the magnetrons described in this article: H. D. Hagstrum, Bell Telephone Laboratories; Ray Rice, Raytheon Manufacturing Company; and Clifford Johnson, Western Electric Company.—D.G.F.



Type 706Y, a close copy of an early British model, with coaxial r-f output lead at right. Note fins surrounding anode for forced-air cooling. About 300 kw peak output at 3000 mc

SUPERSONIC FLAW DETECTOR

Industrial electronic device reveals presence of internal defects in castings and other solid objects. Quartz transducer energized with supersonic pulses as high as 12 mc is held against object, and time for pulses to travel to flaw and back is measured with an oscilloscope

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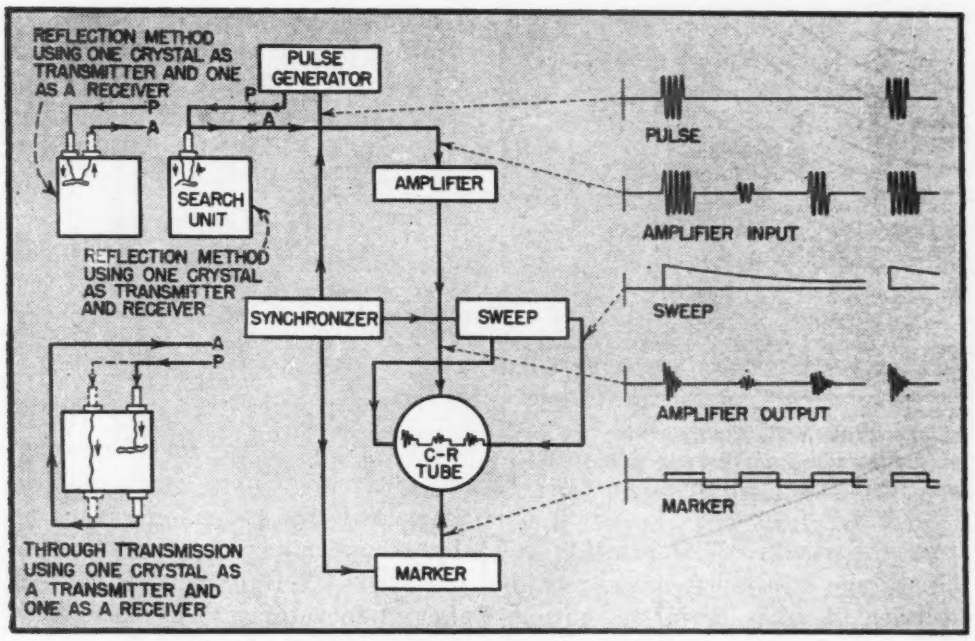


FIG. 1—Block diagram of supersonic flaw detector, with graphs of waveforms at various points in the circuit

THE electronic instrument to be described uses supersonic energy to nondestructively test and measure solid parts. Internal defects are detected by sending supersonic pulses into the material under test and measuring the length of time it takes these pulses to penetrate the material, reflect from the opposite side or an internal defect, and return to the sending point. The instrument will locate defects at distances of the order of 10 feet from the testing surface.

Frequencies Employed

The supersonic frequencies employed are between 0.5 and 12 mc, far beyond the audible range. This band of frequencies is often considered as being occupied only by radio waves. Yet sound waves (which are waves of molecular vibrations) have been generated at frequencies up to and even above 500 mc.

Unlike radio waves, which travel mainly in air and on the surface of metallic objects, supersonic waves are very rapidly attenuated in air and travel best in matter such as metals, liquids, plastics, and wood. In travelling through one inch of air, a supersonic vibration with a frequency of 1 mc is attenuated by one half, whereas in water this same vibration would travel over 130 feet before being attenuated by the same amount as through air.

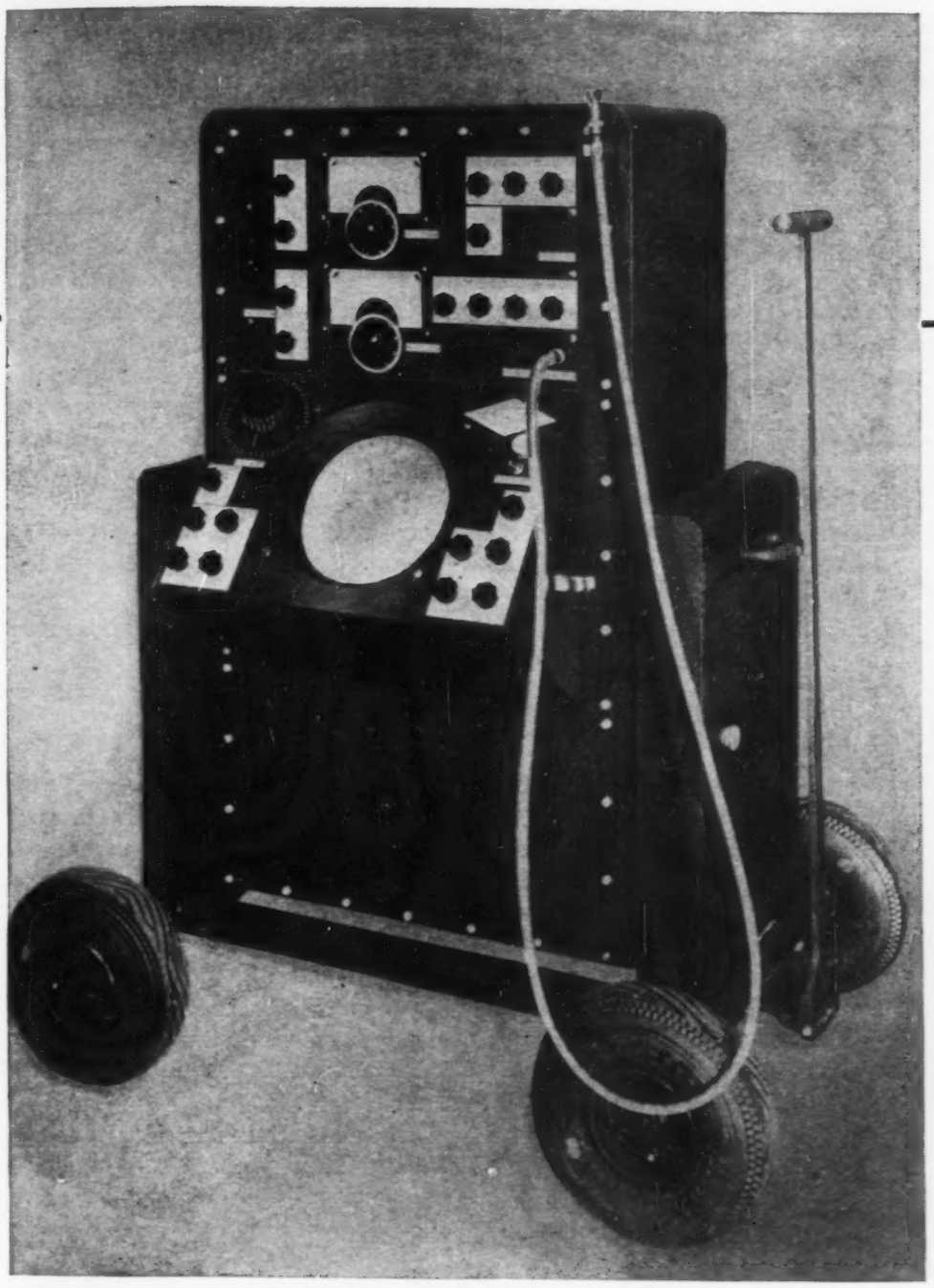
Several types of sound waves may travel in solids, but in air most sound waves are of a compressional character. Since the main applications for supersonic testing have been confined to longitudinal waves (in which the molecules vibrate in the direction of propagation), this article will be concerned only with waves of that type.

Properties of Supersonic Waves

The velocity of longitudinal sound waves travelling through a solid depends upon the density and elastic properties of the material. The two elastic properties involved are the extension or contraction of a body caused by a given force (the ratio of the stress intensity to the strain produced by

TABLE I. Ratio of Reflected Energy to Supersonic Beam Energy at Normal Incidence on Various Interfaces

Interface	W_r/W_i
Steel—oil	0.89
Steel—mercury	0.16
Steel—water	0.88
Glass—water	0.59
Glass—oil	0.68
Liquid—air	1.0
Solid—air	1.0



Type SR04 Supersonic Reflectoscope on rubber-tired carriage, with quartz crystal transducer mounted at end of cable

that stress) and the unrestricted change in width caused by a given change in length (the ratio of the compressional strain to the tensile strain, or vice versa). These two fundamental elastic properties are well known in the field of applied mechanics and are called Young's Modulus and Poisson's Ratio respectively. The velocity of longitudinal sound waves through solids is given by[1]

$$V_l = \sqrt{\frac{E}{\rho}\,\frac{1-\sigma}{(1+\sigma)(1-2\sigma)}} \qquad (1)$$

where

V_l = velocity of longitudinal waves in cm per sec

E = Young's Modulus in dynes per sq cm

ρ = density in grams per cu cm

σ = Poisson's ratio (no units)

Wavelength, velocity, and frequency are related by the same equation as in other types of wave motions.

$$\lambda_l = V_l/f \qquad (2)$$

where

λ = wavelength in cm

V_l = velocity of longitudinal waves in cm per sec

f = frequency in cps

Because of their short wavelength, longitudinal sound waves may be presumed to be travelling through a uniform medium in a beam when they are generated in this form. Sound waves may be easily beamed at these frequencies since suitable electromechanical transducers having radiating surfaces of several wavelengths on a side are commonly available.

The factor which determines reflection at a boundary is the product of the density and velocity, and is known as the specific acoustic impedance. The equation which relates the reflected energy to the incident energy is[2]

$$W_r = W_i\left(\frac{\rho_1 V_1 - \rho_2 V_2}{\rho_1 V_1 + \rho_2 V_2}\right)^2 \qquad (3)$$

where W_r is the reflected energy, W_i is the incident energy, and the subscripts 1 and 2 identify the properties of the two mediums. Any consistent system of units may be used. In the case of a steel-air interface nearly 100 percent of the energy is reflected. A few other common boundary reflection percentages are given in Table I. Table II gives the common acoustical properties of various mediums.

If the medium is composed of uniform particles of dimensions approaching those of the supersonic wavelengths, no reflections are observed due to scattering and absorption of the energy.

Description of Instrument

The functions of the various circuits are shown in Fig. 1.

The synchronizer provides three different output pulses sixty times a second. The first pulse is the rapid rise and exponential decay of the sweep voltage. The second pulse starts and stops the distance markers and may be delayed with respect to the sweep by a variable amount in order that the inch marks may be moved along the sweep. The third pulse is also delayed by a variable amount and determines the point at which the supersonic pulse generator is started.

The sweep amplifier provides push-pull horizontal deflection for the cathode-ray tube. The time of travel of the spot across the screen may be varied from 10 microseconds to 2000 microseconds approximately. Sweep expansion is also

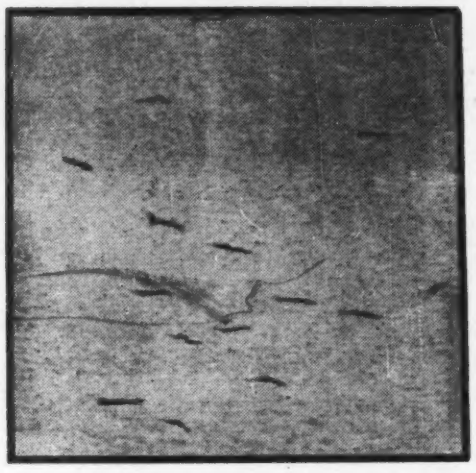


FIG. 2—Macroetch section of a forging, showing defects originally found by supersonic testing

provided to obtain good resolution when inspecting over long distances.

The distance marker generates square waves in which the fundamental frequency is continuously variable from 5 kc to 130 kc. Therefore, one square-wave cycle can be made to represent any desired length from one to more than twenty-five inches of aluminum. The half-cycle points are made to divide the length equally.

The pulse generator provides a 1000-volt pulse and is continuously variable in overlapping ranges of frequency from 0.5 mc to 12 mc. The pulse length may also be varied from one to several microseconds.

Transducers are Quartz Crystals

For longitudinal waves, an X-cut piezoelectric quartz crystal is used as an electromechanical transducer and is normally operated near its thickness resonance to obtain maximum sensitivity. In this case, the crystal is one-half wavelength thick. It is clamped to damp out free vibrations rapidly. This same crystal (search unit) serves as a transducer for the transmitted pulse as well as for the received echoes. A metallic coating is provided on the back side of the crystal. For materials which are not good electrical conductors, an additional metallic coating must also be placed on the side that comes in contact with the material under investigation. Normally, however, the material is the other electrode.

To obtain good supersonic coupling, a film of liquid, usually oil, is applied between the crystal and the work. When inspecting cylindrical surfaces, the crystal is ground to fit the piece being tested; in this case, both surfaces of the search unit must be curved in order to make the crystal a good transducer.

The amplifier is actually a superheterodyne receiver, all the reflected pulses being converted to one frequency before being amplified. Its bandwidth is variable from a fraction of a megacycle to several megacycles. Since sound absorption and the minimum size of the defect to be located vary with each application, the amplifier gain is variable. The indicator is a nine-inch cathode-ray tube.

Although single-crystal operation is usually most desirable, it is also possible to use one crystal as the transmitter and one as the receiver, as shown in the block diagram.

Design Problems

The requirement that the final instrument be as universal in use as possible presented certain problems. To test many materials of sizes ranging from a few inches to many feet, it was necessary to have a variable time scale as well as variable pulse length, sweep length, etc. The different grain sizes of various materials, the varying wavelength of sound in these materials, and the necessity of finding defects of a different size for each testing application also made it mandatory to provide a wide range of testing frequencies for general industrial use.

Fatigue cracks in metal pieces tend to develop near shoulders and keyways. Since these shoulders often give reflections, high resolution was desired in their vicinity and sweep expansion was required. The exponential type of sweep, which is the easiest sweep to generate, was also deemed advisable. From an operational point of view, it provides high resolution near the testing surface and yet, since the end reflection can also be seen, this type of sweep enables the operator to be certain of good supersonic wave transmission. It is necessary to adjust only 6 knobs to change the testing frequency. For ordinary testing, an operator can be trained in two weeks.

The necessity for rapid amplifier recovery immediately after the pulse was a difficult problem to solve. For example, in the case of a flaw $\frac{1}{8}$th inch below the surface of a piece of aluminum, only one microsecond elapses between the time the beginning of the pulse is applied to a crystal and the time when it is reflected back by the flaw. The instrument pictured herein can find small flaws $\frac{1}{2}$ inch below the surface, and special techniques permit even closer testing.

A large cathode-ray tube was provided so that the person manipulating the search unit could see the indications directly. The spot trace

FIG. 3—Checking axle of Tournapull Super Model C tractor with Supersonic Reflectoscope

can be seen easily if direct light does not strike the screen, as the tube has about 3000 volts on its final anode.

For industrial use, the machine had to be easily wheelable and require only a small amount of electrical power. When built with a magnesium case and carriage the complete unit weighs 225 pounds and consumes less than 200 watts of power.

Applications

Present applications mainly involve testing of metals, although

FIG. 4—Etched section of defective axle, showing flaws revealed by supersonic testing

both liquids and plastics will conduct supersonic waves at these frequencies and could be tested or measured by this instrument. The surface to be tested should be relatively smooth. If not already machined, it can be smoothed with a portable hand grinder. A finish equivalent to the GE number f5, which has an average roughness of 125 microinches or a peak-to-valley value of 450 microinches, is satisfactory; however, the maximum allowable surface roughness depends on the minimum size of defect it is desired to detect.

The average maximum distance which can be shot depends upon the material being tested. Steel and aluminum are quite homogenous and permit relatively deep penetration, while metals such as lead and copper are relatively hard to penetrate. Large grain size and the presence of small discontinuities make it hard to shoot long distances in any material when these discontinuities are of the same magnitude or larger than the supersonic wavelength being used.

The size of the indication on the scope is only roughly proportional to the size of the defect, but by moving the search unit over the testing surface and noting the indication on the cathode-ray tube the area of a large flaw may be plotted. For small flaws the testing frequency may be decreased in steps until the supersonic wavelength approaches the major dimension of the defect, whereupon its indication disappears, thus indicating the approximate size of small flaws. In some cases considerable information as to the type of defect may be determined from the shape of the indication.

A great many types of defects have been found with this instrument, such as hydrogen ruptures, coring, inclusions, fatigue cracks, shrinkage cracks, stringers, piping, segregations, and laminations. Relative grain size and thickness of material can also be measured. The latter measurement is especially valuable when one side is inaccessible. Defective bonds between like as well as dissimilar metals can be found, and the soundness of welds may be determined, except in the case of paste welds.

TABLE 2. **Properties of Materials and Wavelengths of Supersonic Vibration in Materials at Three Frequencies**

Material or Medium	V_l cm per sec	Wavelength in cm at Freq. Indicated			Density grams per cu cm	Specific Acoustic Impedance (density x velocity)
		1 mc	2¼ mc	5 mc		
Steel	581×10^3	0.581	0.259	0.116	7.8	4.54×10^6
Aluminum	622×10^3	0.622	0.277	0.124	2.70	1.68×10^6
Brass	443×10^3	0.443	0.196	0.0886	8.4	3.72×10^6
Copper	462×10^3	0.462	0.206	0.0824	8.89	4.10×10^6
Lead	213×10^3	0.213	0.0947	0.0426	11.3	2.41×10^6
Mercury	142×10^3	0.142	0.0632	0.0284	13.6	1.93×10^6
Glass	520×10^3	0.520	0.231	0.104	2.5	1.30×10^6
Quartz	575×10^3	0.575	0.255	0.115	2.65	1.52×10^6
Bakelite	259×10^3	0.259	0.115	0.0518	1.4	0.363×10^6
Polystyrene	267×10^3	0.267	0.119	0.0534	1.1	0.294×10^6
Transformer Oil	139×10^3	0.139	0.0618	0.0278	0.92	0.128×10^6
Water	145×10^3	0.145	0.0645	0.0290	1.00	0.145×10^6
Air	34.4×10^3	0.0344	0.0153	0.00688	0.00120	41.3

Examples of Performance

An etched section of a steel billet containing defects found by nondestructive supersonic testing is shown in Fig. 2. The short, dark semihorizontal lines are the defects which were found before the sample was taken and etched. This type of a defect is known as a hydrogen rupture or flake, and is difficult to find by any other method except that of sectioning and etching.

An on-the-job test on the axle of a tractor is illustrated in Fig. 3. Many of these axles which had been assembled prior to the availability of the supersonic tester were later checked in this manner. There are many similar applications where fatigue cracks can be located in assembled equipment when only the end of the shaft or axle is accessible.

An etched section from a forged axle rejected by supersonic testing is shown in Fig. 4. The defect can be seen as longitudinal segregations with transverse cracks at either end. Figure 5 shows a series of reflectograms (pictures of cathode-ray tube beam traces) of the axle for different positions of the searching unit. The supersonic pulse reflections obtained are identified by the letters P_1, P_2, etc. The cause of these reflections and the corresponding positions of the searching unit on the axle are identified in Fig. 6. The reflection P_1 was due to a $\frac{1}{16}$-inch diameter, 2-inch deep hole drilled in

a test axle for comparison purposes. Note the almost complete scattering and absorption of the beam obtained when shooting along the axis of the axle. Also note the reflected pulses P_4 and P_6 obtained from the transverse defects and the diminution of the end reflection P_3. The search unit crystal was in this case one inch square, hence a supersonic wave train having approximately the same cross-section was formed. Parts of this wave train were reflected by various discontinuities in the axle, causing multiple reflections on the cathode-ray tube sweep.

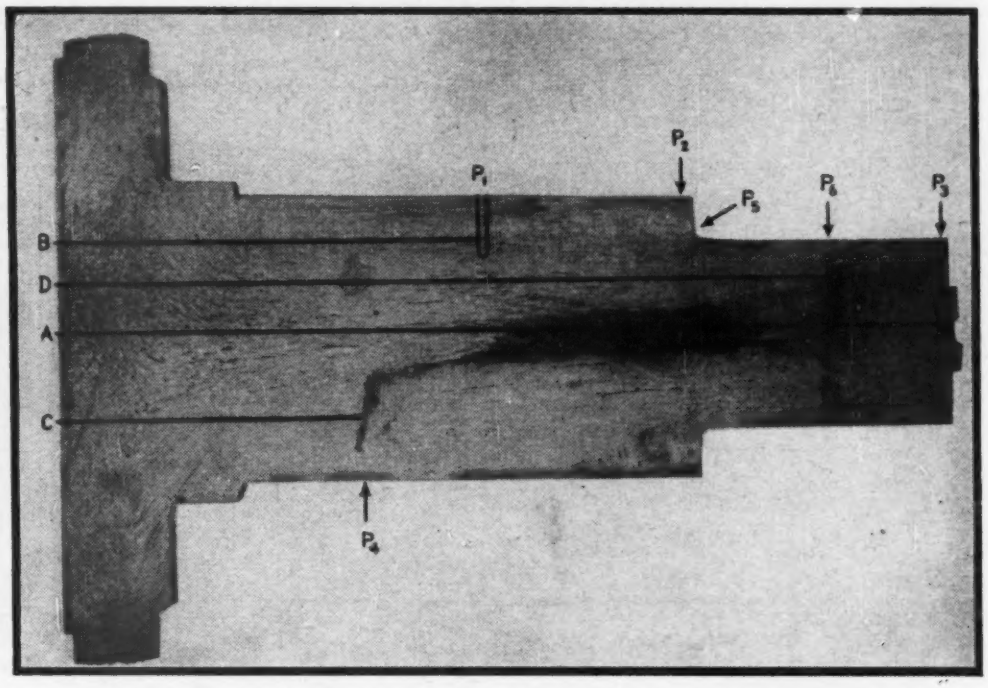


FIG. 6—Identification of scanning lines and reflections in Fig. 5. Crystal transducer was applied in turn to points A, B, C, and D on exposed end of axle

Advantages of Supersonic Testing

The supersonic method of testing does not necessarily require a change in density for its indications, due to the inherent properties of supersonic waves. Since it depends on the reflected energy caused by an area of discontinuity, it enables the operator to detect flaws of infinitely small depth regardless of the thickness of the part being tested.

The minimum area of the discontinuity required to give an indication depends on the distance of the discontinuity from the surface being tested and the homogeneity of the material. In aluminum, for example, a void with surface dimensions of 0.125 x 0.002 inch will give a reflection when this discontinuity is 2 inches from the searching unit, and a $\frac{5}{32}$-inch diameter defect will give a reflection when the flaw is 10 feet from the crystal.

This method of testing gives results immediately to the operator as he tests the part, and the speed of testing does not depend on the thickness of the piece undergoing inspection. In the average application only a small amount of operator interpretation is required and the defect indication is plainly visible.

The fact that axles, shafts, etc. may be inspected for fatigue cracks while still assembled, provided one end is accessible, is an obvious advantage in industrial use. It is not necessary that the flaw extend to the surface, nor is it necessary to remove wheels, etc. from the axle or shaft. Formerly, fatigue cracks, even though they extended to the surface, could be found only by disassembling the part and examining it closely. In the case of some crank pins which have wheels pressed on with great force, the disassembling method ruins the pin.

Such advantages, combined with high accuracy and ease of operation, make supersonic testing an important addition in the field of nondestructive testing.

The instrument described was invented by Dr. Floyd A. Firestone. The technical development of this model was under his direction and that of H. C. Drake, director of research at Sperry Products, Inc.

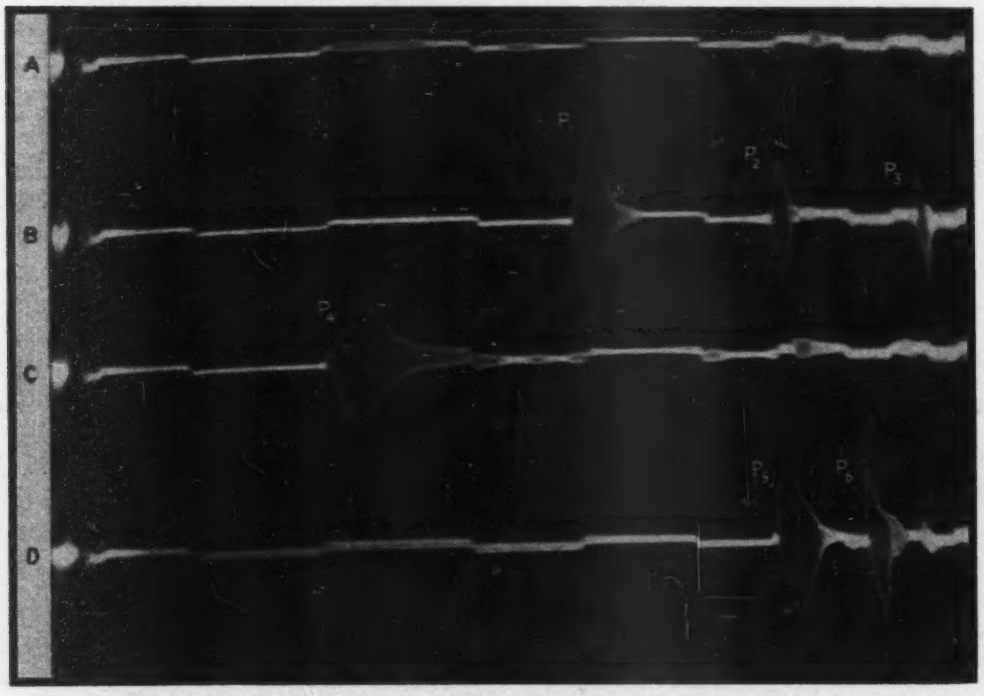


FIG. 5—Reflectograms of axle shown in Fig. 4, as photographed on screen of cathode-ray tube. Operators can be thoroughly trained in two weeks in use of supersonic tester and interpretation of traces like these in terms of flaws

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(2) Wood, R. W., Supersonics—The Science of Inaudible Sound, Brown University, Providence, R. I., 1939.

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	TYPE	MODE	CONFIGURATION	CUTOFF FREQUENCY f_c IN MC PER SEC	ATTENUATION α IN DB PER FOOT OF AXIAL LENGTH	$\frac{f_{MIN}}{f_c}$	α_{MIN} IN DB PER FT OF AXIAL LENGTH
RECTANGULAR	H WAVES	$H_{n,m}$ n≠0	$a \geq \frac{1}{2}\lambda_c n$ $b \geq \frac{1}{2}\lambda_c m$ $H_{2,1}$	$5905.5\sqrt{\left(\frac{n}{a''}\right)^2+\left(\frac{m}{b''}\right)^2}$	$0.01119\frac{\sqrt{m}}{\sqrt{(b'')^3}}\cdot\frac{\left\{\frac{b}{a}+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^3\right\}\left(\frac{f}{f_c}\right)^{\frac{3}{2}}+\left\{1+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^3\right\}\left(\frac{f}{f_c}\right)^{-\frac{1}{2}}}{\left\{1+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^3\right\}^{\frac{3}{4}}\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$	$\sqrt{\frac{6}{2}\sqrt{1+\left(\frac{k_2}{k_1}\right)+\sqrt{\left(\frac{k_2}{k_1}\right)^2+\frac{14}{9}\left(\frac{k_2}{k_1}\right)+1}}}$ $\frac{k_2}{k_1}=\frac{\frac{a}{b}+\frac{\left(\frac{n}{m}\right)^2}{\left(\frac{a}{b}\right)^2}}{1+\frac{\left(\frac{n}{m}\right)^2}{\left(\frac{a}{b}\right)}}$ $\frac{k_2}{k_1}=2\frac{a}{b}$	α_{MIN} IS TO BE DETERMINED SEPARATELY FOR EACH CASE
		$H_{0,m}$ n=0	$H_{0,1}$ LOWEST CUTOFF FREQ	$\frac{5905.5}{b''}$	$0.01119\frac{\sqrt{m}}{\sqrt{(b'')^3}}\cdot\frac{\frac{b}{2a}\left(\frac{f}{f_c}\right)^{\frac{3}{2}}+\left(\frac{f}{f_c}\right)^{-\frac{1}{2}}}{\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$ LOWEST ATTENUATION		
	E WAVES	$E_{n,m}$	$a \geq \frac{1}{2}\lambda_c n$ $b \geq \frac{1}{2}\lambda_c m$ $E_{1,1}$	$5905.5\sqrt{\left(\frac{n}{a''}\right)^2+\left(\frac{m}{b''}\right)^2}$	$0.01119\frac{\sqrt{m}}{\sqrt{(b'')^3}}\cdot\frac{1+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^3}{\left\{1+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^2\right\}^{\frac{3}{4}}}\cdot\frac{\left(\frac{f}{f_c}\right)^{\frac{3}{2}}}{\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$	$\sqrt{3}=1.732$	$0.01804\frac{\sqrt{m}}{\sqrt{(b'')^3}}\cdot\frac{1+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^3}{\left\{1+\left(\frac{n}{m}\right)^2\left(\frac{b}{a}\right)^2\right\}^{\frac{3}{4}}}$
CIRCULAR	H WAVES	H_0	H_0	$\frac{7167.7}{a''}$	$\frac{0.0061769}{\sqrt{(a'')^3}}\cdot\frac{\left(\frac{f}{f_c}\right)^{-\frac{1}{2}}}{\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$	∞	THIS WAVE IS UNSTABLE AND CAN BE USED ONLY IN STRAIGHT TUBING
		H_1	H_1	$\frac{3458.8}{a''}$ LOWEST CUTOFF FREQ	$\frac{0.0042746}{\sqrt{(a'')^3}}\cdot\frac{0.42\left(\frac{f}{f_c}\right)^{\frac{3}{2}}+\left(\frac{f}{f_c}\right)^{-\frac{1}{2}}}{\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$	$\sqrt{\frac{6}{2}\sqrt{1+\left(\frac{k_2}{k_1}\right)+\sqrt{\left(\frac{k_2}{k_1}\right)^2+\frac{14}{9}\left(\frac{k_2}{k_1}\right)+1}}}$	$\frac{0.0041848}{\sqrt{(a'')^3}}$ LOWEST MINIMUM ATTENUATION
	E WAVES	$E_{0,1}$	E_0	$\frac{4520.85}{a''}\rightarrow\frac{4521}{a''}$	$\frac{0.0049}{\sqrt{(a'')^3}}\cdot\frac{\left(\frac{f}{f_c}\right)^{\frac{3}{2}}}{\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$	$\sqrt{3}=1.732$	$\frac{0.0078953}{\sqrt{(a'')^3}}$
		$E_{1,1}$	E_1	$\frac{7167.7}{a''}$	$\frac{0.0061769}{\sqrt{(a'')^3}}\cdot\frac{\left(\frac{f}{f_c}\right)^{\frac{3}{2}}}{\sqrt{\left(\frac{f}{f_c}\right)^2-1}}$	$\sqrt{3}=1.732$	$\frac{0.00995476}{\sqrt{(a'')^3}}=\frac{0.01}{\sqrt{(a'')^3}}$

TABLE I. DESIGN FACTORS FOR COPPER-WALLED AIR-CORED WAVEGUIDES

Minimum Attenuation in Waveguides

Tabulation of design equations covering attenuation, development of simplified versions applying to specific practical air-core guides most often encountered, and graphical construction for determining conditions for minimum attenuation at exciting frequency

By EDWIN N. PHILLIPS
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IN recent literature dealing with the phenomena of signal transmission through air-cored waveguides, the conditions for propagation have been adequately dealt with, but the conditioning of the cross-sectional dimensions to obtain minimum attenuation at the exciting frequency has generally been dismissed with but a brief treatment. Since this aspect is of direct concern to the designer using waveguides, this paper presents the necessary formulas for predicting cross-sections giving minimum attenuation.

Practical formulas concerning attenuation in air-cored guides are listed by Sarbacher and Edson in their book, "Hyper- and Ultra-High-Frequency Engineering." These formulas, as converted into English units, appear in Table I. For material other than copper, multiply the value of attenuation by $\sqrt{\rho/\rho_c\mu}$, where ρ is the specific resistivity of the guide material, ρ_c is the specific resistivity of copper, and μ is the permeability of the guide material. It will be noted that the formulas for the $H_{0,m}$ and the $H_{n,m}$ excitation modes in rectangular cross-section guides and the H_1 mode in circular cross-section guides can be expressed in the form

$$\alpha = k_1\frac{\left(\frac{f}{f_c}\right)^{1.5}}{\left[\left(\frac{f}{f_c}\right)^2-1\right]^{0.5}}+k_2\frac{\left(\frac{f}{f_c}\right)^{-0.5}}{\left[\left(\frac{f}{f_c}\right)^2-1\right]^{0.5}} \qquad (1)$$

where α is in db per foot and k_1 and

It is to be understood that the opinions or assertions contained in this paper are the private views of the author, and are not to be construed as being official, nor as reflecting the views of the Navy Department or of the Naval Service at large.

k_2 are constants dependent on the material and the dimensions of the guide.

For copper guides, the ratio k_2/k_1 for various excitation modes is

$H_{0,m}$ . . . rectangular cross-section . . . . . $2\,a/b$
H_1 . . . circular cross-section 2.38
$H_{n,m}$ . . . rectangular cross-section . . . . . .

$$\frac{\left(\frac{a}{b}\right)+\frac{\left(\frac{n}{m}\right)^2}{\left(\frac{a}{b}\right)^2}}{1+\frac{\left(\frac{n}{m}\right)^2}{\left(\frac{a}{b}\right)}}$$

where a is the thickness of the rectangular cross-sectioned guide in inches or the radius of the circular guide, b is the width of the rectangular cross-sectioned guide, and n and m are subscripts defining the order of the wave propagated within the guide.

By differentiating Eq. 1 with respect to the frequency ratio f/f_c, where f represents the exciting frequency and f_c the cutoff frequency below which the guide will not transmit electromagnetic waves, and setting the resulting expression equal to zero, the following expression for f_{min}, the frequency of minimum attenuation, is obtained

$$\frac{f_{min}}{f_c} = \frac{\sqrt{6}}{2}\sqrt{1+\frac{K_2}{K_1}+\sqrt{\left(\frac{K_2}{K_1}\right)+\frac{14}{9}\left(\frac{K_2}{K_1}\right)+1}} \qquad (2)$$

Because an approximate solution to this equation may be sufficiently accurate in certain cases, and to provide a rough check on computations, the equation has been plotted as Fig. 1.

Because of the relative ease of excitation and the low overall attenuation, the $H_{0,m}$ excitation mode is often used. For this case, Eq. 2 simplifies to

$$\frac{f_{min}}{f_c} = \frac{\sqrt{6}}{2}\sqrt{1+2\frac{a}{b}+\sqrt{4\left(\frac{a}{b}\right)^2+\frac{28}{9}\left(\frac{a}{b}\right)+1}} \qquad (3)$$

While Eq. 2 and 3 are useful for fixed guide dimensions and variable input frequency, often the input frequency is fixed and the guide dimensions must be varied to secure minimum attenuation. Manipulation of Eq. 2 and 3 to isolate the dimensional factor gives for the general case

$$\frac{k_2}{k_1}=\frac{\left(\frac{f}{f_c}\right)^4-3\left(\frac{f}{f_c}\right)^2}{3\left(\frac{f}{f_c}\right)^2-1} \qquad (4)$$

and for the $H_{0,m}$ mode

$$\frac{a}{b}=\frac{\left(\frac{f}{f_c}\right)^4-3\left(\frac{f}{f_c}\right)^2}{6\left(\frac{f}{f_c}\right)^2-2} \qquad (5)$$

In all these formulas, the condition for minimum attenuation is independent of the guide material, and dependent only on the cross-sectional dimensions of the guide and the order of the transmitted wave.

Formulas for *E* Modes

Treating the formulas for the $E_{n,m}$ excitation mode in rectangular cross-sectioned guides and E excitation modes in circular cross-sectioned guides in a similar manner, the general formula is

$$\alpha = k \cdot \frac{\left(\frac{f}{f_c}\right)^{1.5}}{\left(\left(\frac{f}{f_c}\right)-1\right)^{0.5}} \text{ db per ft} \qquad (6)$$

Setting the first derivative equal to zero gives

$$f_{min}/f_c = \sqrt{3} \qquad (7)$$

Note that here the condition for minimum attenuation is independent of both the guide material and the guide dimensions. It will further be noted that Eq. 7 is the limit of Eq. 2 obtained when k_2/k_1 equals zero, indicating in the $H_{0,m}$ mode a zero guide thickness.

Thus, when the thickness a equals the width b of the guide and it is excited in the $H_{0,m}$ mode, Eq. 3 gives f/f_c equal to 2.96; for a cutoff frequency of 1.5 x 10^9 cps, the frequency of minimum attenuation is 4.44 x 10^9 cps. For a guide of the same cross-section excited in the $H_{1,1}$ mode, k_2/k_1 will be unity, and Eq. 2 gives 2.416 for f/f_c; for a cutoff frequency of 2.12 x 10^9 cps, the frequency of minimum attenuation will be 5.12 x 10^9 cps. For the same cutoff frequency, Eq. 7 gives 3.67 x 10^9 cps as the frequency of minimum attenuation in a rectangular guide which is excited in transverse magnetic mode.

Approximate Plots

By means of Eq. 2, 3, and 7, approximate plots of the attenuation-vs-frequency relationship for air-cored waveguides can easily be made. The cutoff frequency f_c forms a vertical asymptote on a log-log grid, the point of minimum attenuation forms a horizontal asymptote, and since at a frequency much higher than that giving minimum attenuation the guide behaves as any metallic conductor whose attenuation varies as the square root of the frequency, a third boundary to the curve is formed. To illustrate this, Fig. 2 gives the attenuation curve of a brass guide (square root of the resistivity-ratio of brass to copper is 2.015) with inside dimensions of 1.34 x 2.84 inches. This data combines with the formulas in Table I to give, for $H_{0,1}$ excitation, f_c = 2,084 mc, f_{min} = 4,960 mc, α_{min} = 9.42 db per thousand feet, and α = 110 db per thousand feet at $f = 10^{13}$ cps. The resultant curve closely approximates a point-by-

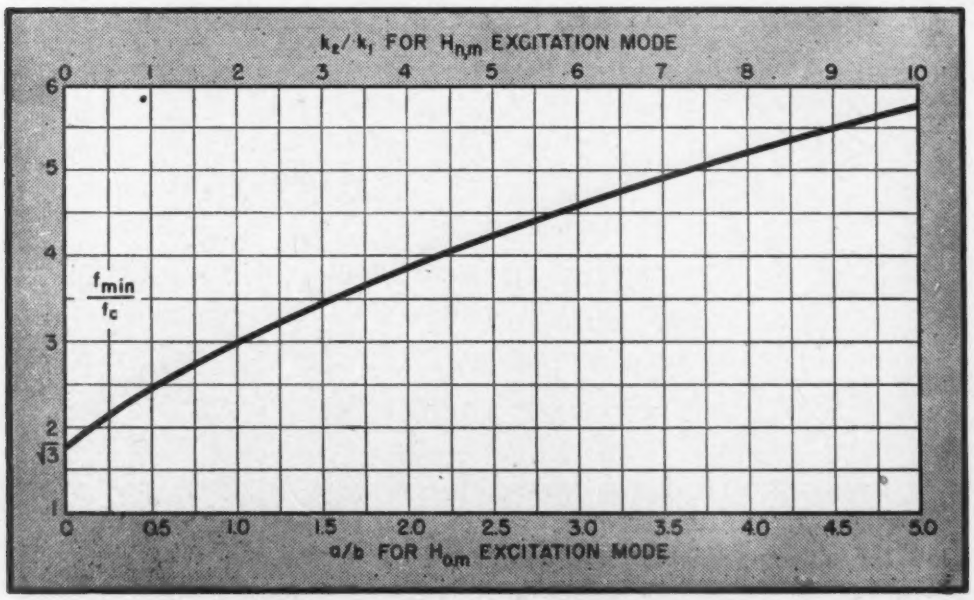

FIG. 1—Minimum attenuation in air-cored waveguide in terms of dimensional parameters

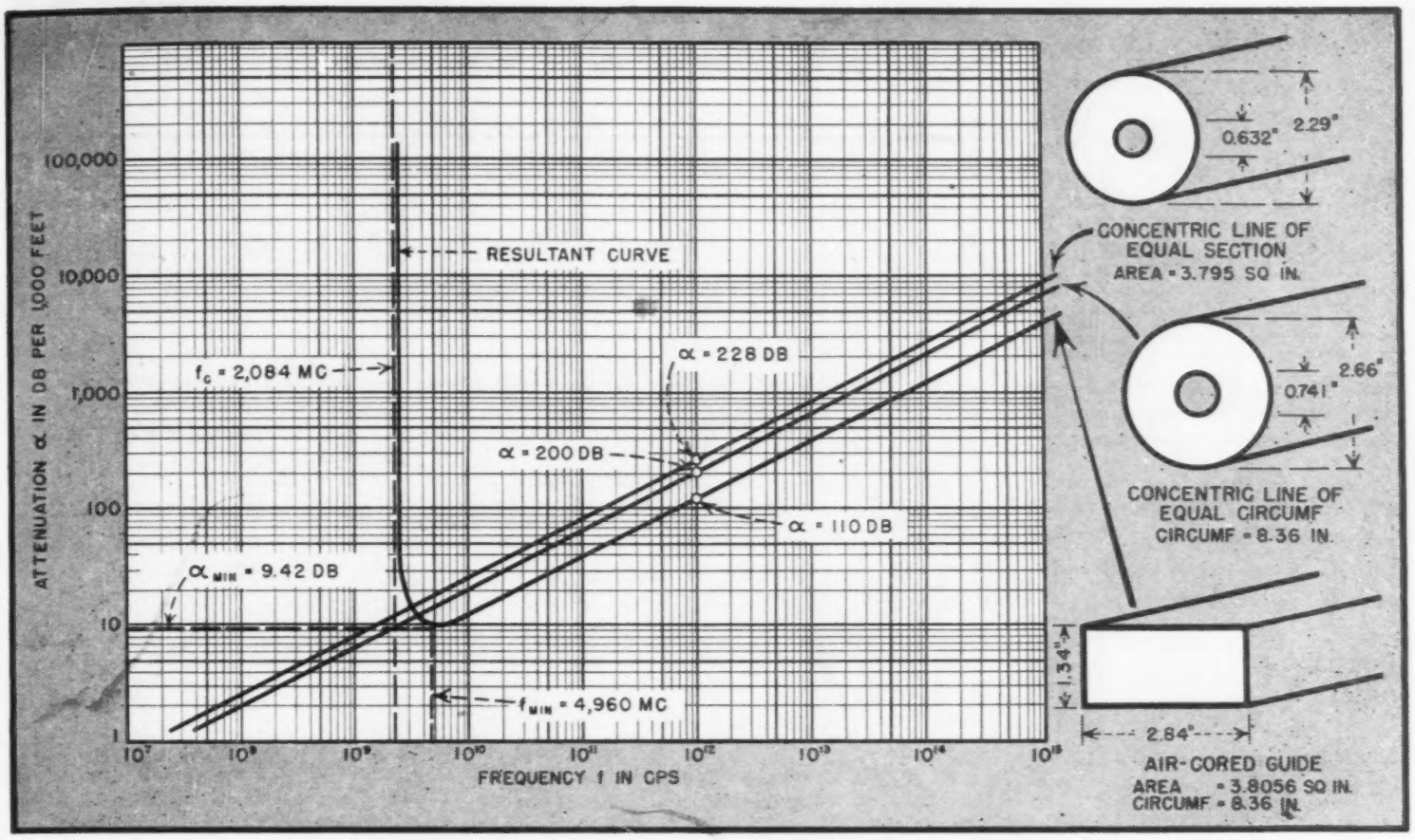

FIG. 2—Graphical construction of attenuation curve for 1.34 by 2.84-inch rectangular waveguide, and corresponding curves for two types of concenctric lines

point plot, and saves much time.

Once the attenuation per unit length and the guide length are known, the transmission-line efficiency in percent is obtained from

$$\eta = 100/\left(\log^{-1}\frac{\alpha l}{10}\right) \qquad (8)$$

where l is line length in feet. A plot of this is shown in Fig. 3.

Also plotted in Fig. 2, as a matter of pertinent interest, are the attenuation curves of a zero-conductance brass concentric line for which $D/d = 3.59$, with the same inner periphery as the guide ($\alpha = 200$ db per thousand feet at 10^{12} cps), and a zero-conductance brass concentric line for which $D/d = 3.59$, with the same dielectric area as the guide cross-section ($\alpha = 228$ db per thousand feet at 10^{12} cps). It will be noted that above the point of intersection of the guide curve with these curves, the guide has less attenuation than concentric lines.

Summary

Pertinent points to be mentioned as general guides in the use of waveguides of this type are (1) an infinite number of cross-sections will give minimum attenuation at any prescribed frequency, but the cross-sectional dimensions are not independent of each other and both depend on the cutoff frequency chosen; (2) the lower the cutoff frequency chosen, the lower will be the attenuation, the larger the guide dimensions will be, and the broader will be the bell of the curve at the minimum point in the H propagational modes, giving less distortion in the output of a complex wave or a modulated carrier.

Concerning this last point, most of the present-day applications use pulsed modulation, and it can be shown that when the ratio of the carrier (or pulsed) frequency to the pulsing frequency is great, the sidebands are negligible, and attenuation can be computed for the carrier alone.

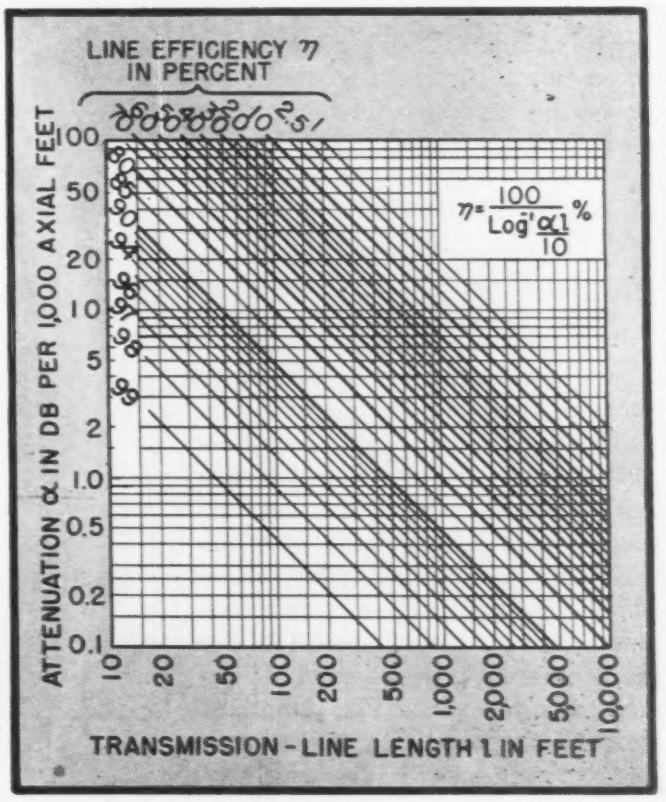

FIG. 3—Transmission-line efficiency in terms of cable length and attenuation per unit length

PICKUP WITH LOW

By HENRY P. KALMUS

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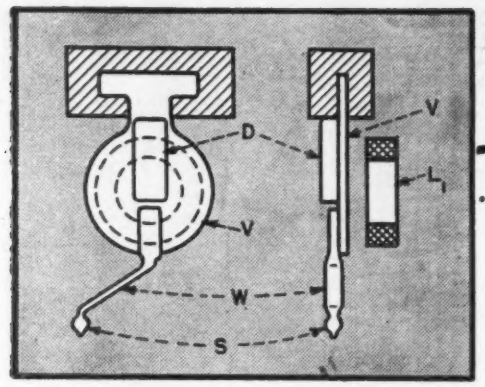


FIG. 1—Side and front views of pickup show the essential mechanical elements and their relation to the oscillator coil

THE MECHANICAL IMPEDANCE of the pickup to be described is so low that not more than 14 grams weight is necessary for tracking, thus giving long record life. Distortion caused by eccentricity of records is approximately 0.5 db. Little mechanical noise is radiated from the pickup itself, giving quiet operation for open record players. The complete reproducer has an inherent high-frequency cutoff at 4,000 cps, which is desirable for high signal-to-noise ratio on commercial shellac pressings. Mechanical proportions of the pickup are shown in Fig. 1.

The pickup stylus moves a high-resistance vane in relation to the inductance of the resonant circuit in a radio-frequency oscillator. This relative motion varies the mutual inductance between vane and coil, thereby changing the resistance reflected into the coil. This action produces amplitude modulation of the oscillator by varying the loss of its resonant circuit. The amplitude modulation is detected to give the audio-frequency output from the electronic reproducer, the circuit of which is shown in Fig. 2.

Mechanical Characteristics of Pickup

The moving element shown in Fig. 1 consists of three parts: the vane V made of resistance, nonmagnetic, stainless steel with an essentially circular form 0.004 inch thick and $\frac{1}{8}$ inch diameter; the compliant member W of wire, 0.02 inch diameter, and flattened to 0.008 inch for vertical compliance, which couples the stylus S to the vane; and a piece of damping material D cemented to the vane. The compliant member is spotwelded to the vane. The stylus is fused to the wire.

The compliance, C_M measured at the tip is 9.6 x 10^{-6} cm per dyne. The mass is distributed over the total length of the moving element. In Fig. 3A a lumped mass M concentrated at the needle tip is assumed. This mass M is 14 x 10^{-3} grams. If we neglect friction, the mechanical impedance for 1000 cycles equals

$$Z = \omega M - 1/\omega C_M$$
$$= (2\pi \times 10^3)\ (14 \times 10^{-3}) - 1/(2\pi \times 10^3)\ (9.6 \times 10^{-6})$$
$$= 71.4 \text{ grams per sec}$$

This is a very small value because of the high compliance and small mass of the moving element. A vertical weight of only 10 grams is necessary to keep the needle in the groove, and needle and record wear are extremely low. In addition, the mechanical impedance and the vertical compliance produced by the member W reduce the noise radiated from pickup and record to such a degree that it is scarcely noticeable even when no cover is used for the record player.

The mechanical impedance becomes zero for

$$\omega = (MC_M)^{-1/2}$$
$$= [(14 \times 10^{-3})\ (9.6 \times 10^{-6})]^{-1/2}$$
$$= 2700 \text{ rad per sec}$$
$$f = \omega/2\pi = 430 \text{ cps}$$

In Fig. 3B a lumped mass is also assumed, but concentrated in the center of the moving element. At a certain frequency the element will be excited in its second mode of vibration. The amplitudes at the tip are then very small, but the center moves with large deviation as shown in Fig. 3B, and at this frequency a high output is obtained. The mechanical constants of the pickup are chosen in such a way that this peak in the response curve is at 4000 cycles. The output drops sharply beyond this frequency. Damping material D is used to keep the resonance peak at a desired level.

Circuit Operation

Figure 4 shows the frequency-response curve, using the Columbia constant-tone record 10003M. The sharp cutoff beyond 4000 cycles is desirable for reproduction of commercial shellac records with minimum scratch. By proper choice of

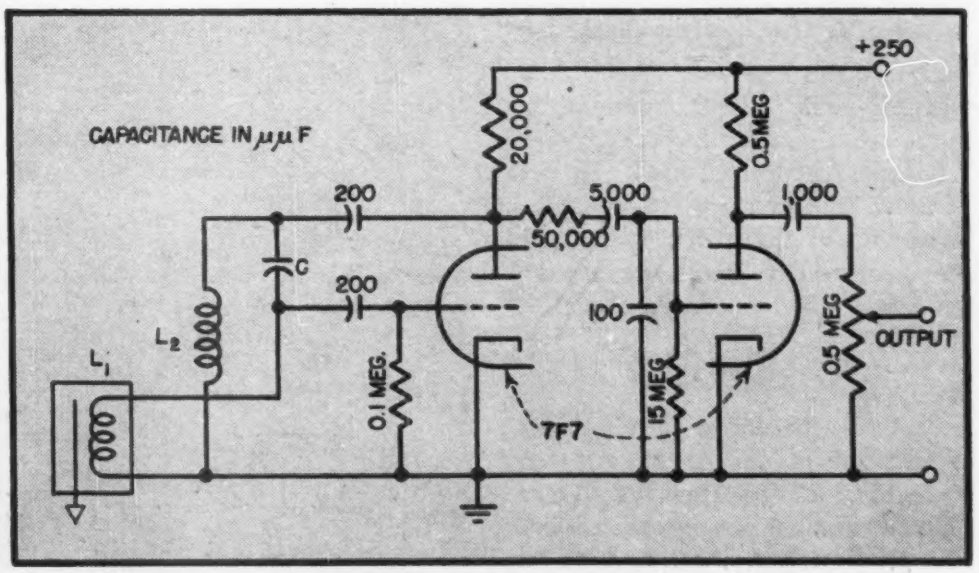
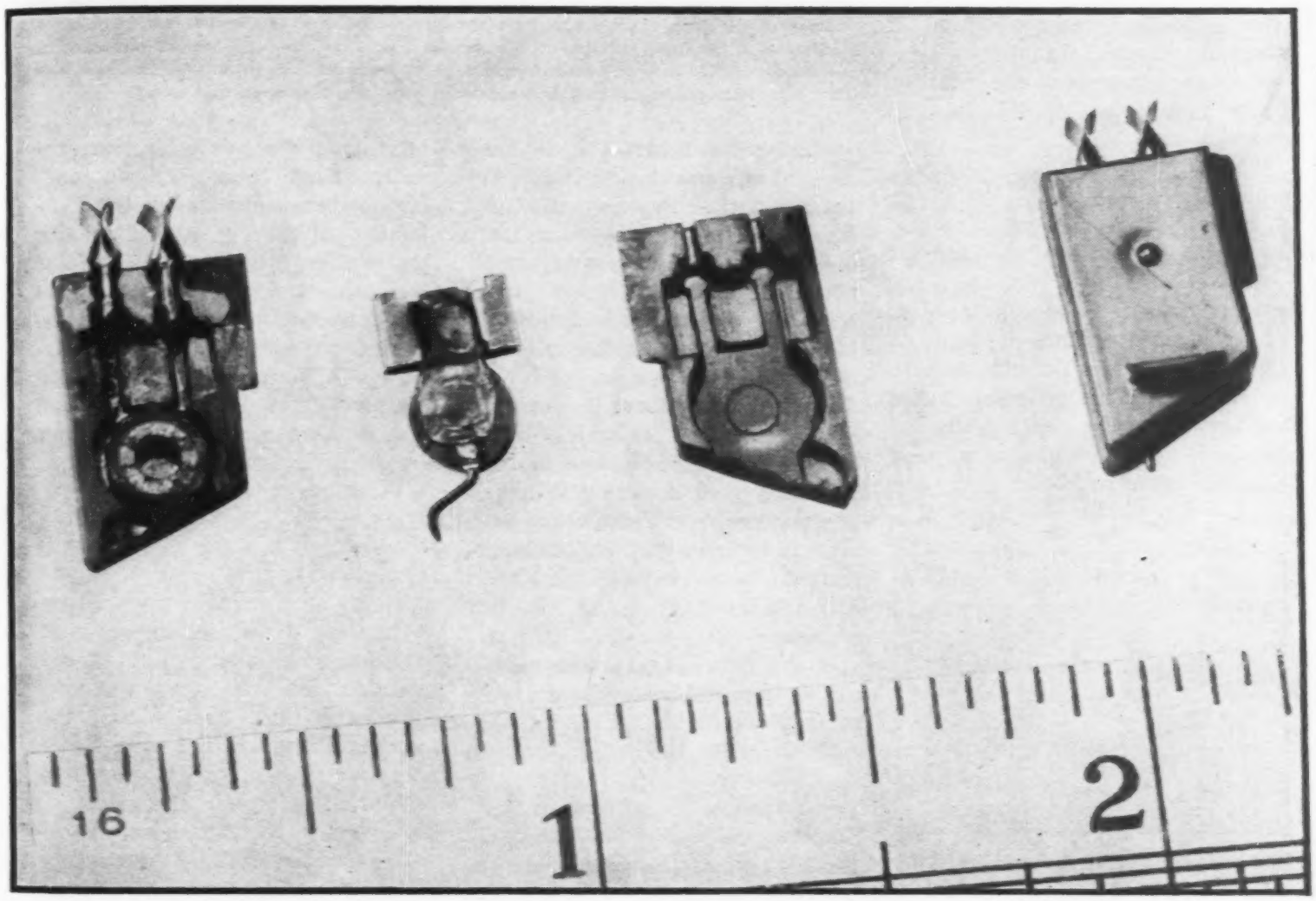


FIG. 2—Stylus movement amplitude-modulates an oscillating detector-amplifier from which audio output is obtained

MECHANICAL IMPEDANCE

Metallic vane attached to stylus varies loss in radio-frequency oscillator circuit, thereby producing amplitude modulation. High compliance and small mass of moving element results in low mechanical impedance, thus reducing pickup weight required for tracking



Pickup cartridge, shown about three times normal size, carries stylus, vane, and coil, and plugs into a tone arm

mass and compliance the cutoff can be shifted between 3000 and 6000 cps.

Figure 2 shows the circuit in which the pickup is used. The two triodes are the two sections of a 7F7 or a 6SL7. The pickup coil L_1 together with a fixed coil L_2 and the capacitor C form a tuned circuit which is excited by the first triode. The reflected vane resistance changes the Q of L_1, thereby varying the amplitude of the oscillations.

The first triode section works simultaneously as r-f oscillator, detector, and audio amplifier. The oscillation frequency is 2.5 mc. The r-f amplitude at the grid is 1.5 volts, but the modulation is less than one percent so that an audio voltage of only 10 millivolts is developed at the first grid. The audio voltage at the plate is 30 millivolts. The audio gain is reduced to less than it could be were the triode used solely as an amplifier because of the action of plate-bend detection. A filter between the first plate and second grid eliminates r-f, passing only audio voltage. The attenuation of the low audio frequencies shown in Fig. 4 is obtained by the coupling elements between the two triodes of the dual tube and by an additional low-pass filter between second plate and volume control.

The second triode section acts as an additional audio amplifier and supplies one volt to the phono-amplifier.

The coil L_1 consists of about 40 turns of No. 40 copper wire, and has an inductance of 20 microhenries. Its impedance is so low that a simple twisted pair can be used to connect the coil to the first tube. In this respect, a pickup using a coil has an advantage over a variable capacitance type which requires a well-shielded low-capacitance line.

Hum voltage induced in the coil L_1 by the motor field or other 60 cps sources does not reach grid or plate of the first triode because of

the filter action of the 200 $\mu\mu$f coupling capacitor and grid and plate resistors. However, care must be taken that the vane is not made of magnetic material. A 60 cps flux changes the permeabilty of the vane so that the hysteresis loss changes during each 60 cps wave. If the hysteresis loss changes, the resistance reflected into the coil changes also, and amplitude modulation at 60 or 120 cps is produced, depending on the initial magnetic condition of the vane material. Stainless steel is almost non-magnetic and thus no hum modulation occurs in this model.

Detection Efficiency

In the first triode section of Fig. 2, which acts simultaneously as oscillator and grid detector, the plate current produces random noise. It is therefore desirable to improve the detection efficiency in order to produce as much audio voltage across the plate resistor as possible so as to obtain a good signal-to-noise ratio.

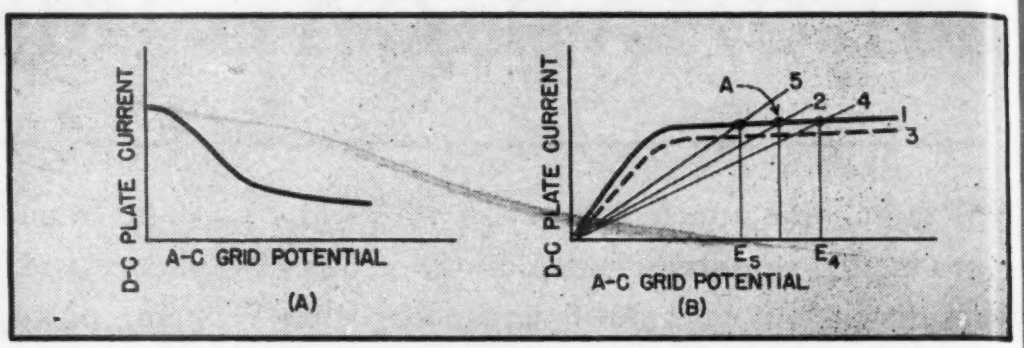


FIG. 5—(A) Conversion characteristics of grid-leak detector indicates the effect of the change of the type of detection from straight rectification to unequal amplification of positive and negative halves of the carrier cycle as the grid signal increases in amplitude. (B) From the characteristics of a feedback oscillator one can see how changes in the feedback efficiency vary the developed grid signal

Figure 5A shows the d-c plate current of a grid-leak detector plotted against a-c grid potential. For small voltages, the tube works like a diode, followed by an audio amplifier. For larger voltages a second form of rectification, plate-bend detection, counteracts the grid-leak detection and is responsible for the flattening of the curve.

It is therefore desirable to maintain small a-c grid voltages in order to obtain good detection efficiency. The percentage of modulation, however, is so small that negligible distortion occurs even for the largest grid voltages produced by this pickup.

As mentioned before, the first triode section develops an audio voltage of 30 millivolts which together with the plate current produces a signal-to-noise ratio sufficiently high for home record reproduction. However, if very high requirements are to be met, as when used in broadcast stations, there are two ways to increase the signal-to-noise ratio. By using a separate oscillator and a diode-detector stage, the detection efficiency is made much higher because no plate-bend detection occurs. The other method is to use a low μ tube such as a 6J5 as oscillator-detector, followed by an audio transformer. Plate-bend detection is reduced because of the remote cutoff characteristic and the higher voltage between plate and cathode. Again the result is higher detection efficiency.

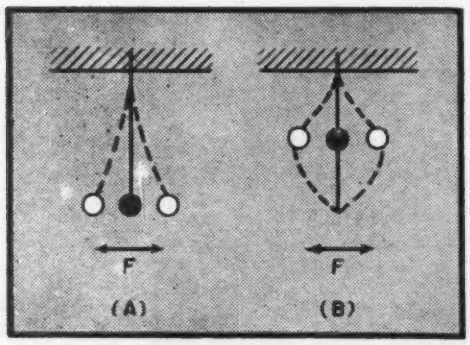


FIG. 3—Depending on the mode of vibration of the mechanical elements of the pickup, the mass can be considered either concentrated at the tip as at (A) or at the center as at (B)

Oscillator Stability

Returning now to Fig. 2, oscillator stability with aging tubes and changing line voltages must be investigated. It is usually assumed that low-amplitude oscillators are unstable. However, if a low-Q, low-impedance tank circuit, and the maximum feedback are used, stable operation is obtained with only 1 to 2 volts r-f voltage at the grid.

Curve 1 in Fig. 5B shows the oscillation characteristic of an amplifier tube. The curve is obtained by connecting the tank circuit to the plate and feeding variable r-f voltage into the grid The resultant r-f plate current is then plotted.

Curve 2 shows the grid voltage produced by any given r-f plate current. It is a straight line because the feedback circuit which produces the grid voltage is a linear network. The slope of curve 2 depends on the Q and the L/C ratio of the plate circuit as well as on the feedback factor.

It is evident that the point of intersection A determines the actual oscillation condition. Now, even though the plate circuit has a low impedance, curve 2 may intersect curve 1 on its flat part if the feedback factor is low enough. In this

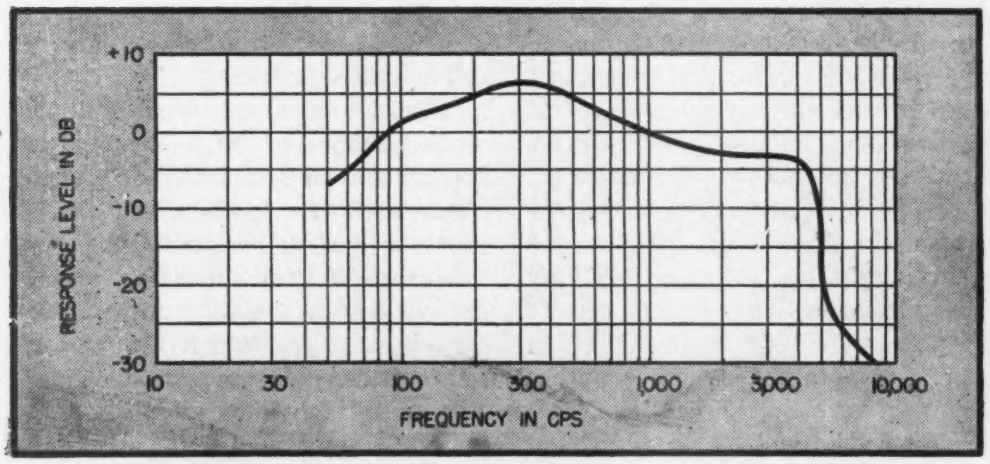


FIG. 4—Frequency response of the pickup, using a constant-tone record as the driving source

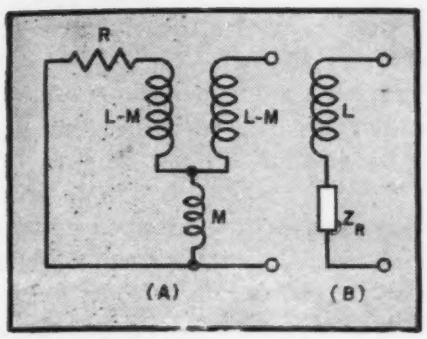


FIG. 6—(A) Equivalent circuit of pickup is that of a resistive coil coupled to the pickup inductance. (B) The equivalent circuit simplifies to an inductance and a variable impedance

case the r-f plate current has reached its maximum, although the produced voltage is still small. At the same time the oscillator tube operates with a transconductance approaching the maximum value for zero plate load. One or two volts on the grid are sufficient to cover the entire range from cutoff to where grid current starts flowing, and so provide the drive needed to reach the maximum amplitude of r-f current in the plate circuit.

If the voltage drops, or if the tube ages, curve 1 will assume a lower position such as curve 3. Curve 2 still intersects the flat part of the oscillation characteristic and stable oscillation is maintained.

As the vane swings, the Q of the circuit changes, and the slope of curve 2 is altered. Curve 2 will fluctuate between curves 4 and 5. Correspondingly, the grid voltage will be modulated between E_4 and E_5.

Linearity

Figure 6A shows the equivalent electrical circuit of the pickup. Series resistance R can be considered as belonging to a flat coil which could be substituted for the vane. The impedance reflected from the vane into the tuned circuit is Z_R as shown in Fig. 6B, and is equal to

$$Z_R = \frac{R(\omega M)^2}{R^2 + (\omega L)^2} - j\frac{-\omega^2 LM}{R^2 + (\omega L)^2}$$

(Do not confuse M used for mutual inductance and M used for mass.) The real part of the reflected impedance changes the Q of the oscillator circuit. The imaginary part shifts the oscillator frequency and need not be considered because no frequency discriminating elements are used in the circuit.

The mutual inductance between coil and vane increases if the distance is reduced. The reflected resistance grows with increasing mutual inductance. Finally, the oscillator amplitude decreases with increasing reflected resistance. The amplitude therefore changes with distance according to some function which is not necessarily linear. The linearity within the working range is sufficient to produce not more than two percent distortion.

Figure 7 shows an amplitude versus displacement characteristic, curve 1, plotted for a large scale model of a pickup with linear dimensions increased by a factor of 10. The point A corresponds to the minimum gap in the actual pickup. Points B and C determine the actual working range. They are 10 mils apart on the model, which corresponds to one mil on the pickup.

For a constant vane displacement, the audio output is proportional to the slope of curve 1 at the operating spacing. Curve 2 shows the first derivative of curve 1. The audio output at any spacing is proportional to the ordinate of curve 2 at that spacing. The pickup would be ideal if this curve were a straight horizontal line, because the audio output would then be independent of spacing.

As mentioned before, the distortion is very small. For certain applications, however, it may be of interest to eliminate it entirely, and this is possible by designing a pickup with push-pull action.

Push-Pull Pickup

Two identical coils are arranged at both sides of the high resistive vane as shown in Fig. 8. Both are tuned to the same frequency and the two circuits are coupled either by their stray field or by external means. We shall neglect changes in inductance produced by the moving vane. This assumption is justified because the vane can be made of high resistive material with some permeability. The loss of inductance due to eddy currents is then compensated by the concentration of the magnetic flux in the vane.

The resistances R_1 and R_2 are now the impedances reflected from the vane into the two circuits. The impedance reflected from the left circuit into the right circuit is Z. If the vane moves to the right side it increases R_2 and decreases R_1, which in turn results in an increase of Z. On moving to the left, the vane decreases both R_2 and Z. The nonlinear amplitude versus deviation for the two circuits are superimposed in such a way as to give a linear resultant curve over a wide range.

The following derivation will demonstrate this linearity. From the circuit of Fig. 8 and for the condition that both circuits are ad-

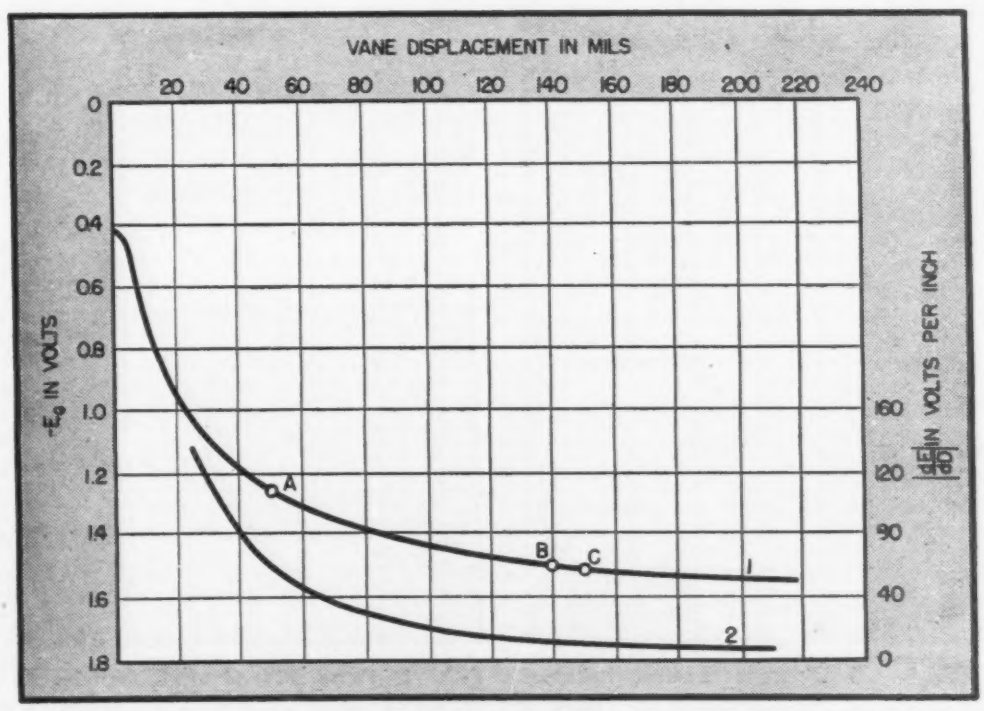


FIG. 7—Measured voltage change versus displacement, for a large scale model of the pickup, indicates linearity

justed to resonance, the impedance Z is

$$Z = (\omega M)^2/R_1$$

Let the ratio $(M/L) = k$ where k is the coupling factor between the two circuits. Then

$$Z = (\omega L)^2 k^2/R_1$$

Now let m be the fractional change of resistance produced by the moving vane in one of the two circuits. If the vane moves to the right

$$R_{2R} = R_2(1 + m)$$
$$R_{1R} = R_1(1 - m)$$

If the vane moves to the left

$$R_{2L} = R_2(1 - m)$$
$$R_{1L} = R_1(1 + m)$$

The changes in resistance and impedance in the two circuits are then

$$\Delta R_{2R} = mR_2$$
$$\Delta R_{2L} = -mR_2$$

The changes in reactance are likewise determined, and are given by

$$\Delta Z_R = (\omega L)^2 \frac{K^2}{R_1}\left(\frac{1}{1-m} - 1\right)$$

$$\Delta Z_L = (\omega L)^2 \frac{K^2}{R_1}\left(\frac{1}{1-m} - 1\right)$$

For small values of m the fraction $1/(1 - m)$ is approximately equal to $1 + m$, and the fraction $1/(1 + m)$ is approximately equal to $1 - m$. Substituting these approximate values into the foregoing equations gives

$$\Delta Z_R = (\omega L)^2 \frac{K^2}{R_1} m$$

$$\Delta Z_L = -(\omega L)^2 \frac{K^2}{R_1} m$$

As stated before, we see that both R_2 and Z will increase if the vane moves to the right and will decrease if it moves to the left.

In Fig. 9 ΔR_2 and ΔZ are plotted versus the deviation D. The curve for $\Delta R_2 + \Delta Z$ becomes most nearly linear if $\Delta R_{2R} = \Delta Z_L$ and $\Delta R_{2L} = \Delta Z_R$, that is for

$$mR_2 = (\omega L)^2 \frac{K^2}{R_1} m$$

If we assume that the vane in its rest position is exactly in the center between the two coils, $R_1 = R_2 = R$. Thus $R = \omega Lk$, and by the definition $\omega L/R = Q$ we obtain the relation

$$(1/Q^2) = K^2$$

This relation is the condition for critical coupling between the two circuits. Therefore we obtain perfect push-pull action if we couple the circuits critically for the center position of the vane. For this condition we also obtain $R = Z$.

Comparing the sensitivity of a double-coil pickup with one with a single coil, we find that there is no difference if we keep the gap the same. If the vane moves to the right, the total resistance that appears in the right circuit changes from $2R$ to $2R(1 + m)$. For a single-coil pickup, Z_R changes from Z_R to $Z_R(1 + m)$. The percentage change, which determines the sensitivity, is therefore the same in both cases.

Effect of Eccentricity

Quite often in commercial records the hole is not accurately centered, and the tone-arm head is driven back and forth in a radial direction during each revolution of the disc. Due to the moment of inertia of the arm, the stylus will swing with a frequency of 1.3 cycles per second, corresponding to the 78 revolutions per minute of the record. Deflection due to this motion is superimposed upon groove modulation. In pickups with large linear ranges (e.g. crystal or dynamic systems) eccentricity produces neither distortion nor amplitude changes. In electronic pickups however it can cause trouble. If the discriminator range of an f-m pickup is not wide enough, distortion appears once or twice during each disc revolution. In an a-m pickup of the type described herein, no distortion occurs, but amplitude changes can still be observed.

Figure 10A shows the rectilinear equivalent of the tone arm and pickup. The tone-arm mass assumed to be concentrated at the end of the arm, is represented by M. The compliance of the moving vane is represented by C_M. Groove displacement at the needle tip produced by record eccentricity at a frequency of 1.3 cycles per second is x_1, and x_2 is the maximum deformation of the vane, considered as a spring, due to the groove displacement. The displacement of the end of the tone arm is designated as x_3.

If the mass were small and the moving element stiff, the arm would follow the needle tip and the vane would not be deformed and displaced relative to the tone arm, i.e x_2 would equal 0. In this case eccentricity would not produce any amplitude change.

However, there is motion and in order to calculate x_2 we use the electrical equivalent of the pickup which is shown in Fig. 10B. The mechanical displacements, x_1, x_2, and x_3, are equivalent in the electrical analogy to charges. However, it is more convenient to use currents (which actually correspond to velocity). This substitution is justified in this particular problem because the frequency is constant, thus charge and current are proportional.

The constant-current generator G of infinite internal impedance delivers a current of 78 cycles per minute or 1.3 cycles per second through the network of Fig. 10B; the electrical parameters of which

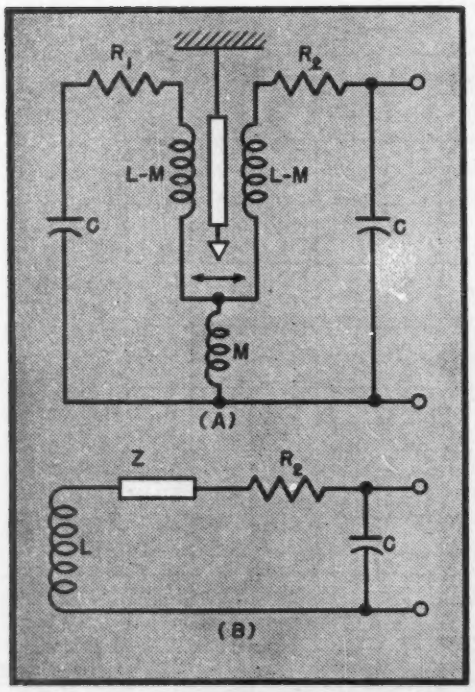


FIG. 8—(A) Electro-mechanical behavior of a push-pull modification of the pickup is analyzed from this circuit of which (B) is the electrical equivalent

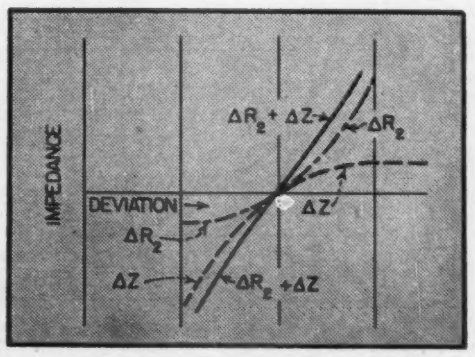


FIG. 9—From this plot of impedance versus deviation the most linear region of operation is chosen for the push-pull type pickup

are proportional to the mechanical parameters of Fig. 10A as listed in Table I.

To determine x_2 as a function of x_1, we have to determine the corresponding relation of i_2 to i_1. From Fig. 10B this latter relation is

$$i_2 = i_1 \frac{\omega^2 LC}{\omega^2 LC - 1}$$

which corresponds to

$$x_2 = x_1 \frac{\omega^2 MC_M}{\omega^2 MC_M - 1}$$

$$= - x_1 \frac{(2\pi f)^2 MC_M}{1 - (2\pi f)^2 MC_M}$$

If we assume an eccentricity of $\frac{1}{16}$ inch, and that $f = 1.3$ cps, $M = 20$ grams, and $C_M = 9.6 \times 10^{-6}$ cm per dyne, we obtain 0.8 mils for x_2.

Output of the pickup in db plotted for different gaps is given in Fig. 11. The curve is almost a straight line and this is not surprising if we again inspect curves 1 and 2 in Fig. 7. These two curves have essentially the same trend. Therefore, they are nearly exponential functions because only in this case does the first derivative have the same slope as the basic function. The curve in Fig. 11 corresponds to curve 2 in Fig. 7, and must therefore appear as a straight line if we choose a logarithmic (decibel) scale for the ordinate.

The gap is adjusted to about 12 mils. In this range the slope of the curve is 0.5 db per mil. If needle deflection due to eccentricity is 0.8 mils, the amplitude change is 0.4 db. This change is so small that it becomes unnoticeable.

Table I—Electrical-Mechanical Proportionalities

(For a fixed-frequency system)

Electrical	Mechanical
i_1	X_1
i_2	X_2
i_3	X_3
C_E	C_M
L	M

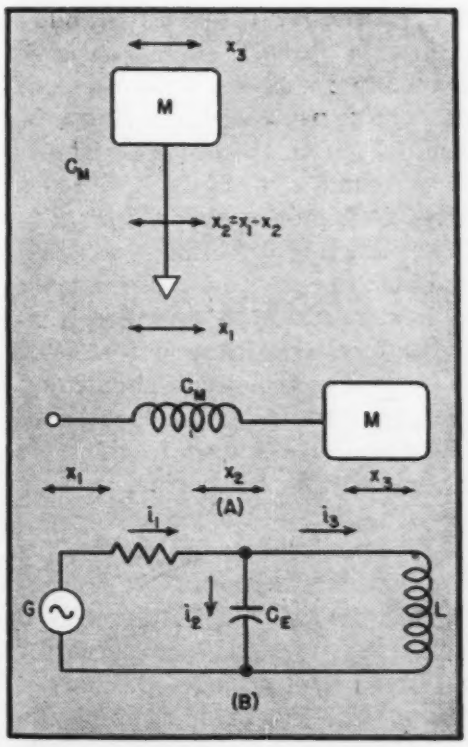


FIG. 10—(A) The mechanical equivalent of the pickup is composed of tip, spring, and mass, each having its own displacement. (B) The electrical equivalent of the pickup consists of constant-current generator, capacitance, and inductance, each having its own current

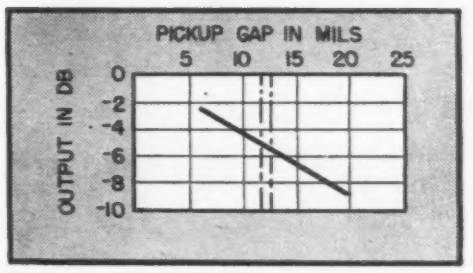


FIG. 11—Decibel output from the pickup is nearly a linear function of gap width

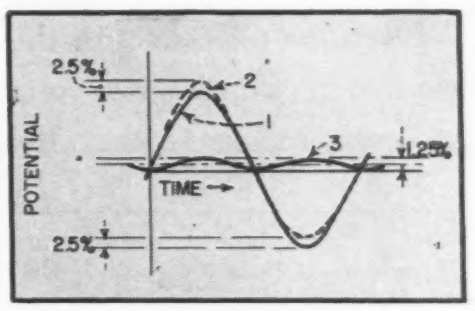


FIG. 12—Due to slight curvature of the pickup output characteristic, there is a small second-harmonic distortion component in the output

Output and Distortion

Figure 11 can also be used to determine distortion. Assuming a sinusoidal movement with a maximum displacement of plus or minus 0.5 mil, the output will consist of a distorted wave in which the maximum and minimum amplitudes are increased and decreased by 0.25 db respectively, which is shown in Fig. 12. Curve 1 shows an undistorted sine wave, curve 2 is the output from the pickup.

The distortion consists of a 2.5 percent change of the amplitude maximum (0.25 db corresponds to 2.5 percent). This distortion can be represented by superposition upon the fundamental wave of a d-c component and a wave of double frequency, curve 3. It can be seen that the peak amplitude of this second harmonic is 1.25 percent of the peak amplitude of the fundamental wave. The distortion factor is therefore 1.25 percent, a very small value compared with the inherent record distortion and the tracing distortion produced by the spherical shape of the stylus surface. This distortion is for maximum needle excursion. For smaller swing, proportionately less distortion would occur.

This reproducer and its electrical circuit have not only been laboratory tested but also put into large scale production. The author expresses his thanks to C. W. Carnahan and Robert Adler for their assistance in preparing this article.

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Pulse Response of DIODE VOLTMETERS

Corrections necessary in the calibration of a typical peak-reading, vacuum-tube voltmeter used to measure pulse amplitudes are determined from theoretical analysis and checked against measurements. Means are suggested for calibrating, and improving performance

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PEAK VALUES of pulse voltages must frequently be measured in such circuits as radar and pulse-modulation systems. There are several methods, the most popular being the cathode-ray oscilloscope and the diode, vacuum-tube, peak voltmeter. Each instrument has its own particular advantages, and is selected partly for its technical factors, partly for its convenience.

The familiar vacuum-tube, peak voltmeter, based upon the rectifying properties of a diode, has been used for many years for measurements of sinusoidal voltages. References in the literature are frequent and the nature of its shortcomings has been discussed.[1, 2, 3] It is only possible to use this type of vacuum-tube voltmeter for pulse voltage measurements if sufficient information concerning the characteristics of the instrument, as well as the nature of the source to be measured, is available.

TABLE I — DEFINITIONS OF SYMBOLS

C	—capacitance
ϵ	—base of natural logarithm
e_1	—maximum value to which capacitor charges
e_2	—minimum value to which capacitor discharges
e_0	—peak amplitude of measured pulse
E_{DC}	—direct voltage resulting from rectification and integration of pulse waveform
α_1	—time constant during t_1
α_2	—time constant during t_2
f	—pulse repetition frequency
k	—correction factor for vtvm
R_1	—total resistance during charging
R_2	—total resistance during discharging
R_D	—diode conducting resistance
R_g	—source resistance
t_1	—duration of charging interval
t_2	—duration of discharging interval
T	—period of pulse

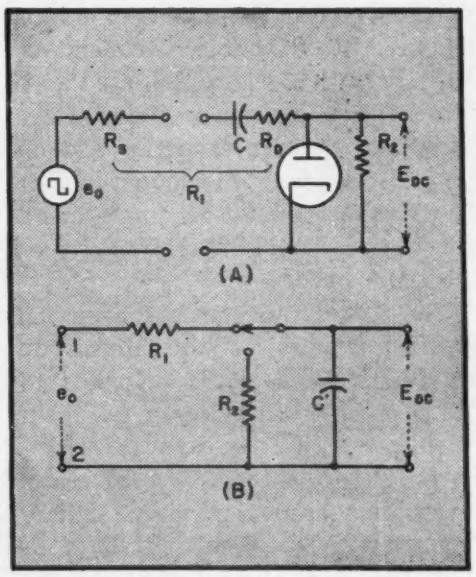


FIG. 1—(A) At the left is the rectifier portion of a typical diode voltmeter. At the right is the equivalent of a circuit under measurement. (B) The simplified, equivalent circuit of a diode voltmeter uses a switch to represent the diode

Diode Voltmeter Action

A circuit diagram of a typical vacuum-tube voltmeter is shown in Fig. 1A. Many variations of the circuit exist but the basic analysis is similar for each type. The magnitude of the direct voltage output (a measure of the source potential) is indicated by a d-c microammeter either directly, or through an amplifier. This part of the circuit is not shown as it has little bearing on the analysis to be presented.

Fig. 1B is the equivalent circuit. In order to derive an expression for the relationship between E_{DC} and e_0, defined in Table I, several assumptions are required. It is assumed that the source resistance R_g can be combined with the diode resistance R_D to form R_1. The diode is regarded as perfect, that is, the resistance in the conducting direction is constant and equal to R_D, and the resistance in the nonconducting direction is infinite. Leakages in capacitor C and in the rest of the circuit shunting C are combined in R_2. The duration of the pulse is assumed to be short compared with the period.

The positive pulse voltage, shown in Fig. 2A is impressed across the terminals marked 1 and 2 in Fig. 1B. During the charging interval t_1, C is charged through R_1 and then for a time t_2 partially discharged through R_2. The phenomenon is repeated periodically and eventually reaches the steady state condition depicted by Fig. 2B.

The maximum value to which C can be charged is

$$e_1 = e_2 + (e_0 - e_2)\,(1 - \epsilon^{-\alpha_1 t_1}) \qquad (1)$$

On top of the pulse generator used to check theoretical conclusions is a peak vacuum-tube voltmeter of the type analyzed in this article

$$= e_0 + (e_2 - e_0)\,\epsilon^{-\alpha_1 t_1}$$

The minimum potential to which C can fall is

$$e_2 = e_1 \epsilon^{-\alpha_2 t_2} \tag{2}$$

Equations 1 and 2 can be manipulated to obtain

$$e_1 = \frac{e_0\,(1 - \epsilon^{-\alpha_1 t_1})}{1 - \epsilon^{-\alpha_1 t_1}\,\epsilon^{-\alpha_2 t_2}} \tag{3}$$

and

$$e_2 = \frac{e_0\,\epsilon^{-\alpha_2 t_2}\,(1 - \epsilon^{-\alpha_1 t_1})}{1 - \epsilon^{-\alpha_1 t_1}\,\epsilon^{-\alpha_2 t_2}} \tag{4}$$

Practical Simplifications

It is desired to obtain an expression relating E_{DC} and the several factors appearing in the equations. The d-c potential across C is

$$E_{DC} = \frac{1}{T}\int_{t_0-t}^{t_1} e_1 dt + \frac{1}{T}\int_{t_1}^{(t_1+t_2-T)} e_2 dt \tag{5}$$

Integrating, evaluating, and simplifying, we obtain

$$E_{DC} = \frac{e_0 t_1}{T} + \frac{e_1}{\alpha_1 T}(\epsilon^{\alpha_1 t_1} - 1) - \frac{e_0}{\alpha_1 T}(\epsilon^{\alpha_1 t_1} - 1) + \frac{e_1}{\alpha_2 T}(\epsilon^{-\alpha_2 t_1} - \epsilon^{-\alpha_2 T}) \tag{6}$$

Equation 6 is an exact expression for E_{DC} limited only by the simplifying assumptions stated above. The equation in this form is unwieldy but may be simplified considerably if

$$\begin{aligned} \alpha_1 t_1 &< 1 \\ \alpha_2 t_1 &< 1 \\ \alpha_2 T &< 1 \end{aligned} \tag{7}$$

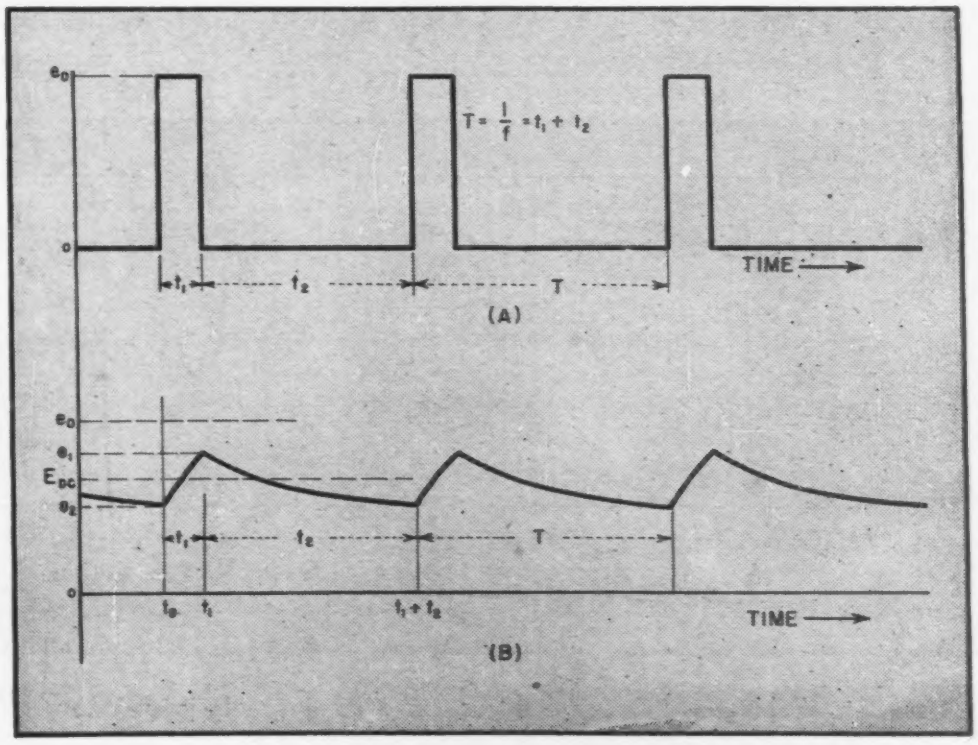
This simplification is justified because the pulses under discussion are relatively short in duration. However the voltmeters will perform accurately with long pulses, as will be demonstrated later. Also α_1 and α_2 may be made as small as desired by increasing the size of capacitor C. This practice will be justified later in the discussion.

If Eq. 7 is correct, the approximation $\epsilon^x = 1 + x$ may be employed. Equation 6 then reduces to

$$E_{DC} \simeq e_1 \tag{8}$$

Equation 3 may be simplified, using the same assumptions, to

FIG. 2—(A) Pulse applied to diode voltmeter, and (B) pulsating d-c developed across meter capacitor



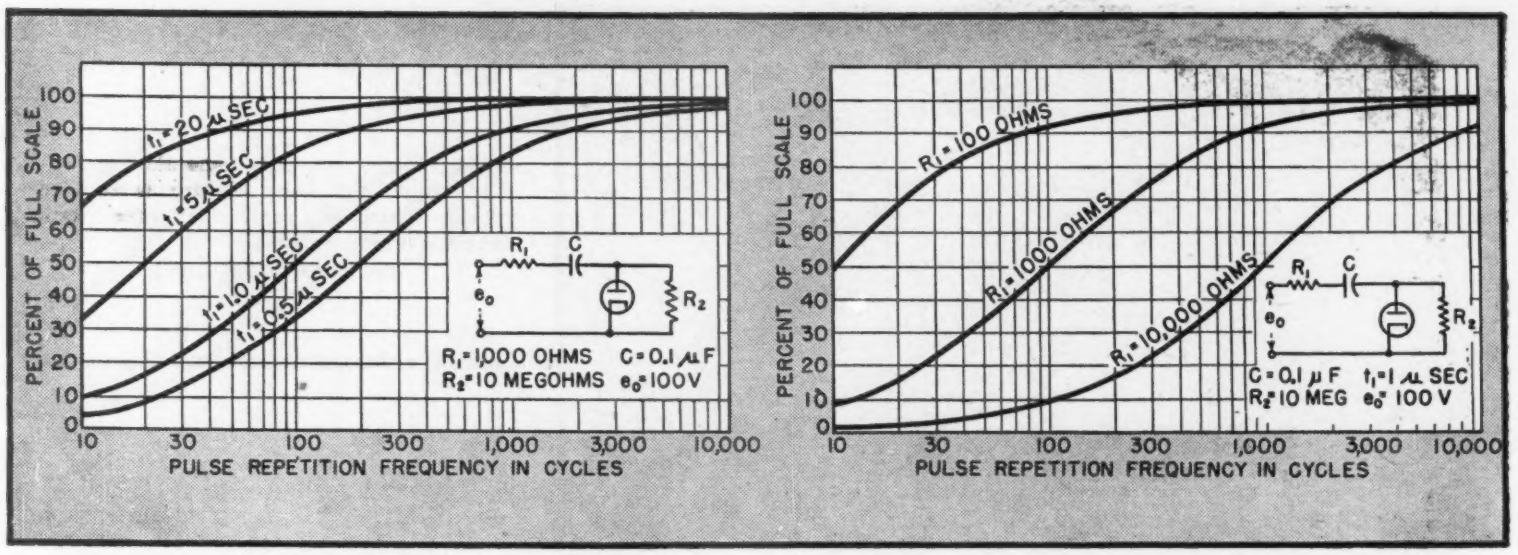

PERCENT OF FULL SCALE
PULSE REPETITION FREQUENCY IN CYCLES
t1 = 20 μ SEC
t1 = 5 μ SEC
t1 = 1.0 μ SEC
t1 = 0.5 μ SEC
R1 = 1,000 OHMS C = 0.1 μ F
R2 = 10 MEGOHMS e0 = 100 V
R1 = 100 OHMS
R1 = 1000 OHMS
R1 = 10,000 OHMS
C = 0.1 μ F t1 = 1 μ SEC
R2 = 10 MEG e0 = 100 V

FIG. 3—Effect of pulse width on meter indication

FIG. 4—Source impedance affects meter indication

$$e_1 \simeq \frac{e_0 \alpha_1 t_1}{\alpha_1 t_1 + \alpha_2 t_2} \qquad (9)$$

From which

$$E_{DC} \simeq \frac{e_0 \alpha_1 t_1}{\alpha_1 t_1 + \alpha_2 t_2} \qquad (10)$$

Equation 10 may be restated, in terms of the original quantities, as

$$E_{DC} = e_1 = \frac{e_0 t_1 R_2}{t_1 R_2 + t_2 R_1} \qquad (11)$$

It should be noted at this point that E_{DC} is independent of C as long as C is sufficiently large to make Eq. 7 valid. Intuitively, one might suspect from examination of the circuit that one could replenish the loss in amplitude of the charging voltage due to the long charging time constant by making C larger. Equation 2 shows this is not the case. Substitution of typical values into Eq. 7 will show that over the larger part of the range of narrow pulse width and pulse repetition frequencies the approximation is not in great error.

Figures 3, 4 and 5 are plots of Eq. 11. Note the effect of the various parameters on the meter calibration. It can be seen from these curves that, unless the four factors can be evaluated, the diode vacuum-tube voltmeter may indeed be quite misleading. The curves of Fig. 4 and 5 are identical in shape indicating that it is possible to improve meter accuracy for any given values of t_1 and t_2 by either making R_1 small or R_2 very large. In practice, R_2 is not usually made more than 10 megohms because of leakage difficulties. In certain military equipment destined for operation in humid climates, use of resistors larger than one megohm is undesirable. The limiting value of R_1 is the diode resistance plus the source impedance.

Corrections

A typical commercial diode-type vacuum-tube voltmeter uses approximately the following constants

$R_1 = 1000$ ohms
$R_2 = 10$ megohms
$C = 0.01$ μf

Reference to Fig. 4 indicates that this instrument will read approximately half the peak pulse value when a one microsecond pulse of 100 cps repetition frequency is impressed on its terminals, therefore a correction factor should be determined.

Equation 11 may be restated in a

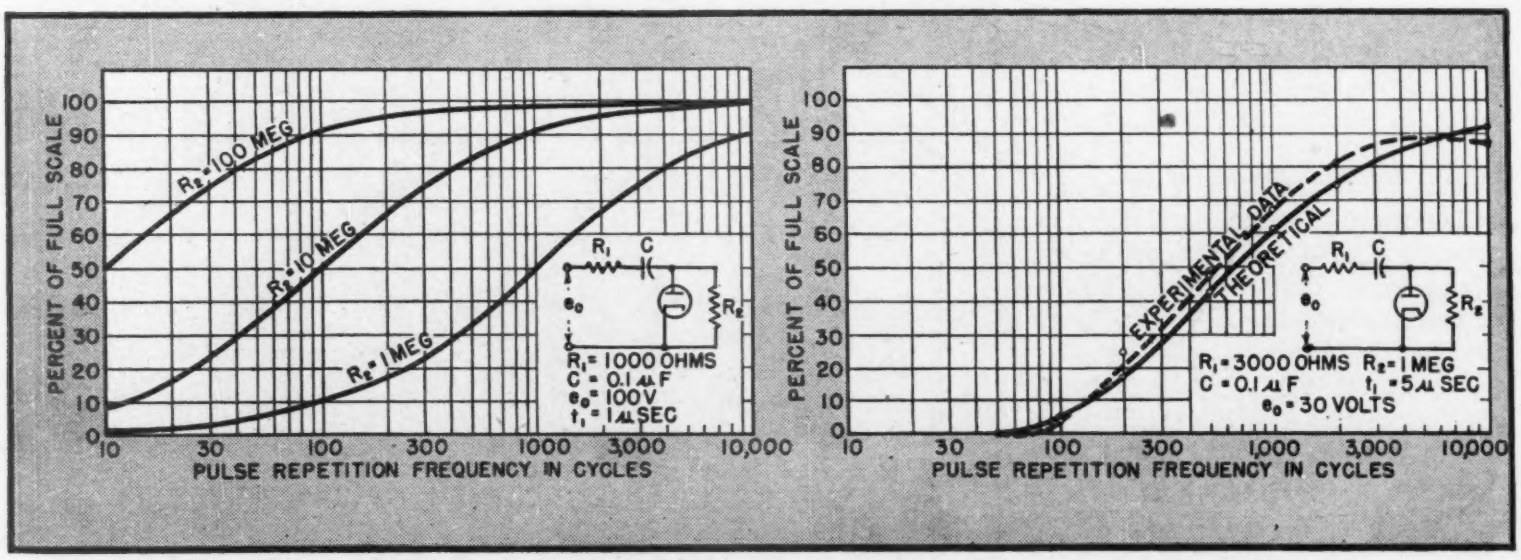

PERCENT OF FULL SCALE
PULSE REPETITION FREQUENCY IN CYCLES
R2 = 100 MEG
R2 = 10 MEG
R2 = 1 MEG
R1 = 1000 OHMS
C = 0.1 μ F
e0 = 100 V
t1 = 1 μ SEC
EXPERIMENTAL DATA
THEORETICAL
R1 = 3000 OHMS R2 = 1 MEG
C = 0.1 μ F t1 = 5 μ SEC
e0 = 30 VOLTS

FIG. 5—Magnitude of capacitor discharging resistor affects meter calibration

FIG. 6—Experimental data verifies the results of the theoretical circuit analysis

slightly different form

$$e_0 = kE_{DC} \quad (12)$$

$$e_0 \simeq E_{DC}\left[1+\left(\frac{t_2}{t_1}\right)\left(\frac{R_1}{R_2}\right)\right] \quad (13)$$

and if $t_2 \simeq T$

$$e_0 \simeq E_{DC}\left[1+\left(\frac{T}{t_1}\right)\left(\frac{R_1}{R_2}\right)\right] \quad (14)$$

Because the usual commercial, diode, vacuum-tube voltmeter is calibrated in rms values, that is 0.707 of peak, Eq. 14 should be modified to

$$e_0 \simeq \sqrt{2}\, E_{DC}\left[1+\left(\frac{T}{t_1}\right)\left(\frac{R_1}{R_2}\right)\right] \quad (15)$$

and

$$k = \sqrt{2}\left[1+\left(\frac{T}{t_1}\right)\left(\frac{R_1}{R_2}\right)\right] \quad (16)$$

Equation 16 gives an expression for a multiplying factor by which the meter reading must be corrected to indicate the pulse amplitude.

A circuit was set up similar to that shown on Fig. 1A. A 6AL5 diode with an R_D equal to 300 ohms, was used. Figure 6 shows the experimental results compared with the theoretical values computed from Eq. 16. The agreement is fairly close, within plus or minus ten percent.

It can be seen from the above discussion and from study of the curves that this type of voltmeter has anything but high input impedance under pulse conditions. The impedance of the diode voltmeter may be arbitrarily specified as being equal to the source impedance when

$$E_{DC} = \frac{1}{2}\, e_0 \quad (17)$$

It is difficult to define this impedance precisely because it is dependent on the pulse duration, repetition frequency, and amplitude, as well as circuit constants. Reference to Fig. 3, 4 and 5 shows that the input impedance may be well below 1000 ohms even though R_2 is equal to 10 megohms.

Improving Voltmeter Performance

Utility of Eq. 16 is limited by the difficulty in evaluating the magnitude of R_1. In most cases it is possible to estimate the source resistance with fair precision; but the diode resistance is difficult to determine. The diode resistance differs from tube to tube and, what is worse, is dependent on current. Thus R_1 depends on the magnitude of e_0 as well as values of α_1 and α_2.

A simple cathode follower similar to that shown in Fig. 7 will help improve the diode performance in some measure. Variations in R_3 will have no effect on meter calibrations, providing R_3 is small compared with the input resistance. Under this condition R_1 will be quite small, possibly 600 ohms or less. This low resistance will enable calibration of an instrument and computation of a correction curve based only on pulse width and repetition frequency.

Pulse stretching can sometimes be utilized as a means of making the instrument usable on narrower pulses. Figure 8 illustrates a possible arrangement. It is necessary to proportion R_3 and C_3 so that complete recovery, 99 percent, can occur between successive pulses. At the highest repetition frequency, C_3 should be small enough to permit e_3 to reach at least 99 percent of final value during the narrowest pulse to be measured. These precautions are necessary to minimize loss of peak amplitude and consequent loss of accuracy.

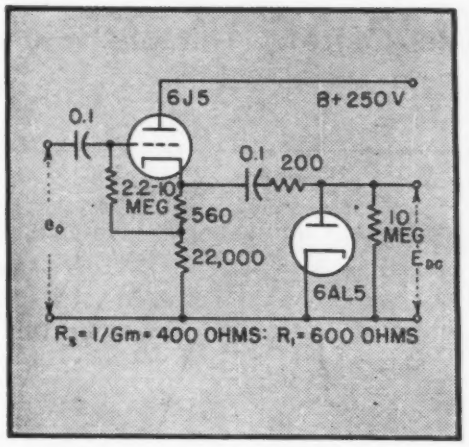


FIG. 7—A cathode-follower reduces circuit loading by the meter and prevents circuit impedance from affecting the meter indication

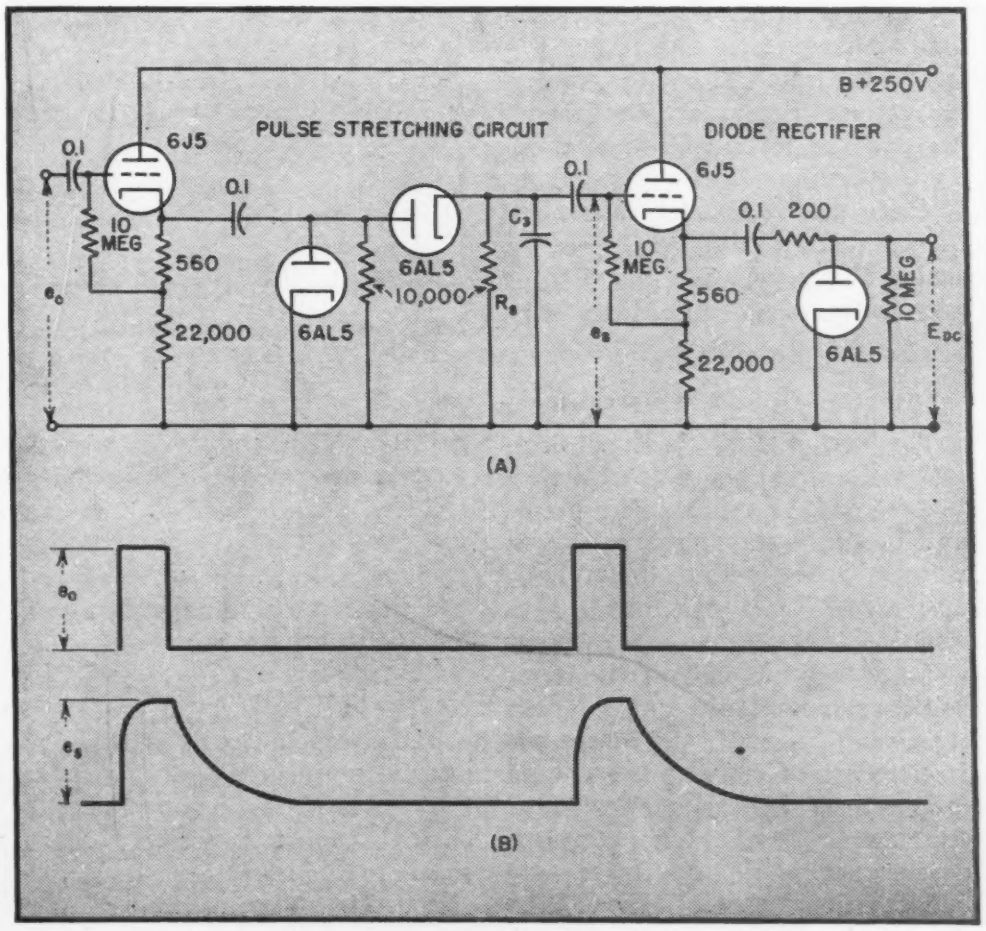


FIG. 8—(A) When measuring exceedingly short pulses, the reliability of the meter can be improved by introducing a pulse-stretching circuit ahead of the diode rectifier. The circuits are connected through cathode followers to prevent interaction. (B) The applied pulse shown at the top is stretched into a longer pulse by the circuit

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Electronic A-C VOLTAGE

Line voltage fluctuations are absorbed by a circuit using ordinary radio tubes and components, giving a regulated output suitable for calibration of a-c instuments. Attenuation at 180 cycles improves wave form by suppressing third harmonic

THIS PAPER describes an electronic regulator circuit developed to deliver a stabilized a-c voltage by absorbing the fluctuations of the input voltage. One of the important features of this circuit lies in the fact that the wave form of the output voltage is not altered appreciably by the regulator.

While such a device may have other applications, it was primarily developed as a power source for calibrating a-c instruments, where a constant voltage of sinusoidal shape is highly desirable. Those who must make precise alternating-current measurements know of the aggravation, inaccuracy, and wasted time caused by voltage fluctuations of the ordinary a-c power source. Voltage fluctuations at the output of a regulator based on the circuit under discussion were found to be only about six percent of the voltage fluctuations at the input.

General Principle of Operation

The block diagram in Fig. 1 illustrates the principle of operation of the regulator circuit. The circuit automatically stabilizes the output voltage by adjusting the voltage drop across series transformer T_1 to compensate for any change of input voltage. For example, if the input voltage should increase by one volt, the circuit automatically makes adjustments that increase the voltage drop across the primary side of transformer T_1 to nearly neutralize the increase of input voltage.

A fractional part of any voltage change that appears at the input terminals will appear across the output terminals. This change is detected and amplified by the rectifier-filter unit and the voltage-change detector, and an amplified voltage change or correction voltage is applied to the grid of a vacuum tube that acts as a load on the secondary side of T_1. A change of grid voltage on this tube changes its power consumption, thereby changing the voltage drop across both primary and secondary sides of T_1. The resultant voltage change across the primary side of T_1 nearly neutralizes the original voltage change at the input terminals.

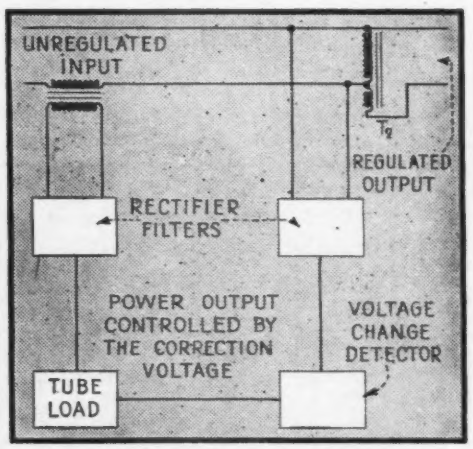

FIG. 1—Block diagram of voltage regulator for general laboratory use

Description of Circuit

The complete circuit of the regulator is given in Fig. 2.

The voltage rectified by V_2 is supplied through transformer T_2 from a part of the regulated output voltage. After filtering, this rectified voltage appears across the series combination of R_1, R_2, and R_3. A portion of this rectified and filtered voltage is compared with a constant voltage across regulator tube V_3. The difference between voltage drops A-B and B-C is the grid-cathode voltage on voltage-change detector V_3. By this method a voltage change appearing at the regulated output produces a corresponding voltage change at the grid of V_3. This voltage change is amplified and applied to the grid of V_4, the loading tube. A change in the grid voltage of V_4 produces a change in the impedance of winding 1 of T_1, changing the voltage drop across it enough to neutralize most of the change in input voltage.

If the voltage drop across winding 1 of T_1 is to approximate a sinusoid, special consideration must be given to the values of circuit elements L_1, L_2, C_1, C_2, and C_3. The network involving these impedance elements and rectifier tube V_1 is used to convert the d-c resistance of V_4 into a nearly linear a-c impedance appearing between the terminals of winding 1 of T_1.

The values of the circuit elements were determined by an experimental process. While making rough adjustments of these circuit elements, a cathode-ray oscilloscope was used to observe the wave form of the voltage across winding 1; for closer adjustments, the harmonic content of the regulated output voltage was measured by means of a wave analyzer. It was found not too difficult to make adjustments such that the harmonic content of the regulated output was essentially the same as that of the input.

A regulator based on this circuit may be built from simple radio receiver components. In the case of the regulator now in use in the

REGULATOR

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Standards Laboratory of the University of Utah, T_1 and T_2 are conventional radio power supply transformers. Winding 3 of T_1 is the 115-volt primary winding; winding 2 is the 1000-volt center-tapped secondary; winding 1 consists of two 5-volt windings connected in series. Winding 1 of T_2 is a 115-volt winding; winding 2 is a 800-volt center-tapped winding; winding 3 is a 6.3-volt winding. Windings 1 and 3 of T_2 are connected as an autotransformer to boost the voltage so that the output voltage has nearly the same value as the average of the input voltage.

Tubes V_4, V_5, V_6, and V_7 are cold-cathode diode-type voltage regulator tubes. Tubes V_4 supply a stabilized d-c voltage for the screen of V_3, and tubes V_6 and V_7 supply stabilized bias voltages for both of the grids of V_3.

Performance

A test of the regulator was conducted in conjunction with a model 326 Weston laboratory standard wattmeter, which has a 12-inch scale with 750 divisions. The following test was conducted with the regulator feeding the wattmeter circuit. When the input voltage to the regulator changed from 120 volts to 122 volts, the wattmeter indication changed from 740.0 watts to 741.5 watts. Within the limits of the accuracy of these measurements, the corresponding change of wattage is 0.2 percent. This change in wattage would require a change of 0.1 percent in output voltage of the regulator. Then a change of input voltage to the regulator of (2/120) x 100 or 1.67 percent produces a change of output voltage of 0.1 percent. The ratio, therefore, of output voltage fluctuation to input voltage fluctuation for this regulator is 6 percent.

The most troublesome variations in the voltage supplied by utility companies are often those continually fluctuating about a constant voltage to the extent of plus or minus 0.5 volt or less. When the regulator is not used, such a change of voltage is 0.42 percent and the corresponding change in wattmeter reading is 0.83 percent. The use of the regulator reduces the change of wattmeter reading to 0.05 percent, which is well within the limits of accuracy of many calibrations.

The regulator suppresses the third harmonic found in the input and leaves the other harmonics essentially unchanged, so that the output wave form is actually slightly better than that of the input. It has proved to be a time saver and an aid to accuracy in precision calibration. When the local utility company voltage was used as a source voltage for the calibration of a particular rotating standard against an indicating wattmeter, approximately 20 man-hours of time were required to complete the calibration. With the aid of the regulator this same calibration job now requires four man-hours of time. By using the regulator one man can perform a calibration in less time than was required previously of two men.

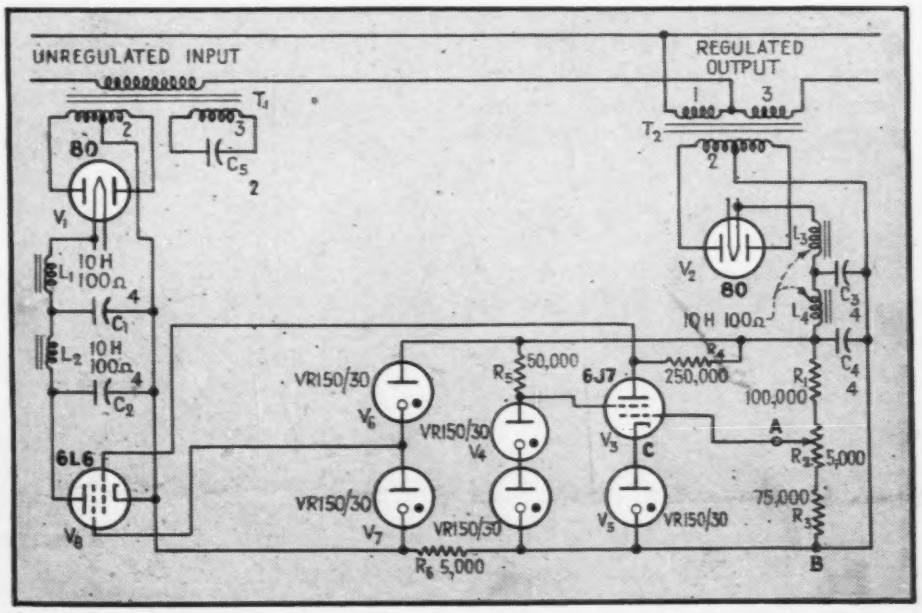
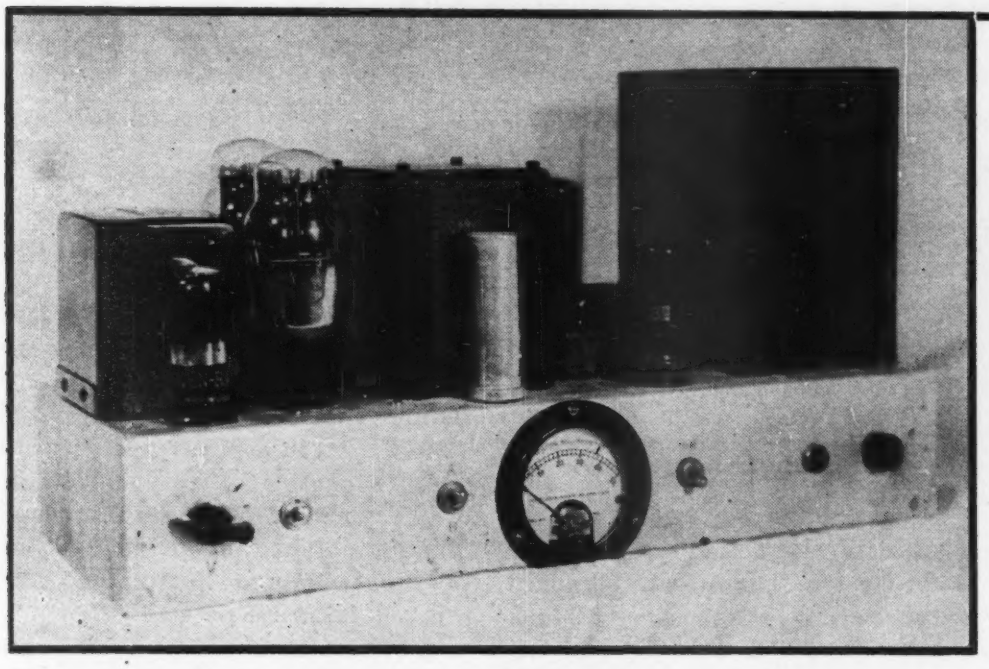


FIG. 2—Circuit diagram, with values of components. Transformers T_1 and T_2 are ordinary radio receiver power types. With the regulator, voltage changes at the output are only 6 percent of line voltage changes at the input

BRIDGING

Complete data on a monitor amplifier whose signal-to-noise ratio at maximum power output is approximately 80 db. It contains a combination differential amplifier and cathode-coupled phase inverter for balanced line input and a response flat within 0.5 db from 20 to 25,000 cps



The bridging amplifier includes a meter that indicates plate current of either output tube

INCREASING DEMAND for higher quality audio transmission requires apparatus capable of gaging the true quality of transmitted programs, especially in frequency modulated radio stations.

The amplifier to be described is capable of 15 watts output at frequencies as low as 20 cycles per second and as high as 25,000 cycles per second, having response flat within better than ± 0.5 decibel between these limits. This is obtained by the use of an all pushpull triode circuit particularly adapted to monitor bridging amplifier use, but also usable with single-ended input sources. Adequate gain for operation from a standard 500-ohm, 0-v-u (0 v-u = 1 mw) line and good signal-to-noise ratio further justify its use in monitor applications.

Components and circuit details are such that the unit can be duplicated at relatively low cost with assurance that good performance will be obtained. The addition of a single-ended preamplifier would make the unit adaptable to applications other than monitor use, such as quality reproduction in homes.

The unit has excellent response, as shown in Fig. 1, low effective source impedance for loudspeaker or cutter operation, differential amplifier, cathode-coupled phase-inverter input in place of a bridging transformer, balanced inverse feedback, low noise level, and adequate gain.

Push-pull triode operation is utilized throughout, as shown in Fig. 2. The input stage, comprised of a 6SC7 dual triode, is employed in place of a standard bridging input transformer. Since it has 500,000-ohm resistors in the grid-return circuits, the load imposed on the standard 500-ohm line is negligible. An input voltage of the order of 0.3 volt rms grid-to-grid will drive the amplifier at full output.

This input stage has several useful features. By utilization of a 25,000-ohm unbypassed cathode resistor in place of the usual 1,000-ohm resistor, a combination differential amplifier[1] and cathode-phase-inversion amplifier[2] is obtained. In normal operation, when a push-pull signal is introduced to the grid, the cathode remains at a fixed potential. However, if an in-phase signal is introduced to both grids, it will tend to drive grid *A* in a positive direction, and the cathode, due to the large value of cathode resistor, will also tend to go in a more positive direction. This is equivalent to grid *B* going in a negative direction. However, since grid *B* is also going in a positive direction, the in-phase signals cancel out, which is the desired characteristic to allow cancellation of hum pickup in balanced lines, etc. In spite of the high value of cathode resistor, full gain of the 6SC7 stage is realized with the circuit outlined.

Single-Ended Input

The amplifier may be supplied with a single-ended input signal by grounding the unused grid. A signal fed to the other grid, the signal voltage being developed between ground and that grid, will appear as a push-pull signal in the plate circuits of the input tube. This is occasioned by a similar action to that outlined above, that is, if grid *A* goes in a positive direction, the cathode follows the grid, giving an equivalent negative signal to grid *B*, thus providing a push-pull signal reproduced in the plate circuits. With the circuit constants shown, a balance of five percent or better can be obtained between the two signals, that is, the ratio of a-c plate signal *A* to plate signal *B*, will be approximately 1.05 or less.

The ratio of the plate signal magnitudes in the two plate circuits is given by the equation[3]: $E_{pA}/E_{pB} = 1 + 1/G_m R_k$ where G_m is the mutual conductance of the tube and R_k is the cathode resistance.

When the amplifier is used with single-ended input, it is important that provision be made to ground

AMPLIFIER for F-M Monitoring

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the unused grid directly, rather than depend upon the 500,000-ohm, grid-return resistor to furnish adequate ground return. If the grid is left at 500,000 ohms above ground, high-frequency response of the amplifier suffers (curve 3 of Fig. 1 results). Apparently the effective capacitance between grid and cathode causes the unused grid to follow the cathode at frequencies where the impedance of the capacitance approaches the value of the grid resistor, the effect becoming progressively more pronounced at higher frequencies. This decreases the overall gain at these frequencies by decreasing the input to the unused grid, that is, the differential voltage between the two grids is reduced.

In a sense, this effect is similar to the Miller effect[3] in usual triode operation, where the effective input capacitance C_i of a triode is many times the grid-cathode capacitance C_{gk}. The actual value of C_i is given by: $C_i = C_{gk} + (M + 1)C_{gp}$ where M = stage gain at mid-frequency and C_{gp} = grid-plate capacity.

For a 6SC7 with plate load and plate voltage conditions noted on the circuit, this capacitance is about 90 $\mu\mu$f, which is an impedance of 500,000 ohms at about 3,000 cps.

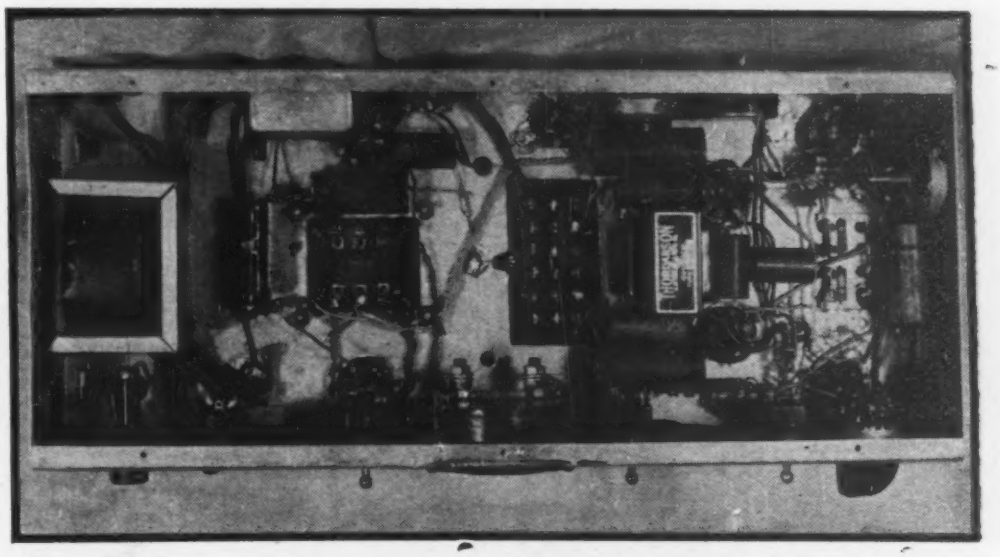
Driver Stage

The basic circuit of the driver stage is conventional with certain exceptions. The gain control of the amplifier, which might be omitted under certain application conditions, is a push-pull gain control. In the circuit shown, a two-circuit, six points per circuit, switch was utilized to give five steps of gain and an off position, in steps of approximately six db, except for the last step which was made approximately 10 db. It is probable that a standard dual gain control, with a resistance of 250,000 ohms per section, could be used.

The plate circuit of the driver stage and the grid circuit of the output stage can best be described jointly. A UTC type HA-107 interstage transformer having split secondaries is utilized to allow the application of fixed bias, with individual adjustment, to the output tubes. The two secondaries are individually loaded to reflect an impedance of approximately 10,000 ohms per plate to the 6SN7 driver sections. This serves to provide proper operation of the transformer over a wide range of frequencies. In the primary of the transformer, a 22-henry choke is utilized in the plate supply lead, to provide a high a-c impedance to ground from the center point of the driver transformer primary. This insures that the 6SN7 will maintain class A operation, since any tendency for the sum of the two plate currents to depart from a constant value (true class A operation) is offset by the a-c impedance of the choke. Since there is no necessity to drive this tube into the grid-current region to obtain adequate driving voltage for the output tubes, class A operation is to be desired under all circumstances for low-distortion operation, and is assured by the inclusion of this choke.

Output Section

The output tubes utilized are type 6A5G'. These tubes are similar to the 2A3 triode, but have two advantages over 2A3's. First, since they operate with 6.3 volts on the heater, no 2.5-volt transformer is needed. Secondly, the inclusion of a cathode internally tied to the center tap of the heater makes an equipotential electron source available. This has proved of great assistance in reducing the residual hum level present in the amplifier. The unit was previously constructed with 2A3's but the change was made to reduce hum. The output tubes are operated with a fixed bias of approximately 68 volts derived from the two potentiometers and the voltage divider network sup-



Components underneath the chassis are positioned for minimum pickup of stray fields

plied by the 5W4 bias rectifier. Individual adjustment of these tubes, so that they draw equal plate current, is accomplished by use of the two potentiometers and the 50-ma meter. The switch allows checking current of either tube independently. Each tube is adjusted to draw 40-ma with no output.

Since the cathodes are grounded, the center-tap ground return for all heaters is accomplished through the internal connection within the 6A5's, a common 6.3-volt supply being used for all three amplifier tubes. The output transformer utilized is an LS-55 UTC type.

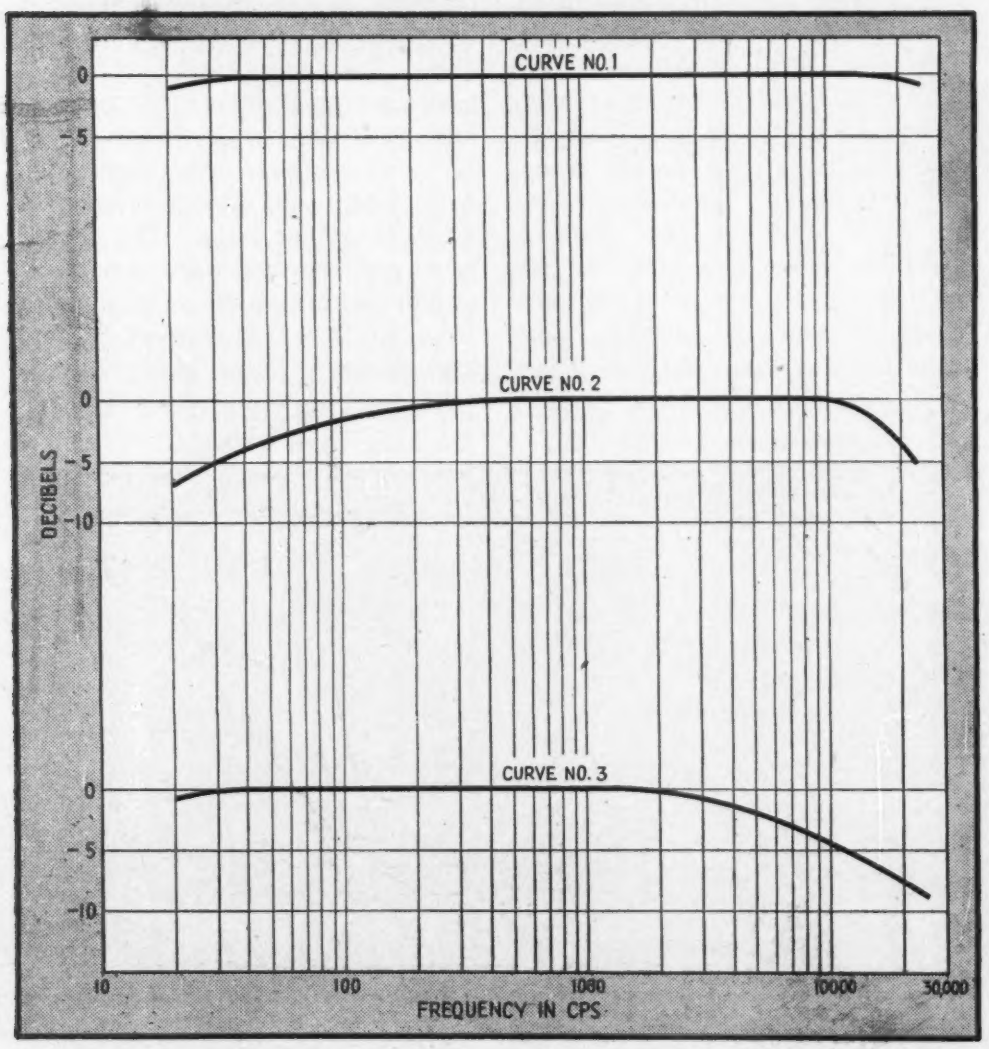
Feedback

Monitor amplifier application requires excellent response over a wide range of frequencies. It is most desirable to obtain a response curve which is flat over a wide range and which contains no positive slopes within or near the range of frequencies normally supplied to the amplifier. This last characteristic improves transient response.

Feedback is incorporated in the amplifier in a balanced fashion, in keeping with the overall push-pull characteristics, by feeding a portion of the output voltage from each output plate back to the appropriate driver cathode, through a large blocking capacitor, to reduce phase shift in the feedback circuit to negligible proportions. These bypass capacitors are 8 μf electrolytics further bypassed by paper capacitors to reduce the impedance effect of electrolytics at high frequencies. A $\mu\beta$ of 4 is provided [5] (corresponding to about 14-db feedback) without any tendency to oscillate.

Feedback could have included the output transformer under certain conditions but it was thought desirable to leave the output lines entirely independent of any relation to ground so that they could be operated with associated equipment in which one side of the output line might be operated at ground or the output line might be floating or center tapped to ground. The frequency response does not appear to suffer from the omission of the output transformer from the feedback loop.

FIG. 1—Curves made at one-volt rms input to amplifier feeding a 500-ohm resistive load. For curve 1, the input was a balanced line and feedback was used in the amplifier. A similar curve was obtained with single-ended input and one grid grounded. Curve 2 resulted when feedback was omitted. Curve 3 was produced with single-ended unbalanced input to one grid and the other grid not directly grounded, with feedback



Frequency Response

The characteristics of the amplifier without feedback, with push-pull balanced line input, are shown in curve 2 of Fig. 1. Under these conditions, the amplifier is flat within ± 2 db from 40 cycles to over 20,000 cycles, which is reasonably good. However, with feedback, as shown in curve 1 of Fig. 1, response is within better than ± 0.5 db from 20 cycles to 25 kc. The usual low impedance of the 6A5G's is further reduced by the addition of feedback. Under these conditions, a power output of 20 watts at 1,000 cps may be realized before overloading.

Output capabilities at low frequencies are sufficient to allow 15 watts (+ 42 v-u) to be developed at frequencies as low as 20 cps, while a similar output may be obtained at 25 kc. In addition, the introduction of feedback lowers the overall noise and hum level of the amplifier to a point where the signal-to-noise ratio at maximum power output of the amplifier is approximately 80 db. The actual background noise at full gain, with the input open but shielded, is —40 v-u below zero (approximately seven mv across the 500-ohm output line), or 82 db below maximum signal. A noise level of this magnitude is practically inaudible even when the amplifier is utilized with speakers capable of excellent wide-band response [6,7,8].

Other Features

The time constant of coupling circuits in the first two stages has been made quite large, 0.05 second, to maintain adequate low-frequency response. Input capacitors are included to allow coupling to sources having d-c components present.

The negative bias supply, in addition to being used to supply bias to the two output tubes, allowing operation as a fixed-bias amplifier, also supplies a point of return for the cathode circuit of the input tube. Since a large voltage drop occurs across the 25,000-ohm cathode re-

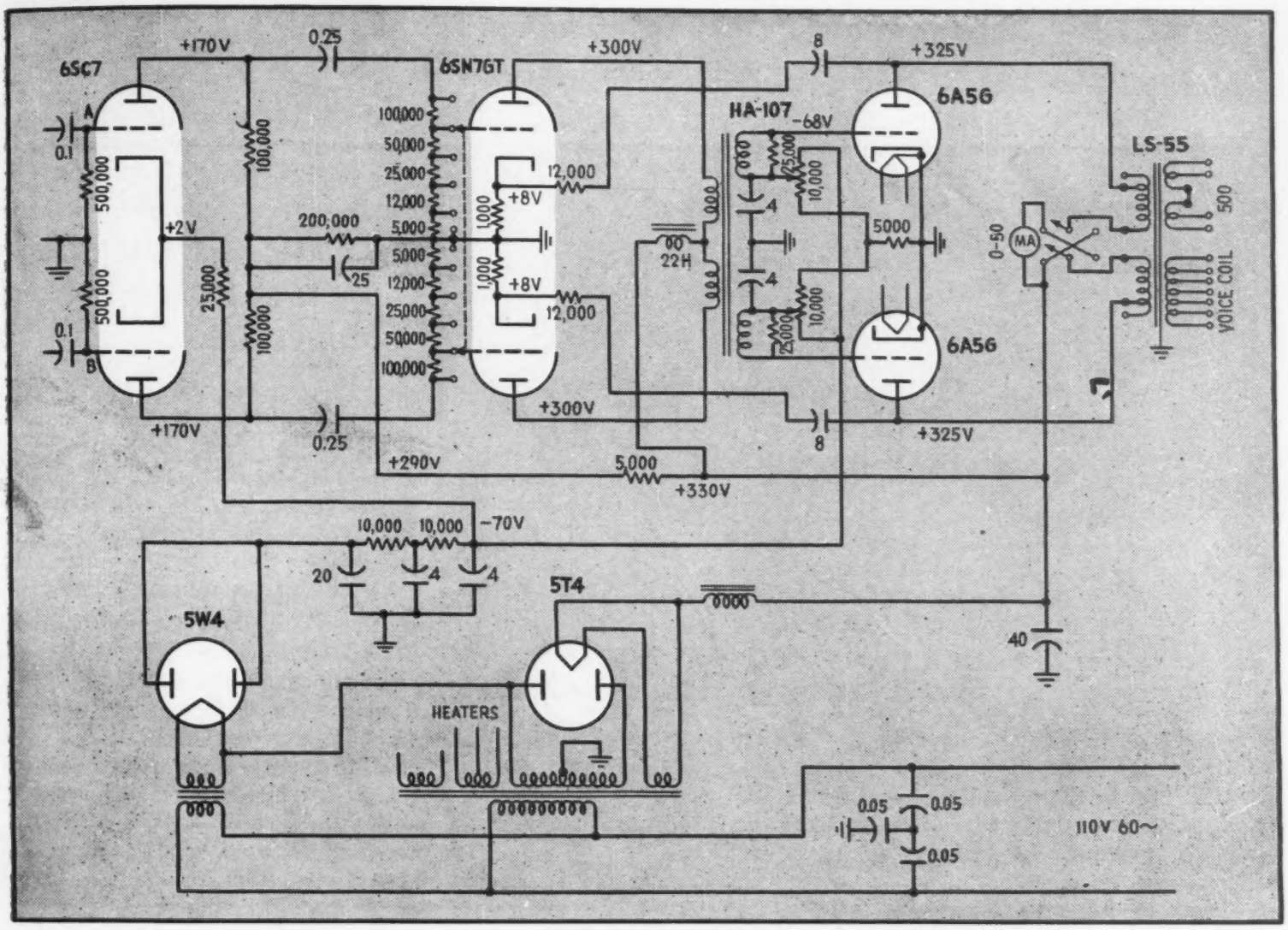


FIG. 2—Complete circuit of amplifier. For single-ended input, the unused grid is connected to ground

sistor (approximately 72 volts), it is desirable to return the cathode to a point at −70 volts, so that the effective bias on the input tube is approximately two volts, the grids being returned to ground. While this is not essential, returning the cathodes to ground through 25,000 ohms would reduce the effective plate voltage on the tube by 70 volts, and necessitate returning the grids to a point above ground, thus reducing the overall gain of the tube and also creating a rather high heater-cathode potential which is not a favorable operating condition.

A study of the circuit diagram will indicate that upon applying power to the amplifier, the output tubes are initially biased practically to cutoff until all the heaters in the amplifier warm up, since the 6SC7 input tube initially does not draw any current, resulting in a smaller drop across the filter resistors in the bias supply, or effectively increasing the bias on the output tubes. This makes for longer life and lack of surge current conditions in the output tubes during heating.

Individual adjustment was provided in an early model for the return point of the 6SC7 cathode but it was found that this was not required despite various tube changes and a fixed value was used which, with the other resistors, gives appropriate potential.

The output transformer is sufficiently well shielded so that it does not pick up appreciable hum from the power transformer. Grounding practice of returning single-stage grounds to a single point, with special care being taken to exclude any portion of the chassis from grid and and grid-return circuits within a single stage, has been followed. This is straightforward r-f and audio practice but is particularly necessary to obtain the noise level realized in this amplifier. The output transformer was chosen to allow use with standard line impedances or voice-coil impedances, adequate match being obtained from 500 ohms down to approximately one ohm. Since triodes are utilized in the output stage, with feedback, the damping characteristics of the circuit are excellent for applications in which the impedance of the driven device may vary over wide ranges. This is the case with almost all electroacoustic devices such as recording cutters and direct-radiator or horn-type loudspeakers. In addition, with direct-radiator loudspeakers, the damping provided by the amplifier tends to reduce distortion present at low frequencies to some degree. This unit has been successfully used with compact horn-type loudspeakers[6, 7, 8].

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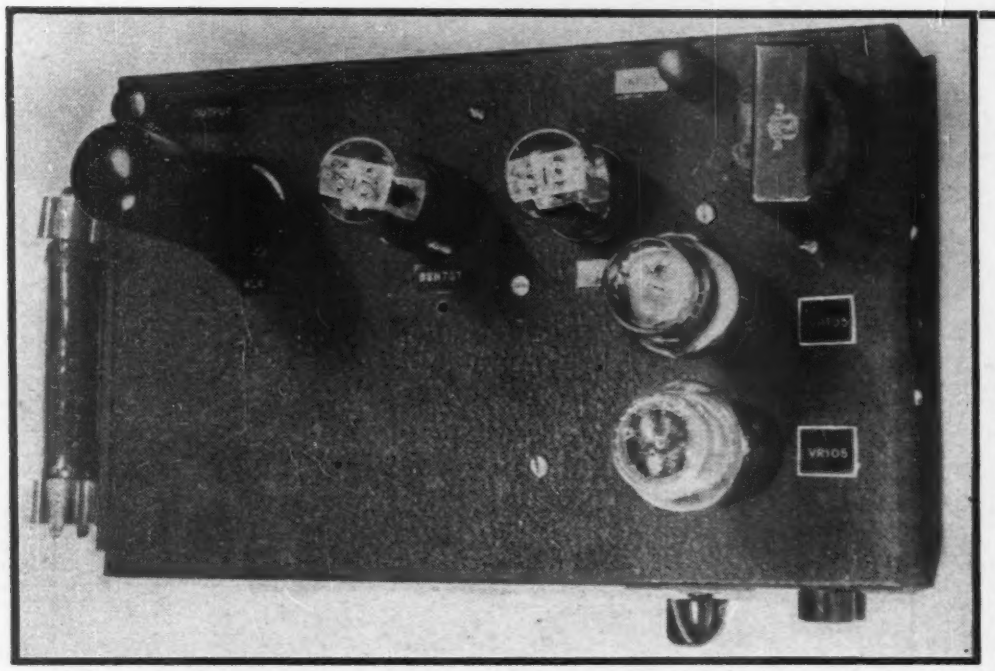
BETATRON



Input stages of orbit-shift pulse system used with Ohio State University betatron

By I. PAUL and T. J. WANG

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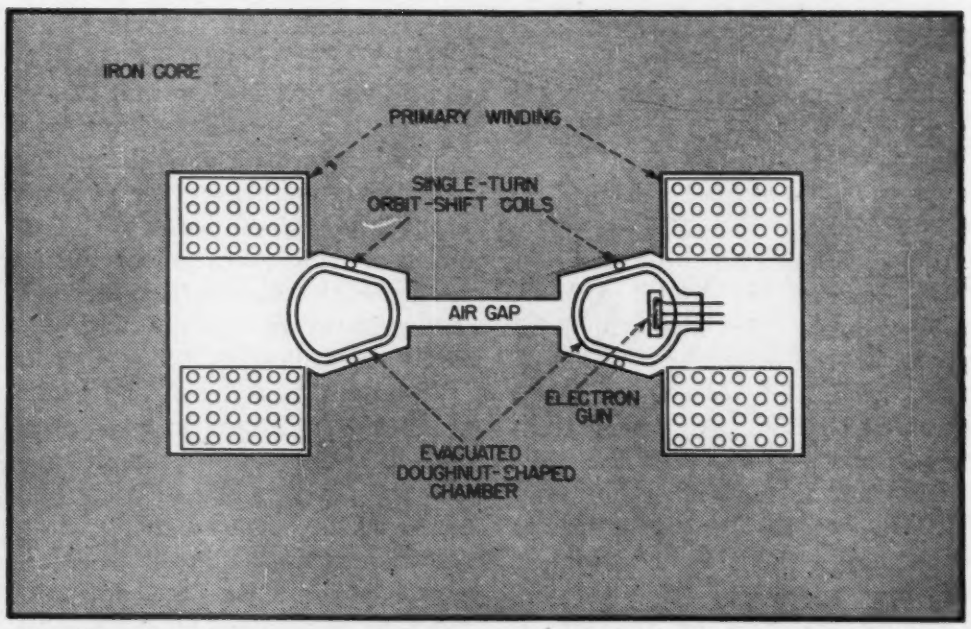
THE BETATRON is an electron accelerator that employs the principle of electromagnetic induction for its accelerating action. Basically, the betatron is roughly equivalent to a high-voltage transformer. However, in the betatron the customary metallic secondary conductors are replaced by a toroidally shaped evacuated chamber in which the electrons traverse circular orbits in space. The evacuated chamber, or doughnut as it is called, is placed around the central leg of the transformer core. Electrons inside the doughnut are accelerated by the electric field which is associated with the time-changing magnetic flux of the primary. The force resulting from the interaction of magnetic field and electron velocity serves to confine the whirling electrons to a circular orbit.[1, 2, 3]

Orbit-shift Coils

The electron injector is located within the doughnut at a point in the median plane just outside or just inside the circle in which electrons are ultimately confined and accelerated. Electrons are shot into the chamber in bursts by pulsing the injector. The electron beam avoids striking the injector as a consequence of a rapid shifting of the beam circle to the ultimate equilibrium orbit together with a damped oscillatory motion of the individual electrons about their instantaneous circle. Acceleration of the electrons transpires for approximately a quarter cycle of the magnetic field, that is, for a period of 1/240th second for a 60-cycle field or 1/4,000th second for a 1,000-cycle field. At the instant in the cycle when the electrons have reached the desired energy, the field-flux conditions for circular confinement are upset by the electronically triggered discharge of a capacitor through a set of orbit-shift coils.

The complete pulse-generating circuit with values of the components used is shown in Fig. 1. The arrangement of the final stages is similar to Kerst's,[2] employing a pair of gas tubes, a thyratron FG-41 and an ignitron GL-415, the thyratron firing the ignitron. The triggering unit which precedes the gas tubes consists of pulse generator V_2 followed by pulse amplifier V_3-V_4. The triggering unit is coupled front and back through cathode followers V_1 and V_5.

In one possible arrangement the orbit-shift coils are single-turn loops mounted just outside the doughnut parallel to and directly above and below the electron orbit. The momentary surge of current

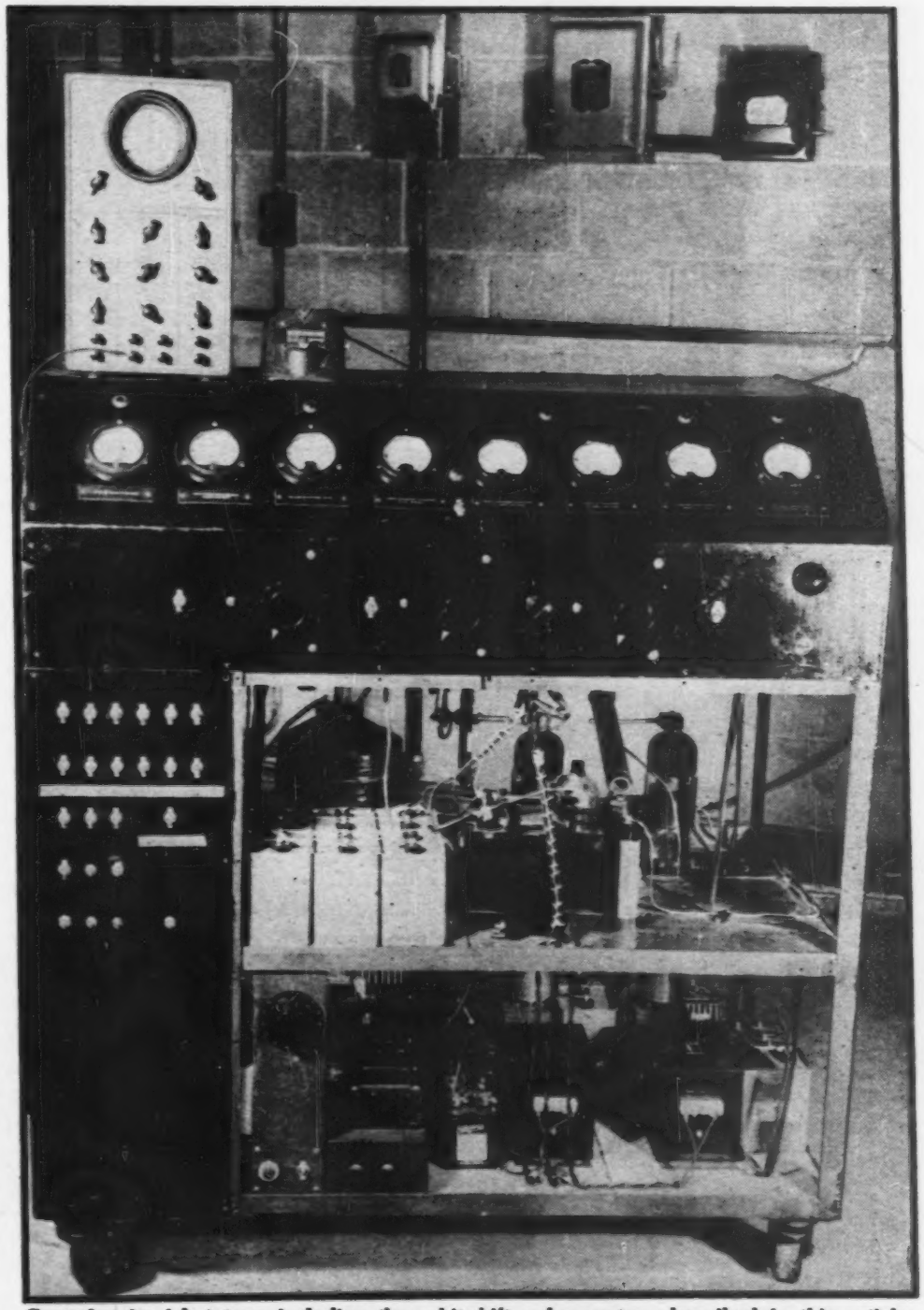


Cross-section through betatron accelerating unit, showing positions of orbit-shift coils that are energized by the 1,000-ampere pulse

PULSING SYSTEM

Circuit used to generate high-current pulses for orbit-shift coils of betatron has many industrial applications, including resistance welding and stroboscopic inspection. Pulse generator having cathode followers and flip-flop amplifier triggers a thyratron that fires an ignitron, and the resulting capacitor discharge gives a 1,000-amp peak pulse

through these coils causes a strengthening of the central flux, with little change, however, in the field at the position of the orbit. The result is that the electrons receive an incremental acceleration without a correspondingly increased centripetal force, and the beam spirals outward.



Control unit of betatron, including the orbit-shift pulse system described in this article. Input stages of orbit-shift system are mounted directly back of the meter panel

Pulse-generating Circuit

The pulse system of the Ohio State University betatron is described herein. The basic circuits are conventional; however, the design details are felt to be of general interest. The system has a wide variety of possible applications beyond the immediate use for which it was intended. It may be employed in any device which requires periodic high-power pulses at frequencies up to a few thousand cycles per second. Conceivable applications are to be found in television, in pulse-time modulation, in stroboscopic studies, in high-speed photography, and in resistance welding.

The 1-megohm resistor and 0.03-microfarad capacitor going from the betatron exciting coil to ground form an integrating circuit such that the voltage developed across the capacitor is in phase with the flux through the betatron coil. Pulse generator V_2 is operated as a limiter so that a rectangular output wave is obtained corresponding to the sinusoidal input signal. The phase of the leading edge of the rectangular wave (and, incidentally, the width of the wave) are controllable through the variable bias on this stage. A differentiating circuit after the limiter provides a negative pulse corresponding to the leading edge of the rectangular wave and a positive pulse corresponding to the trailing edge. The latter pulse is ultimately clipped by high negative biasing of subsequent stages. However, the trailing-edge pulse is of no concern since it does not affect the thyratron following.

The pulse amplifier is connected as a flip-flop circuit with a very short time constant. Negative pulses trigger it and cause it to pro-

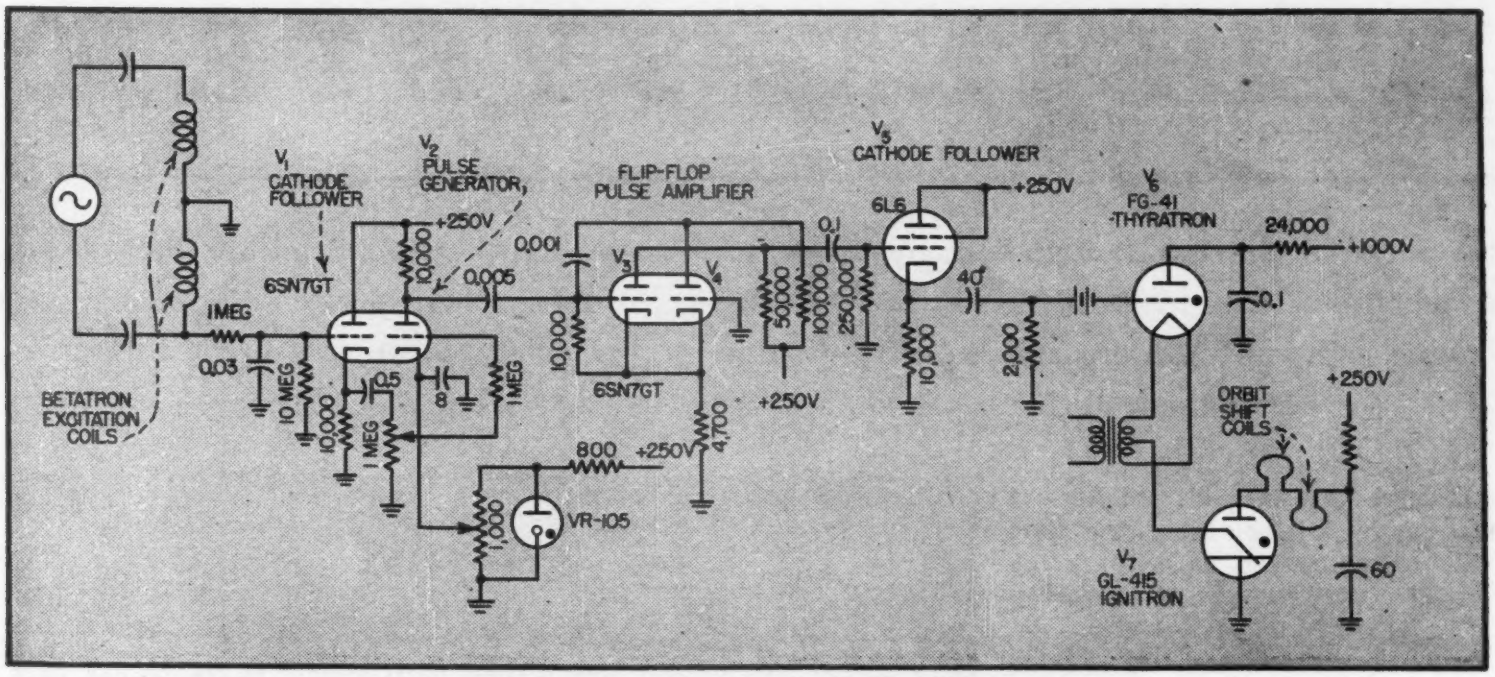


FIG. 1—Circuit diagram of complete pulsing system

duce positive pulses of larger amplitude. The amplitude of the output pulse is practically independent of the exciting pulse, so that the operation of subsequent circuits is independent of the magnitude of the exciting voltage. The peak value of the 6L6 output pulse is approximately 70 volts.

Thyratron Circuit

The FG-41 thyratron is biased with a battery so that the anode will not conduct before a pulse is applied. When a positive pulse large enough to overcome the grid bias is applied to the grid, the tube conducts. The capacitor discharges through the FG-41 anode and the GL-415 ignitor more rapidly than it can charge through the series resistance. The capacitor voltage falls to a value sufficiently low to extinguish the tube, and the grid regains control.

The sudden discharge of the thyratron capacitor serves to fire the ignitron, and the ignitron in turn discharges its own capacitor through the orbit-shift coils. This latter pulse is approximately 40 microseconds wide at half-amplitude and is of the order of 1,000 amperes peak value, as determined by a cathode-ray oscillograph across a measured small resistance.

The simplified circuit of Fig. 2 will serve for a study of the principles involved in the thyratron circuit. Figure 3 shows the lower extremities of a family of static characteristics for a typical thyratron with various values of grid bias. The current scale in this figure is abnormally enlarged. The portions of the characteristics shown are for currents below 20 milliamperes for a tube which is designed to carry several amperes in conventional operation. The higher curves are for more negative values of grid bias. It will be noted that for any particular value of current the steepness of the characteristics varies with the grid bias, the steepness decreasing from one curve to the next as the grid bias becomes more negative. For current values much higher than those shown in Fig. 3 the various current-potential loci all merge into one single curve because, except for the lowest current values, the tube behavior is independent of grid bias.

Initially, capacitor C in Fig. 2 is presumed to be charged nearly to the supply potential E_{bb}, and the system is temporarily in an essentially stationary state while awaiting the positive grid pulse. Point A in Fig. 4 depicts the situation. The solid curve is the e_c-i_a static characteristic for the quiescent grid bias E_{cc}.

Analysis of Firing Action

On the application of a sufficiently strong brief positive pulse to the grid, the tube characteristic assumes a new position as indicated by the dotted curve of Fig. 4. For the instantaneous value of grid voltage E_{cc}' corresponding to the dotted characteristic, the existing anode potential is sufficient to fire the tube, and operation jumps to point B.

Point B is actually on the E_{cc}' characteristic, and operation might be presumed to follow the E_{cc}' characteristic if the grid-cathode voltage were to remain at E_{cc}'. But since only a brief pulse is applied to the grid, the grid voltage returns to its quiescent value E_{cc} after a short time. It may be regarded that the characteristic followed is an intermediate one which coincides with that of E_{cc}' at the beginning of the pulse and with that of E_{cc} at the end of the pulse. However, in view of the approximate coincidence of the characteristics at all values of grid bias for the relatively high anode currents which obtain immediately after the firing, the characteristic for E_{cc} will be assumed as adequately describing the tube behavior.

Inasmuch as R_1 is very much greater than the sum of R_s and the tube resistance, the discharge current of the capacitor, denoted by i_d in Fig. 2, exceeds the charging current i_c for an appreciable time following breakdown of the tube. During succeeding moments the capacitor potential falls, and the course of events is as described by successive points C and D in Fig. 5, where D, the point of tangency of the R_s load line with the tube characteristic, is the extinction point. Further decrease of capacitor voltage with time results in a transfer of operation to point E.

The tube in the dark stage presents a high resistance so that i_c rapidly exceeds i_d, and the capacitor recharges approximately to the potential E_{bb}, where the system awaits a new grid impulse.

Capacitor-charging Equations

As long as the capacitor charges nearly to the supply potential E_{bb} in each cycle, the peak value of the current pulse through R_2 depends only upon the values of R_2 and of the potential E_{bb} and not upon the size of the capacitor. This is evident on examination of the exponential charge-time decay curve of the capacitor. Neglecting the current i_c, the current i_d is the time derivative of the capacitor charge, and has its maximum value at the time $t = 0$. Thus, if the potential of the charged capacitor be taken as E_{bb},

$$i_d = \frac{d}{dt}\left(CE_{bb}\epsilon^{-t/R_2C}\right)$$

$$= -\frac{CE_{bb}\epsilon^{-t/R_2C}}{R_2C}$$

$$\text{For } t = 0\,,\ i_d = -E_{bb}/R_2$$

The duration of the pulse is lengthened by the use of a high value of the product R_2C since with high R_2C, currents of small magnitude obtain only at large values of t.

The amplitude of the pulse will be lessened at high-frequency operation if the charge circuit time constant, R_1C, is too long to permit the capacitor becoming charged nearly to the potential E_{bb} before the tube triggers.

Continuous conduction, that is, failure of the tube to extinguish at some time following breakdown, will occur in case (contrary to the previous assumption) i_c equals i_d in magnitude before the extinction point is reached on the down swing along the characteristic (Fig. 4 and 5). Once the condition is reached wherein i_c equals i_d, the source E_{bb} feeds the tube directly, and the plate current cannot be stopped thereafter unless E_{bb} is reduced substantially to zero. The particular point on the characteristic wherein i_c equals i_d is that point, hereinafter referred to as the critical point, where the tube characteristic intersects the $R_1 + R_2$ load line. This load line is drawn through the point (E_{bb}, 0). The position of the critical point as located above is evident because for equality of charging and discharging currents, the capacitor is effectively out of the circuit, and the tube is simply being supplied from the source E_{bb} through R_1 and R_2 in series.

Dynamic Operation

The foregoing remarks are concerned only with tube characteristics obtained under static or steady conditions, and the analysis is not entirely correct for transient states (dynamic operation). However, in a practical setup where some experimental control of the parameters is possible, the method serves successfully to indicate orders of magnitude and appropriate directions for adjustment.

One effect of dynamic operation which must be taken into account is the steeper slope of the dynamic characteristic which causes the extinction point (point of tangency of the R_2 load line) to occur at a slightly lower current value than that which would be predicted from static considerations. The steepness of the characteristic depends upon the size of the capacitance used, and for very small values of capacitance the extinction point decreases as the capacitance is reduced. As can be observed from Fig. 5, the anode potential rises with the discharge of the capacitor. With this rise of anode potential there occurs a current flow i_g through the grid resistor (in addition to any current which may flow during continuous static operation) by virtue of the presence of the inherent grid-anode capacitance of the tube, as indicated in Fig. 6. The current i_g is in such a direction that the associated drop across the grid resistor changes the grid potential

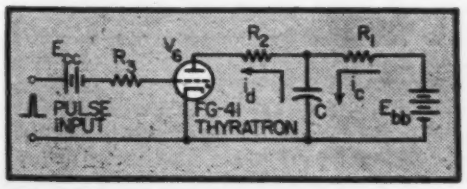


FIG. 2—Idealized thyratron pulse circuit. Arrows indicate current flow

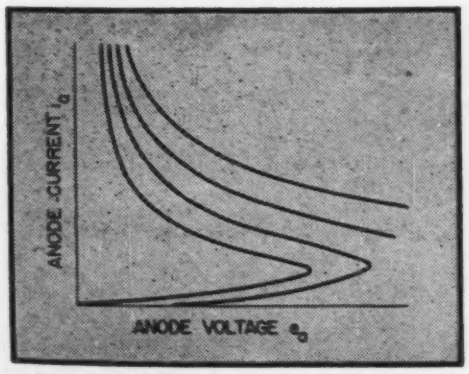


FIG. 3—Lower portions of the family of anode current-anode voltage characteristics of a thyratron

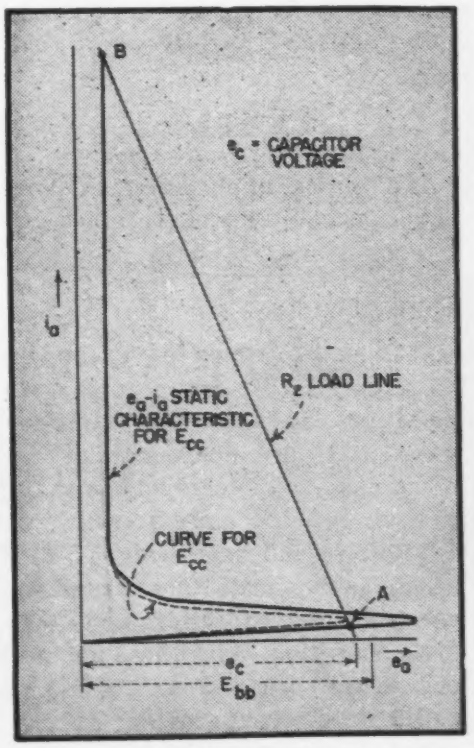


FIG. 4—Operation immediately before firing (point A) and immediately after firing (point B) of the thyratron in the pulse circuit

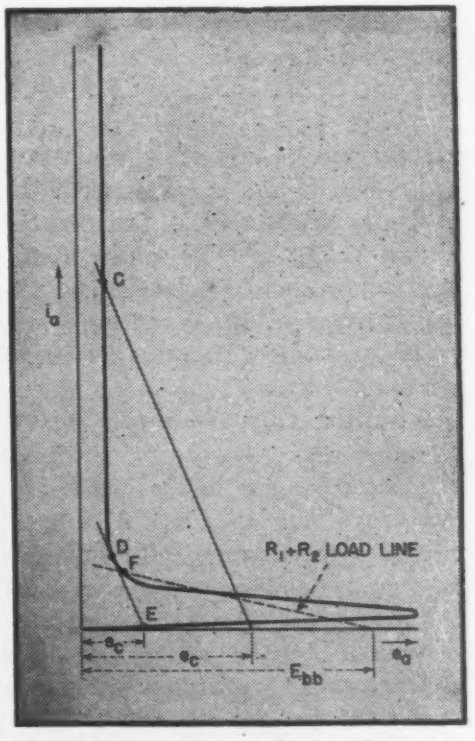


FIG. 5—Operation at intermediate time during capacitor discharge (point C), just before extinction (D), and after extinction (E)

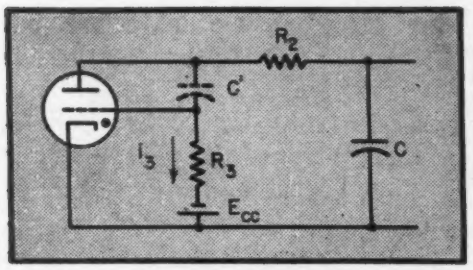


FIG. 6—Current flow through grid resistor associated with rapid increase of anode potential

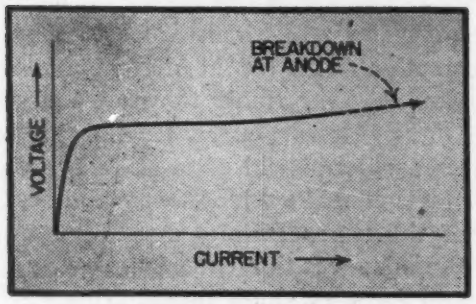


FIG. 7—Voltage-current characteristic of ignitor rod for d-c operation

in a positive sense. As has been noted in connection with Fig. 3, this serves to steepen the characteristic and therefore to lower the extinction point. In a similar gas tube arrangement Puckle[4] was able to enhance the depression of the extinction point with reduction of the main capacitance by deliberately shunting additional capacitance across the stray grid-anode capacitance.

Ignitron Characteristics

A specific distinction between the ideal circuit of Fig. 2 and the actual circuit which was employed lies in the thyratron load resistor R_2. Actually, R_2 consisted of the ignitron striking element, whose resistance was found to vary with the current through it in the manner shown in Fig. 7. For values of current below about ½ ampere the ignitor behaves as an essentially uniform resistance of about 25 ohms, whereas for larger current values the ignitor presents a varying resistance which decreases with the current in such a manner as to produce a voltage drop which is practically independent of the current. Hence, when during the course of the cycle the point of operation moves downward along the tube characteristic toward smaller currents, the slope of the R_2 load line steadily decreases and R_2 approaches its maximum value of about 25 ohms. The bend in the experimentally determined static characteristic curve for the FG-41 thyratron sets in below about 14 milliamperes (above this value of current the curve is nearly vertical); thus, when operation on the downward swing of the characteristic reaches the neighborhood of the extinction region the ignitor has attained its maximum resistance.

Design Procedure

The first step in the design of the actual circuit was the choice of an appropriate value for C. A small value of C was deemed to be advantageous since for any suitable time constant R_1C a small value of C would permit a large value of R_1, the latter being desirable to ensure bringing the critical point below the extinction point. On the other hand, the capacitor could not be too small because small capacitances serve to lower the extinction point. A compromise value of 0.1 microfarad was selected somewhat arbitrarily, and values for R_1 and E_{bb} were determined as follows:

During the charging period, and neglecting the dark current, the voltage across the capacitor is given by

$$e_c = E_{bb}\,(1 - \epsilon^{-t/R_1C}) \qquad (1)$$

A second relation involving R_1 and E_{bb} arises from the requirement of fixing the critical point below the extinction point. If the critical point has the coordinates e_o, i_o, it is seen from Fig. 8 that

$$\frac{E_{bb} - e_o}{i_o} = R_1 + R_2$$

For all practical purposes the above equation may be written simply as

$$\frac{E_{bb} - e_o}{i_o} = R_1 \qquad (2)$$

since $R_2 \ll R_1$.

Numerical values of e_o and i_o were selected to correspond to a point safely below the extinction point on the experimentally determined static characteristic of the thyratron, and a value of 1/500th second was assigned to t as the approximate time alloted for charging the capacitor with 500-cycle operation. With these numerical values of e_o, i_o, and t, Eq. 1 and 2 were solved together graphically. Several curves of E_{bb} versus R_1 for Eq. 1 were plotted with e_c as a parameter, and the intersections with the graph of Eq. 2 yielded the e_c-versus-E_{bb} relation shown in Fig. 9. The knee of the curve corresponds roughly to 1,000 volts for E_{bb}, indicating the desirability of using approximately 1,000 volts for the source. This value provides a high capacitor voltage just prior to the discharge, which in turn assures an adequate pulse for firing the ignitron. Any higher source potential does not appreciably increase the capacitor voltage. The value of R_1 follows at once from Eq. 2.

A similar set of considerations holds for the ignitron circuit, except that here the load resistance is constant and of sufficiently low value that the tube action is not critical to wide variations of circuit parameters. It was found with the ignitron, just as with the thyratron, that too low a value of the charging resistor resulted in continuous conduction. For very high values of charging resistance the ignitron fired spasmodically, indicating that the capacitor voltage was not adequate to insure spread of the discharge to the anode at the time of the ignitor pulse.

References

(1) Kerst, D. W., *Physical Review*, **60**, p 47, 1941.
(2) Kerst, D. W., *Review of Scientific Instruments*, **13**, p 387, 1942.
(3) Wang, T. J., The Betatron, ELECTRONICS, p 128, June 1945.
(4) Puckle, O. S., "Time Bases", John Wiley & Sons, Inc., 1943, p 172.

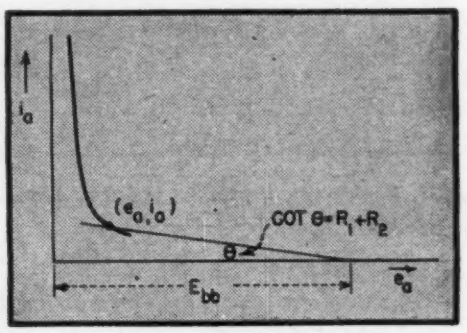


FIG. 8—Critical point, safely below extinction point

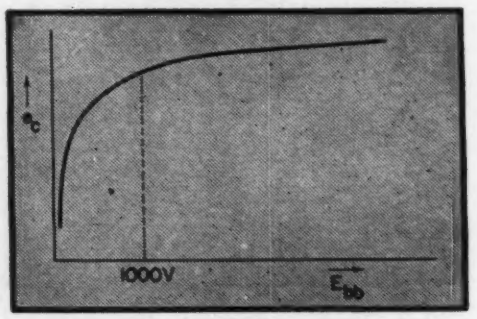


FIG. 9—Effect of supply voltage on fully charged capacitor voltage

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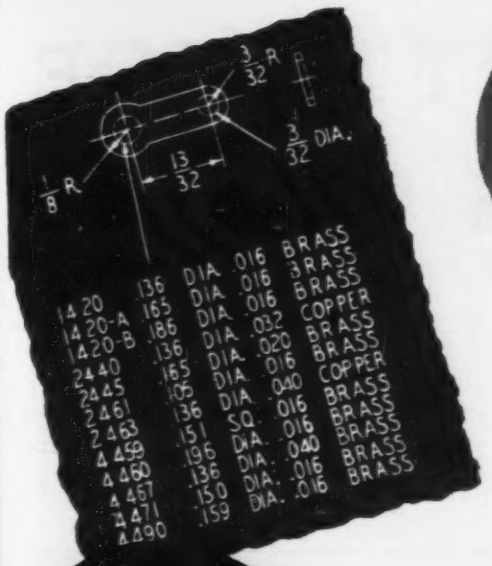
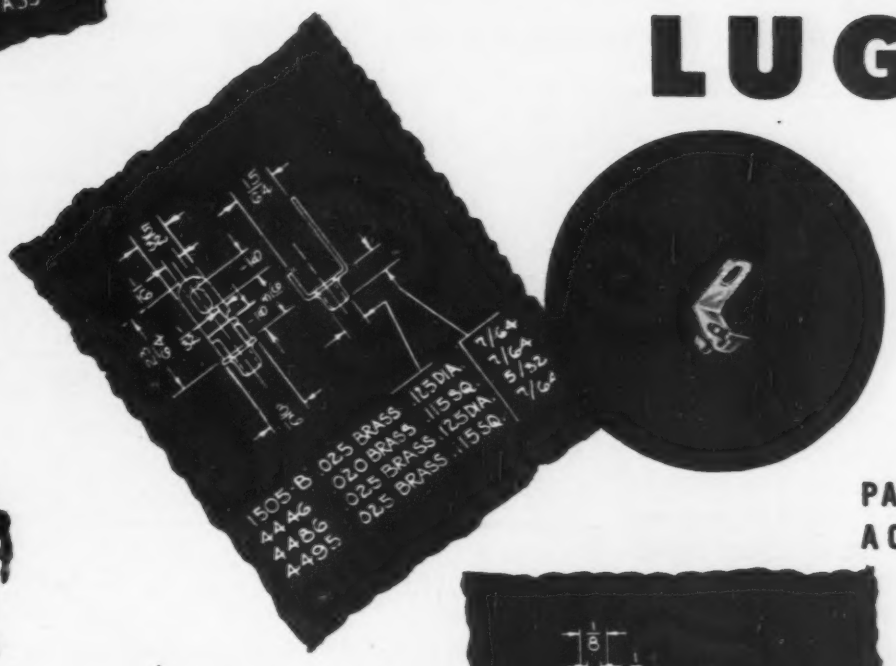
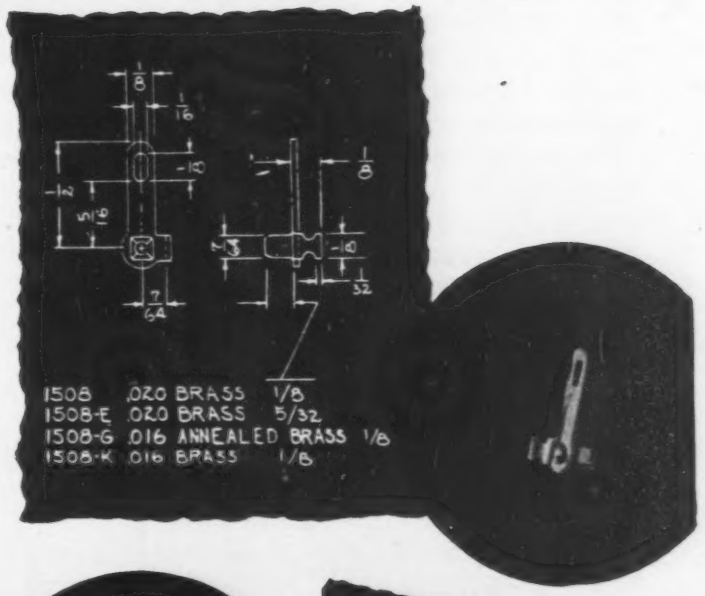




PARTS SHOWN
ACTUAL SIZE



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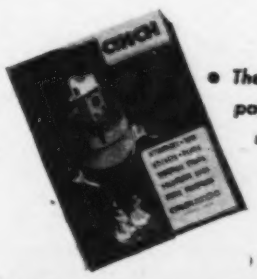

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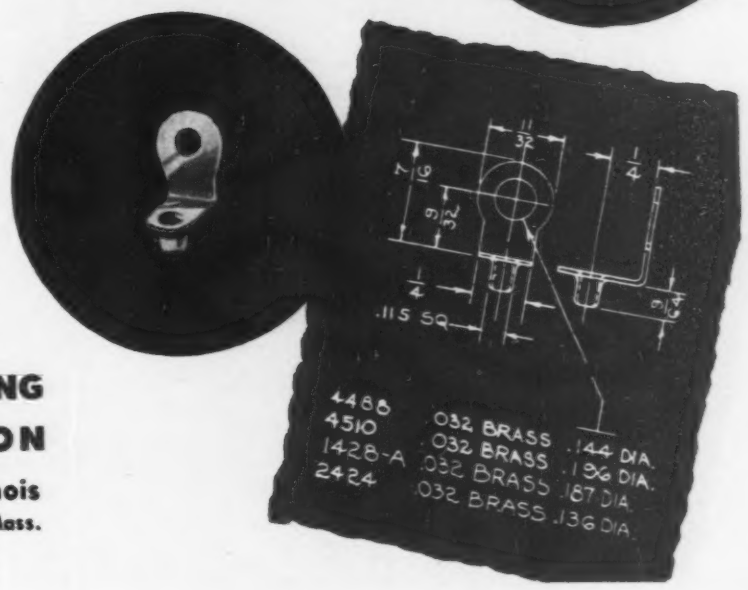

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IMPEDANCE-ADMITTANCE CONVERSION CHART

Simple chart for quickly converting $Z = R \pm jX$ to $Y = G \mp jB$ and vice versa when combining impedances in parallel. One-decade range requires use of multipliers for most problems, but accuracy is adequate for quick, rough estimating work

By ROBERT C. PAINE

Boonton, New Jersey

WHEN impedances are to be combined in parallel it is necessary to convert them to admittances so they can be readily added together. Such conversions can be conveniently performed with reasonable accuracy by the aid of the chart shown. Resistance R is plotted in a horizontal direction and reactance X in a vertical direction, being considered positive or negative depending on whether it is inductive or capacitive. Passing through each impedance point $R \pm jX$ is an arc centered on the horizontal axis for conductance G, and another arc centered on the vertical axis for susceptance B of the corresponding admittance $Y = G \mp jB$. The susceptance is read as negative for inductances and positive for capacitances.

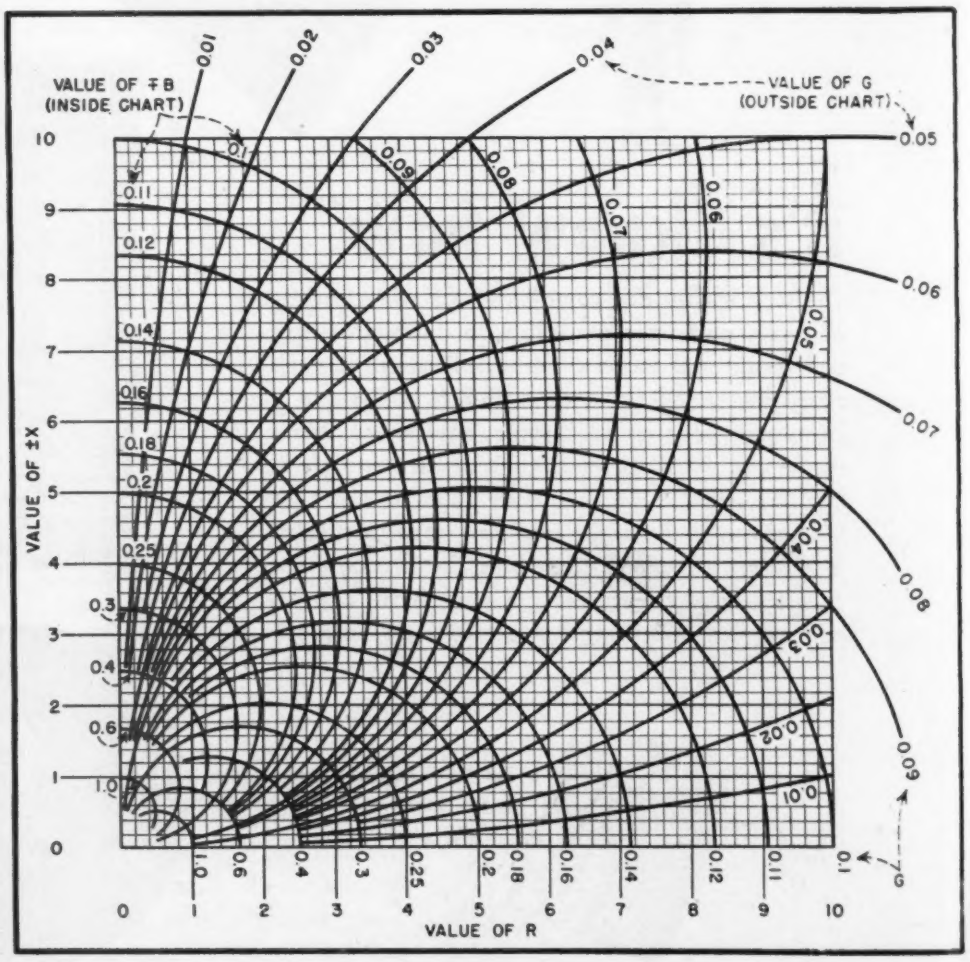
Examples of Use

(1) Impedance $Z_1 = R + jX = 8 + j5$. On the chart this corresponds to an admittance $Y_1 = G - jB = 0.09 - j0.057$.

(2) Impedance $Z_2 = 2 - j1.5$. The corresponding admittance is $Y_2 = G + jB = 0.32 + j0.24$. Because Z_2 lies near the origin, greater accuracy can be attained by using a multiplier such as 4. Then $4Z_2 = 8 - j6$, the corresponding value of Y is $0.08 + j0.06$, and this value multiplied by 4 gives the correct value, $Y_2 = 0.32 + j0.24$.

(3) The impedances of examples 1 and 2 are combined in parallel. The sum of their admittances is $Y_3 = Y_1 + Y_2 = 0.41 + j0.183$. Changing the scale by a factor of $\frac{1}{4}$ we obtain $\frac{1}{4}\ Y_3 = 0.102 + j0.046$, and $Z_3 = \frac{1}{4}\ (8.2 - j3.6) = 2.05 - j0.9$.

(4) Impedance $Z_4 = 20 - j15$. Multiply Z_4 by 1/10 to get values within range of chart, find admittance values as in example 1, and multiply them by 1/10 to get $Y_4 = 0.032 + j0.024$.



When known values are outside range of chart, multiply by any convenient value that brings them within range, find equivalent values from chart, and multiply them by that same multiplier to get final answer. Values of R and centers of circles for corresponding G values are on horizontal reference axis. Values of X and centers of circles for corresponding B values are on vertical reference axis

ELECTRONICS REFERENCE SHEET

MALLORY engineers are first to solve the tremendous technical problems involved in developing a tab-type vitreous enamel resistor that completely meets JAN (Joint Army and Navy) Specification, Characteristic F!

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JAN-R-26 Style	WATTAGE 275° C. New	TUBE DIMENSIONS Length	Outside Dia.	Inside Dia.	RESISTANCE—OHMS Using Wire Dia. .010 Min.	.0025 Max.	Tap Available
RW 30 F	7	1	½	5/16	2	630	No
RW 31 F	10	1½	½	5/16	10	2000	No
RW 32 F	12	2	½	5/16	18	3100	Yes
RW 33 F	18	3	½	5/16	35	6300	Yes
RW 34 F	30	3	¾	9/16	50	9000	Yes
RW 35 F	38	4	¾	9/16	71	12000	Yes

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INDUSTRIAL CONTROL

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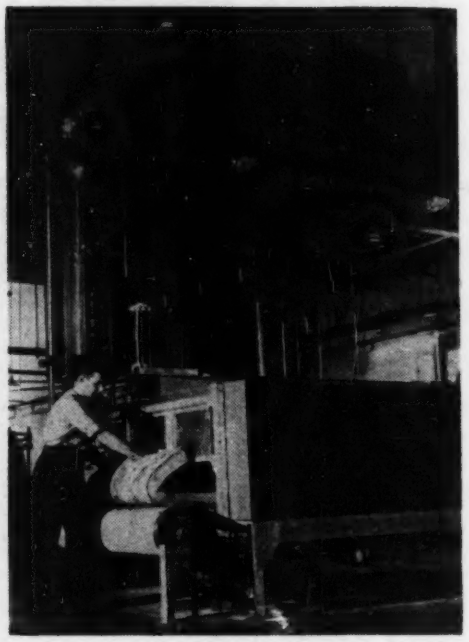
X-Ray Snapshots Used in Atomic Bomb Production

EQUIPMENT WHICH makes possible millionth-of-a-second x-ray photographs was used extensively in connection with the atomic bomb experiments, according to Dr. Charles M. Slack, assistant director of research for the Westinghouse Lamp Division, Bloomfield, N. J. The specific role of high-speed x-ray pictures in atomic bomb experiments must remain confidential at this time.

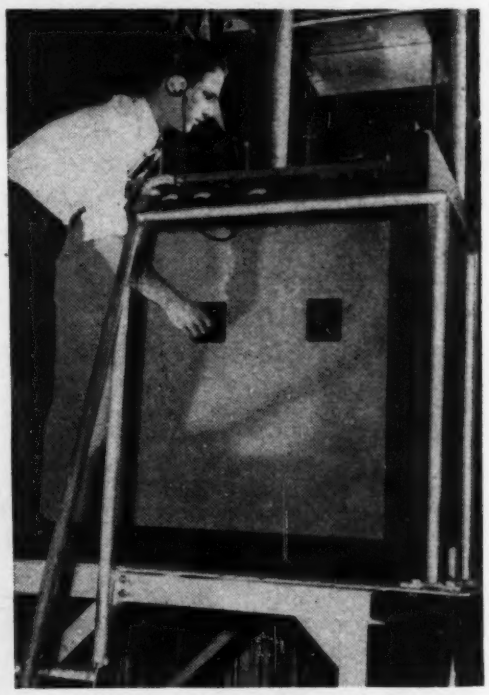
The new picture-making technique also contributed to the science of ballistics during the war by making possible studies such as determining the realignment of a bullet as it zipped down the bore of a gun barrel and the swelling, bursting and disintegration of a high explosive shell resulting when it chewed through steel plate.

The evidence was obtained by comparing photographs of the original missiles with several high-speed sequence pictures made as the bullet raced down the bore of the gun or as it entered the steel plate.

Typical of the amazing discoveries uncovered by high-speed x-ray pictures in ordnance experiments during wartime was one which at Aberdeen Proving Grounds, Md., cleared up the mystery of why certain armor-piercing bullets failed to penetrate the armor. The reason, the pictures disclosed, was that the cap on the bullet, designed to break up the case hardening on the armor plate at the point of penetration, ripped free of the bullet after it left the barrel. Later the bullet caught up with the cap and shattered it just prior to reaching the target.



Cured by electronic heating, a Foamex mattress is removed from the drying oven where dielectric heating is also used to reduce moisture. The three levels of the equipment include the vulcanization chambers, remote control panel and power unit



A Firestone engineer adjusts the remote control panel on the second level of the electronic vulcanization unit

Electronic Vulcanizing on Commercial Basis

THE FIRST commercial electronic vulcanizer in the rubber industry has been placed in operation in the manufacture of Foamex mattresses at the Fall River, Mass. subsidiary plant of Firestone Tire and Rubber Co.

Designed for mass production, the three-story machine uses 125 kilowatts of r-f power to service two complete vulcanization chambers. It was designed by Firestone engineers and the Westinghouse Electric Company in Baltimore, Maryland. With the electronic machine, large double-bed mattresses that previously required a 35-minute cure by the old steam jacket method, are completely cured in 5 minutes by 13.6-mc r-f power.

Close examination of the electronic-cured mattress reveals that the final product is far superior structurally to the old type. This is due largely to instantaneous heat supplied uniformly to all parts of the mattress, as contrasted to the conventional steam method in which the foamed rubber is vulcanized slowly from the outside of the mattress to the interior. Since in this method, the process cannot always be precisely controlled, parts of a steam-cured mattress may be com-

Power tubes and tuning capacitor of the 125,000-watt electronic vulcanization unit installed at the Firestone plant in Fall River, Mass.

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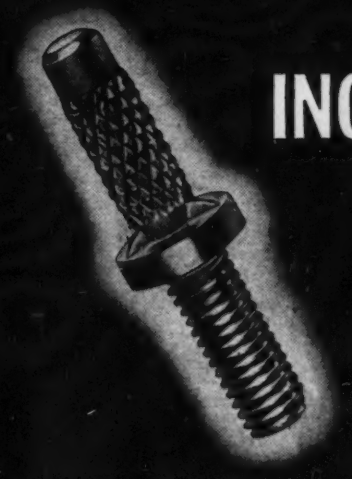

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pleted and other parts of the same product under-cured.

Electronic curing is being applied to the manufacture of other rubber products at Firestone. Large hard rubber wheels that required five hours of curing by steam may be vulcanized electronically in 18 minutes and brake blocks that required seven full hours by steam are done in 48 minutes by electronic heating.

Electronic Package Inspector in Post Office

Fig. 1—Military mail clerks at the right are placing packages from soldiers abroad into the examining bay of the Inspectoscope which x-rays them for illegal objects

X-RAY APPARATUS is now used by the army to forestall such GI activities as the mailing home of a jeep, piece by piece, as reported of one

Fig. 2—Operating position of the Inspectoscope. Contents of the packages are viewed through the window above the panel

enterprising soldier. Installed at The New York and San Francisco Army Post Offices, the equipment has a window through which the homebound parcels are passed as shown in Fig. 1.

While they are on the central counter they are exposed to x-rays under control of the operator at the left whose booth is shown in more detail in Fig. 2. He also acts as observer, examining the packages for illegal material of which considerable has been recovered. Built by The Sicular Co. of San Francisco, the Inspectoscope does away with opening all packages.

Synchronizing Electric Motors

APPLICATION OF ELECTRONIC controls in industry sometimes involves the synchronizing of several operations that are ultimately accomplished by electric motors. A simple method of maintaining synchronism of motors is suggested by the system employed in the SCR-502, a semiportable radio direction finder. It employs two fixed Adcock antenna arrays, one being used from 1.5 to 10 mc and the other for 10 to 30-mc operation. Instantaneous visual azimuths are indicated on the screen of a five-inch cathode-ray tube by means of deflection coils driven in synchronism with motor-driven goniometers that scan the output of their respective antenna arrays.

The motors that drive the goniometers operate on 115 volts, 60 cps, and are single-phase types rated at 1/50 horsepower. The motor that rotates the deflecting coils of the cro is a ¼-horsepower synchronous 1800-rpm unit. The motors are some distance apart and one of the problems is to maintain synchronism between them.

Contactors

Open-close contacts are operated by eccentric cams on the drive shafts of the goniometer and the bearing indicator drive motors. These two sets of contacts are wired in series with one end grounded and the other end terminal connected to the grid of a 6SL7 tube. The contacts are so arranged that the indicator contactor is closed for about 270 degrees of its rotation and the goniometer contacts close for about 30 degrees of rotation during the time the contacts at the indicator are open. In normal synchronous operation, the two contactors never close at the same time and, therefore, the grid of the 6SL7 is not grounded. In fact, the tube is biased to cutoff during the

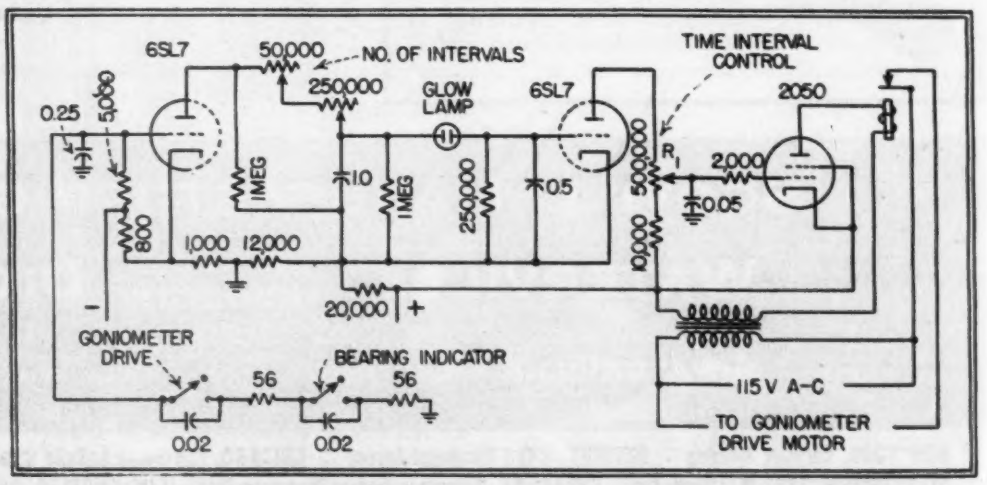
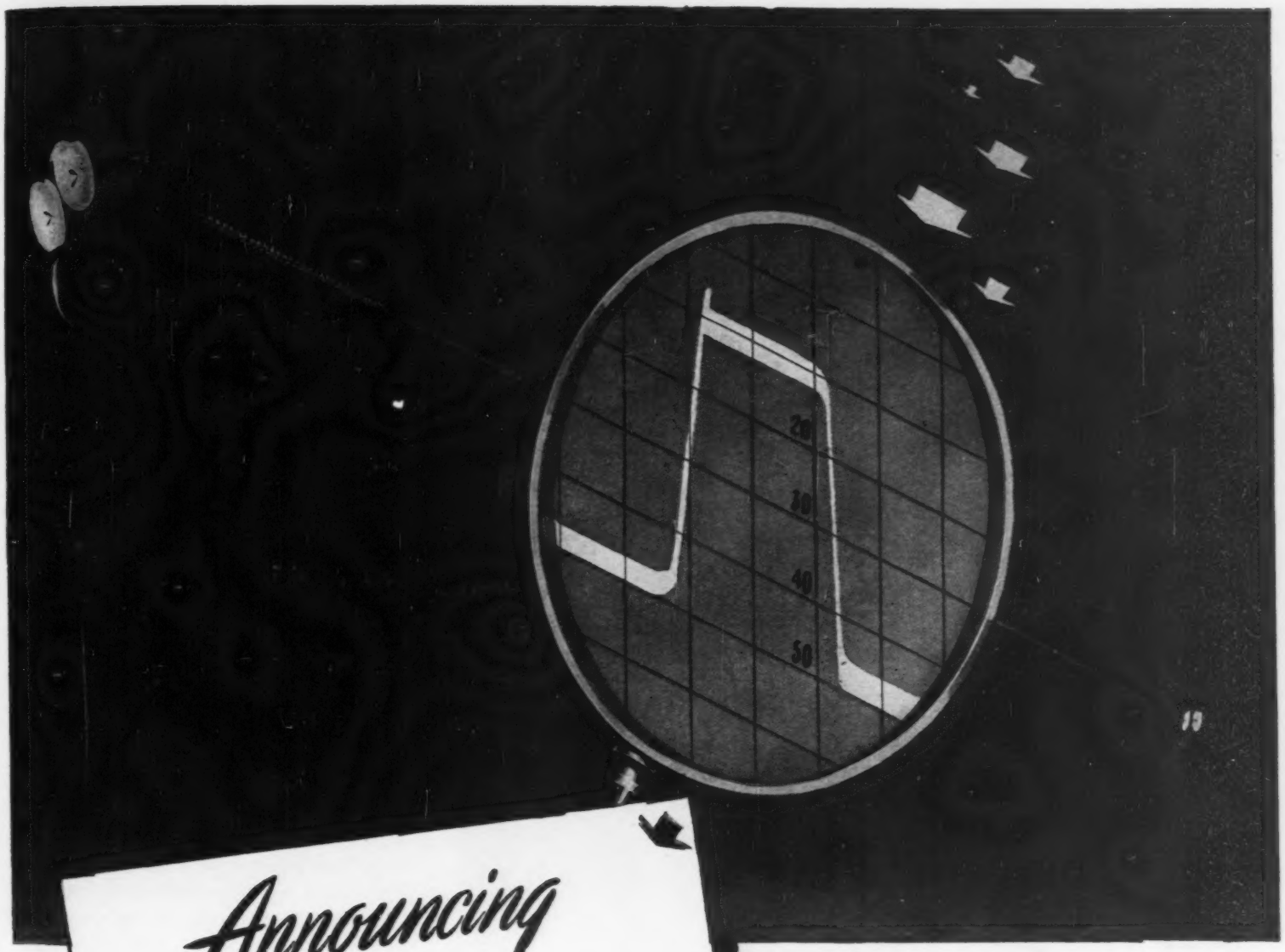


Fig. 1—Circuit of electronic unit for synchronizing electric motors



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synchronous condition.

The circuit of the synchronizing unit is shown in Fig. 1 and consists of a twin-triode 6SL7 and a 2050 thyratron. When the first grid of the 6SL7 is cut off, the grid and cathode of the second triode are at the same potential so that this tube draws current through resistor R_1 and produces bias at the 2050 grid so that it does not draw current. The plate circuit of the 2050 contains a relay with its contacts arranged so that when the 2050 does not draw current, the a-c supply is connected to the goniometer drive motor.

Operation

If the motors are not synchronized on the same pole, then closing of the goniometer contactor will not occur during the open interval of the indicator contactor and, during some part of the revolution, the grid of the first half of the 6SL7 will be grounded. When this occurs, this tube draws plate current. Since each time the grid is grounded the tube draws current, pulses of voltage will appear across the load resistor and an associated RC circuit. After several pulses, the number depending upon the setting of the *No. of Intervals* control, the potential across the capacitor reaches the flashing voltage of the glow lamp.

When the glow lamp conducts, current flows through the resistance in series with it, overbiasing the second triode of the 6SL7. With this tube cut off, bias is removed from the 2050 which fires and operates the relay in its plate circuit, opening the power line to the goniometer motor. As the negative bias leaks off the second tube, its plate current rises and ultimately the thyratron is again shut off. The time interval during which current is cut off the goniometer motor depends upon the setting of the *Time Interval* control.

With the power circuit to the goniometer motor open, this motor slips behind the rotation of the indicator drive motor until the series contactors are again situated so that the series circuit in which they are connected no longer grounds the grid of the first half

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Type of Cooling	water

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of the 6SL7. In other words, the synchronizing cycles continue until the two motors are in step. When the controls are properly set, there will be one, three, or five synchronizing cycles requiring approximately one to three seconds.

• • •

Control and Recording with Floating Grid

By E. L. Deeter

Naval Ordnance Laboratory
Washington, D. C.

THE BLOCKING EFFECT of a floating grid charged to a high potential has long been known but until recently little use has been made of the phenomenon in electronic devices. Some applications of the blocking effect in connection with a low-impedance input relay and a recording system will be described.

If the control grid of a vacuum tube is disconnected, and the proper voltages applied to the filament and plate, the grid will assume a negative charge. This charge is the result of velocity emission from the cathode, grid leakage current, gas ions and true contact potential. The electrostatic potential of the grid under such circumstances is low, usually around one volt negative. When an alternating potential is applied to the grid through a small ca-

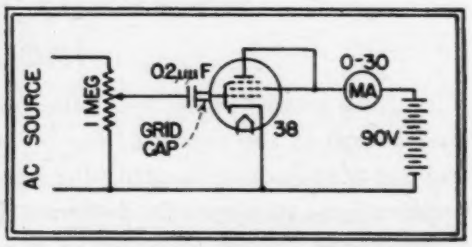

Fig. 1—Basic grid-blocking circuit from which the relay unit and recorder were developed

pacitor, electrons from the grid collect on the inner capacitor plate during the time that the outer plate is positive. During the next half cycle, when the outer plate is negative, the accumulated electrons are trapped in the circuit between the inner plate and the grid. The electron accumulation continues on the following cycles, and the plate current in the tube becomes blocked, the extent depending upon the potential applied to the grid and the leakage resistance from grid to

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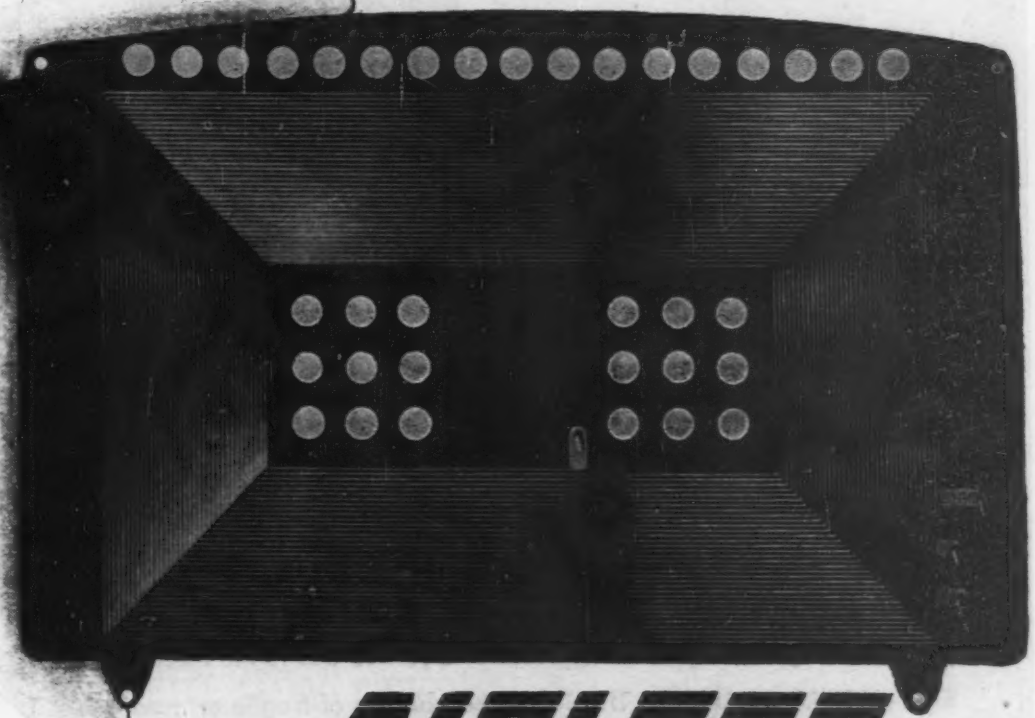

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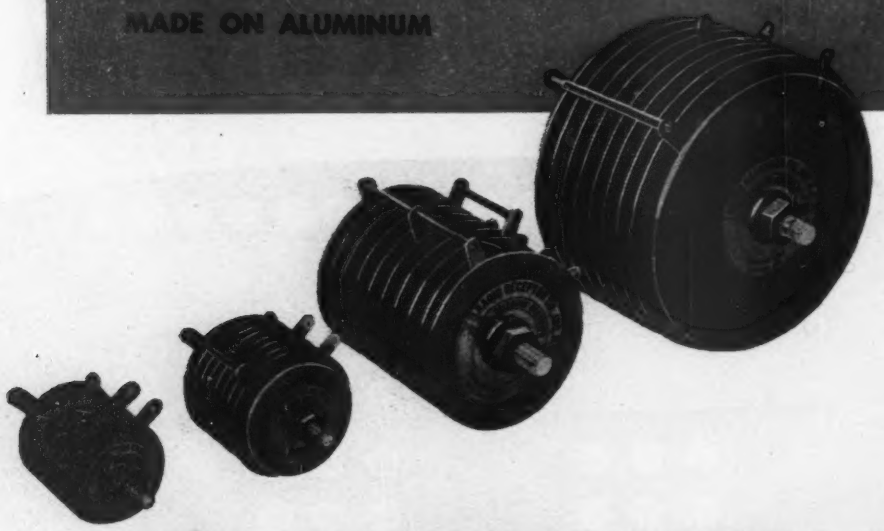

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cathode or ground over the grid supports and surface of the tube. Grid-leak detection employs this principle, but with low applied potentials and an essential leakage path of proper value to control the time constant of the blocking effect.

Basic Circuit

When the maximum blocking effect is desired, tubes with low leakage between grid and cathode, or other elements, must be used. Tests revealed the type 38 tube to be superior for this application. With the control grid brought out to a cap on the apex, surface leakage is reduced to a minimum. Since the tube is a power-amplifier type, an adequate supply of plate current is available for control operations.

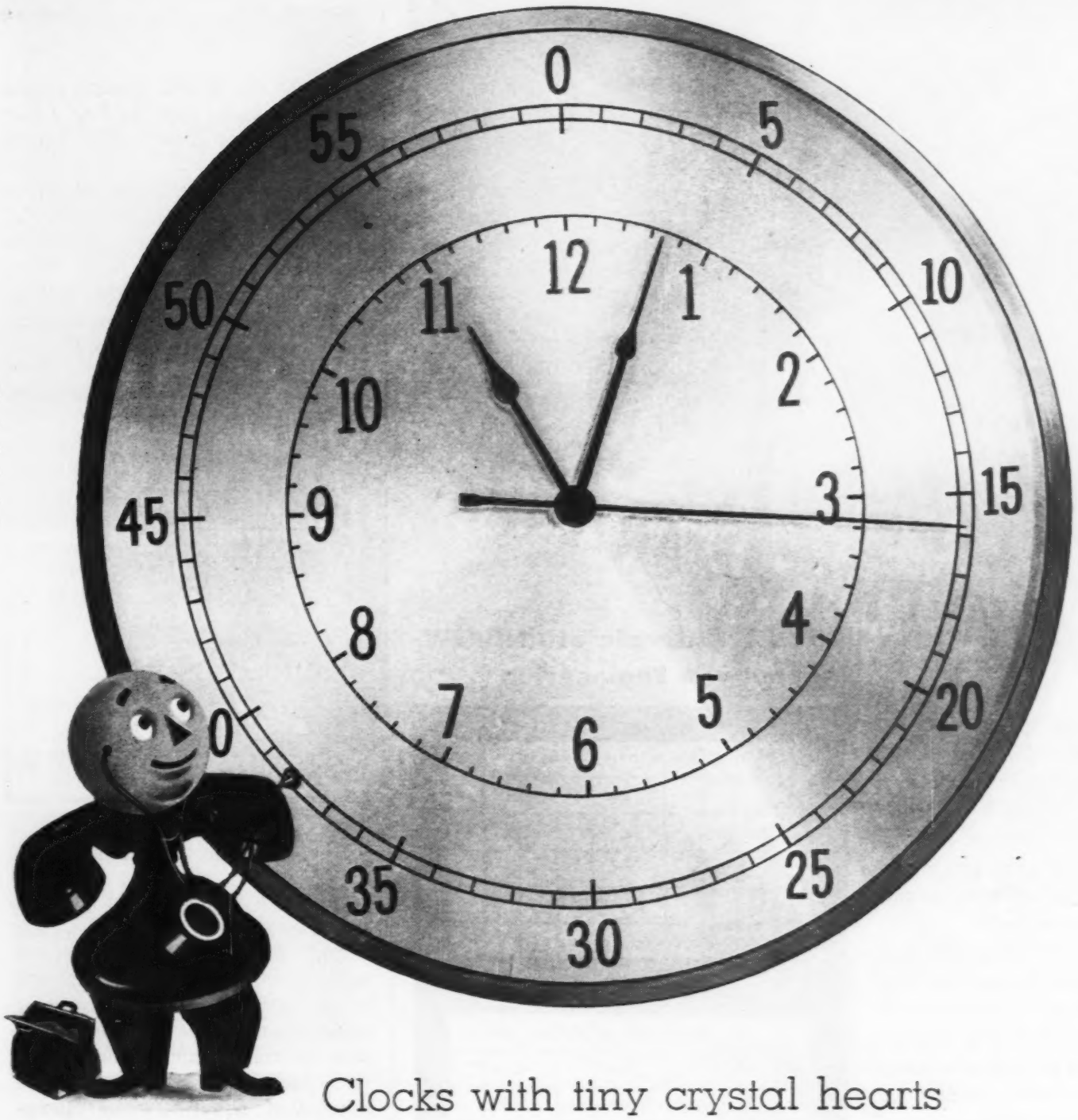
A fundamental circuit is shown in Fig. 1. The special capacitor in the grid circuit consists of two metal plates, ½-inch square and separated about ¼ inch. This results in a capacitance of about 0.2 μf. One plate is attached to and supported by the tube grid cap and the other plate is supported by other means that assures a high-resistance path between grid and cathode or ground.

A variable source of 60-cycle potential is applied between the outer capacitor plate and cathode. The tube and capacitor plates are enclosed in a metal cabinet about eight inches square, to shield them from alternating potentials other than those purposely applied, as well as capacitance changes introduced by moving objects near the unit.

Operating Point

Figure 2 presents the plate current vs applied potential. Capacitance existing between the tube grid capacitor and metal cabinet, as well as the d-c plate potential, will affect the shape of the curve, which will also straighten to some extent when a load resistor is inserted.

With the set-up described, the plate current changes rapidly with applied potentials between 10 volts and 80 volts, rms. Transfer characteristics taken on the tube in the metal cabinet indicate an equivalent change in plate current when a d-c negative bias from 1 to 5 volts is used. Alternating potentials over



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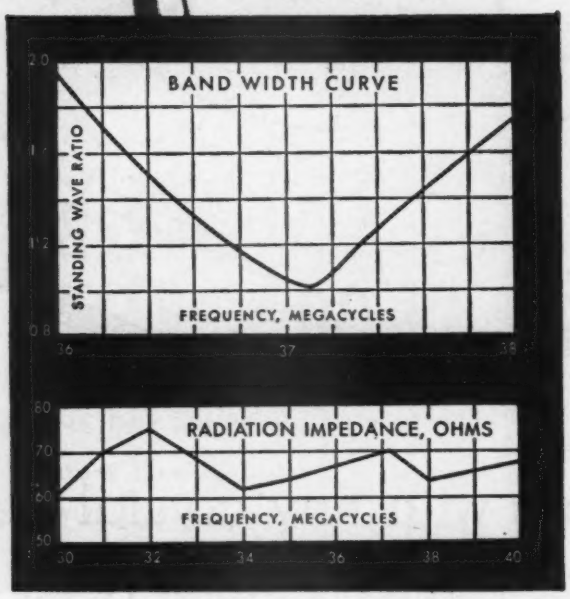

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80 volts cause less change as the value increases and the lower part of the curve becomes a nearly straight line.

If an a-c potential of about 60 volts (point *A* in Fig. 2) is applied as a bias potential, the swing of the plate current will be fairly linear over the slope of the curve, with an increase or decrease of capacitance in the grid capacitor, while the

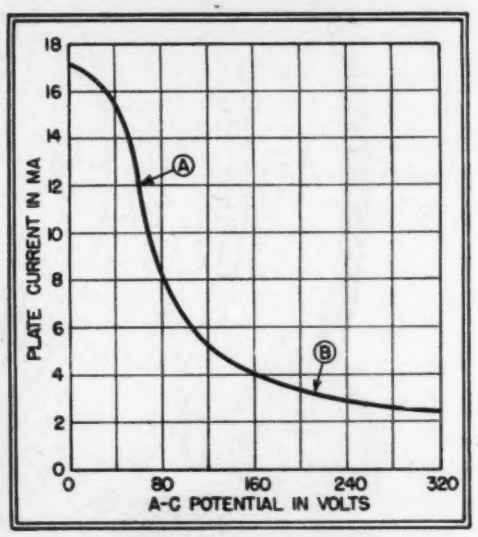

Fig. 2—Remote cutoff characteristics obtained with circuit of Fig. 1

plate current remains on the relatively straight portion of the upper slope. Furthermore, the capacitance change of the grid capacitor is proportional to the distance between the metal plates. Thus a variable controlling current is obtained from the plate circuit of the tube, a total of about five ma change, when very small capacitance changes affect the grid.

Mechanical Features

The small value of the grid capacitor makes it possible to use small lightweight plates on the device, and the movable plate may be supported by a d'Arsonval meter indicating arm, or the arm itself may be used as the variable plate. In this instance, the arm should pass close to the grid-cap plate, which may be a short piece of wire, parallel to the moving arm at maximum capacitance.

The plate current change may be increased by mechanically arranging the meter arm to pass quite close to the grid plate. Conditions may be improved also by increasing the length of the meter arm. This

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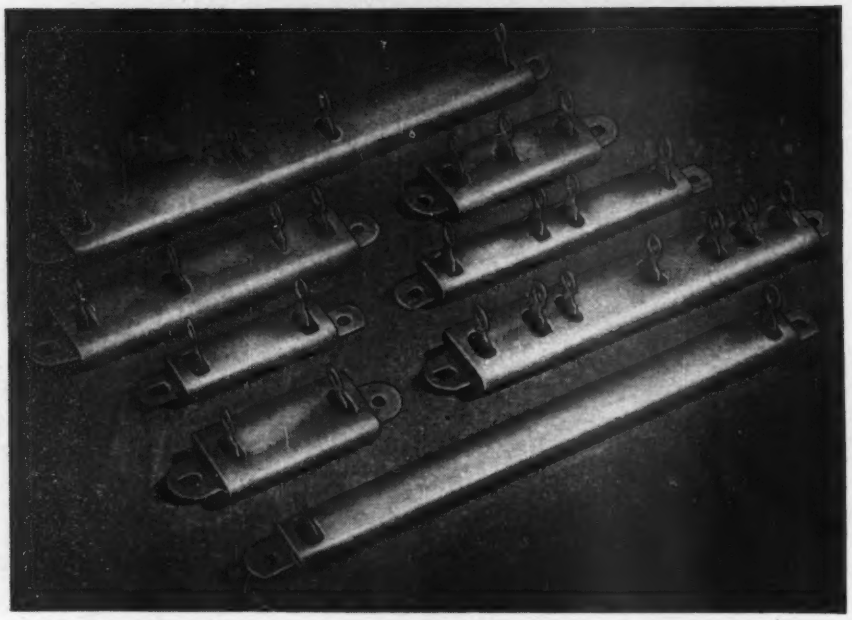

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will increase the arc over which the arm travels, and allow the meter movement to be placed at a greater distance from the tube, reducing capacitance and improving efficiency. The length and weight of the meter arm will be limited by the movement construction, jewel bearings, etc. Any additional weight should be balanced.

The meter arm must be connected to the high potential. Since the arm is electrically connected to one side of the coil in some instances, only certain circuit inputs could be tolerated, and it will be necessary to employ a movement in which the arm is isolated.

The high impedances involved demand special consideration if stability is to be expected. Even slight changes in humidity cause the input impedance to vary due to dielectric changes in the capacitor and leakage over the surface of the tube. Unless shielded, with controlled humidity, the device becomes a virtual hygrometer. Used in the manner described, the capacitor arm has continuous control over the plate current, unless the travel arc of the

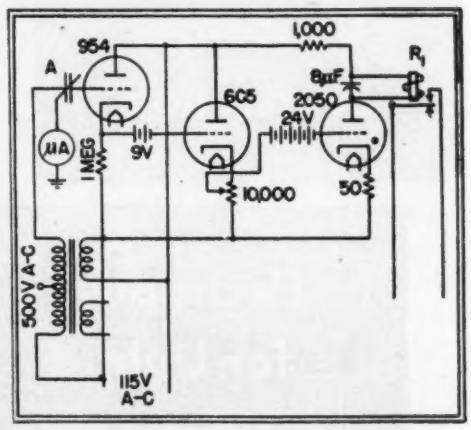

Fig. 3—Improved stability resulted when the three-plate capacitor shown at *A* was used in this control circuit

arm is great. This is not always an advantage in control work, especially when relays are employed and a trigger action is desirable.

More Stable Version

Improved stability and an instantaneous operating point were obtained in a second model by employing a three-plate capacitor in the circuit shown in Fig. 3. One plate connects to the grid and the other stationary plate is connected to the high potential a-c to bias the

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A typical "postwar" item is this Silex Steam Iron illustrated. Like so many new products which are a part of our long-awaited civilian economy, this revolutionary home appliance makes prudent use of Durez phenolic plastics.

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Notice the eye-appealing, heat-resistant handle. Exhaustive tests proved that plastics were better suited for this purpose than any other material known. Their use resulted not only in a better finished product but also in a product which could be produced very economically.

Why Phenolic Plastics?

Of all plastics, the phenolics are the most versatile. Naturally, this makes them the logical starting point for the design engineer with a materials problem. The handle of this Silex Steam Iron, for example, called for heat resistance, smart appearance, pleasant and hand-fitting "feel," moisture resistance, and excellent moldability . . . all inherent characteristics of phenolic plastics.

Why Durez Phenolic Plastics?

The more than 300 multi-propertied Durez phenolic molding compounds available today are the direct result of a quarter century's continuing research and product development on the part of Durez laboratory technicians. This rich background and the high quality of the materials themselves are convincing reasons why manufacturers in every field of industry look to Durez for the plastics that fit their jobs.

Experienced Assistance Available

Do you have a plastic materials problem? If so, see your custom molder. Wartime developments have endowed him with many new molding methods and processes so that his services are even more valuable than before. Should further assistance be necessary, the competent advice of experienced Durez service engineers and a wealth of proved product development data await your request.

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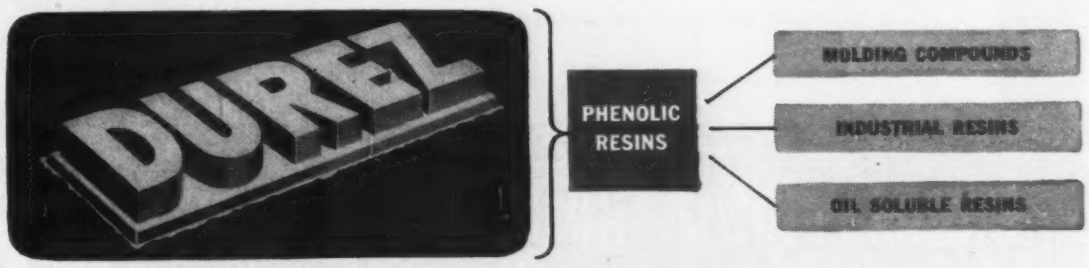
Export Agents: Omni Products Corporation, 40 East 34th Street, New York 16, N. Y.



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MODEL 402 MULTIPLIER

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The important feature of logarithmic scale indication in the Ballantine Voltmeter provides the same degree of accuracy at 1 as at 10. Also the simplicity of this scale reduces errors in visual observation, common with most multirange instruments. Finally, the care taken in overall calibration combined with the inherent stability of the circuits used permits reliable readings within the 2% specified tolerance over the complete range of operation.

Write for descriptive technical Bulletin 8

BALLANTINE LABORATORIES, INC.

BOONTON, NEW JERSEY, U. S. A.

tube. The third plate is the movable arm which is grounded and passes between the outer and inner plates to reduce the capacitance and blocking bias.

The applied a-c potential is increased to a value that reduces the plate current to near cut-off so that the tube operates on the lower part of curve *B* in Fig. 2. Small voltage changes due to imperfect regulation, tube leakage, and capacitance variations become of less consequence.

Trigger action is obtained by virtue of the fact that in this system the capacitance of the grid capacitor is not altered or reduced by the moving arm until it traverses the path directly between the two plates, when considerable capacitance change occurs. This assumes that the plates are closely spaced. Additional stability is secured by employing a cathode resistor of high value.

Compactness was obtained by using the 954 acorn tube and both

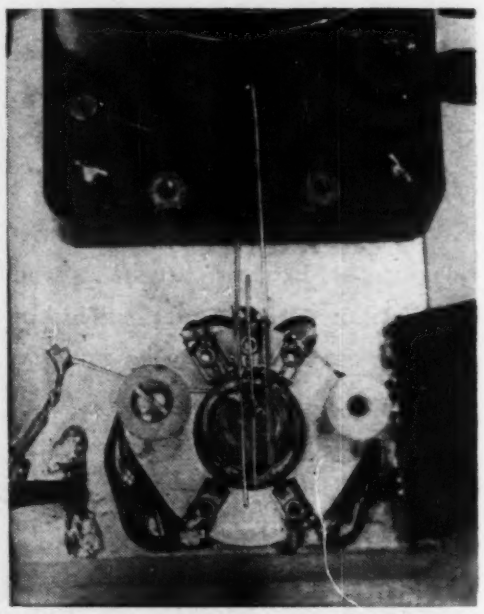
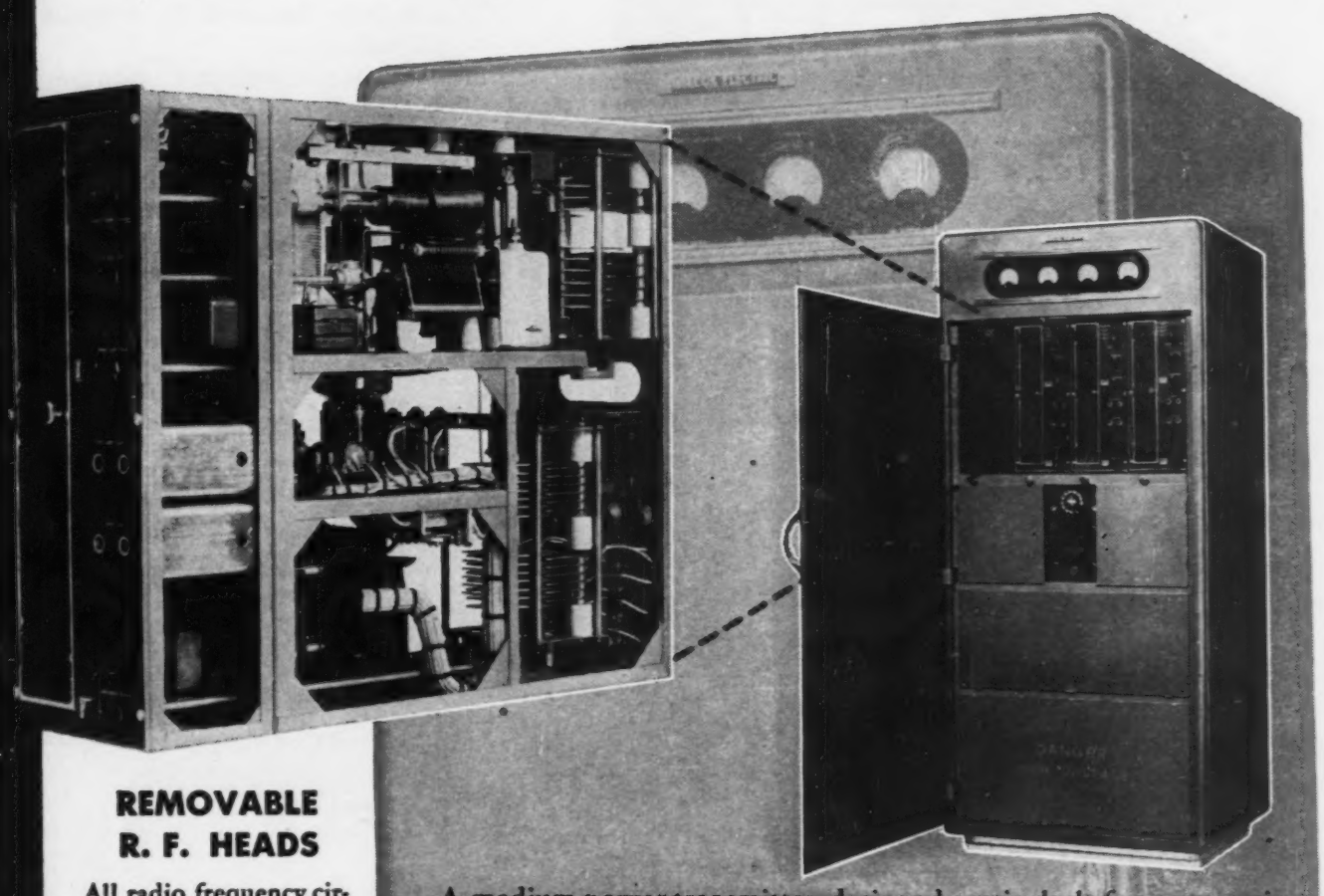


Fig. 4—Details of the three-plate capacitor. Each plate is a short section of wire

the tube and socket are ideal from the leakage viewpoint. The grid capacitor comprises two pieces of tinned copper wire about two inches long. One piece is soldered directly to the grid connection of the tube. A Weston relay, type 534, was modified to operate the variable capacitor. The extended arm is a section of aluminum tubing of the type manufactured for meter indicator arms. The tubing was slipped over

Wilcox Type 99A Transmitter



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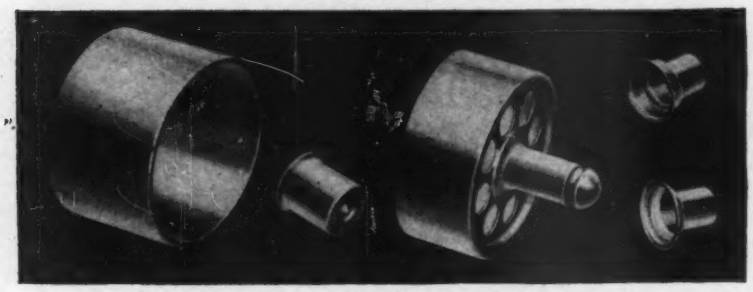

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the balance end of the relay contact arm, after removing the balance coil weight, and the relay case slotted to accommodate the movement of the arm. The end of the extended arm was flattened to increase capacitance.

In this model, the plate current change of the blocked tube has been reduced to a very low value, with increased stability. The plate current change operates a 2050 thyratron and a relay capable of handling a one-kw output is used for control purposes. The input required to secure control operation is around five microamperes.

Temperature Control

The relay could be made to operate from a thermocouple. This system might be operated from the arm of a galvanometer indicating on a scale calibrated in degress of temperature, and the position of the tube and capacitor made adjustable

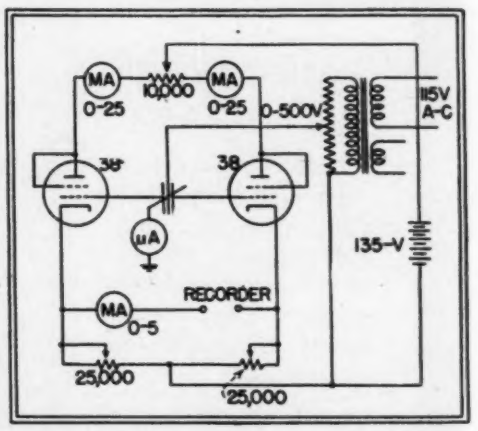

Fig. 5—Balanced bridge circuit used with a recording milliammeter

over the arc to realize operation at any desired point on the scale.

Figure 4, a close-up view of the capacitor plates and grounded arm, shows how the outer plate is offset to allow overswing of the moving arm without losing control. This action is imperative in many control operations. When one side of the input is connected to ground, it will not be necessary to isolate the arm and conventional types of galvanometers may be used with modifications.

An attempt was made to adapt the grid-blocking principle to a low-impedance recording system. The circuit is shown in Fig. 5. This circuit makes use of two type 38

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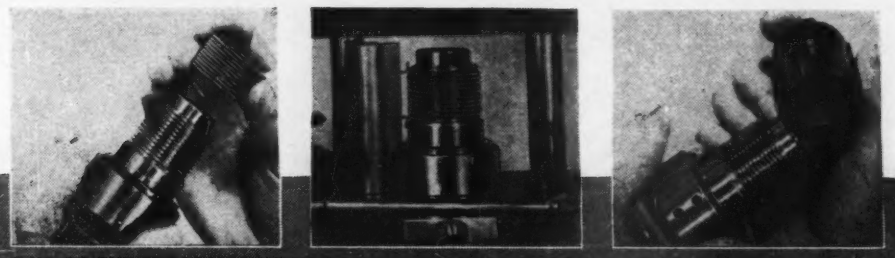

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tubes connected as triodes in a bridge circuit, the blocking potential being applied to a slotted plate to provide a push-pull connection.

Recording Application

The first model constructed as a recorder is shown in Fig. 6. The grid caps are removed from the tubes to reduce capacitance from grid to ground, increasing the effectiveness of the grid wires or plates.

Since a balanced bridge circuit is used, it was assumed that humidity

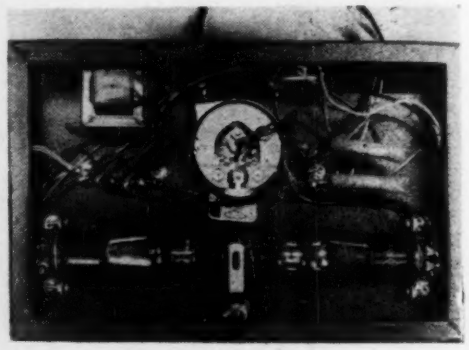

Fig. 6—Mechanical arrangement of tube and components in the balanced bridge circuit. The a-c blocking potential is applied to the slotted plate mounted on a standoff insulator

changes would affect both tubes in a similar manner and the effects would cancel out. The model exhibited promising results; however, lack of adjustment of essential parts prevented attainment of linearity between input and output. Sensitivity was also reduced, because of the high capacitance between grid plates and cabinet.

Improved Model

A third model was constructed with all essential elements made adjustable as to position. The tubes and capacitor unit were isolated in a large cabinet. The power supply and monitoring meters were located on a separate chassis.

The tubes are mounted on a track as shown in Fig. 7 and each may be positioned by means of a screw shaft terminating in an external knob. The meter movement and arm, and the a-c potential plate, may be moved in a vertical direction. The capacitor plates, attached to the grid caps with screws, may be adjusted for alignment with the grounded arm.

Since it was not necessary to util-

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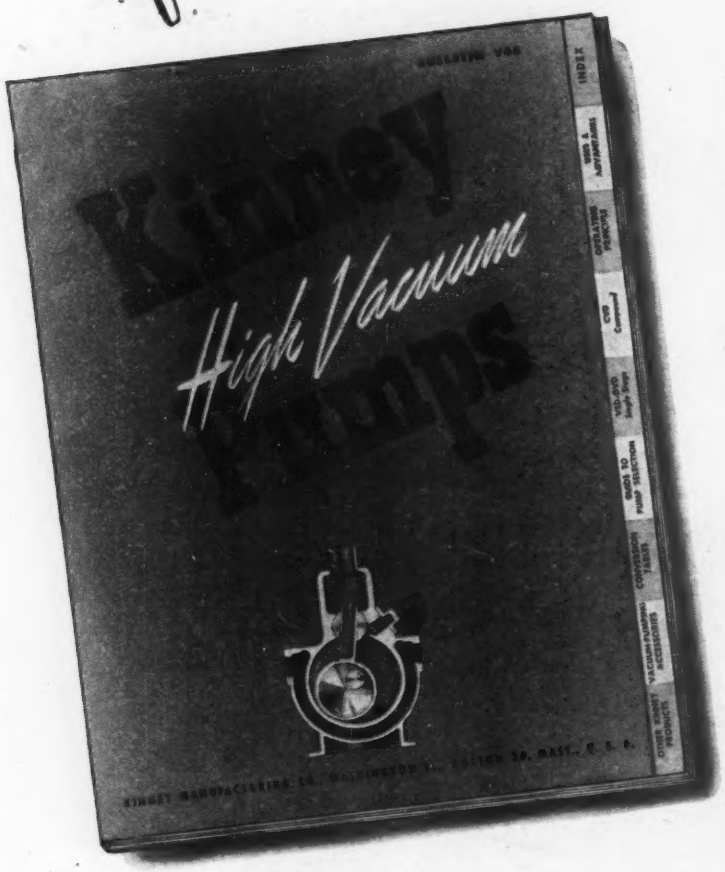

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ize an isolated arm, a precision type of meter movement was used. The permanent magnet was rotated 90 degress with respect to the armature coil, to a vertical position to decrease capacitance to ground.

An Esterline-Angus recorder with a full-scale sensitivity of one milliampere was operated from the output taken from across the cathodes of the tubes.

A systematic analysis, involving complicated formulas, is necessary to arrive at a correct procedure for obtaining linearity, with minimum adjustments. It is perhaps more logical to adopt the trial-and-error method, in steps as follows:

(1) If the recorder is to be used to register positive and negative peaks, with zero center, the

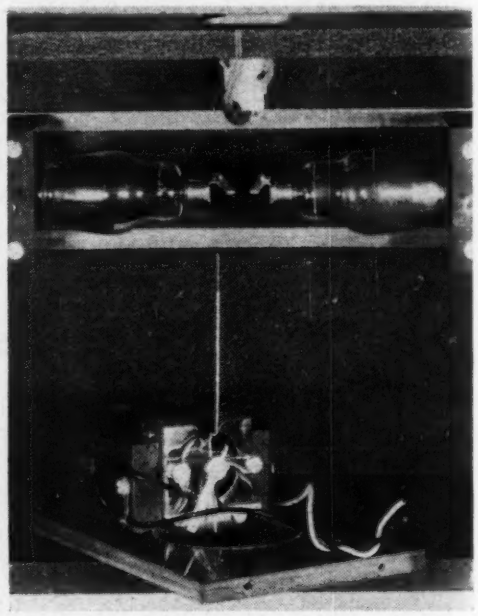

Fig. 7—In the final version, the tubes were mounted on a track and the meter movement made rotatable so that stray capacitance effects could be compensated

first step involves centering the grounded arm on the blocking recorder, as well as the Esterline-Angus writing pen.

(2) Adjustment should be made to obtain a plate current balance in the two tubes.

(3) An input to the arm movement is applied sufficient to deflect the grounded arm to a position over each capacitor plate in turn. While the arm is over a plate, the a-c potential is adjusted to give full-scale deflection on the recorder.

(4) Adjustments are made on applied potential, position of poten-

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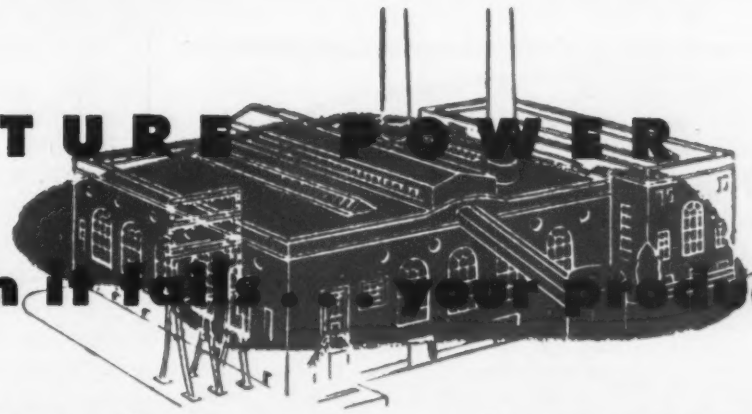

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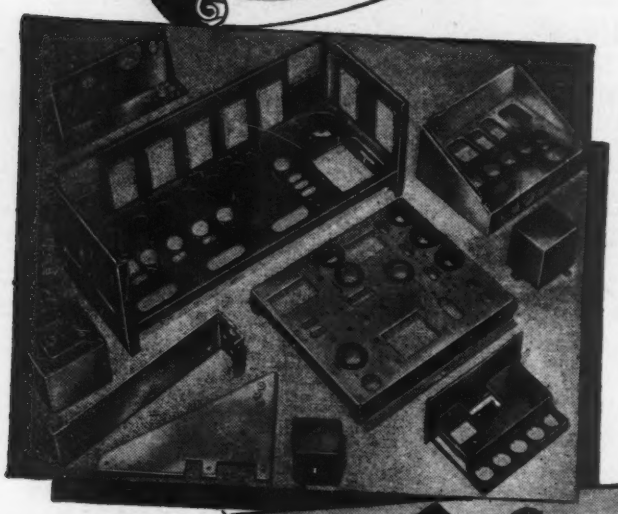

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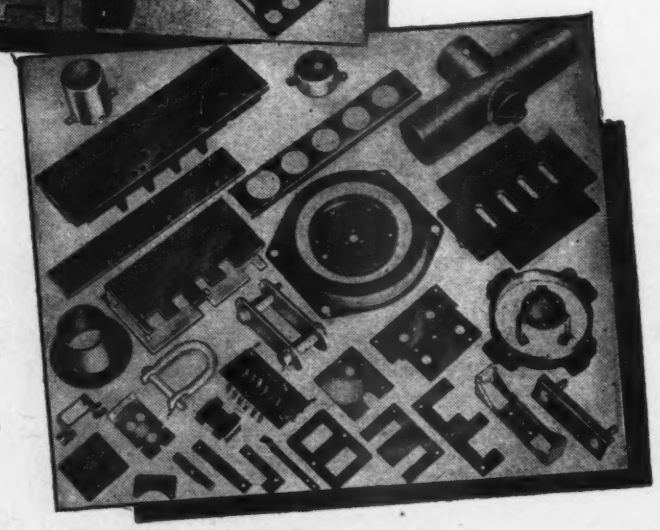

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tial plate and distance between the moving arm and grid capacitor plates. By following the method outlined, good linearity was obtained.

The sensitivity of the recorder will depend entirely on the sensitivity of the movement employed for moving the arm.

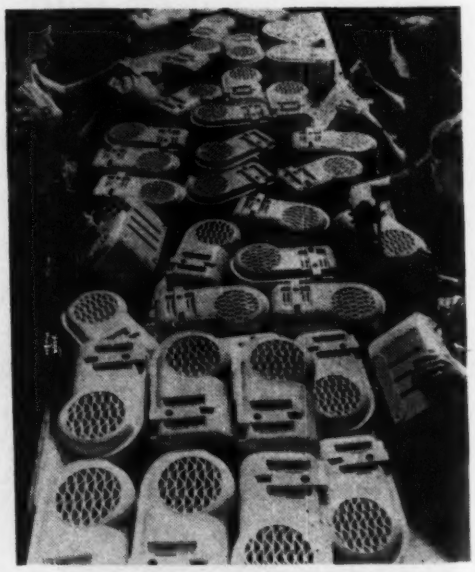
The period or resolving power of the system is dependent on the period of the meter movement and the inherent period or time constant of the electrical system which is associated with the blocking principle. With the model described, the resolving power was equal to that of the recorder, about one second for full scale deflection.

• • •

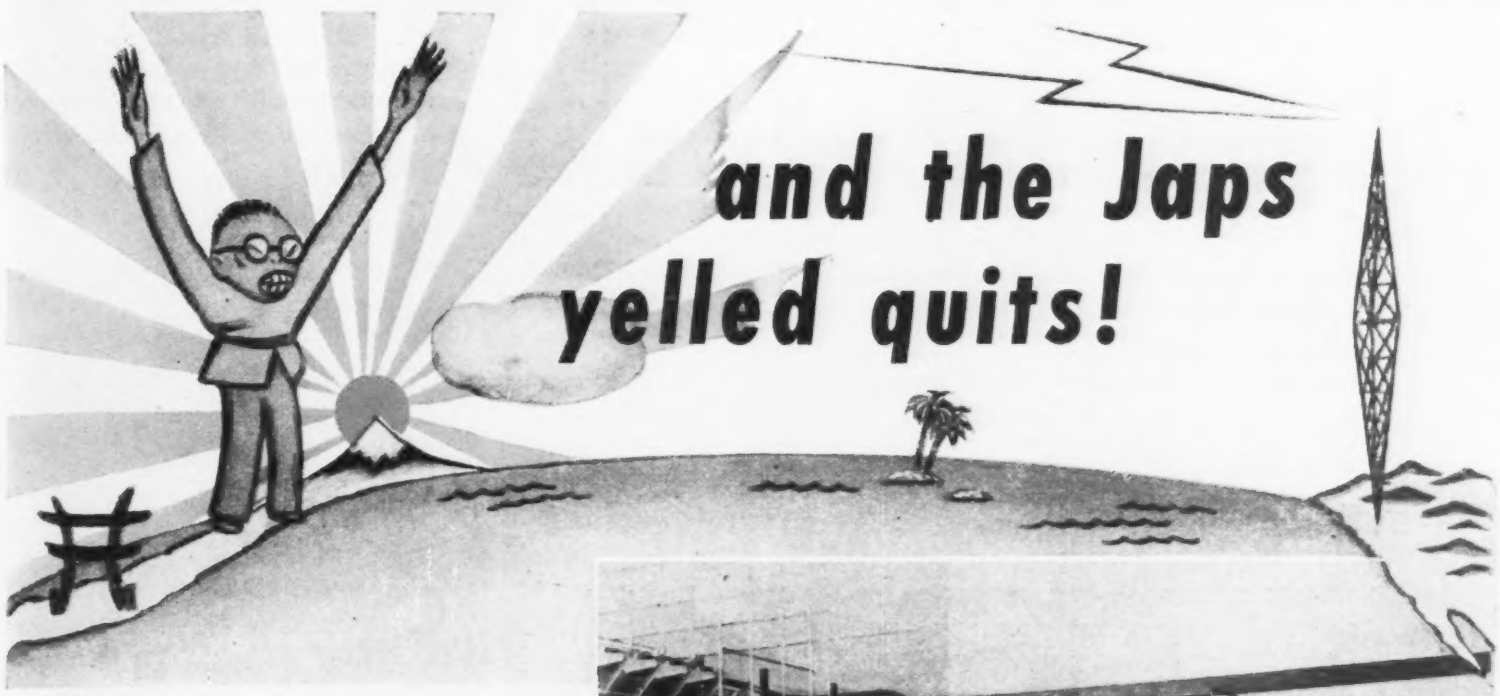
SCR-545 Radar

THE ATTENTION of the editors has been called to the fact that certain specifications of the SCR-545 Radar listed on page 118 of the November 1945 issue of ELECTRONICS are in error, or not computed on the same basis as the data listed for other equipments. The correct values are: weight 26,000 pounds; frequency 2700-2900 mc (t); receiver noise figure 15 db (t); maximum range 46 statute miles (s), 23 statute miles (t); range accuracy 18 yards (t); angular accuracy 1 degree (s). The SCR-545 equipment has completely automatic circuits for tracking in range as well as in angular coordinates.

RECONVERSION CABINETS



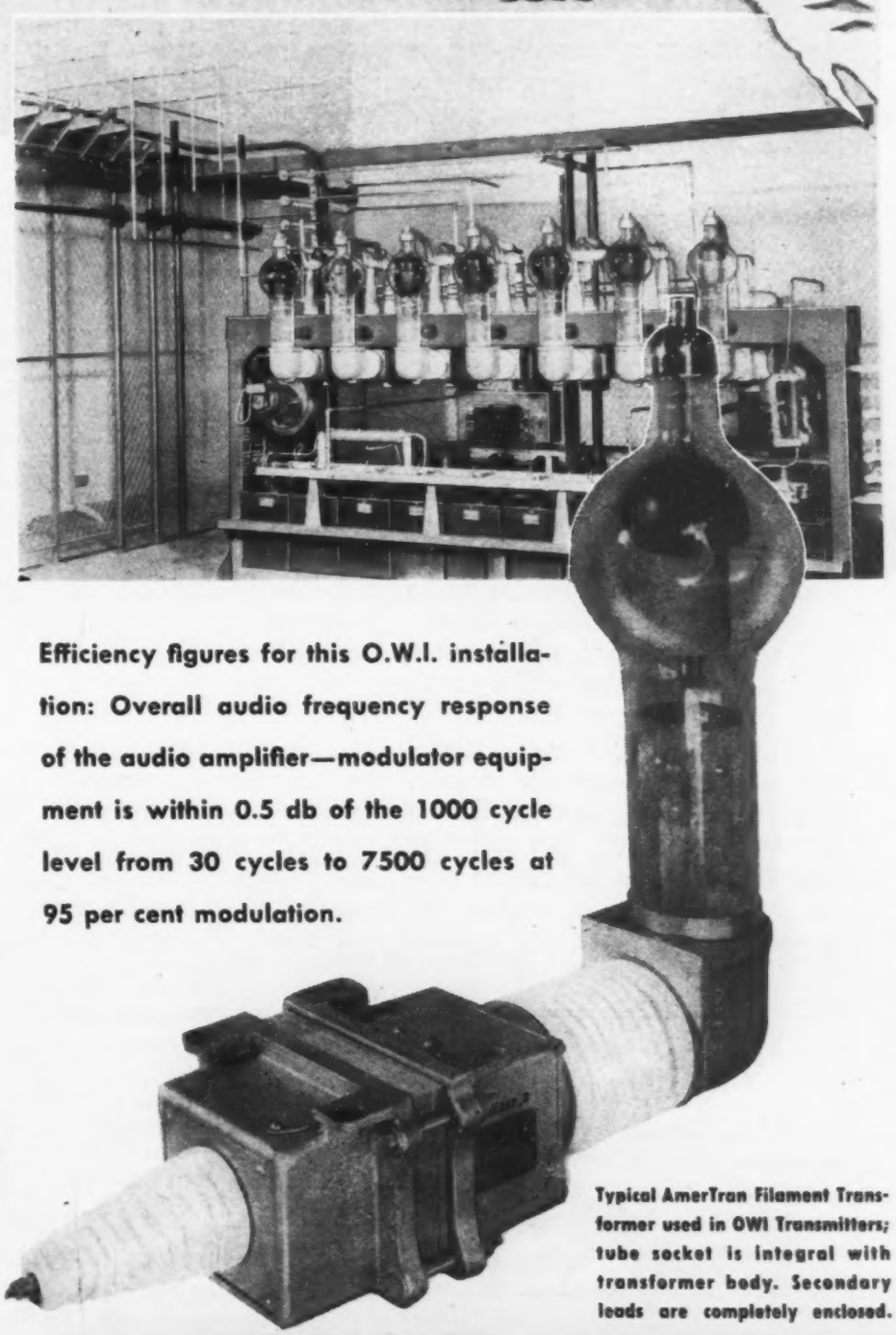
Plastic cabinets roll off the assembly line at Firestone Tire & Rubber Co. where plastic helmet liners were made prior to V-J Day



...and the Japs yelled quits!

● The OWI did a neat and decisive job with their 200 KW short wave trans-Pacific radio stations. Their Tokyo broadcasts are credited with a definite part in ending the war in the Pacific.

AmerTran transformers and reactors were used in the rectifier. plate and modulation circuits. The rectifier filament transformers shown here are typical of AmerTran adaptability to a unique problem. Both stations were designed and equipped by Federal Telephone and Radio Corp.



Efficiency figures for this O.W.I. installation: Overall audio frequency response of the audio amplifier—modulator equipment is within 0.5 db of the 1000 cycle level from 30 cycles to 7500 cycles at 95 per cent modulation.

Typical AmerTran Filament Transformer used in OWI Transmitters; tube socket is integral with transformer body. Secondary leads are completely enclosed.

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TUBES AT WORK

Visible Speech Patterns Transmit Intelligence..... 200
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Photoelectric Aid for the Blind..... 204
Remote Indicating Antenna Ammeter..... 210
Electronic Sounds Effect Circuit..... 214

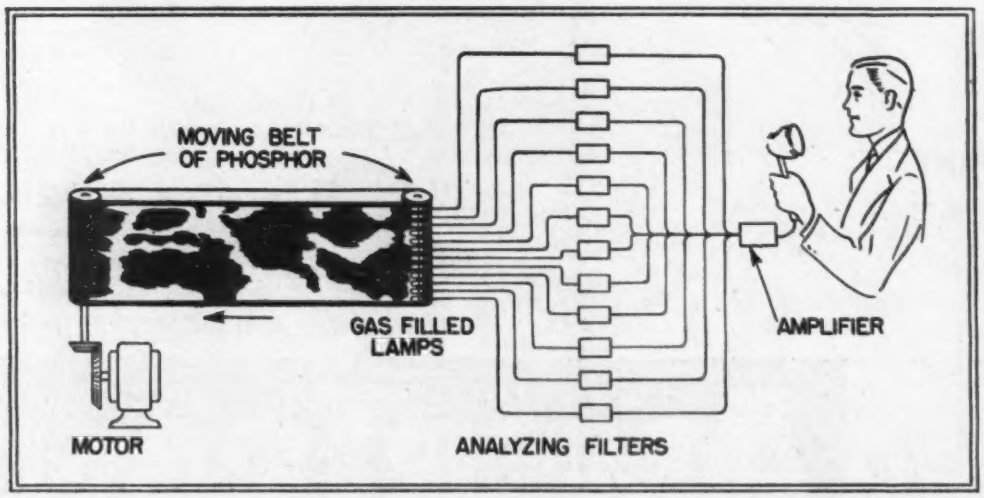
Visible Speech Patterns Transmit Intelligence

AN ELECTRONIC METHOD of making speech sounds visible in patterns that can be read by a totally deaf person was demonstrated recently by Bell Telephone Laboratories. The patterns appear on a moving belt for observation by a large group or on the screen of a special cathode-ray tube for viewing by a small group.

The system involved in the moving belt arrangement is shown in the accompanying diagram. Speech sounds picked up by the microphone are amplified and fed to a number of analyzing filters or cut-off amplifiers. These operate at different bands of frequencies up to about 3000 cps and feed their output to gas-filled glow lamps mounted vertically next to the surface of the belt. The belt is continuous and is coated with a phosphor material held in place by a plastic lacquer. As each lamp brightens and dims with modulation, the belt moves to the left and holds the speech pattern visible until it reaches the take-up drum. When the pattern passes out of sight, it is erased by infrared illumination to leave a new surface for following patterns.

Typical sound patterns produced by the instrument are also illustrated. That these can easily be read by persons trained in interpreting them back into intelligence was demonstrated by personnel of the Laboratory. An engineer who has been deaf since birth watched the sound patterns and repeated aloud members and words which had been spoken by members of the audience. Although he has practiced lip reading all his life, he has become more proficient in reading the visual representation of speech after ten months practice than in lip reading.

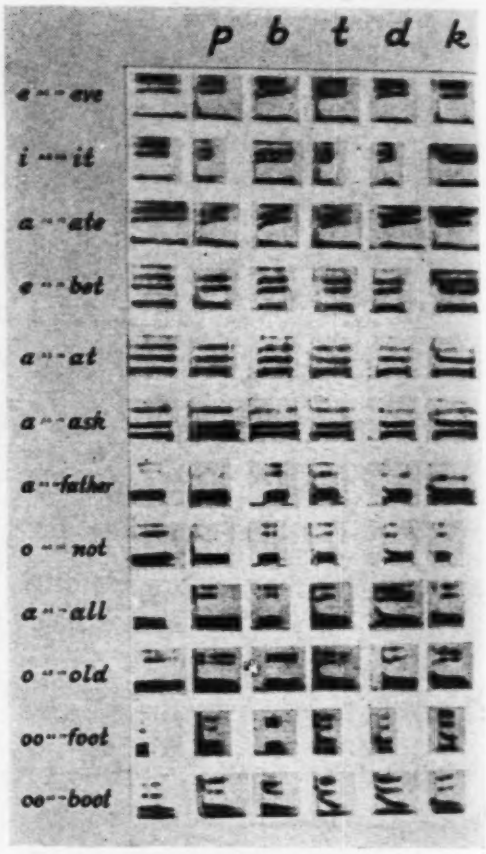
A serious problem to the deaf person is learning to talk. Unable to hear the shadings of his own voice, he usually speaks in a monotone and often his voice is distinctly unpleasant and difficult to understand. This has proved a serious social handicap to the totally deaf. The man who took part in the demonstration has greatly improved his ability to talk, and has also learned to add emotional color to his speech.



Basic elements of the translator that converts speech sounds into visible patterns on a moving belt

Special C-R Tube

For observation by an individual or a small group of people, another instrument that contained a special cathode-ray tube was shown. The tube contains a continuous band of phosphor on the inside of the glass envelope and the tube revolves slowly to expose this band to the viewer. The electron beam sweeps the phosphor band vertically; synchronized with the sweep is a selector which gives each filter control of the intensity of the electron stream as the latter reaches the appropriate point in its sweep. Hence the beam, as it passes a point where a filter is passing no current, is extinguished and the tube surface is dark. When a filter is passing current, the beam increases and a bright spot is recorded. The spots blend into a speech pattern as the tube rotates and visible speech appears.



Typical speech patterns for representative syllables as they appear on the moving belt or the fluorescent screen

Another type of visible speech instrument was demonstrated that made permanent records. Speech

wherever a tube is used

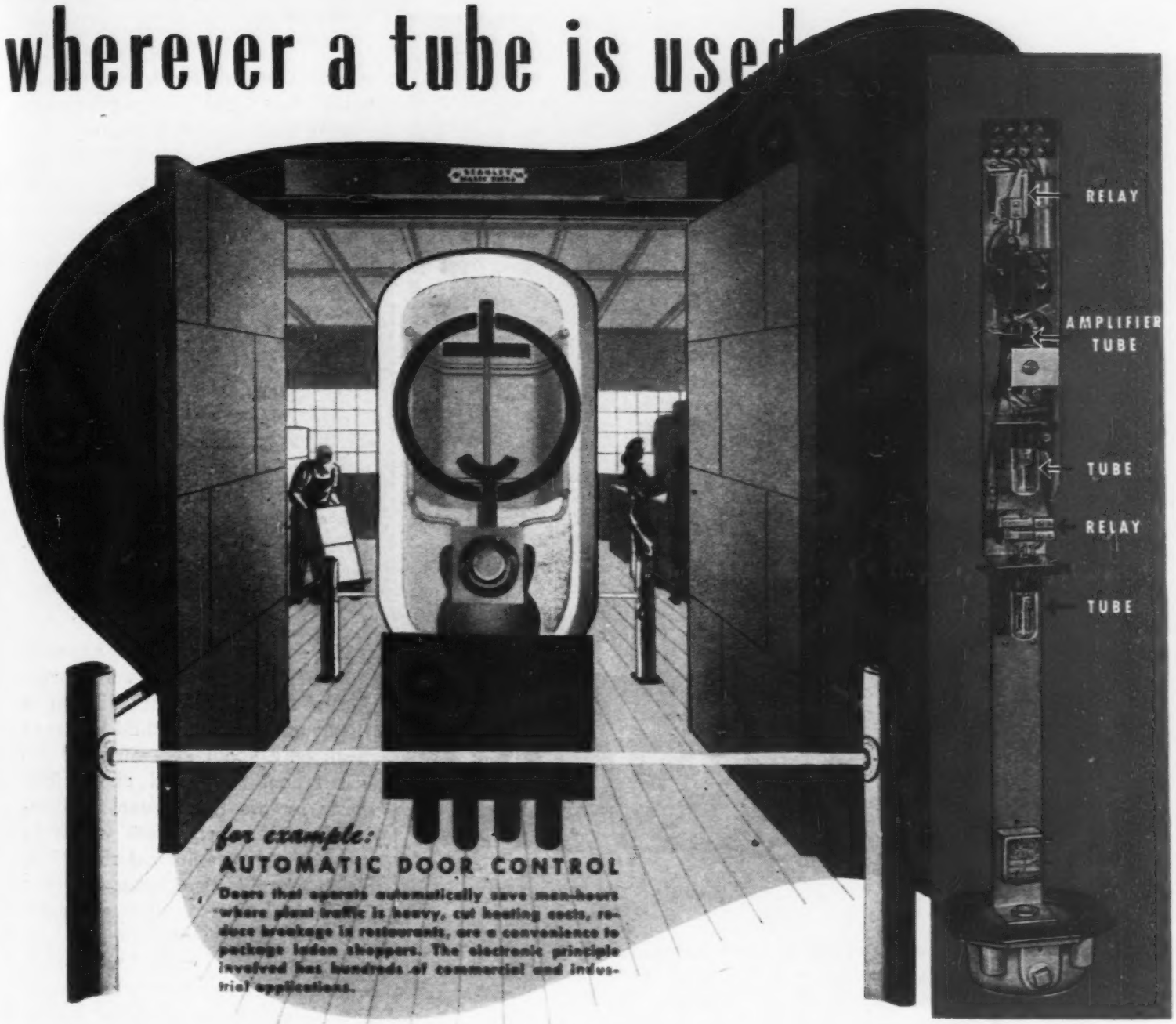

RELAY
AMPLIFIER TUBE
TUBE
RELAY
TUBE
for example:
AUTOMATIC DOOR CONTROL
Doors that operate automatically save man-hours where plant traffic is heavy, cut heating costs, reduce breakage in restaurants, are a convenience to package laden shoppers. The electronic principle involved has hundreds of commercial and industrial applications.

PHOTO-ELECTRIC DOOR CONTROL
Above unit manufactured by General Electric Co., is a part of STANLEY "MAGIC DOOR" CONTROLS.

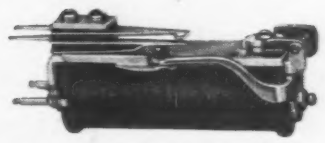
THERE'S A JOB FOR

Relays BY GUARDIAN

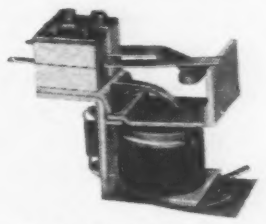
★ The "Magic Door" made by The Stanley Works of New Britain, Conn., uses a General Electric control unit which operates automatically at the approach of a pedestrian or vehicle. In this unit a beam of light focused on the cathode of a phototube causes a tiny current to flow. Enlarged through an amplifier tube this current operates a sensitive telephone type of relay such as the Guardian Series 405. Another phototube with an auxiliary relay, Guardian Series R-100, is employed to hold the doors open for anyone standing within the doorway.

The telephone type of relay is extremely sensitive and able to operate on the small current supplied through the electronic circuit. The auxiliary relay, Series R-100, is required to handle a greater current. It is a small, efficient relay having a contact capacity up to 1 KW at frequencies up to and including 28 megacycles. Contact combinations range up to double pole, double throw. Standard coils operate on 110 volts, 60 cycles, and draw approximately 7 V. A. Coils for other voltages are available. For further information write for Bulletin R-6.

Consult Guardian whenever a tube is used—however—Relays by Guardian are NOT limited to tube applications but are used wherever automatic control is desired for making, breaking, or changing the characteristics of electrical circuits.

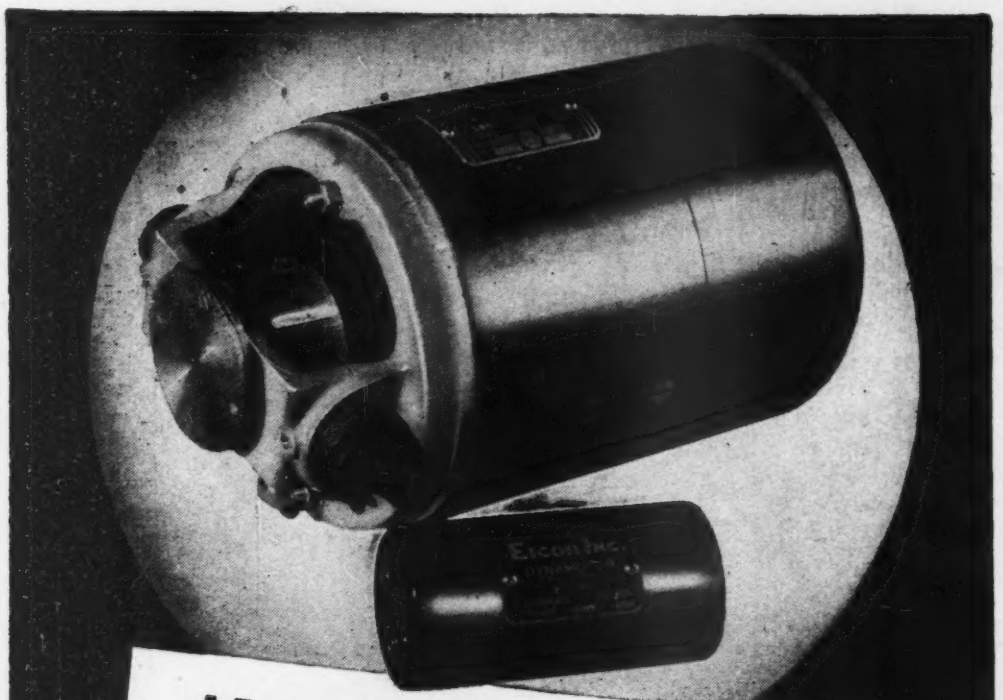


Series 405 Telephone Type Relay



Series R-100 H. F. Relay

GUARDIAN ELECTRIC
1625-A W. WALNUT STREET CHICAGO 12, ILLINOIS
A COMPLETE LINE OF RELAYS SERVING AMERICAN INDUSTRY



A roster of all Eicor products, in their various types and sizes, shows an astonishing number and diversity. But of special interest to users of rotary electrical equipment is our ability to produce units unusual in design or performance . . . and do it quickly, accurately, and at reasonable cost.

Serving in an endless list of special applications, these developments include . . . the smallest commercially produced dynamotor, for 10 watts continuous output, in a 2-5/16" diameter frame and weighing only 34 ounces . . . a motor rated 1/5 hp at 3800 rpm for intermittent duty, 2-5/16" in diameter, weight 38 ounces . . . an aircraft inverter to supply output of 100 va, 400 cycle, single or three phase, in a 3" frame and unit weight of 5¾ lbs. . . . a .6 hp, 4000 rpm, intermittent duty motor, 4" in diameter and 9½ lbs. weight . . . a dynamotor 4-1/16" in diameter which supplies 32 watts continuous output per pound weight . . . a 12 vdc motor rated ¼ hp at 1700 rpm with 150 in. lbs. lock torque in a 5¼" frame.

These highlights are an indication of what EICOR has done in the past. In the days to come our creative engineering will solve similarly difficult problems involving motors, dynamotors, and generating equipment for industry. Your inquiry is invited.

EICOR, INC. 1501 W. Congress St., Chicago, U.S.A.
DYNAMOTORS • D.C. MOTORS • POWER PLANTS • CONVERTERS
Export Ad Auriema, 89 Broad St., New York, U.S.A. Cable Auriema, New York

was first recorded on a loop of magnetic tape and then repeatedly played back at the same rate as a paper-covered recording drum was rotated. On each play-back, a variable filter was tuned to a different band of frequencies to control the intensity of the blackening of successive strips caused by a stylus contacting the paper. Thus after the tape had made a number of repetitions and the drum an equal number of rotations, there appeared on the paper a pattern of lines which expressed the sound in terms of visible speech. In this machine, as in the others, the filter determines whether the pattern will exhibit either fine detail or broad outline.

Applications

The sound patterns are expected to be of great value in the instruction of the 20,000 deaf children in this country. Such a child learns to speak only about six words in his first school year and about fifty words by his third year, as compared with nearly 3000 words in that time for the normal child. The device should also be useful for linguistic studies. Visible speech records can be printed in books as halftones along with text which points out dialect peculiarities and may ultimately be used for speech improvement and correction and to teach pronunciation of foreign languages. It was found that the women trained in reading speech patterns of English could pronounce French words quite understandably although they did not know the language.

• • •

A Cathode-Coupled Isolating Amplifier

BY EARLE TRAVIS
Chief Engineer, KVEC

THE CATHODE-COUPLED AMPLIFIER whose circuit is shown in Fig. 1 is designed to isolate a monitoring line from the network line. This monitoring line is more than three miles long and is part cable and part open wire.

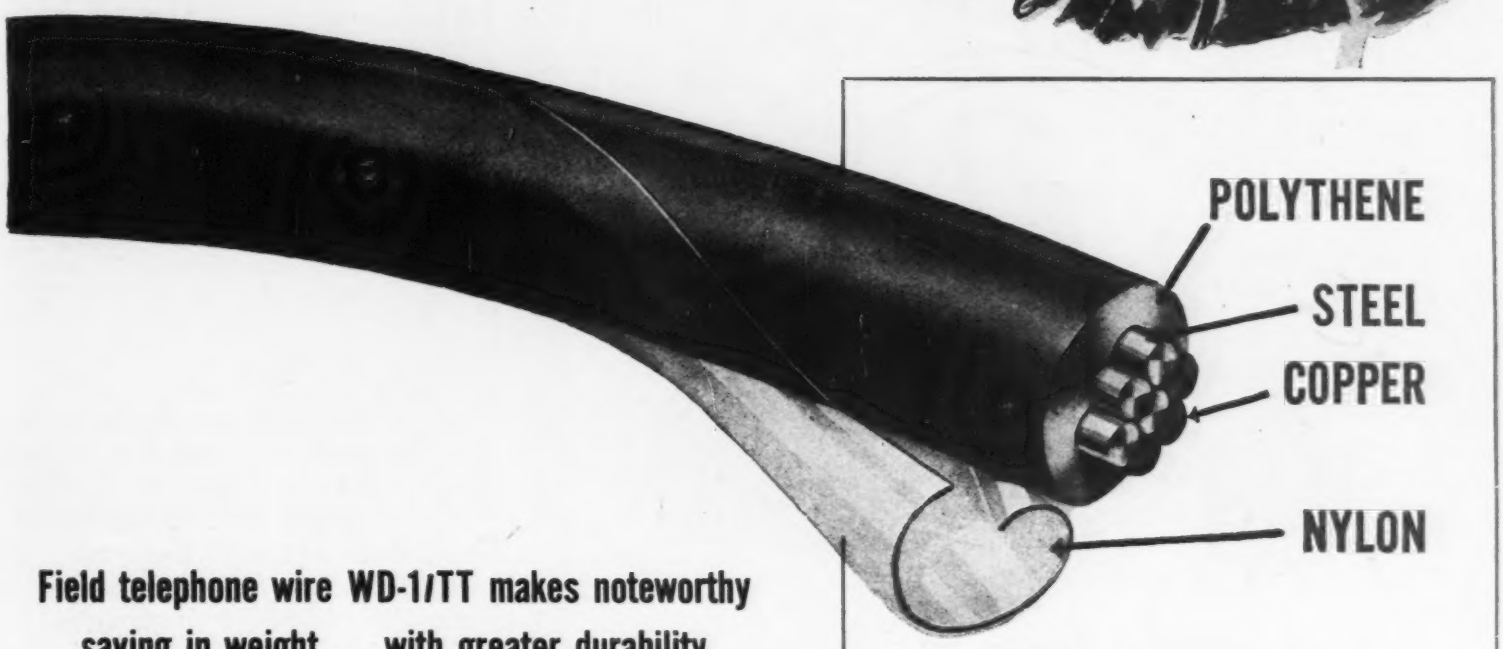
If the cathode resistors are matched carefully, the noise level of

Insulated with

Du Pont POLYTHENE

Jacketed with

Du Pont NYLON



Section of Field Wire WD-1/TT, magnified approximately 12 times. Wall thickness of polythene insulation; 18 mils. Wall thickness of nylon jacket; 5-6 mils. Overall diameter; 80-82 mils. Average weight per mile: 48 lbs.

Field telephone wire WD-1/TT makes noteworthy saving in weight . . . with greater durability

Here's where *two* Du Pont plastics teamed up to fill a recent urgent need of the Signal Corps. Specifications called for a considerably lighter field telephone wire, which would possess the necessary durability as well as a talking range equal to that of the heavy Army field wire.

Du Pont polythene, because of its outstanding electrical properties, needs less weight to do a given job of insulation. Polythene alone is widely used as insulation on some of the lightest types of Army wire. But for this field wire it was necessary to protect the insulation against all kinds of weather, and abrasion from rocks, trees, sharp corners. That's where the toughness of nylon was called on to help. A wall thickness of only 5-6 mils of nylon, extruded rapidly over the polythene insulation, gave the necessary protection—and the finished wire weighs a little more than a third as much as the heavier type of Army field wire—though the talking range is the same.

For information on nylon, polythene and other Du Pont plastics, write E. I. du Pont de Nemours & Co. (Inc.), Plastics Dept., Arlington, N. J.

***Share* in the Victory—Buy Bonds**

FOR PLASTICS.. CONSULT DU PONT

DU PONT

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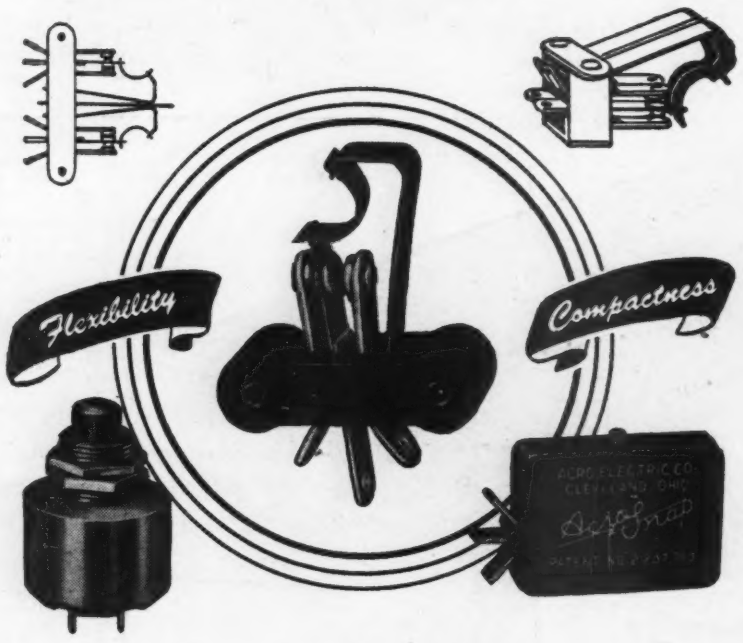
BETTER THINGS FOR BETTER LIVING

. . . THROUGH CHEMISTRY

LOOK to ACRO

for the solution of your *Switch Problems*



Conventional Types or Specials . . .

Hundreds of switch problems are being solved the ACRO way because the ACRO-SNAP engineering design with its patented beryllium ROLLING SPRING permits countless variations in shapes and sizes. With the ACRO, you can design for greater COMPACTNESS and longer life. Mount them vertically, horizontally, or at any angle—singly or in multiples. Pressure ranges from 2 GRAMS to 1¼ lbs. Open blade, button, leaf bracket or fully enclosed types. Ratings: 10 and 15 amps. at 125 volts A.C. Circuits: single pole, normally open, normally closed and double throw. Send us details of your problem now if you want better switch performance with prompt service.

THE ACRO ELECTRIC COMPANY

1316 SUPERIOR AVENUE, CLEVELAND 14, OHIO

the amplifier and line will be satisfactory. The advantages of using a cathode-coupled amplifier for this purpose are; availability of parts, small size and low cost.

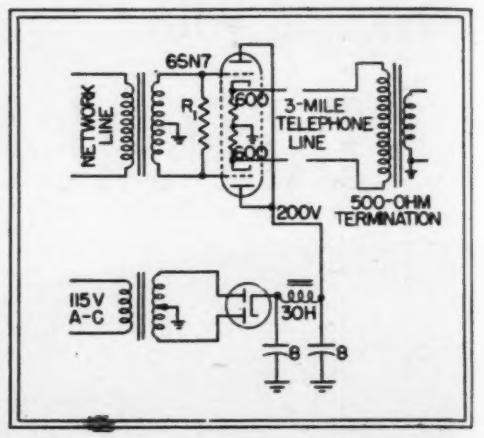


Fig. 1—Isolation of a long monitoring line from a network line is accomplished by this circuit

Note that no coupling capacitors are used. This is possible where the receiving end of the line is not grounded and this is the recommended termination. If a grounded termination must be used, coupling capacitors will be necessary and should be four μf or larger.

The input transformer T_1 is an ordinary inter-stage push-pull transformer. The value of the load resistor R_1 depends upon the transformer. In this case, a value of 75,000 ohms increased the square-wave response from 1,000 to 3,000 cycles.

• • •

Photoelectric Aid for the Blind

AN ELECTRONIC CANE for blind persons that is carried like a lunchbox and is turned from side to side to scan the path ahead has been under development by the Signal Corps. The device projects a beam of light that is reflected by objects within a 20-foot radius. The reflection is detected by a photoelectric cell which creates coded tone signals in an earphone.

Development of the device was initiated by the Signal Corps last April at the request of the Surgeon General. Other development pro-

SYLVANIA NEWS

ELECTRONIC EQUIPMENT EDITION

JAN. *Published by* SYLVANIA ELECTRIC PRODUCTS INC., Emporium, Pa. 1946

NEW T-3 TUBE FILLS NEED FOR SMALLER UNIT IN TINY BROADCAST RECEIVERS

For any further details, or questions you may want answered about this tiny, sturdy vacuum tube, do not hesitate to write or call Sylvania Electric Products Inc., Emporium, Pa.

Commercial Version of Proximity Fuze Tube Is Rugged, Has Long Life

Following Sylvania Electric's recent announcement about the sensationally small vacuum tube—originally developed for the now-famous proximity fuze transceiver—have come many inquiries concerning this super-midget.

SET MAKERS ESPECIALLY INTERESTED

Since the commercial version of the "war-baby" is being produced, many set manufacturers are extremely interested in its qualities—with a view toward making radios about the size of the average wallet or package of cigarettes, miniature walkie-talkie sets and other units.

This new tube, then, is being made in a low-drain filament type and is able to operate at 1.25 volts. This takes advantage of a new, small battery developed during the war which, of course, is a further aid in the manufacture of remarkably small radio sets.

WILL BE AVAILABLE FOR ALL TYPES

Future designs of this versatile tube can be incorporated in radios ranging in size from tiny pocket sets up to deluxe receivers. It has a life of hundreds of hours, is rugged and exceptionally adaptable to operation at high frequencies.

SYLVANIA ELECTRIC

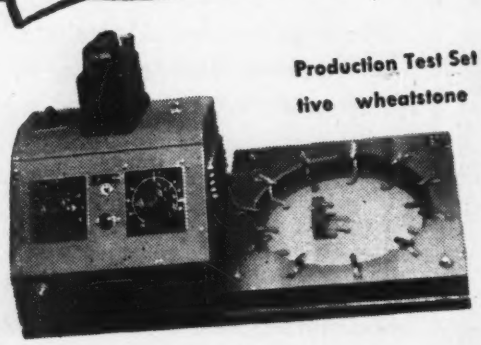
Emporium, Pa.

MAKERS OF RADIO TUBES; CATHODE RAY TUBES; ELECTRONIC DEVICES; FLUORESCENT LAMPS, FIXTURES, WIRING DEVICES; ELECTRIC LIGHT BULBS

Unusual and Vital
ELECTRONIC EQUIPMENT
PRODUCED BY
TECH LAB
SUBCONTRACTING DEPT.

Special Short Wave Transmitter Range: 1-16 megacycles on one dial.

Development of special hearing aid analyzer for Western Electric Company.



Production Test Set to test varisters. A sensitive wheatstone bridge arranged with switching means for quickly checking a number of varisters in rapid sequence.

Manufacturers have continually called upon the modern facilities of Tech Lab Subcontracting Department to assist them in the production of unusual and vital electronic equipment. Our Engineering Department is ready to assist you with your production problems.

Carbon filling machines for precision filling of carbon microphones.

TECH LABORATORIES

MANUFACTURERS OF PRECISION ELECTRICAL RESISTANCE INSTRUMENTS

15 LINCOLN STREET, JERSEY CITY 7, N. J.

grams, utilizing both supersonic and radar waves for the same purpose, are also being conducted by the Signal Corps. The experimental model of the photoelectric unit comprises a nine-pound case the size of a loaf of bread. It was designed and constructed by Lawrence Cranberg, a civilian physicist at Evans Signal Laboratory, Bradley Beach, N. J.

Experimental model of the sensory aid for the blind developed at the Signal Corps Engineering Laboratories

Light from a three-watt lamp is focused by a lens into a very narrow ray. When that ray is directed upon any object within 20 feet of the device, a spot of light about two inches in diameter appears upon that object and the reflected light is directed back toward a second lens in the front of the case.

The second lens focuses the reflected ray to a revolving disk mounted in front of a photoelectric cell. A ray reflected from an object at the maximum range of 20 feet would be focused just below the center of the disk, and would gradually move downward on the disk as the reflecting object neared the device.

The revolving disk is used to code the reflected light signal in order to indicate the distance of the reflecting object. It is divided into five concentric rings, each of which contains one or more holes through which the light can shine upon the photoelectric cell.

The inner ring, on which the

BH EXTRA FLEXIBLE FIBERGLAS SLEEVING

MAKE THIS

FLAME TEST

Hold a match under a piece of BH *Extra Flexible* Fiberglas Sleeving. The flame does not burn, char or otherwise affect it. And temperatures much higher than usually encountered electrically are just as readily resisted by the inorganic Fiberglas!

BH *EXTRA FLEXIBLE* FIBERGLAS SLEEVING

2 WAYS BETTER

NON-FRAYING • NON-STIFFENING

FEW electrical insulations can double in brass as heat insulations. Yet so effectively heat resistant is BH *Extra Flexible* Fiberglas Sleeving that actual service records show it refuses to burn even in direct contact with heat units. The reason—both yarns *and* impregnation are non-inflammable!

A special gum base and dye applied by an exclusive BH process is responsible for many more features. It permanently *prevents fraying, stiffening and abrasive wear.* The sleeving is unusually flexible and takes the roughest handling without fraying. It does not harden and crack with age—lasts indefinitely without deterioration. It is also non-crystallizing at low temperatures.

Fiberglas is non-absorbent and unaffected by moisture, oil or grease—qualities ideally suited to appliance manufacture for instance. And it has high dielectric and tensile strength.

"Punishment" tests prove that BH *Extra Flexible* Fiberglas Sleeving is the most logical insulation for a host of tough jobs. Why not see for yourself? It's available in all standard colors and all sizes from No. 20 to ⅝", inclusive. Write for samples today and compare!

BH SPECIAL TREATED FIBERGLAS SLEEVING UNAFFECTED BY HEAT UP TO 1200°F!

This is a high quality sleeving that will not fray when cut and withstands heat up to 1200°F. Yet no saturant is used in the exclusive BH process! Flexible as string, too. Made in natural color only—all standard sizes. Try it!

BH PRODUCTS

ALSO MANUFACTURERS OF ELECTRIC HEATING COILS • FLEXIBLE METALLIC TUBING
FLEXIBLE VARNISHED TUBING • SATURATED AND NON-SATURATED SLEEVING

BENTLEY, HARRIS MANUFACTURING CO.

Dept. E, Conshohocken, Penna.

JOB AHEAD

". . . as brilliant Sol through stubborn clouds his dauntless will imposes"

RECONVERSION to national peacetime economy is on the march . . . perhaps not as swift as you and we would like it. Vast displacements, inevitable as we "shift our gears", must be absorbed and neutralized . . . in short, there's a JOB ahead.

Each day supplies of famous CORWICO Wire, so important in the war and so important now, will be made available for civilian use.

Patience! We've moved mountains before . . . . . .

CORWICO

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WIRE COMPANY, INC.
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"Made by Engineers for Engineers"

ray would be focused if the reflecting object were from 17 to 20 feet distant, contains only a single small hole. The next ring, on which the ray would be focused if the object were about 11 feet distant, contains a single large hole which is shaped like a crescent and follows the curve of the disk. The next three rings, upon which the reflection would be directed from distances of eight, five, and three feet respectively, contain successively two small, a small and a large, and three small holes.

A reflection from an object five feet distant would be directed upon the fourth ring, which contains one small hole and one large hole. As the disk revolves, at a rate of once a second, the light would shine through upon the photoelectric cell once briefly as the small hole came in line with the ray and again for a longer time when the larger hole intersected the ray.

As the light penetrated to the photoelectric cell, the cell would produce one short and one long impulse each time the disk revolved. These two bursts would create tone signals—one short and one long—in the earphone.

Operation

The handle of the case is parallel to the direction of the first light ray, and the blind user can sense the position of his hand in determining the direction from which reflections are received.

It has been found that a blind person with little practice easily determines the exact distance, to within one foot, of obstacles detected by the device. While a reflection from an object 11 feet distant would produce a tone signal of one dash, and a reflection from eight feet distant would produce a signal of two dots, a reflection from an object ten feet distant would produce a strong dash and two weaker dots, and a reflection from nine feet would produce a weak dash and two stronger dots. Distances between the five principle distance signals can similarly be measured by means of intermediate combination signals.

One difficulty encountered was

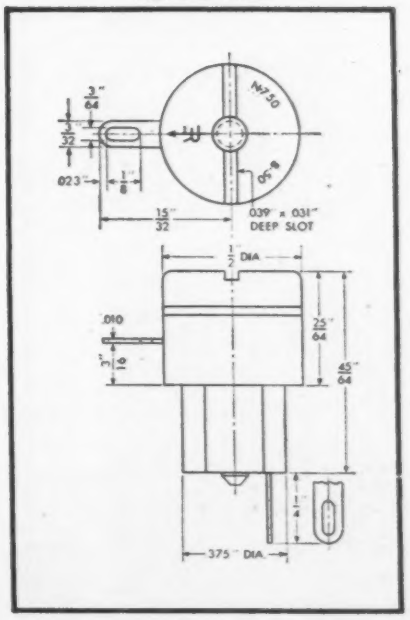
ACTUAL SIZE

Wherever shown, the new Type 554 Ceramic Trimmer has attracted the attention and admiration of the radio industry. First, its original and compact design, its obvious ease of installation and adjustment; later, its demonstrated superb performance—these qualities have not only aroused curiosity and interest, they have won immediate acceptance. Type 554 will be standard equipment on many receiving sets, from now on.

Note, in photograph and drawing, that the metal rotor completely covers the stator track. Contact surfaces of both rotor and stator are lapped, providing a high degree of stability, excluding dust, and keeping noise level to a minimum at high frequencies.

Capacity change is essentially constant per degree of rotation, and full range is covered in 180° rotation. Type 554 Trimmers will be available shortly in production quantities in the following capacity ranges: in zero temperature coefficient, 3-12 MMF and 5-25 MMF; in −750 parts/million/°C, 5-30 MMF and 8-50 MMF. They will also be available in an intermediate temperature coefficient. Trimmers are held firmly in place in a D-hole in the chassis by means of a multiple-tooth spring clip, furnished with the trimmer.

Specifications and capacity ranges are given in the table at right. For complete information, contact our nearest representative or write us direct.



ERIE 554 CERAMICON TRIMMER

Voltage Rating: 350 volts D.C.
Flash Test: 700 volts D.C. for 15 seconds
Initial Q Factor at 1MC: 500 minimum
Initial Leakage Resistance: 10,000 meg. min.

Electronics Division

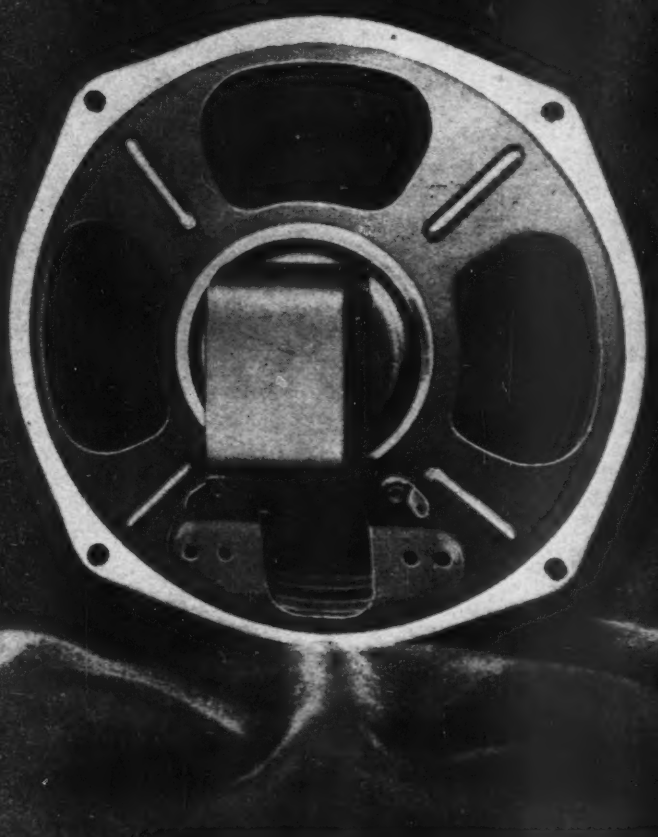
ERIE RESISTOR CORP., ERIE, PA.

LONDON, ENGLAND TORONTO, CANADA.

Big News About Small Speakers

Permoflux leads again with the development of a new permanent magnet dynamic speaker providing maximum performance with minimum magnet weight.

With less than a 1½ ounce Alnico Five magnet weight, Permoflux now achieves performance only obtainable before by using a much heavier Alnico Five magnet.



Setting a new standard for speakers up to 6", this new unit is particularly adaptable to portables and farm radios—in fact to all receivers, battery or power line operated, wherever weight is an important factor.

TRADE MARK
PERM-O-FLUX
PERMOFLUX CORPORATION
4900 WEST GRAND AVE., CHICAGO 39, ILL.

PIONEER MANUFACTURERS OF PERMANENT MAGNET DYNAMIC TRANSDUCERS

the problem of filtering out signals from sunlight and incandescent lamps. This was solved by making the amplifier sensitive only to modulated light, and then modulating the light ray to the exact frequency at which the circuit is most sensitive.

It is hoped that the weight of the unit can be reduced to about two pounds and that it can be sold around the hundred-dollar mark. Although it is not expected to be of value in distinguishing approaching cars, until they reach the 20-foot limit, the ability to determine the color of traffic lights would be a distinct help to the blind user when crossing intersections.

• • •

Remote Indicating Antenna Ammeter

By C. R. Cox
Andrew Co.
Chicago, Illinois

ONE OF THE FCC REQUIREMENTS relating to broadcast station operation is that the magnitude of the antenna current be recorded in a log book at intervals of one-half hour. Since in the majority of standard broadcast installations the antenna is located at a distance of several hundred feet or more from the transmitter, some means of indicating the antenna current at the transmitter is highly desirable. To fill this purpose, the type 708 remote indicating antenna ammeter was developed by Andrew Company engineers.

In designing the instrument, it was thought desirable to avoid the use of thermocouples, because of their susceptibility to lightning damage. While the occasional loss of a thermocouple due to lightning presents no problem to the owner of a broadcast station, the time spent in shutdown, while thermocouples are being replaced, is very serious and expensive because it means loss of revenue to the station. Accordingly, the ammeter was designed on entirely electronic principles, and contains no thermocouples. In several instances, remote ammeters of this type have withstood without damage a direct

No more scenes like this when you use the *new*

HALLOWELL KEY UNBRAKO KIT

Pocketsized, Versatile, Useful, Rugged, Durable

Think how much easier, how much more convenient to have all your bits assembled in one kit, . . . so small that it will fit in the palm of your hand. It saves space, eliminates carrying around a lot of bulky, heavy wrenches and screw drivers, lessens the chance of misplacing them. The "Hallowell" "Unbrako" Key Kit is a compact, complete screw-driver service. The kit itself is only about 6" long, has an indestructible, hollow, shiny black plastic handle which holds the bits. To use, simply remove the top, select the bit you want, and insert it in the chuck. The swivel chuck enables you to have the direct drive of a straight handle screw driver, or, you can lock it at an angle, or at a right angle position for reaching those "hard-to-get-at" places.

Incidentally, this is a rugged, durable tool. The bits and swivel chuck are made of the finest alloy steel to assure you of long and satisfactory service.

The "Unbrako" Speed Tool Key Kit is made in two sizes:

No. 25: contains seven hex, two Phillips, one slotted screw bit.
No. 50: contains six hex, two Phillips, one slotted screw bit.

If your Distributor does not carry it, send his name to us, along with yours, and you will be taken care of promptly.

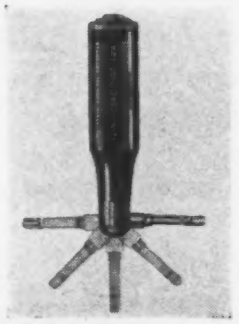


Illustration shows different positions to which swivel chuck permits bit to be swung.

Easily managed, fits comfortably in hand. Bits are placed in hollow plastic handle.

ARMY NAVY E

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STANDARD PRESSED STEEL CO.

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INSULATION'S MOST DEPENDABLE SPECIFICATION

Insl-x Means Modern Insulation

NOW, more electrical and electronic manufacturers than ever before specify INSL-X insulation coatings. They know that INSL-X Research Leadership is their guarantee of unfailing effectiveness.

Literature Upon Request

INSL-X MEANS MODERN INSULATION

THE INSL-X COMPANY INC.
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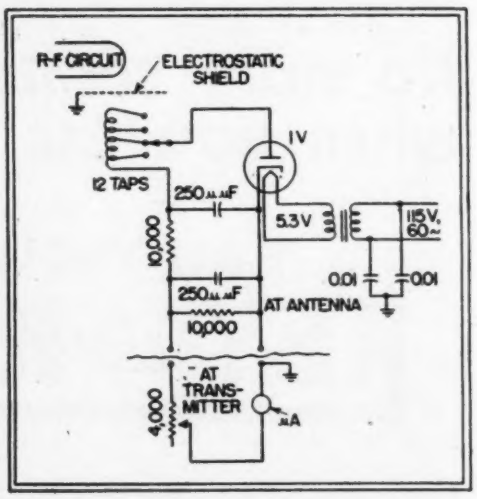
CLEVELAND • ST. LOUIS

stroke on the tower, while in other instances the remote ammeter has been burned to a crisp by lightning, but no loss of broadcasting time resulted.

Rectifier Circuit

In the circuit shown, a current transformer with a primary consisting of a heavy copper rod and an electrostatically shielded secondary feeds a 1-V diode rectifier tube which actuates a d-c microammeter.



Circuit of electronic antenna ammeter for reading antenna current at the transmitter

The microammeter is calibrated in r-f amperes and provides an approximately linear calibration because the diode load resistance is made large in comparison to the tube impedance. An instrument of this type is not absolute and must be adjusted to agree with the regular thermocouple ammeter. FCC regulations provide that this adjustment must be made once a week. When not in use for calibrating purposes, the thermocouple may be switched out of the circuit.

The time constant in the rectifier circuit is such that the instrument is essentially peak reading but this is much smaller than the period corresponding to the highest modulation frequency. Thus, the indication of the meter remains constant with amplitude modulation, provided there is no carrier shift. In contrast, the reading of a thermocouple ammeter increases with modulation.

The photograph shows a rectifier

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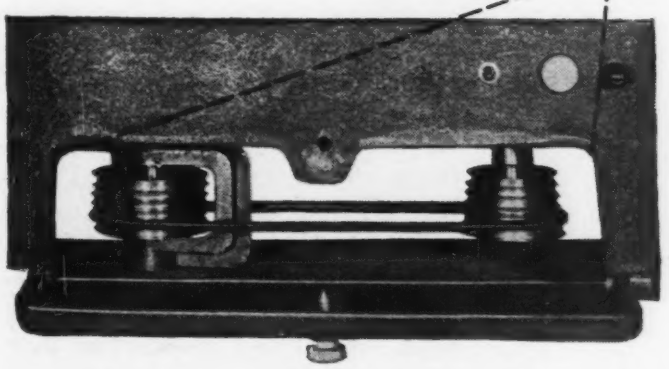

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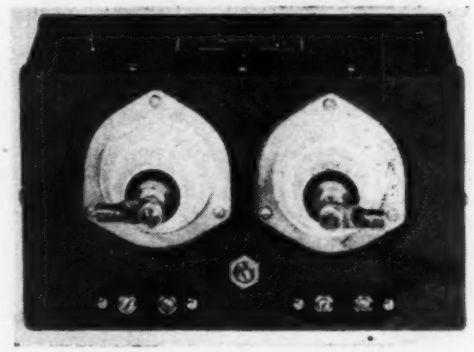
assembly for a five-kw installation. The two large insulators support the ends of the primary coil, which



Rectifier assembly of the electronic ammeter. A half-turn of copper rod mounts on the insulators and forms the primary of a current transformer

is a half turn of ¼-inch diameter copper rod offering negligible impedance to the circuit in which it is connected. The small terminal strips receive a pair of conductors from the remote instrument, and a pair of 115-volt leads for the filament transformer. For a 50-kw station, a larger rectifier assembly incorporates the same circuit but with more insulation between the primary circuit and ground.

• • •

Electronic Sound Effects Circuit

BY HOWARD SYZLING
England

DETAILS HAVE RECENTLY been released of an interesting piece of apparatus constructed for use in training men for the army and to give them a good representation of what battle sounds are like. The reproduction of these sounds is entirely electronic and involves consideration of the precise analysis of real sounds.

Essentially the equipment consists of sixteen separate panels of gear, eight being involved in generating the sound. Four are power amplifiers, four are mixing amplifiers and in addition there is a power pack and two loudspeakers. The total output is 200 watts and the operation is entirely automatic when once started.

The best method of describing the apparatus is to deal with specific sound effects and explain how they are produced. Many of them

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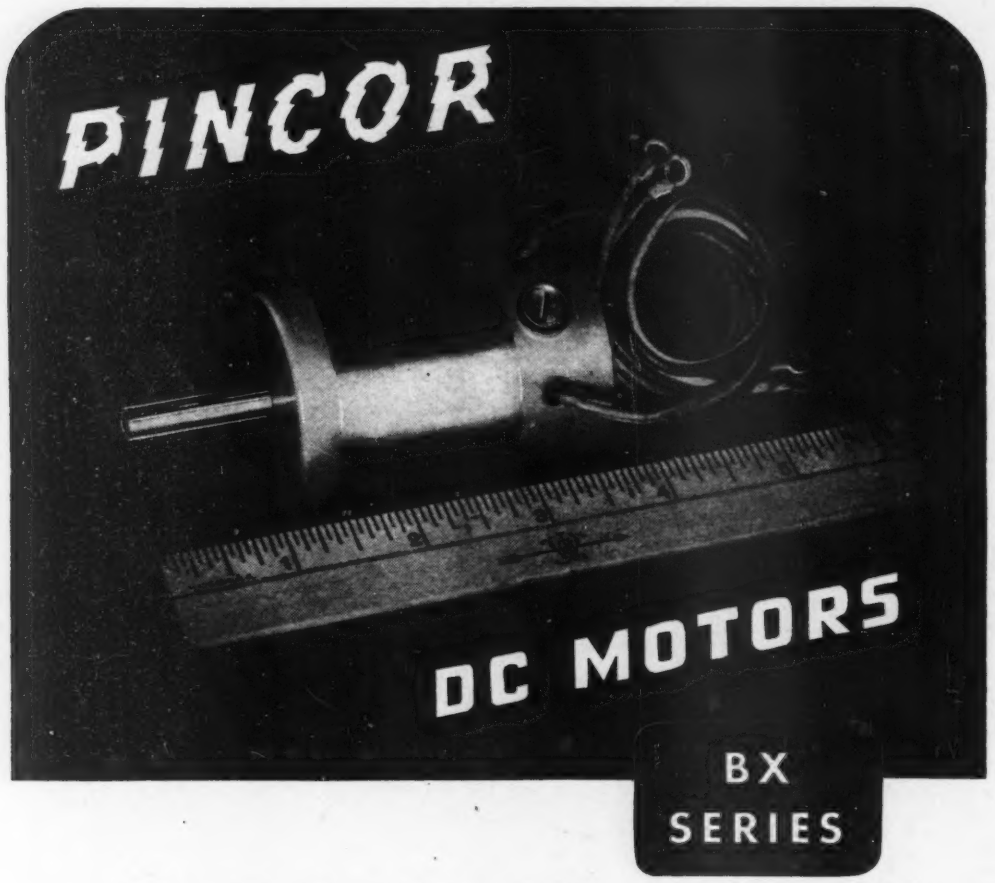

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introduce the use of an aperiodic generator, so called because its output consists of a signal with an irregular waveform and created by amplifying the random noise voltages appearing across a small neon tube.

Distant Shell Burst

Typical shell bursts are obtained from the circuit shown in Fig. 1. The output of the aperiodic generator is fed in at the point shown and is filtered by the resistance-capacitance combination of R_2 and C_1 to remove the higher frequencies. It is then passed to the grid of tube V_1 via capacitor C_2, the tube being normally biased beyond cutoff by resistors R_4 and R_5 so that no signal appears on the anode.

A rotating cam wheel having shaped teeth operates the switch SW so that in one direction C_3 is charged via R_1 to a positive potential and in the opposite direction is connected in parallel with capacitor C_4 via resistor R_3. This results in the grid of V_1 swinging fairly rapidly in a positive direction (since R_3 is small) and then returning more slowly to cutoff as R_5 is comparatively large, so producing a voltage at V_1 grid with a steep-front waveform and a slow exponential recovery and containing also a measure of the aperiodic noise. In other words, R_3 controls the "attack" and R_5 the "decay".

Capacitor C_3 is small compared with C_4 so that the decay is constant even if C_3 is disconnected from C_4 during the decay period. The combination of resistor R_1 and capacitor C_3 is designed to have a

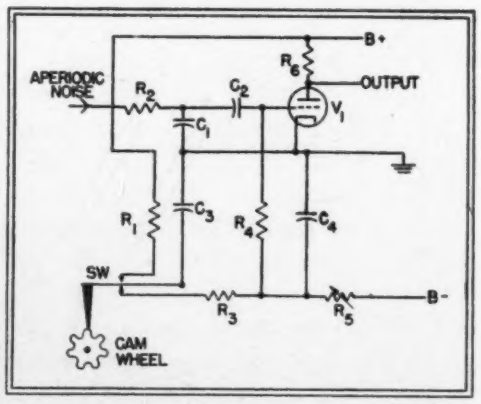

Fig. 1—Wave-forming circuit for simulating shell bursts. The rotating cam wheel changes the connections of capacitor C_3

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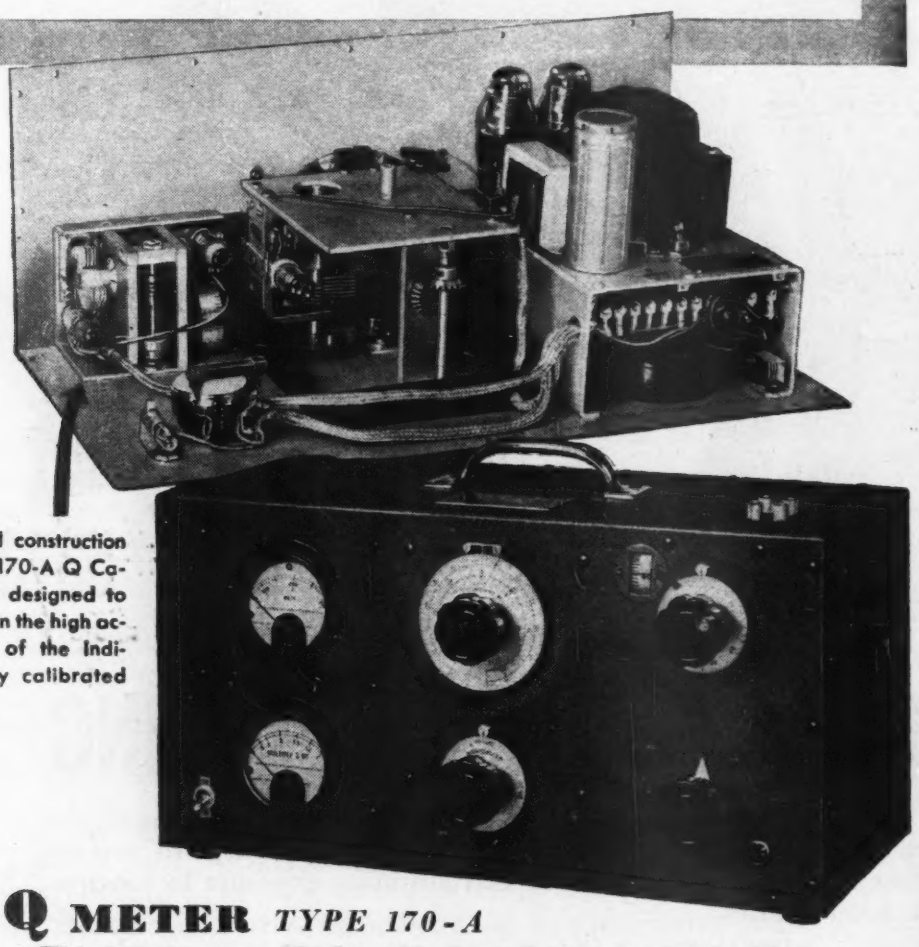

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long time-constant so that C_3 does not have time to charge fully to the anode potential but varies according to the shape of the teeth on the cam and so varies the volume of the shell burst.

Capacitor C_2 and resistor R_4 are proportioned so that R_4 presents a fairly high impedance to the aperiodic noise frequencies while C_2 presents a similar high impedance to the envelope shape of the effects. Resistor R_5 is made variable so that it can be used to vary the echo produced.

Local Shell Burst With Whine

In the case of a nearby shell burst, the explosion is far louder and is always preceded by a whistling which falls in frequency and then stops suddenly as the shell explodes. The whine is generated by a triode connected as a squegging oscillator and the frequency is caused to fall by varying the grid bias so that the audio output varies from the supersonic region down to a few hundred cps. When the oscillations stop, the cathode current drops to zero and gives a pulse into a cathode-follower tube and thus discharges a negatively charg-

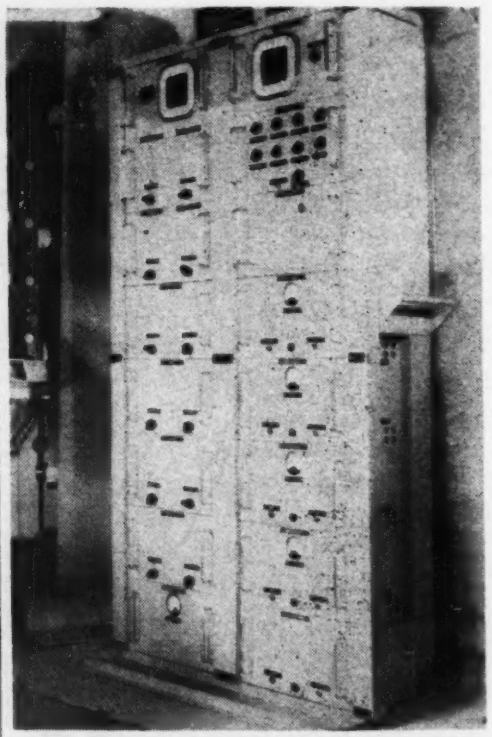

Fig. 2—Electronic sound-effects machine. The panels at the left provide a choice of the sounds of distant and local machine-gun fire, shell bursts, tank, airplane and motorcycle noise

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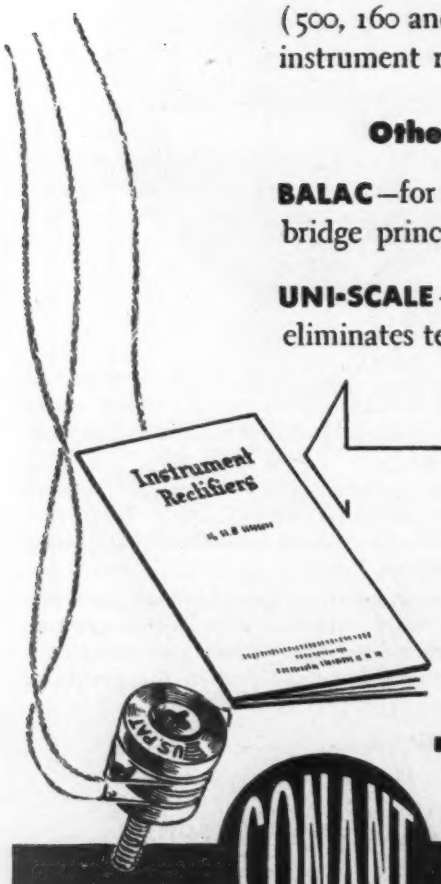

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ed capacitor into the grid circuit of the output.

The sound of nearby machine guns is obtained from two triodes used as a multivibrator with a common cathode resistor and unequal time constants in the grid couplings. This setup gives a large positive pulse across the cathode resistor followed by a smaller positive pulse and is applied to the grid of a triode normally biased to cutoff. A small measure of aperiodic noise is also mixed in. The effect of the double pulse is to create the characteristic "tutter" of an automatic gun while the aperiodic noise generates the sharper "bark".

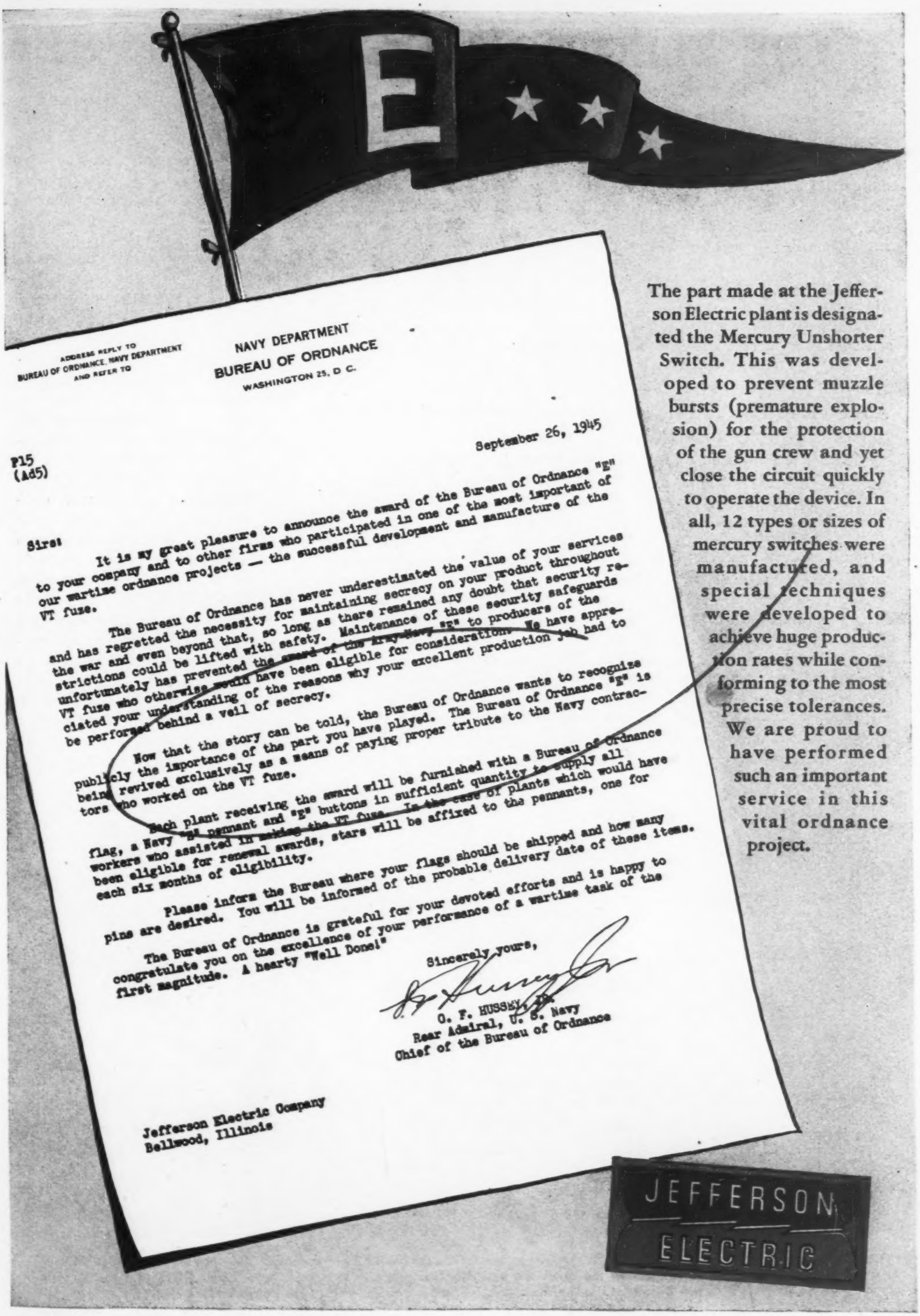
Other effects which can be produced by this equipment are distant machine-gun fire, teleprinters, tanks, aeroplanes, cars, motorcycles and include such effects as starting up, driving off, changing gears, fading away, exhaust, varying engine speed and the like.

Prevents Repetition

The mixing and power amplification of the effects follows conventional lines but there is a simple though ingenious mechanism to prevent any combination of sounds from being repeated too often. It consists of five synchronous motors driving cams running at speeds varying from one revolution in 30 seconds to one in 18 minutes. Four of the motors have an extra cam which opens the motor circuit of an adjacent motor, so that one runs continuously while the others stop at irregular intervals. By this means various sounds can be mixed and yet particular combinations or sequences will only repeat at very long intervals.

Figure 2 shows a photograph of this sound-effects machine, the product of the Mervyn Sound and Vision Co. Ltd. of Great Britain.

It contains two monitor speakers mounted behind the grilles near the top of the cabinet, four power amplifiers and a power supply behind the right-hand panels. Choice of noise facilities is made from the panels at left which are labeled aeroplane, tank, etc. Mixing of these is accomplished by the various controls at upper right.



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September 26, 1945

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It is my great pleasure to announce the award of the Bureau of Ordnance "E" to your company and to other firms who participated in one of the most important of our wartime ordnance projects — the successful development and manufacture of the VT fuze.

The Bureau of Ordnance has never underestimated the value of your services and has regretted the necessity for maintaining secrecy on your product throughout the war and even beyond that, so long as there remained any doubt that security restrictions could be lifted with safety. Maintenance of these security safeguards unfortunately has prevented the award of the Army-Navy "E" to producers of the VT fuze who otherwise would have been eligible for consideration. We have appreciated your understanding of the reasons why your excellent production job had to be performed behind a veil of secrecy.

Now that the story can be told, the Bureau of Ordnance wants to recognize publicly the importance of the part you have played. The Bureau of Ordnance "E" is being revived exclusively as a means of paying proper tribute to the Navy contractors who worked on the VT fuze.

Each plant receiving the award will be furnished with a Bureau of Ordnance flag, a Navy "E" pennant and "E" buttons in sufficient quantity to supply all workers who assisted in making the VT fuze. In the case of plants which would have been eligible for renewal awards, stars will be affixed to the pennants, one for each six months of eligibility.

Please inform the Bureau where your flags should be shipped and how many pins are desired. You will be informed of the probable delivery date of these items.

The Bureau of Ordnance is grateful for your devoted efforts and is happy to congratulate you on the excellence of your performance of a wartime task of the first magnitude. A hearty "Well Done!"

Sincerely yours,

O. F. HUSSEY, Jr.
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THE ELECTRON ART

1945 Rochester Fall Meeting

RADIO ENGINEERS had their first opportunity en masse at the 1945 annual Rochester Fall Meeting held November 12 and 13 to begin their postwar reorientation which will occupy them for some time to come. Papers on war developments, and on the present and future states of communicational and navigational applications of electronics were presented before a registered attendance approaching one thousand.

Dr. Lee DuBridge, who "as head of the Radiation Laboratory of MIT, operated under NDRC, contributed in an outstanding manner to the development of radar apparatus for all branches of the armed services in all fields in which this new science was employed," was awarded the fourth Fall Meeting Plaque.

Abstracts of Papers Presented at the Rochester Fall Meeting

Coaxial Modification of the Butterfly Circuit

By E. E. GROSS
General Radio Company
Cambridge, Mass.

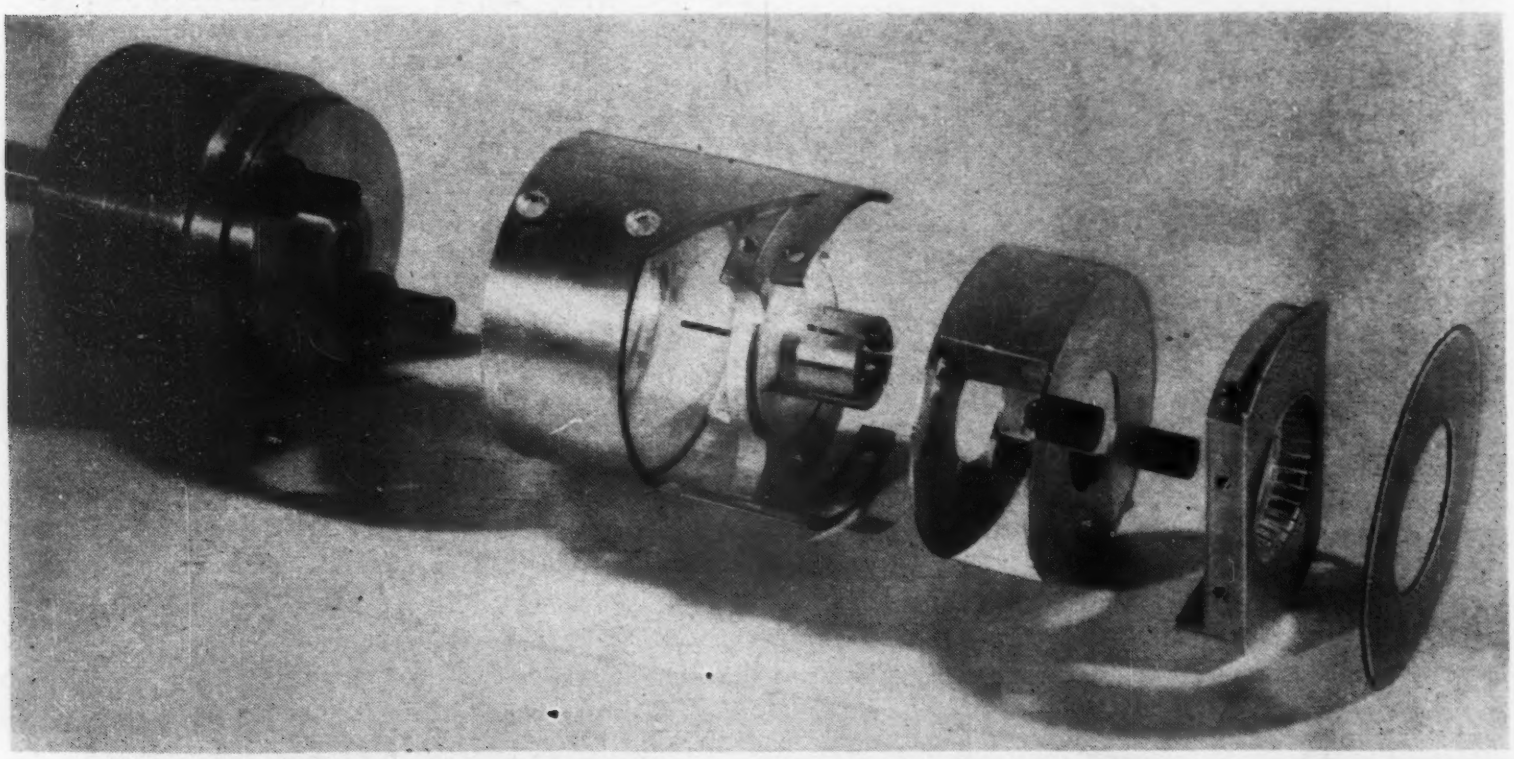
OSCILLATORS FOR THE UHF RANGE of variable frequency can be built for general laboratory use by using a coaxial modification of the butterfly circuit and, in particular, the 2C43 lighthouse tube (McArthur, E. D., Disk-Seal Tubes, ELECTRONICS, Feb. 1945, p. 98.).

Although lumped, inductance-capacitance resonant circuits can be used in the vhf range if compactly constructed, such circuits can not be used at higher frequencies. Feedback to sustain oscillation can be obtained within the tube itself by using grid-cathode and plate-cathode capacitances thus eliminating a troublesome external coupling circuit, but tube lead inductances in the uhf range vary the effect of interelectrode capacitances. Best results are obtained if grid-cathode and plate-cathode capacitances are approximately equal.

Uhf triode oscillators were made possible by the introduction of such tubes as the Western Electric 368-A and 703-A doorknob types with their high resonant frequencies. However these tubes were designed for operation with transmission-line tanks which are not readily adjustable for variable-frequency oscillators. Butterfly circuits were used with these tubes, but because of the high current requirements for their thoriated tungsten filaments the tubes themselves are not particularly suitable for laboratory test equipment.

The General Electric lighthouse tubes such as the 2C43 and the 2C40, which have higher resonant frequencies than earlier triodes, are designed for coaxial-line circuits. The low ratio of plate-cathode to grid-cathode capacitances limited their upper frequency as an oscillator, however. The 6F4 acorn type tube, having a more favorable internal capacitance ratio will oscillate at higher frequencies than these lighthouse tubes if just a tuned circuit between plate and grid is used, but lacks the power of the lighthouse tubes. Therefore the problem became that of designing a tunable circuit in which a lighthouse tube would oscillate up to its highest possible frequency.

A coaxial variation of the butterfly circuit was developed. Basically it is a coaxial line shorted at one end, open at the other. The outer



Coaxial butterfly circuit is built, starting from the left, of the rotor supports on their mounting; the plate-end assembly, which is the shorted end of the line in this model; the rotor; the grid disc, sections of which have been sliced off so that the feedback ring can be coupled to the rotor; and the feedback ring itself at the extreme right

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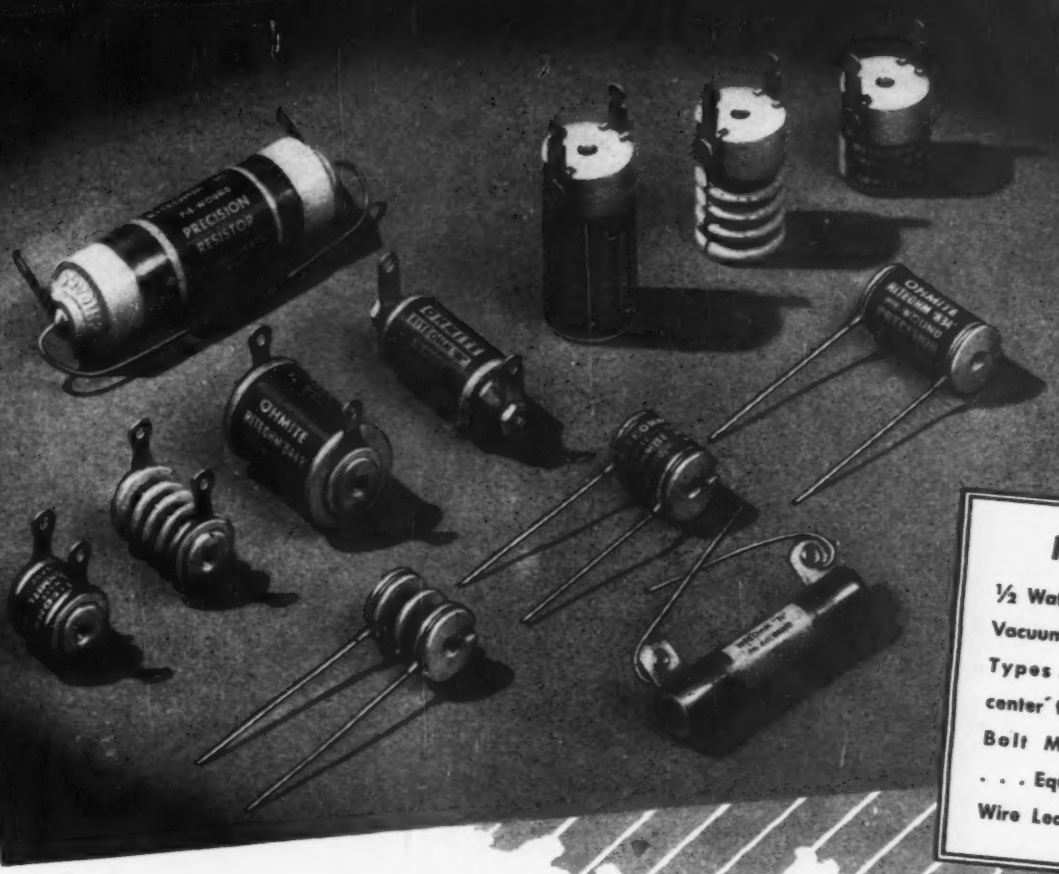

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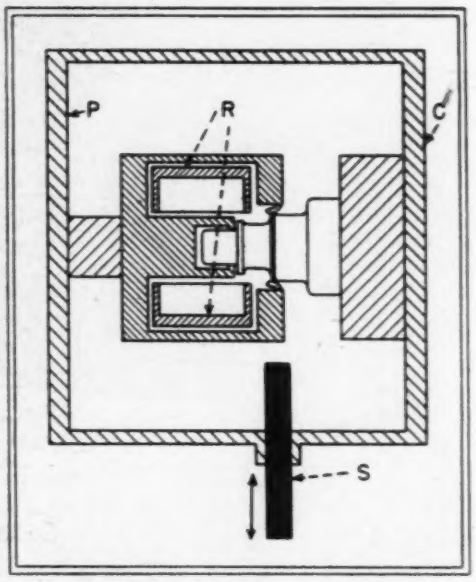
conductor has two sections that are cut away. Two smaller sectors rotating between inner and outer conductors vary the frequency. The tube, connected across the open end of the line, electrically foreshortens the line in proportion to the amount of capacitance loading presented by the tube. This loading depends upon the characteristic impedance of the line. Rotating the intermediate sectors varies the characteristic impedance of the line and thus changes the loading thereby controlling the electrical length of the line and therefore the oscillating frequency of the circuit.

One possible form that the coaxial butterfly circuit for a lighthouse tube can take is shown in the accompanying drawing. Disks *P* and *C* are located for proper feedback at



The coaxial butterfly resonator is mounted inside a cavity the end plates (*P* and *C*) of which form part of the feedback loop. Plunger *S* adds to the feedback at the low frequency end of the tunable range. Rotors *R* vary the frequency

the highest oscillating frequency. Additional feedback at lower frequencies is obtained by adjusting plunger *S*. The coaxial tank with the variable sectors *R* fits over the lighthouse tube. If the tuning drive that operates the rotating sectors also controls the feedback plunger, a fairly wide frequency range with a single control can be obtained.

There are other variations of this construction. The sliding contact between chamber and plunger can be eliminated. For pulsed oscilla-

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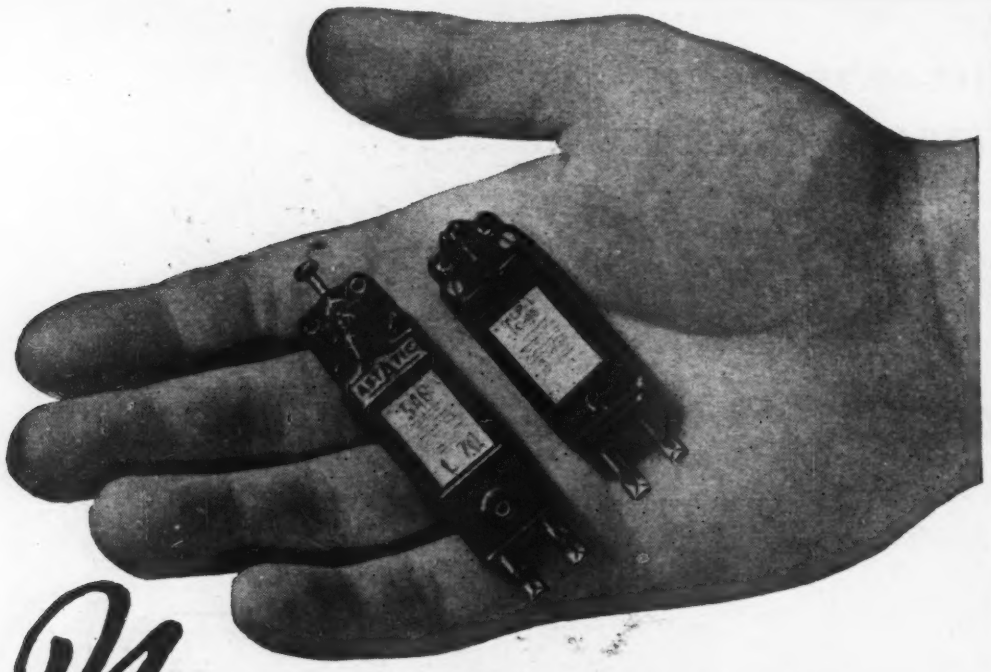

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tors, greater feedback than for continuous wave oscillators must be provided. Output coupling loops can be variously placed but give greatest output if located opposite the feedback loop. Special designs give wider frequency range.

Radio Proximity Fuze

By HERBERT TROTTER, JR.
Eastman Kodak Co.
Rochester, N. Y.

THE IDEA of proximity fuzes is old. Patent offices of the United States, Germany, England, and Switzerland contain references to such devices. These patents are only ideas.

Work on fuzes started in England as far back as 1940 and in Germany, in 1930. Germany was still working on the fuzes at the end of the war, but had put none into service. German scientists said that Germany was not interested in the development of fuzes for shells because, even if engineering difficulties could be overcome, production difficulties were beyond the scope of any country in the world.

Work in the United States on the proximity fuzes was started in the fall of 1940, at the Carnegie Institute, Department of Terrestrial Magnetism, in Washington, by Dr. Tuve and his associates (Proximity Fuze, ELECTRONICS, Nov. 1945, p 110.) This paper deals with subsequent work done on shell fuzes.

The problem of building a radio device that could be shot out of a gun meant the development of special components. To test component parts and to see if the fuze assembly would withstand firing forces, an experimental test field was set up with a 57-mm gun. This gun was mounted vertically so that the shells would land in a relatively small area and could then be recovered and the contents examined. In this way, the various components were studied and designs changed until the unit was mechanically dependable.

The problem of getting the fuze to operate properly in flight was more difficult because it was not possible to study the unit while it was being tested. From results, one could only postulate the causes of failure and try to correct the faults one at a time.

Later, fuzes were tested in 5-inch

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Electro-chemical corrosion—the corrosion that occurs when current-carrying copper wire and insulation get together with humidity—is a prime cause of electrical equipment failure.

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Electronic Engineers and Technicians appreciate the fine quality of Glaser Solders and Fluxes—the result of metallurgical research extending over a period of twenty-three years.

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Navy guns over water and finally in August of 1942, radio-controlled planes were shot down in Chesapeake Bay. Every time a change was introduced into the fuze, elaborate testing had to be undertaken to insure that the performance of the fuze was not impaired. Throughout the war, rigid quality control was followed, and each lot of fuzes turned over to the services had samples tested by firing.

Proximity Fuze Tubes

By MARCUS A. ACHESON
Sylvania Electric Products, Inc.
Kew Gardens, N. Y.

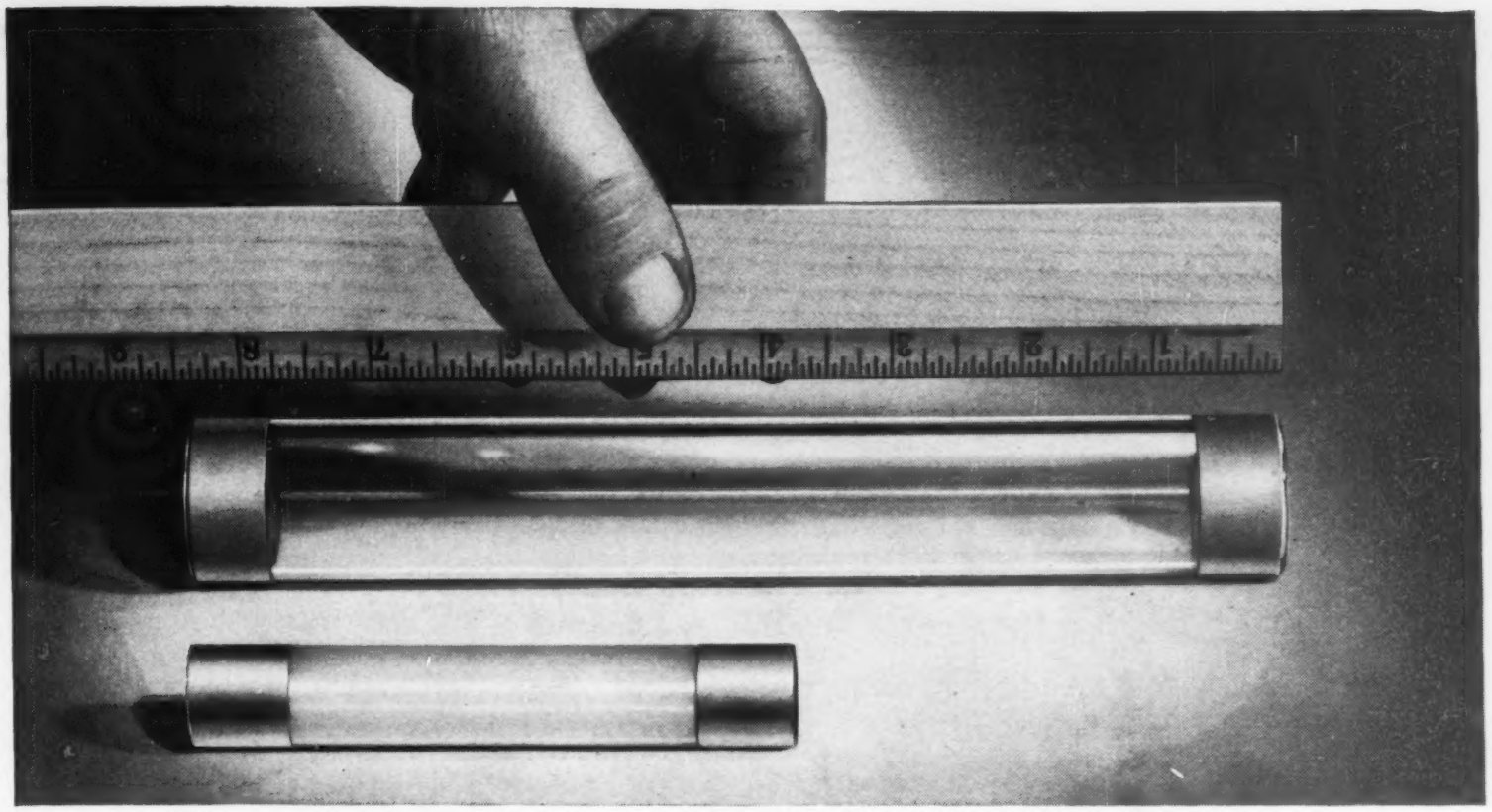
ALTHOUGH SYLVANIA developed and manufactured tubes for non-rotating projectiles, our major contribution to proximity fuze tubes was the development and manufacture of triodes, pentodes and thyratrons required for rotating projectiles. Some of the design problems of such tubes is the subject of this report.

While there is nothing unique about the idea of a proximity fuze for rotating projectiles, there is interest in the successful development achieved in this country. Basic fuze development was finished in the United States less than a year after Pearl Harbor.

German failure to develop a successful fuze lay in their inability to develop the ultra-rugged tubes as was indicated in discoveries by American scientists in Germany since VE-Day. They had, however, developed an ampule-activated battery, a development in this country which had been considered as equal in importance to the tube development.

Several of these British and German attempts to develop a manufacturable proximity fuze were known to Section T of OSRD when developmental work was undertaken by request of the Navy in 1940. Section T searched the tube industry for small, battery tubes which could withstand great stresses and strains. Hearing-aid tubes received attention because they were both small and available, had quick heating time, and were good amplifiers. However their mechanical strength was far below that needed.

Tubes for rotating VT fuzes were



THESE GLASS TUBES HAVE A RULE AGAINST LETTING THE OUTSIDE IN!

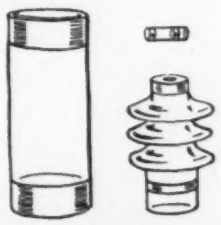
THE point is—they're metallized at each end with the famous Corning process that makes a permanent bond between the glass and metal. This means these tubes for resistors, capacitors, etc. can be soldered into place to form a permanent hermetic seal. No dust, moisture, or corrosive atmosphere can get in.

Besides, these are pretty tough babies. They have two to three times the strength of ordinary glass. Assembled with heavy metal end caps, they will withstand thermal shock up to 275°C.—spot heat to ice water. Their electrical properties are good, too, with high surface and volume resistivity. They're no Johnny-come-lately's either. Millions have been made and used successfully. 20 standard sizes from ½" x 2" to 1¼" x 10" can be mass-produced for immediate shipment.

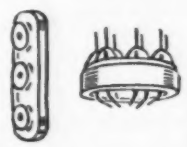
If metallized glass can improve your product through hermetic seals or faster assembly, Corning can help you. Look at the Corning Electronic Products below. If something like these are what you have been looking for, write, wire or phone The Electronic Sales Department, E-1, Technical Products Division, Corning Glass Works, Corning, New York. We'll have an engineer at your doorstep in nothing flat ready to help you work out your problem.

NOTE—*The metallized Tubes and Bushings, Headers and Coil Forms below are all made by the famous Corning Metallizing Process. Can be soldered into place to form true and permanent hermetic seals. Impervious to dust, moisture and corrosion.*

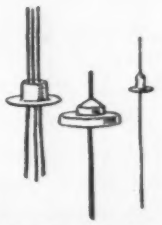
Metallized Tubes *for resistors, capacitors, etc. 20 standard sizes ½" x 2" to 1¼" x 10". Mass-produced for immediate shipment.*



Metallized Bushings. *Tubes in 10 standard sizes, ⅜" x 1⅝" to 1" x 4½" in mass production for immediate shipment.*



Headers — *The best way to get a large number of leads in a small space for assembly in one operation.*



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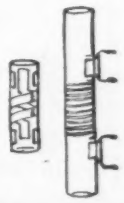

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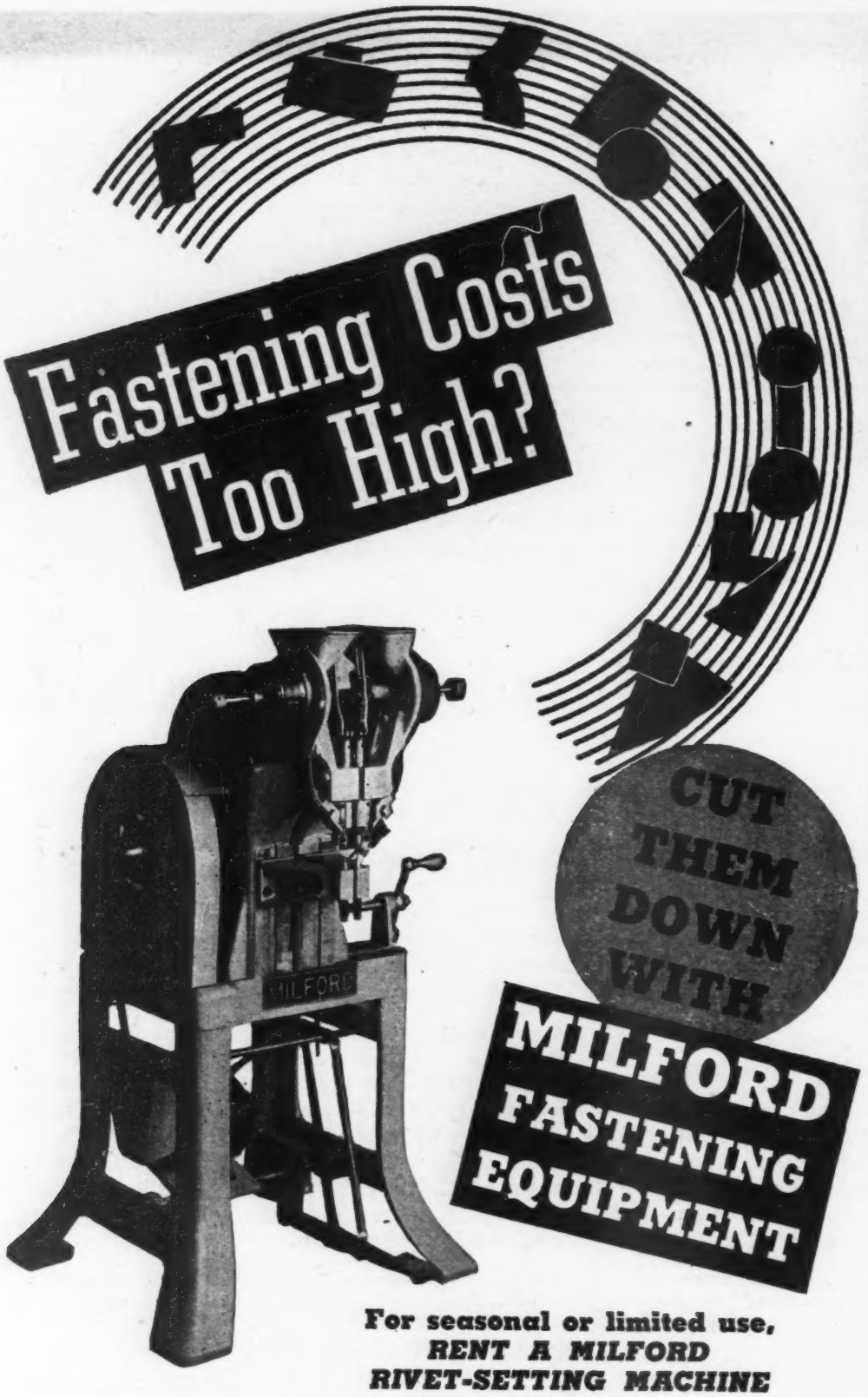

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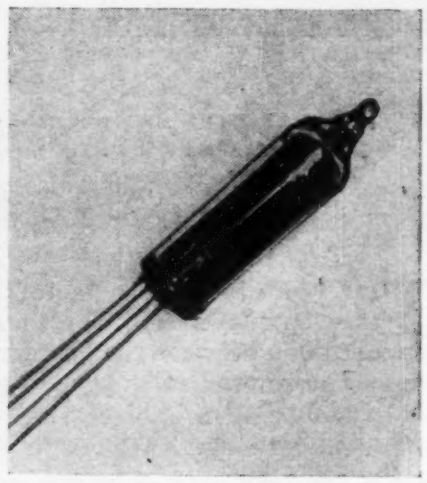
expected to endure an acceleration of 20,000 g's, and centrifugal forces caused by rotations of approximately 30,000 rpm. The problem was so complex that an entirely new approach was necessary.

Of the three main requirements —mechanical and electrical features and adaptability to high-speed production—it was necessary to determine the exact relationship of each to the other. Uppermost importance was placed on mechanical features.

An object shot from a gun is not only subjected to the violent impact of explosion; it is also forced to travel through the gun barrel in a spin that is not smooth, but is more like a shudder. This combination of

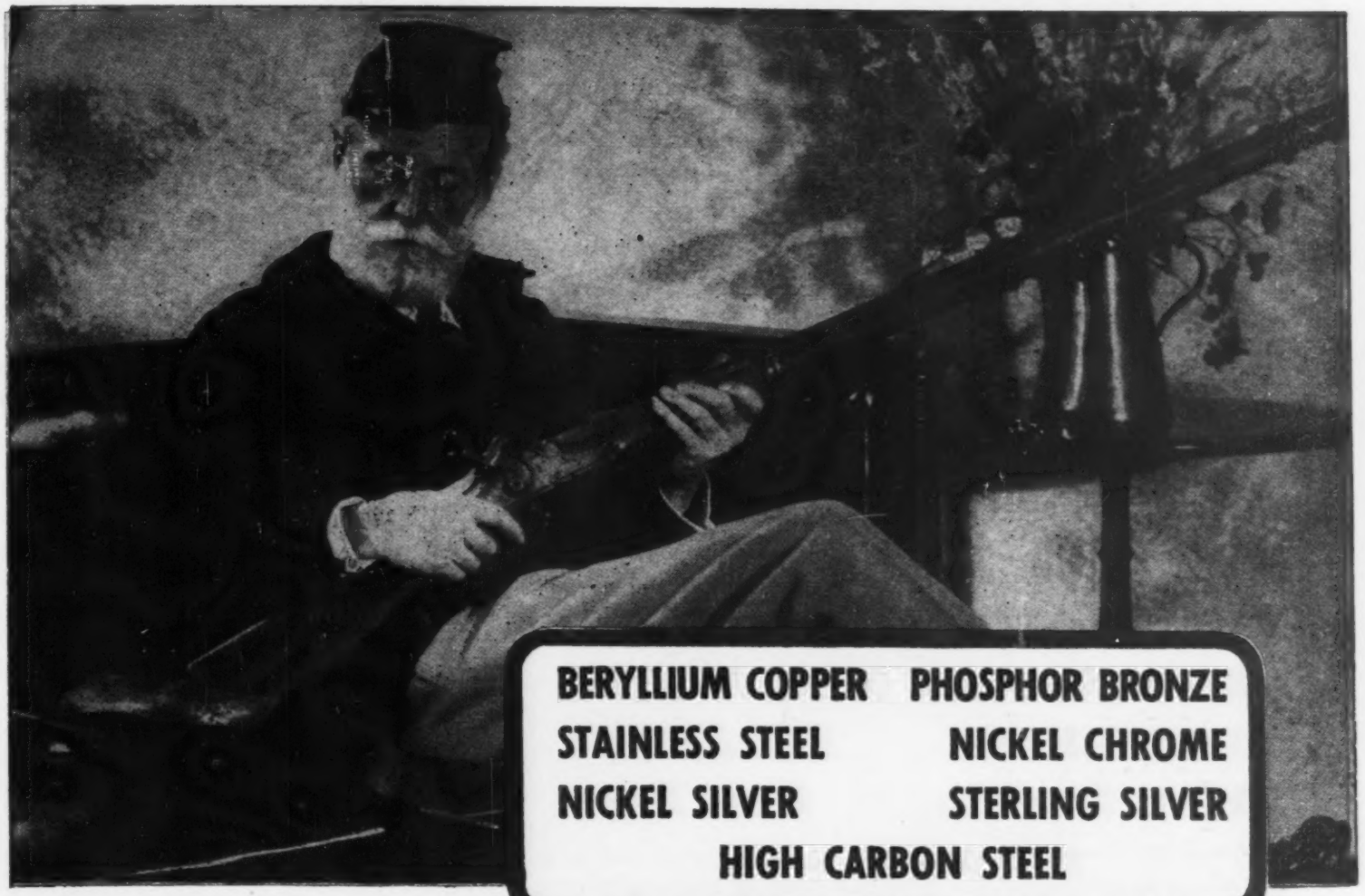


Full size photo of triode designed for mechanical strength and quantity production

spin and shudder (termed "yaw") produces a rapid side-slap, which, when applied under conditions that increase the weight of the parts 20,000 times, greatly complicates the tube design problem.

Space limitation in the shells added another problem to tube design. There was no limit to the minimum specification, except that the size must be manufacturable.

A review of the fundamental knowledge on hand revealed valuable techniques for solving four of the structural problems. These techniques were: (1) glass strain control and fracture analysis by polarized light, (2) a method for making bulbs from tubing, and a method of tubulation that gave unusual strength around the tip, (3) the cataphoresis method of coating very fine tungsten filaments, and (4) the



Grandpa did all right . . . in '63!

Muzzle-loaders, like crudely drawn wires, were acceptable in the old days. Now, however, the stringent requirements of modern products and modern engineers demand that perfection be built into every hank, coil and spool of wire. The Spencer engineering department is staffed with capable men to produce quality steel and alloy wire of exacting standards in all fine sizes. Write Dept. EL-11 for engineering information on your fine wire requirements.

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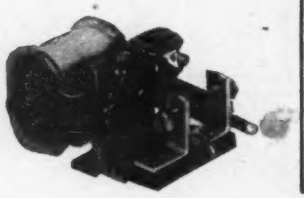
PRECISION — RUGGEDNESS SENSITIVITY

TYPE 4F

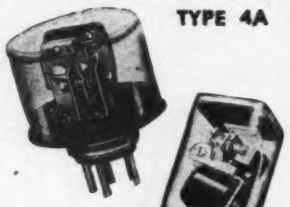


TYPE 4F

This is the Series 4 relay with open frame construction. It is compact (1⅝" x 1⅜" x 1 5/32"), fast (2 - - 3 milliseconds with sufficient power), sensitive (.010 watts minimum).

★

TYPE 4A



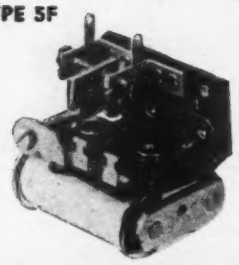
TYPE 4R

TYPES 4A AND 4R

These are the basic Series 4 construction with enclosures. The 4A is 2⅛" diameter and 2⅛" high, the 4R ½" x 1½" x 2⅝". Both have a 5-pin plug in base. The 4A can be sealed (4AH).

★

TYPE 5F

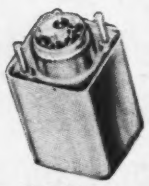


TYPE 5F

This is the Series 5 relay with open frame construction. It is exceedingly sensitive (.0005 watts minimum), very rugged, precise in operation even under extremes of temperature and vibration.

★

TYPE 5R



TYPE 4 (or 5) RJLB2

TYPES 5R AND 4 (or 5) RJLB2

5R is the Series 5 construction in an enclosure (1½" x 1½" x 2¼") with a 5-pin plug in base. The RJLB2 is a hermetically sealed enclosure with two mounting bolts available for both the Series 4 and 5 relays.

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use of straight-through leads as mount supports.

The first important design consideration was to select the material best suited for the envelope. Sylvania decided that an all-glass tube could be expected to endure the severe abuse imposed by gunfire, and at the same time avoid difficulties that had been experienced in making filamentary tubes of other than a glass construction, especially problems of outgassing.

The second consideration was to determine the smallest size that could fulfill the combined, physical, electrical and production requirements. Coupled with this problem was also the one of shape. The flat-press hearing-aid type of tube used in early experimental fuzes was unsatisfactory because of (1) weakness on the flat side which made the tube crack at the press, (2) microphonics caused by mount movement, because only two sides are supported in the flat seal, (3) the high capacitance and leakage problems between leads in the same plane, (4) leads in only one plane are not the most suitable foundation for a rigid mount structure, and (5) all leads in the same plane are not adaptable to streamlining to the tube structure so as to make the most manufacturable type of tube. For these reasons a cyclindrical bulb and leads on more than one plane are essential.

The first size selected was ⅜ of an inch outside diameter, and 1¼ inches overall length. A triode in this size has a volume of slightly more than one-tenth of a cubic inch, and weighs a little more than one-tenth of an ounce, but at 20,000 g's the effective weight of the tube is 125 pounds.

An abnormally strong header construction was required for this size because the header was subjected to the strain of overcoming the inertia of the mount structure as its weight was increased 20,000 times. Header design was not only a success from the mechanical standpoint, but also (1) provided a method of processing small tubes that prevents excessive heat from reaching the elements, (2) freed the mount from dependence for support upon perfectly fitted micas, which in turn

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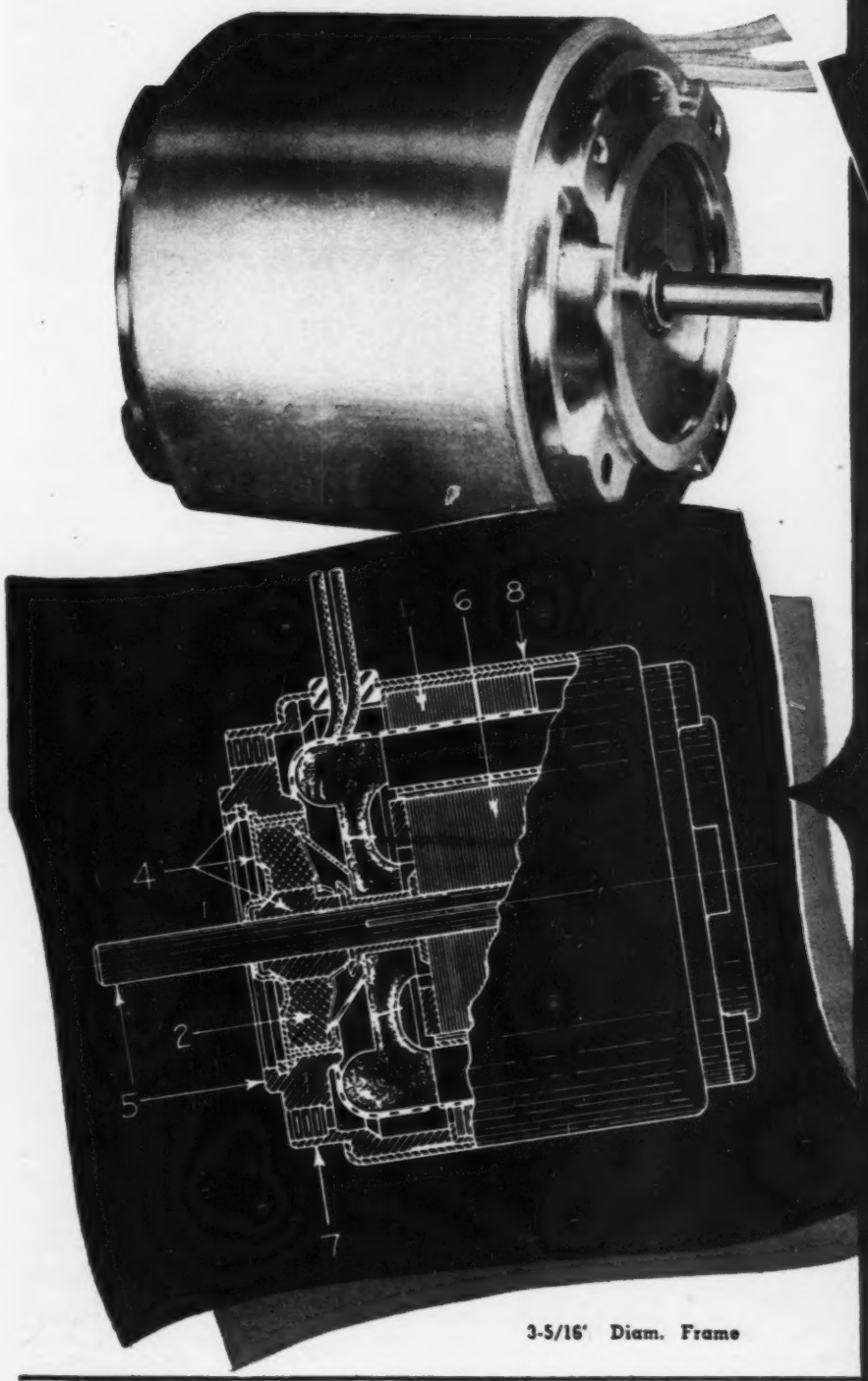

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115 VOLTS - 60 CYCLES
1/200 HP to 1/15 HP

- *Capacitor*
- *Shaded Pole*
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TECHNICAL FEATURES

1. Noiseless operation—Hydraulically locked lamination stacks.
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CLASSIFICATION	H. P.	SPEED	REMARKS
INDUCTION (CAPACITOR START AND RUN)	1/20, 1/30, 1/50	1600 RPM 3200 RPM	Continuous Duty Fully Enclosed
SHADED POLE	1/20, 1/30, 1/50, 1/100	1425 RPM	"
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These motors are ideal for use in original equipment such as recorders, business machines, automatic devises, record players, on circulators and driving fans and blowers.

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Prompt deliveries ranging from one to four months. Further data on request.

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CML 1420
ELECTRONIC GENERATOR

Frequency stability of the CML 1420 is better than 2% after initial warm-up. Maximum distortion at full output into resistive load is 10%. Regulation no-load to full load within 4%. Nominal regulated voltage output 80-120-135-215-255 or 270 volts. Power input 115 volts 60 cycles 1200 watts single phase.

TEST POWER for Continuous Duty

FREQUENCY RANGE:
50 to 6,000 Cycles in 4 bands

POWER OUTPUT:
250 Watts Continuous Duty

FREQUENCY CONTROL:
Single dial, direct reading, linear scale in 4 ranges—50-180; 170-600; 500-1800; 1700-6000 cycles.

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Rotobridge • Electronic Generators • Power Supply Units

required absolutely uniform glass thickness, (3) eliminated waste caused by selecting glass tubing to such fine tolerances, and (4) permitted using automatic equipment for high-speed operations both in assembly and glass working.

A smaller size—¼ inch outside diameter and 1 inch overall length—was later developed and manufactured in large quantities. This size weighs slightly more than one-twentieth of an ounce and has an effective weight of 70 pounds at 20,000 g's. This type could be made on a wafer header, because strain on the header is well within the strength of wafer construction.

The internal parts of these tubes are, of course, very small. Consequently, it was necessary to make studies on simplification of parts, and to find means of assembly that mounting operators could perform. (Sylvania operators were presented and acclaimed for their skill and patriotism by Dr. L. Grant Hector before the Radio Club of America in New York, N. Y., on October 11, 1945 during his talk on the Radio Proximity Fuze. Ed. Note.) Mounters learned to combine touch with sight to produce as many assemblies per hour as are produced for conventional tube types.

Electrical requirements were strict. For example, (1) low microphonic output in the presence of mechanical shock and vibration, (2) less than a second allowed for the tube to reach full and stable operation after application of voltages, (3) low power drain to permit use of small batteries, and (4) dependable performance after months of storage, regardless of climatic conditions.

Coating of the fine filament wire required both chemical and electrical research and development because uniform coating weights and exact control of emission characteristics were necessary for meeting the first three electrical requirements.

Another phase of the filament problem was the structure of the tungsten wire itself. A better grade of fine wire was needed than had been made up to that time. The metallurgical laboratory accordingly developed materials and con-

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The new Franklin Delano Roosevelt $200 Bond—better than actual cash because it earns interest—is a strong building stone toward the secure future of every employee-purchaser!

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12 CHANNEL CAPACITY

...yet as easy to take with you as a brief case!

HATHAWAY TYPE S-12 OSCILLOGRAPH combines portability and small size convenience with excellence of results—without sacrificing capacity, accuracy or record quality

Measures only 18"x 9"x 8", yet accommodates up to 12 channels and gives permanent record on paper or film 6" wide by 100 feet. Interchangeable galvanometers available in natural frequency from 200 to 10,000 cycles per second to a sensitivity of 50 milliamperes per millimeter at one meter.

Automatic features: record length control; record numbering device; and time coordinate device which produces time lines clear across the record, quickly adjustable to either 1/100 or 1/10 second spacing. Viewing screen permits simultaneous viewing and recording.

Performance plus portability makes this instrument indispensable for field testing where accuracy and ease of control are imperative.

DELIVERY—60 to 90 Days

●Write for bulletin SP-167. Hathaway Instrument Co., 1315 S. Clarkson Street, Denver 10, Colorado

Hathaway INSTRUMENTS

tinued quality control of them. It was also necessary that this group work directly with the manufacturers of fine tungsten wire, because the very secret nature of the OSRD project prevented giving wire manufacturers reasons for the strict requirements which at times seemed impossible to meet. The filament tensioning problem was one of the most crucial in the tubes' development.

Dependable performance after storage and under various climatic conditions was provided by the all-glass structure and flexible leads which eliminated the need for bases. Sealing and exhaust techniques prevented leakage paths and provided a thorough removal of occluded gases from the metal parts.

Microwave Radar

By DONALD G. FINK
McGraw-Hill Pub. Co.
New York, N. Y.

ADVANTAGES and disadvantages of the microwaves in radar was the subject of this paper. These very short waves are used, first, to obtain a suitably narrow beam from a reflector of convenient size, and second, to minimize the gap in coverage in the horizontal direction. Also, as the wavelength is decreased, the maximum range of detection increases when a reflector of given size is used (The Radar Equation, ELECTRONICS, April 1945, p 92.).

Disadvantages, largely overcome during the war, were lack of high-power r-f generators for the transmitters, sensitive r-f mixers and reliable local oscillators for the receivers.

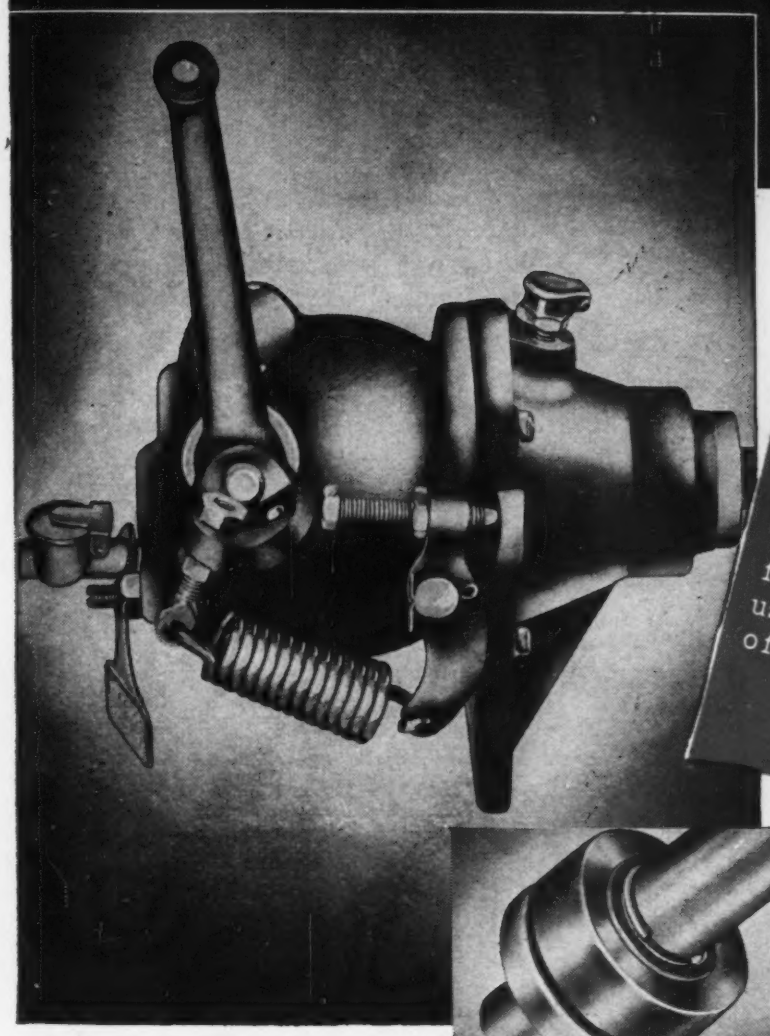
Microwave radar equipment was illustrated by slides of the SCR-584 and AN/MPG-1 radars, which are currently being described elsewhere in this publication. Radar oscilloscope photographs taken with microwave equipment were also shown.

High Quality Sound Recording on Magnetic Wire

By LYNN C. HOLMES
Stromberg-Carlson
Rochester 3, N. Y.

MAGNETIC WIRE high-quality recording, like photographic film or embossed disc recording, is judged on

5 REASONS WHY PIERCE GOVERNOR SPECIFIES WALDES TRUARC RETAINING RINGS



"Waldes Truarc Retaining Rings on our precision flyball governors had five distinct advantages over wire snap rings previously used:
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2. elimination of injury to shafts when installing and removing—
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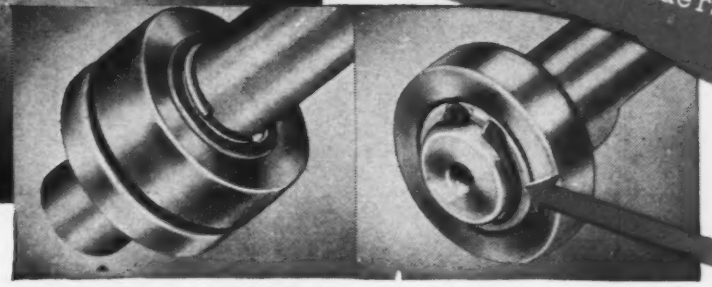
4. narrower groove—
5. greatly reduced production costs.
Truarc rings are completely satisfactory, and we plan to extend their use to other models that were out of production during the war."
Signed: Pierce Governor Co.
Anderson, Ind.

THE PIERCE UNIVERSAL GOVERNOR used on most industrial applications of automotive motors.



1 BEFORE TRUARC
—Bearings on rocker arm were pressed against shoulders at the bottom of the cast iron housing. Wire snap rings were required, adjacent to inside face of bearings to position the shaft and keep it from moving sideways.

2 AFTER TRUARC
—It was possible to slip-fit the bearings. Truarc Retaining Rings have greatly simplified the design by performing the dual function of *positioning the shaft* and *holding the entire assembly* securely in place.

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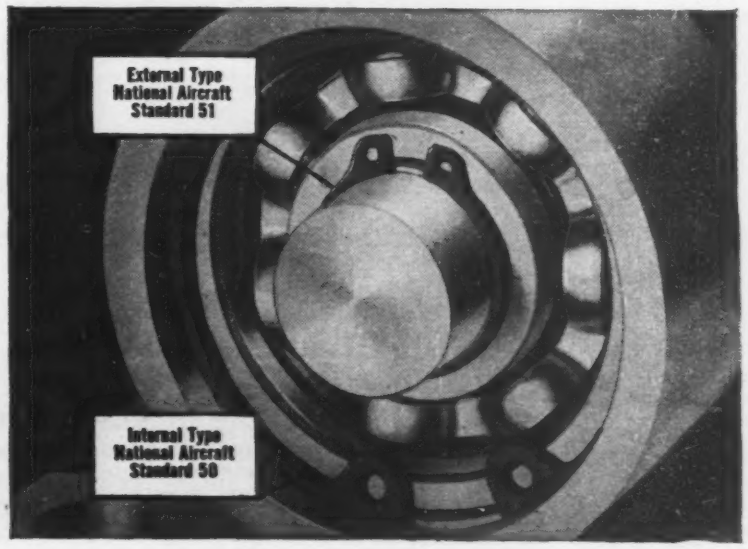

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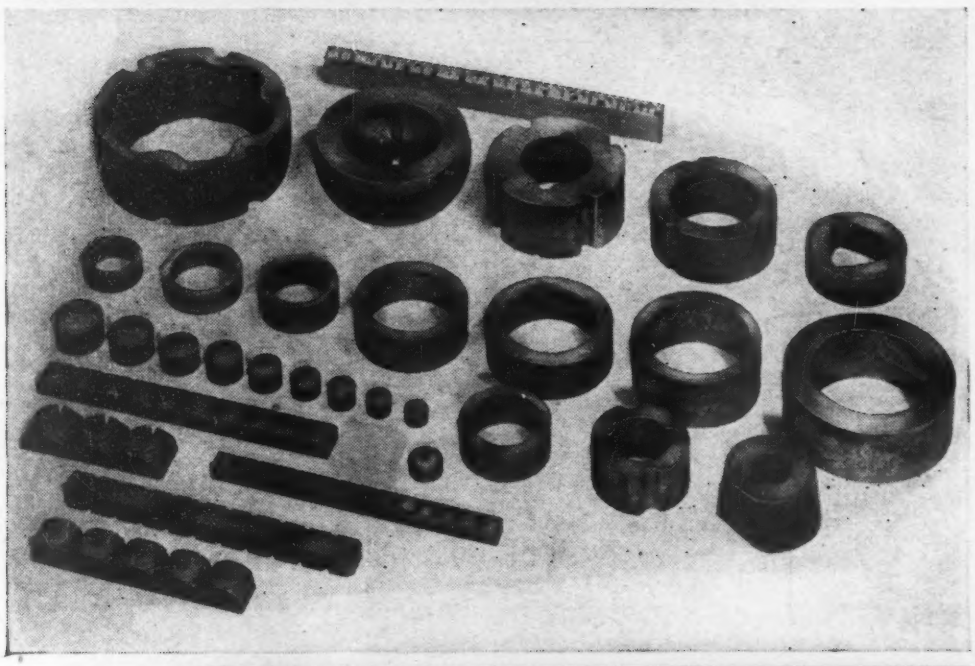

The speaker magnets illustrated above are representative, yet are just one type of permanent magnets described in the new 24-page Arnold bulletin, "Permanent Magnets for Industry."

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its distortion, frequency response, signal-to-noise ratio, and wow.

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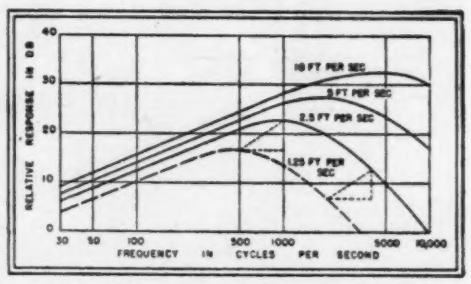
Aural demonstration of the effect of supersonic bias (given at



To obtain high-frequency response from a given wire, it must be run at high speed (Supersonic Bias for Magnetic Recording, L. C. Holmes and D. L. Clark, ELECTRONICS, July 1945, p 126)

the meeting) show the correspondance between optimum experimentally determined bias and that bias determined from the slope of the magnetization curve. The difference so determined is an indication of the effectiveness of the recording head in magnetizing the wire.

Also demonstrated were the effect on frequency of wire speed (illustrated in the accompanying figure), gap length in recording and playback heads, and pre-equalization to improve high-frequency response.

The magnetic property of greatest importance in determining response at higher audio frequencies is coercive force. Self-demagnetization, which is more effective for stubby magnets than for slender ones, tends to decrease the recorded strength of high-frequency signals. Use of finer wires has improved high-frequency response. Also the coercive force has been increased since wire recording began from 16 to 850 oersteds.

Tests to determine the perma-

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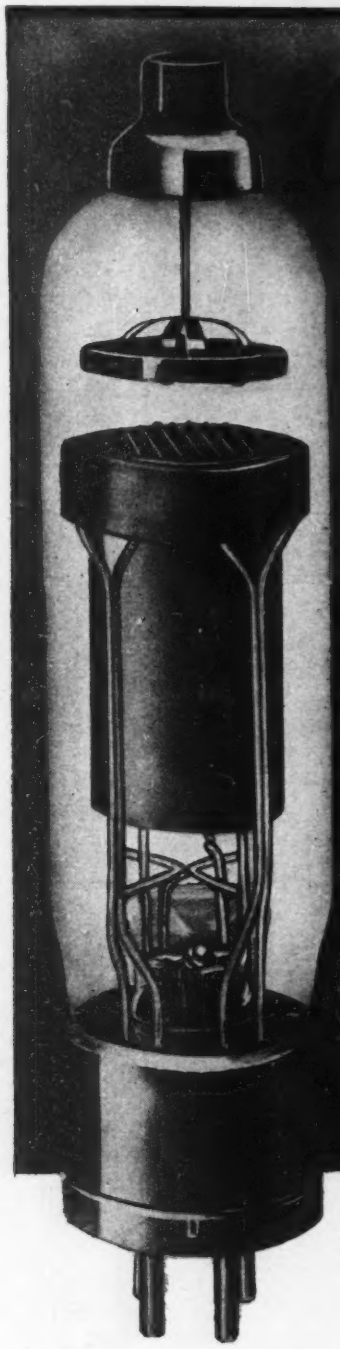

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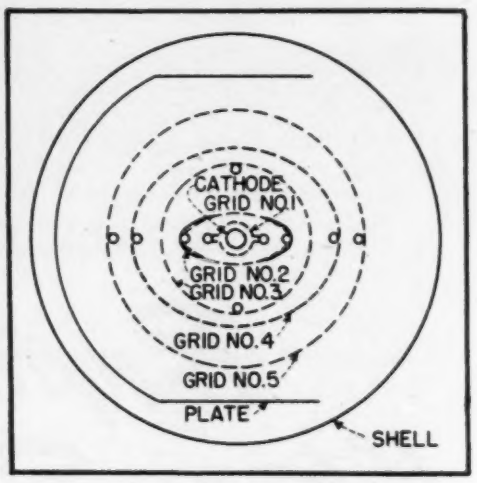
EL Rectifier TUBES

nency of magnetic records have been made by constantly running a closed loop of wire through a playback head. After 1400 playings, a drop in response of 7 to 8 db, occurring mostly during the first few playings, was measured. This test is believed to be more severe than would occur in practice.

Recent Developments in Converter Tubes

By W. A. Harris & R. F. Dunn
Radio Corp. of America
RCA Victor Division
Harrison, N. J.

Type 6SB7Y is a developmental converter tube similar to the 6SA7 in basing and structure. The conversion transconductance is 950 micromhos and the oscillator transconductance is 8,000 micromhos. The tube gives high conversion gain and improved signal-to-noise ratio



Electrode structure of 6SB7Y

in the standard and short-wave broadcast bands. The oscillator circuit should be adjusted to develop an oscillator-frequency voltage at the cathode of approximately 0.8 volt rms at the low-frequency end of each frequency range. The avc system should be arranged so that the gain of the converter does not decrease more rapidly than the gain of the r-f or i-f stages decreases with increase in avc voltage.

Some tests were made in the 88-to-108 mc band using the 6SB7Y as the converter and the 6SG7 as an r-f amplifier. In this band, best results were obtained with the cathode tap on the oscillator coil a little below the position giving maximum oscillator grid current. The

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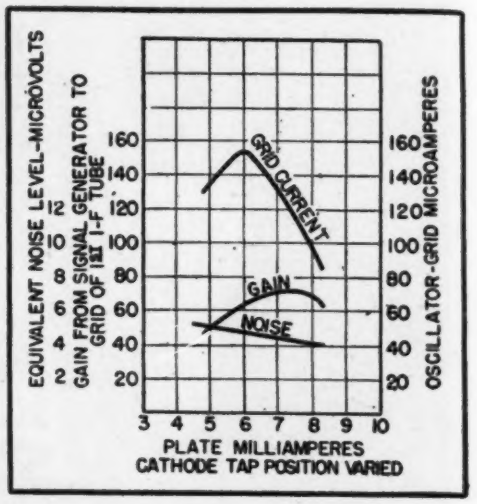

Characteristics of 6SB7Y

Input resistance at the signal grid is negative, and therefore would cause the input circuit to oscillate under some conditions of adjustment. However, such oscillations can be avoided by connecting the plate of the 6SG7 and the signal grid of the 6SB7Y to a tap on the interstage coil. Use of a series resistor of about 3 ohms at the signal-grid terminal was found to be helpful in maintaining uniform and stable gain.

The Aurora and Geomagnetism

By C. W. GARTLEIN
Cornell University
Ithaca, N. Y.

THE AURORA POLARIS, like geomagnetism and the ionosphere, is believed to be a manifestation of particles radiated from the sun. The sun is a variable star, the total radiation varying only about two percent during the 11 year sun-spot cycle, but near enough for detailed observation. Most of the cyclonic sun spots have magnetic fields, some as strong as 9,000 oersteds. Typical spots appear in pairs with one magnetic pole in one hemisphere and the other pole in the other hemisphere. The sunspot latitudes rotate with a period of about 27 days which is the period of magnetic storms and auroras, thus their relation is clear.

Magnetic storms are produced by changes in the electric currents circulating in the earth's upper atmosphere. The total current in the northern hemisphere may exceed a million amperes and changes at the rate of 100,000 amperes per minute. It is this changing current that dis-

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. . . is now available, from stock, in sizes from .007—.010—.015—.020—.025 and .030 inches. Standard sheets are 36 x 24". The grain is parallel to the second dimension.

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	.015-.020	pH	6.5-7.0
	.025-.030	Ash % — less than	1
Specific Gravity gms/cc	1.30-1.40	Methanol Extractable %	.02-.12
Tensile MD #/sq. in.	15,000-19,000	Naphtha Extractable %	.02-.08
Tensile CMD #/sq. in.	6,000-7,000	Chlorides-parts per million	8-12
Elongation MD %	9-11	Water abs. 15 min. approx. %	100
Elongation CMD %	15-18	Water abs. 24 hrs. approx. %	120
Dielectric Volts/.001" thickness	400-600	Approx. Wgt. #/M sq. ft./.001" thickness	7.6
Shrinkage MD%	.4-.7		
Shrinkage CMD %	1.4-1.9		
Mullen #/.001" thickness	34		
Elmendorf Tear MD gms/.001"	60-75		

MD = parallel to grain = Machine Direction
CMD = across the grain = Cross Machine Direction

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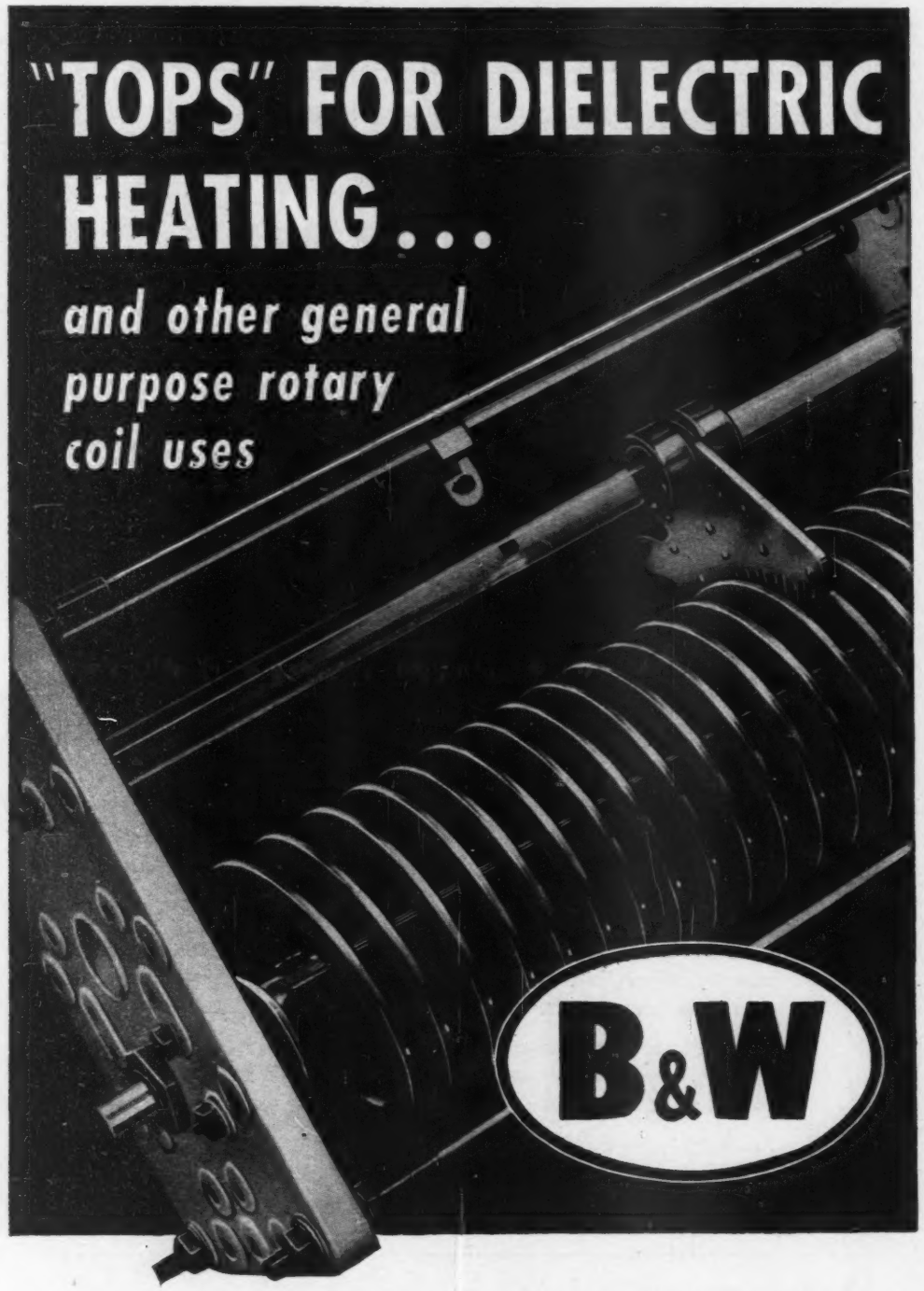

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rupts wire communication and occasionally interferes with power transmission.

The layers of the ionosphere, upon which all long distance radio communication depends, vary seasonally in relation to the sunspot cycle, and the mean ion density of the F_2 region follows the sunspot numbers. Disturbances of the ionosphere follow the sun spots as do the aurora.

A photoelectric recorder has been used in studying the aurora, and with it a spectograph from which correlations with magnetic field changes have been detected.

Industrial Standardization Work in Television

By DAVID B. SMITH
Philco Corporation
Philadelphia, Pa.

THE RADIO TECHNICAL PLANNING BOARD was established about two years ago with a panel on television to make recommendations to the FCC on both allocations and standards. RTPB recommended, and FCC accepted in general, that commercial operation be resumed after the war on 6-mc channels below 275 mc with a few changes in the prewar standards. A higher frequency band was set aside in the 500 to 1,000 mc region for experimentation and eventual utilization when standards could be agreed upon (ELECTRONICS, July 1945, p 92, and Aug. 1945, p 304.).

Basic standards for 6 mc channels have been adopted. Space higher in the spectrum has been specifically set aside for experimentation. Finally, the commission has designated all of the microwave region as experimental which means that there is room for a third television band.

The lowest frequency band, developed and authorized for commercial operations, gives excellent black and white pictures. Standards work is starting on a second band which offers the possibilities of color. This band may be ready for commercial operation in a few years.

Those designing television receivers for the 1946 market will probably be wondering whether changes may be made in the 6 mc channel standards. The answer is that there

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ACCURACY — Line Voltage Adjustment gives accuracy on capacity measurements.
Matched-pair multiple resisters are ±1% accurate.
Circuit eliminates all errors due to line voltage fluctuations.
Constant accuracy of low resistance ohmmeter ranges assured by test of ohmmeter battery under load to determine need for battery replacement.
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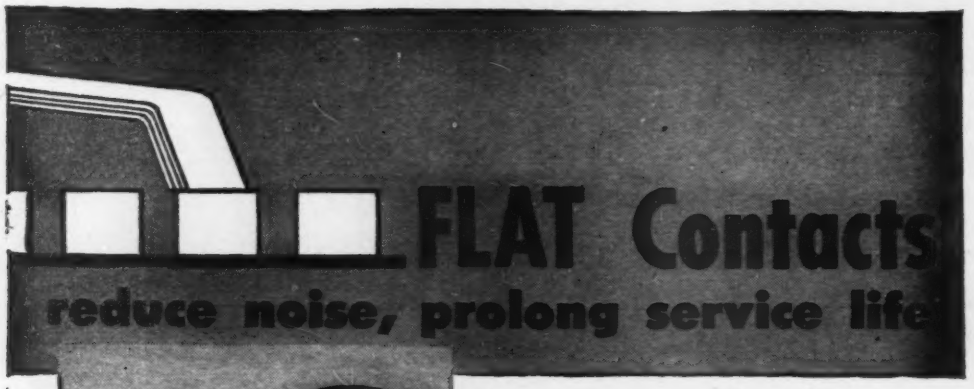

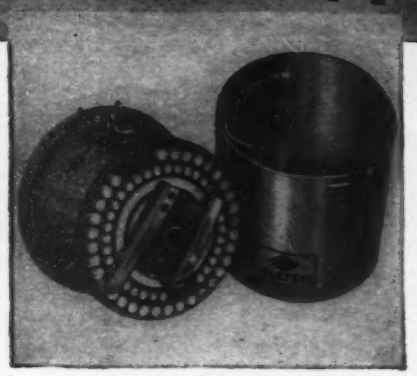

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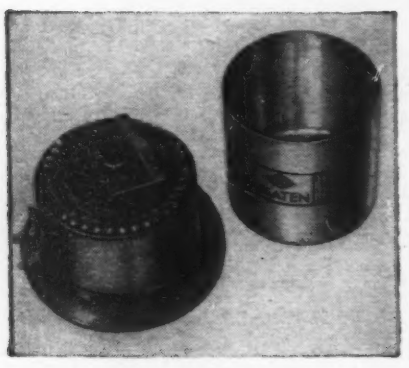

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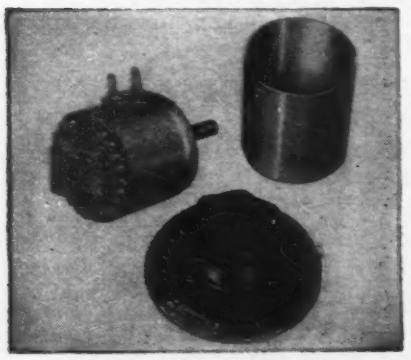
Ladder Circuit—1½ db per step; 30 to 600 ohms impedance. Price, F.O.B...$12.50



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No carbon resistors are used in any Variaten Mixer...
All are of stable, wire-wound construction. Most are step type. Where quiet operation is the major consideration, we recommend ladder type mixers because the circuit requires only one contact brush operation on the input side of the circuit and any possible brush noise is therefore attenuated along with the signal.

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will be no further change in the basic standards for the commercial 6 mc channels in the foreseeable future. We will undoubtedly find, however, as experience is gained in their use that some additional refinements will be necessary.

Some such problems have already arisen. For example, the present standards propose an inverse logarithmic relation between light output and modulation level. But there are numerous logarithmic relations which might be employed. The exact one will depend upon the best camera and picture tube characteristics. Again some difficulty has been reported recently with frequency-stabilized sync circuits as a result apparently of lack of continuity in transmitted sync in switching remotely-located cameras. The standards now establish a tolerance for sync stability but apparently continuity of sync will also be required. As circuits are developed, standards will become more strict.

Improvement in performance is being sought through this standardization, but lack of particular standards is no bar to the commercial use of the system. In general the two criteria by which new standards will be judged are, first, that their use will result in improved service and, second, that their use will not interfere with the operation of existing receivers in the hands of the public. Hence, the net result of this work should be a gradual improvement in the quality of the service with, however, no changes which would obsolete existing receivers in those channels. Although different standards may be adopted for higher channels, such standards will not affect the present channel.

War Influence on Acoustic Trends

By HUGH S. KNOWLES
Jensen Radio Manufacturing Co.
Chicago, Ill.

IN COMMUNICATION of intelligence by either voice or code, the acoustic link between sender and receiver has an important bearing on security. The spatial pattern of such a link, compared to radio, is predictable—the sound does not go too far. Attenuation is frequency selective, the higher frequencies falling off

TALK IS QUICK!

Pick up a 'phone and talk—to an airplane; a speeding train; an inter-city bus; a boat at sea.

Aireon's radio 'phones make this as simple, sure and easy as using a conventional telephone.

Aireon radio equipment for airlines is used by twenty domestic, four foreign companies; Aireon railroad radio, introduced under war-time restrictions, is *already in use* by four leading railroads. Aireon truck, taxi and bus communications equipment has been proved in service on the trucks of one of the nation's largest fleet operators. It's now in production. Aireon marine equipment will be available soon.

On the crowded highways and skyways of the future, radio 'phone communication will keep traffic moving under quick, efficient control.

Aireon MANUFACTURING CORPORATION

Radio and Electronics • Engineered Power Controls

NEW YORK • GREENWICH • CHICAGO • KANSAS CITY • OKLAHOMA CITY • BURBANK • SAN FRANCISCO

1 ACETATE LABEL
Tougher, cleaner, more durable covering. Protects winding from mechanical handling. Prevents lugs from piercing label. Affords fungus protection to winding.
2 ADDED INTERNAL INSULATION
Cross-over wires are insulated from rest of winding with acetate cloth. An exclusive Mepco feature (patent pending).
3 DOUBLE SEAL IMPREGNATION
Prevents breakdowns and shorted turns. Withstands humidity and salt spray tests better. Is fungus resistant.
MEPCO
TRU-TOLERANCE RESISTORS

Triple improved-Better than Ever!

When the war ended, ONE THIRD of all unfilled orders for precision resistors specified Mepco—according to WPB reports! A few of the reasons: Non-Hygroscopic Ceramic Forms. Highest grade Alloy Wire. Extra Terminal Protection. Standard Tolerance of 1%. Special to 0.10%. Severe Breakdown text on every unit. Careful Calibration.

THE MEPCO LINE

THE SAME high quality of material and construction which boosted Mepco "Tru-Tolerance" Resistors to the front rank in the radio and electronic industries, is evident in the entire Mepco Line.

I.F. TRANSFORMERS, COILS — R.M.A. Standard or to your specifications.

R.F. TRANSFORMERS—R.M.A. Standard or to your specifications.

RESISTORS-POWER WIRE WOUND—See Mepco Resistor Guide.

METER MULTIPLIERS—Up to five megohms (wire wound).

ELECTRONIC TEST EQUIPMENT—Tube checkers, VT Voltmeters, Signal Generators, Circuit A.T.C. Analyzers, Etc.

COILS—Layer wound, bobbin wound, paper wound or honeycomb. Most modern winding methods. Get our quotations.

ANTENNA KITS—All types made up with your name on carton.

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TRIMMER CONDENSERS
To your specifications.

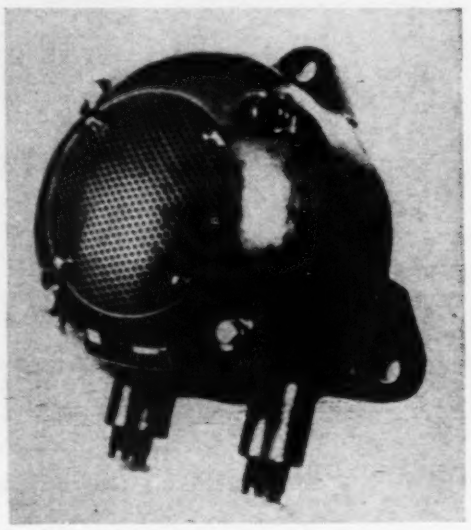
Get your copy of the Mepco Resistor Guide —complete data.

MADISON ELECTRICAL PRODUCTS CORPORATION
MADISON, NEW JERSEY

sooner so that at distances beyond the horizon intelligibility decreases rapidly.

Various sounds have been used as signals during the war, notably warble notes and frequency jumps for certain warning purposes. While these are new uses of sounds, older sounds—like the boatswain's pipe—were continued, especially in the Navy. Gongs, bells, Klaxon sounds, all were utilized in amplified form for one purpose or another.

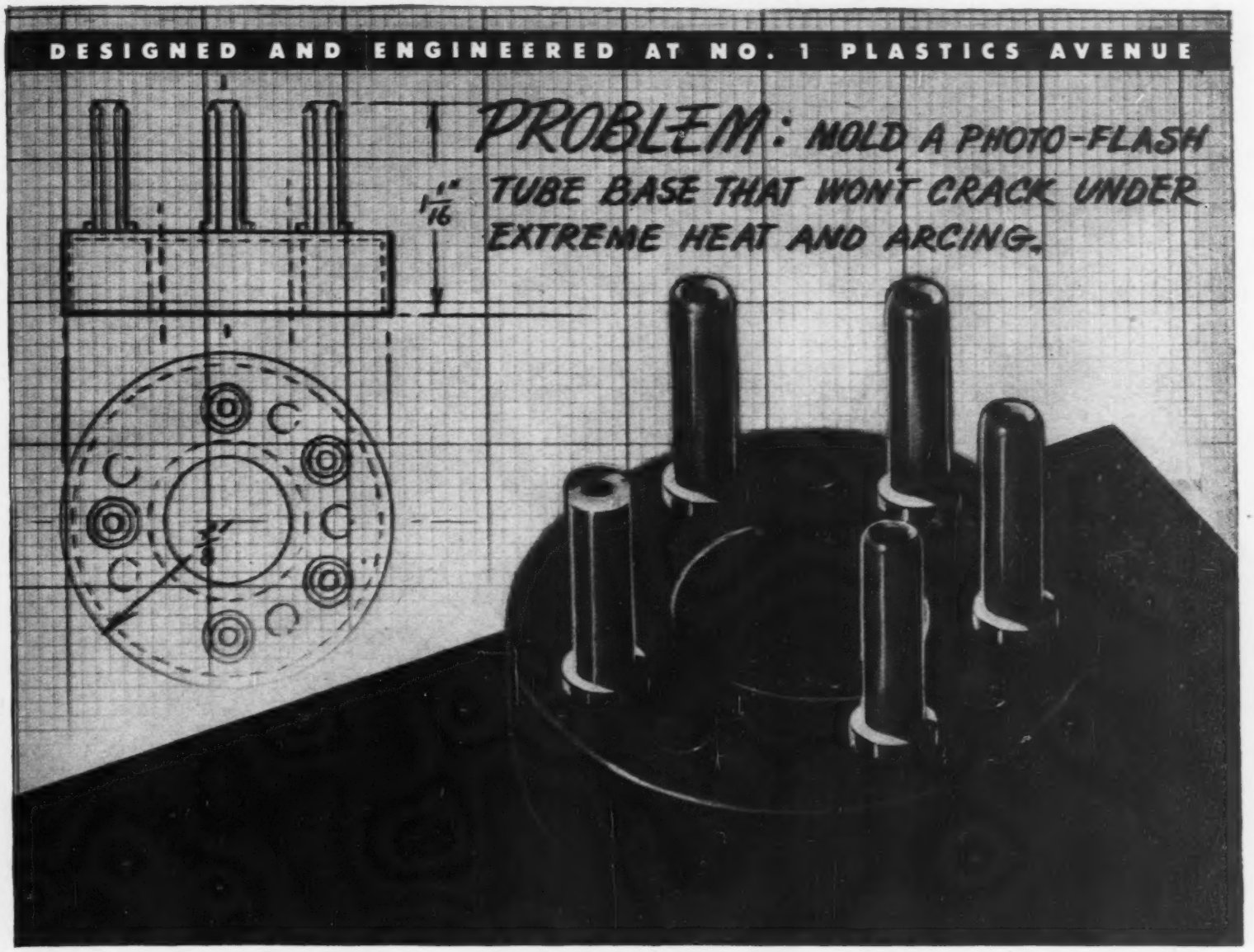
Another useful and interesting phase of acoustics in wartime hinged upon the use of sounds for



Pressure proof submarine reproducer

morale purposes, positive in the case of our own troops, negative in the case of the enemy. Thus the early use by the Germans of sounds which engender fear—loud noises, screaming bombs, noises of dive bombers —was a particular usage in this category.

Application of acoustics to sound ranging and detection was still the preferred medium at the beginning of the war no matter whether the medium was the air where the velocity of sound is slow or water where it is much faster. Even at the end of the war some acoustic work for ranging was continuing in spite of the advent of radar. Acoustic detectors have some advantages in that the low velocity makes ranging somewhat simpler at short distances compared to radio methods, and due to the fact that with acoustic methods, the enemy makes the noise, we detect it. Acoustic systems were used for detecting


DESIGNED AND ENGINEERED AT NO. 1 PLASTICS AVENUE
PROBLEM: MOLD A PHOTO-FLASH TUBE BASE THAT WON'T CRACK UNDER EXTREME HEAT AND ARCING.
1 1/16"

NEVER FAILS TO GET THE PICTURE

● More dependable photo-flash tube bases were needed for war photography—too many of them failed. They cracked under high heat and arcing conditions.

The problem was brought to No. 1 Plastics Avenue and was solved by specifying G-E mycalex—compound of glass and powdered mica with a unique combination of properties.

Again G-E mycalex did an outstanding job. This material could be molded to a five-prong design with firm anchorage for accurately placed contact pins. And its resistance to high temperatures and electrical arcs assured freedom from cracking.

G-E mycalex photo-flash tube bases have stood up so well that not a single failure has been reported.

It will pay you to investigate G-E mycalex as a solution to your special insulation problems. It is available to all manufacturers in standard sheets and rods or custom-molded to your own design. For complete information, write Section S-6, Plastics Divisions, General Electric Company, 1 Plastics Avenue, Pittsfield, Mass.

G-E MYCALEX

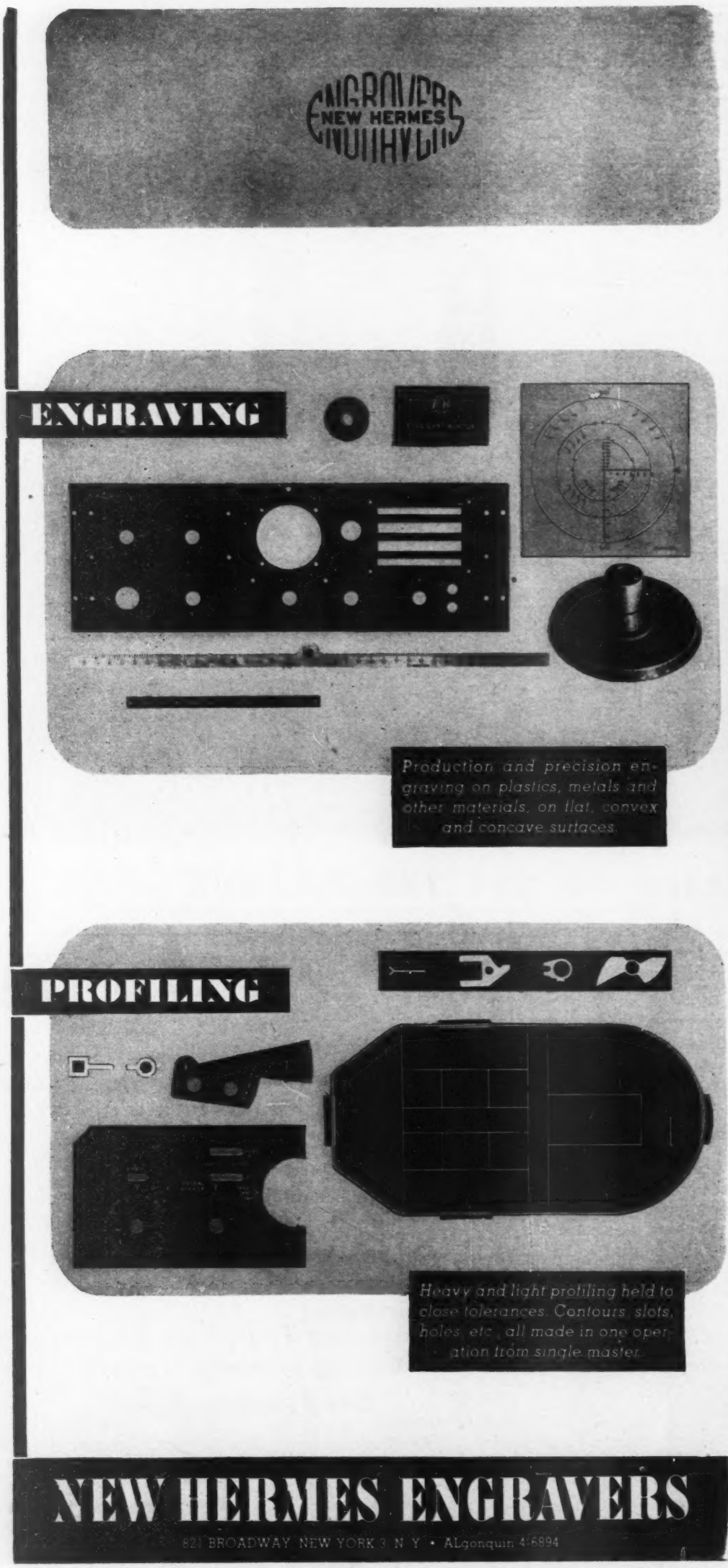
A Unique Combination of Properties

1. High dielectric strength
2. Low power factor
3. Prolonged resistance to electrical arcs
4. Chemical stability—no deterioration with age
5. Dimensional stability—freedom from warpage and shrinkage
6. Imperviousness to water, oil, and gas
7. Resistance to sudden temperature changes
8. Low coefficient of thermal expansion
9. High heat resistance

Samples Supplied on Request

GENERAL ELECTRIC

CD-46-381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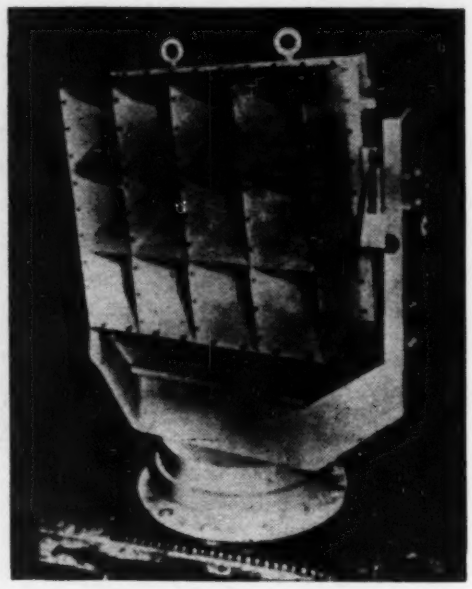
enemy artillery and motor boats.

Another aspect of acoustics which got considerable wartime attention was the problem of quieting noises not only to prevent detection by the enemy but to reduce ambient noises surrounding communication operators, and to get preferred sounds out of locations where the noise is high. The question of the inherent limitations of the ear enter into this picture as do the problems of getting intelligent sounds to a man in the intervals between other unwanted but high amplitude sounds.

Another phase of the subject is that of training men to be familiar with sounds of battle and of segregating men by their nervous characteristics. Some can tolerate noises while other men seem unable to adapt themselves to noisy conditions.



Navy high-power bull horn

The whole problem of acoustic materials adapted to high altitudes and wide humidity or temperature ranges came to have an important bearing on the production of devices made from these materials and acoustic components.

Among the materials developed during the war are: Alnico 5 which is about five times lighter than Alnico 3 for the same flux, diaphram materials which will not corrode, and new adhesives to attach voice coil to diaphram.

Interesting devices such as the bull horns on carriers, of decoys such as that used at Quetarra Depression near Alexandria where the

As the aviation industry again swings into peacetime production, Amphenol is ready to offer practical aid in many forms. Amphenol components helped to fight a winning war . . . and now Amphenol engineers with their "know-how"—deepened and strengthened by wartime experience—are cooperating in creating peacetime applications for aviation communications, electrical circuits and electronic controls. Amphenol connectors, cable assemblies and other parts provide positive electrical contacts within all types of equipment.

For detailed technical data on Amphenol products—send for Condensed Catalog No. 72.

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Case Histories from the RICHARDSON files

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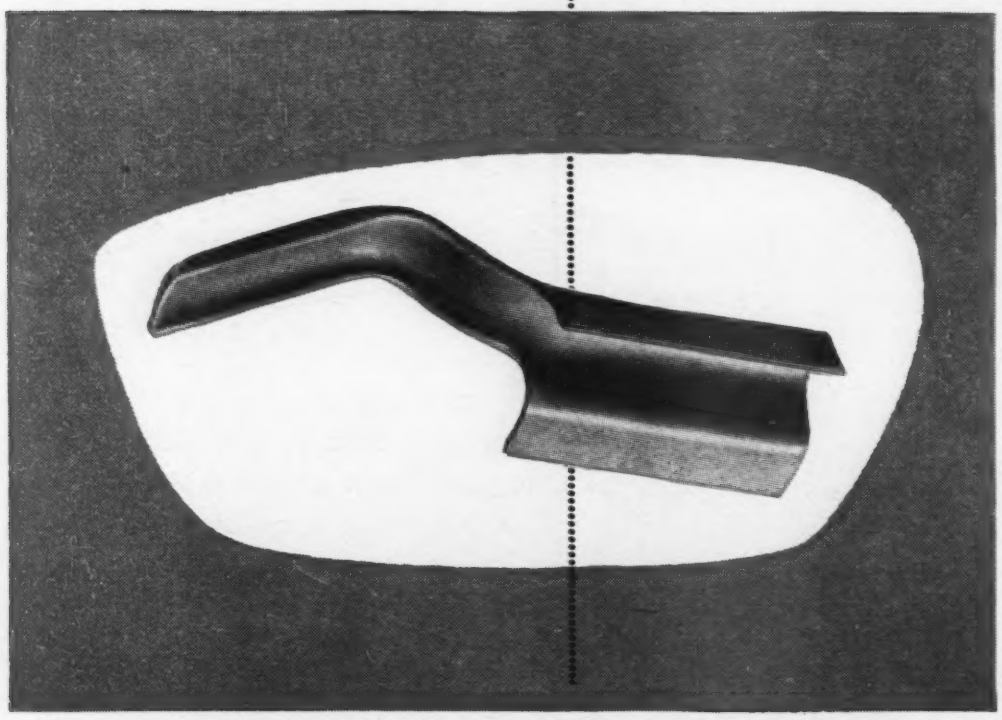
Problem: *Produce an aircraft cable guard for use where the cost of hand-working metal is prohibitive. Must be light, strong, and rigid.*

Solution: *Richardson Plasticians* used laminated thermosetting post forming materials. Specially designed tools, plus precision production methods, resulted in a laminated INSUROK cable guard that is light, strong, rigid, economical, and easy to install.

Why not discuss your product design plans with The Richardson Company? Here you will find the expert personnel, diversified facilities, and manufacturing skill to help solve your plastics problems — whatever they may be. Write today for full details.



INSUROK *Precision Plastics*

The RICHARDSON COMPANY

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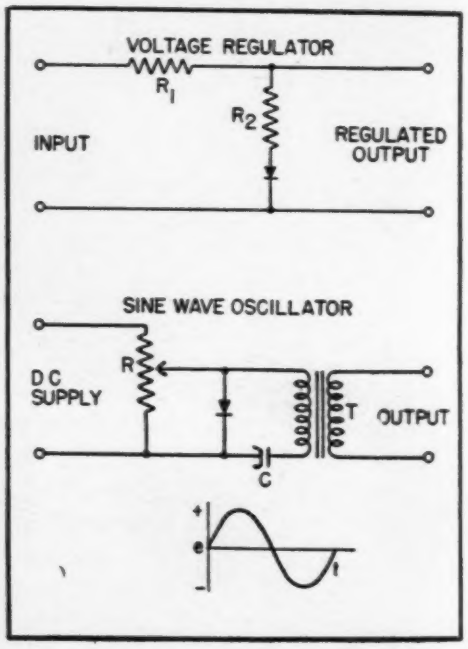
Germans were enticed into sending men and planes to a portion of the seacoast where fake sounds indicated landing operations were taking place, of new microphones, ear plugs, the beach master system of directing traffic—are acoustic instrumentalities of contemporary war.

Germanium Crystal Diode

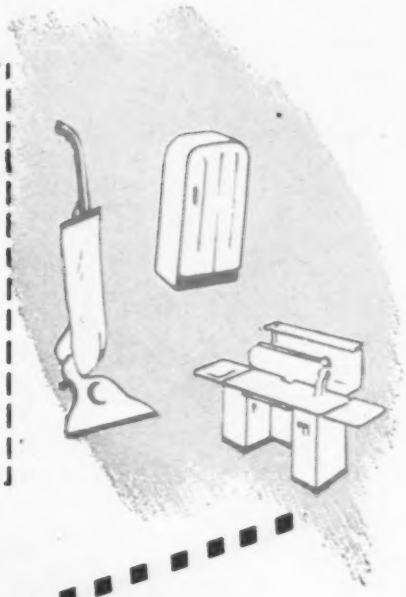
By EDWARD C. CORNELIUS
Sylvania Electric Products, Inc.
Emporium, Pa.

NONLINEAR CURRENT-VOLTAGE characteristic of the contact barrier layer between a metal and a nonmetal is partially explainable by modern subatomic physical theories. These phenomena can be utilized in a germanium crystal diode by using a fine tungsten wire whisker which has the strength to hold itself against the crystal and the conductivity not to introduce excessive loss into the circuit. The germanium is alloyed with a small amount of tin which partially dissolves in the crystals, forming a lattice-imperfection semiconductor necessary for best rectification, and also collects at the grain boundaries to lower the resistivity of the semiconductor thereby minimizing con-

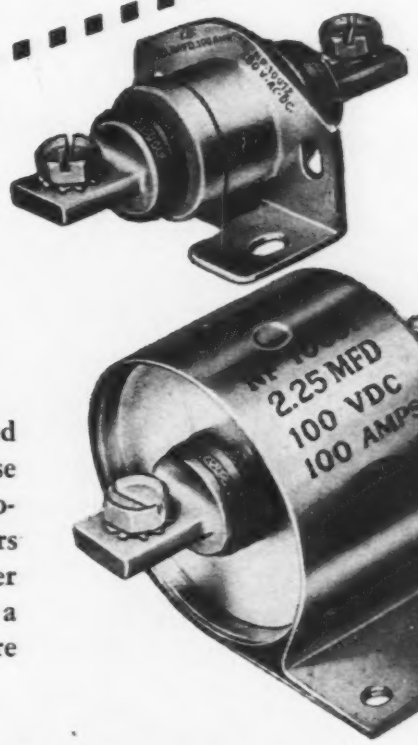


Nonlinear germanium crystal can be used as modulators, detectors, d-c restorers in television receivers, and, shown above, as voltage regulators, and low-frequency oscillators for small-current applications

stop that **noise!**



Capacitor Engineering That Licked Radio Noise



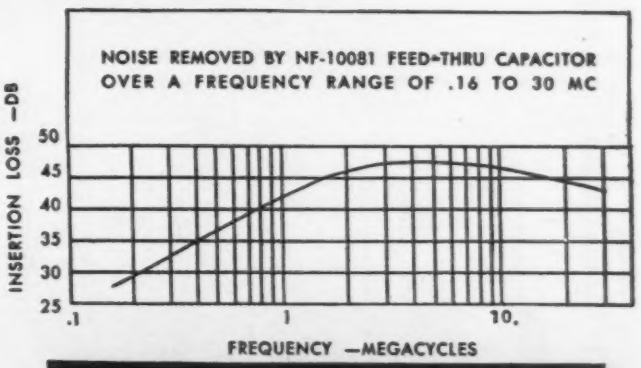
Vehicular radio equipment manufactured for military use was free from radio noise . . . thanks to the engineering that produced small compact capacitors and filters for generators, inverters, motors and other equipment. Now these can be adapted to a multitude of peace-time products where noise suppression is a "must".

The NF series of C-D feed-thru capacitors is specially designed and built for this service . . . to reduce radio noise.

Small and compact, they can be mounted in any position and will operate over a temperature range of plus 85° to minus 55° C. One power line can be fed through the unit; reducing internal inductance and resistance and increasing filtering efficiency. Rated up to 250 V AC-DC, 100 amps., in sturdy, round metal containers.

Other types of filters and feed thru capacitors are available in a range of sizes and ratings.

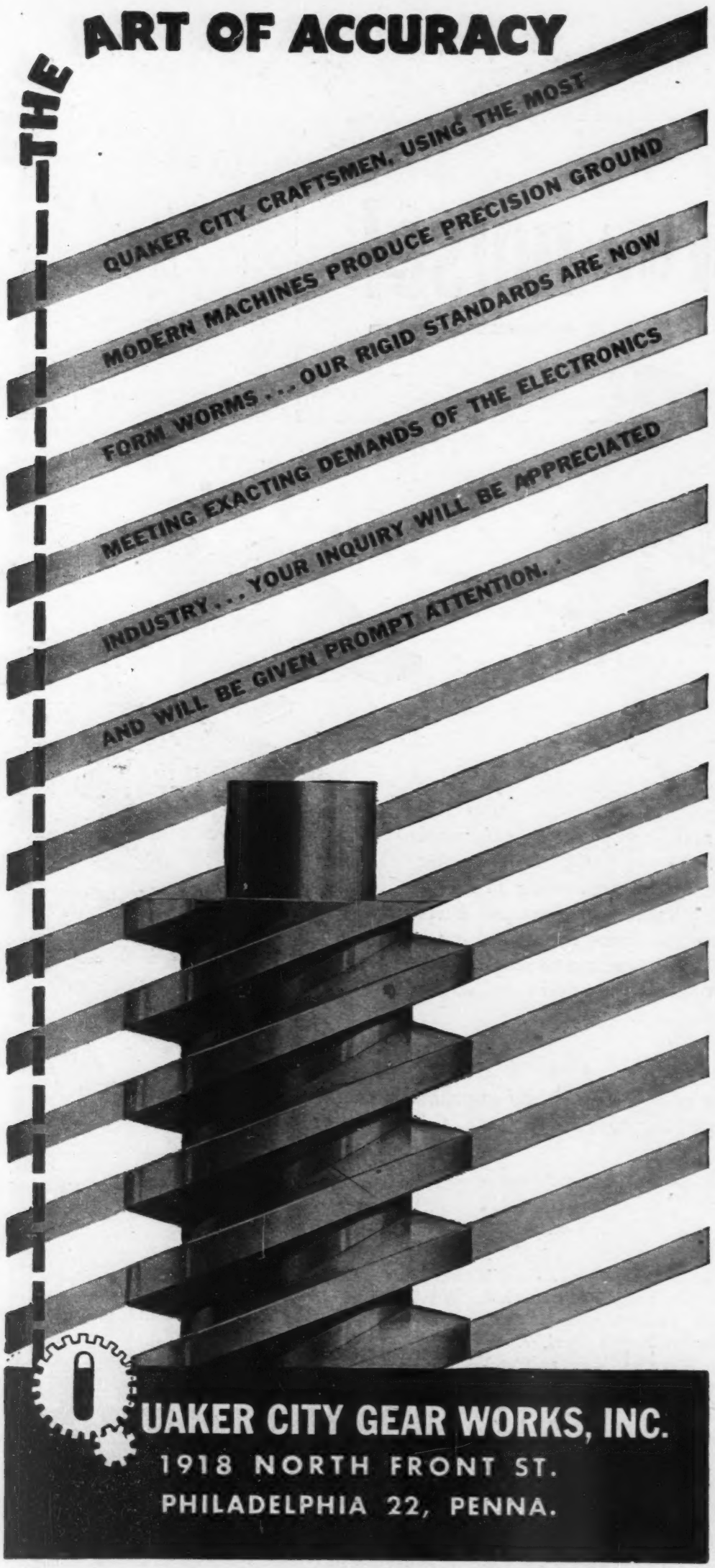
Write for information. Cornell-Dubilier Electric Corporation, South Plainfield, N. J. Other plants at New Bedford, Brookline, Worcester, Mass., and Providence, R. I.



1910 1946

MICA ★ DYKANOL ★ PAPER ★ ELECTROLYTICS

CORNELL-DUBILIER CAPACITORS


THE ART OF ACCURACY
QUAKER CITY CRAFTSMEN, USING THE MOST
MODERN MACHINES PRODUCE PRECISION GROUND
FORM WORMS . . . OUR RIGID STANDARDS ARE NOW
MEETING EXACTING DEMANDS OF THE ELECTRONICS
INDUSTRY . . . YOUR INQUIRY WILL BE APPRECIATED
AND WILL BE GIVEN PROMPT ATTENTION. .
UAKER CITY GEAR WORKS, INC.
1918 NORTH FRONT ST.
PHILADELPHIA 22, PENNA.

duction losses in that part of the circuit.

The nonlinear electrical element so produced has several uses in circuits. Its low shunt capacitance suits it for use in f-m and television receivers which have high intermediate frequencies. For operation into low-resistance loads this crystal rectifier is superior to vacuum-tube diodes.

Cathode-to-anode capacitance of the 1N34 germanium diode is 3 $\mu\mu$f. Life tests of 1,000 hours on this type rectifier indicated neither failure nor deterioration, thus indicating the feasibility of soldering the unit, as it is intended to be, into the circuit. In the audio range, production units vary less than $\pm$ 5 percent, and above 20 mc the units vary about $\pm$ 15 percent; operation above 100 mc is not recommended.

Future of Radar

By L. A. DuBridge
Massachusetts Institute of Technology
Cambridge, Mass.

RADAR WILL HAVE two important fields of application—navigation of ships, and navigation and traffic control for aircraft. Techniques developed during the war for radar will find use throughout the field of electronics and radio. These electronic and high-frequency techniques may, indeed, be the most important peace-time result of the radar war research. But the peace-time applications of radar itself will be the main subject of this discussion.

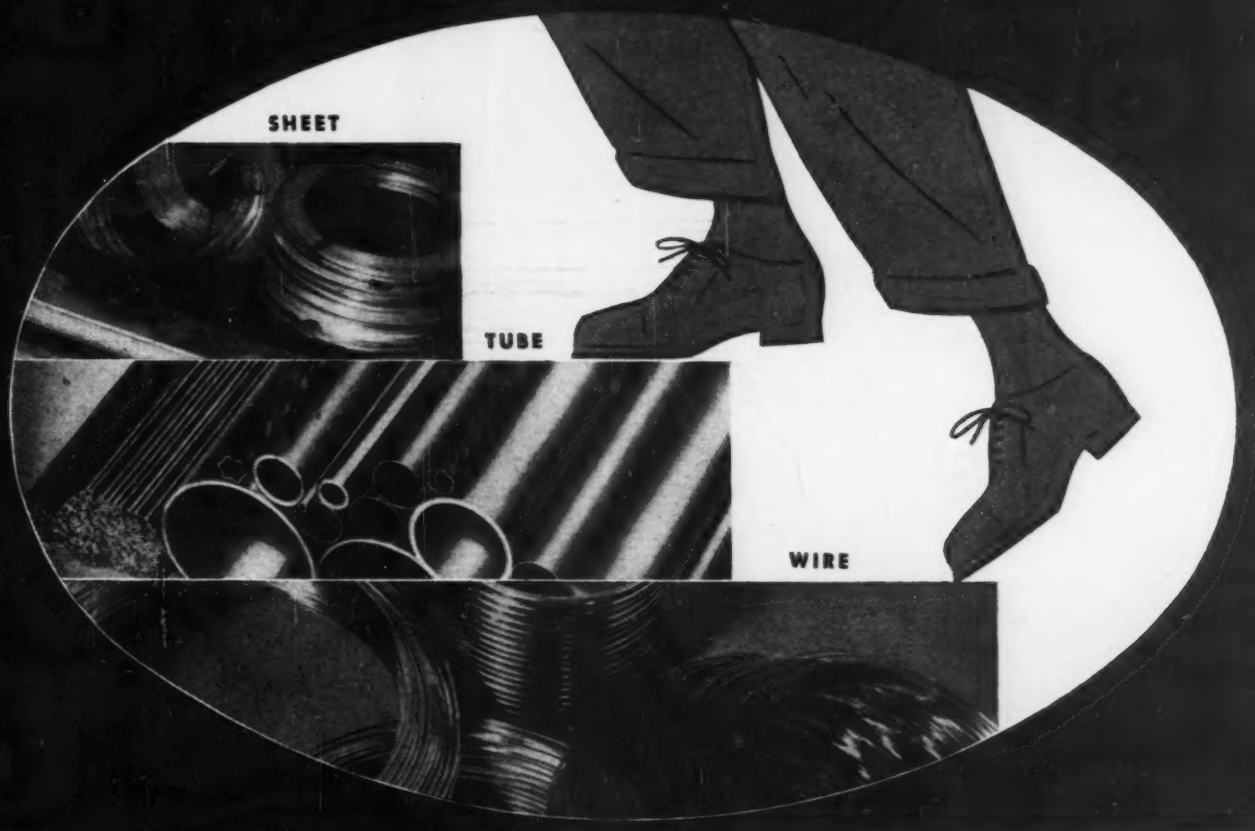
Radar for navigation is in such widespread use in naval vessels that it is hard to realize that commercial ships do not yet have this facility. With a suitable radar set a ship may sail safely in the thickest weather or the darkest night through congested harbors, narrow waters, and iceberg infested seas.

Ship radar for this purpose should, for best results, use a centimeter wavelength. Techniques for using these wavelengths have now been highly perfected. For special purposes, wavelengths of around one centimeter may be used.

Very short wavelengths are needed because they provide sharp beams with a small antenna. Thus

3 Easy Steps

FOR SOLVING A METAL PROBLEM



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General Plate Laminated Metals will increase production and cut costs in such applications as electrical contacts, giant turbines, radar and radio tubes, instruments, chemical apparatus, mobile equipment.

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A few typical combinations produced by General Plate

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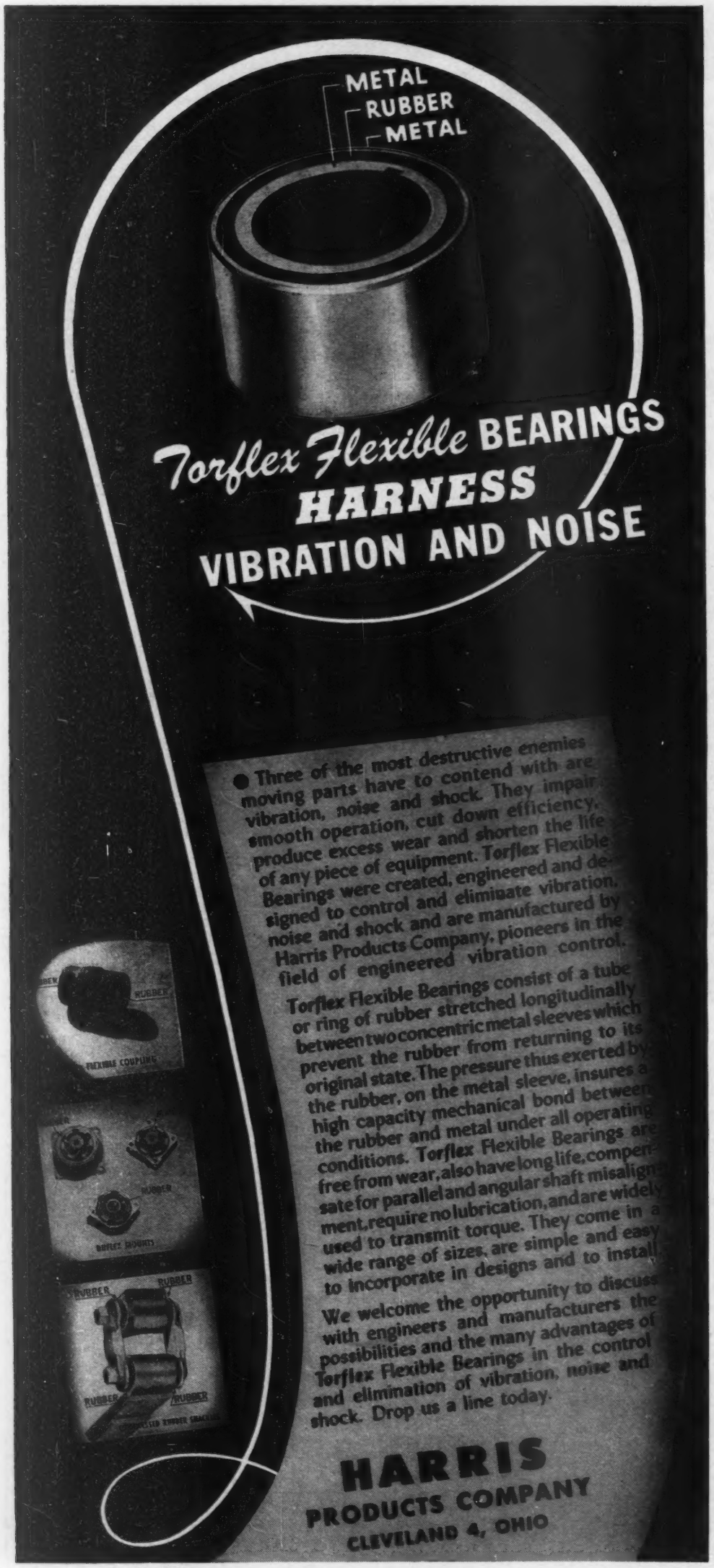
	PLATINUM	PALLADIUM	GOLD	SILVER	ALUMINUM	BRASS	COPPER	BERYLLIUM COPPER	IRON	INVAR	STAINLESS STEEL	PHOS. BRONZE	MONEL	NICKEL	SILVER SOLDER	STEEL SAE 1010	NICKEL SILVER
PLATINUM			SWT	SWT		SWT	SWT		SW	SW		SWT	SWT	SWT	SW	SW	SWT
PALLADIUM			SWT	SWT		SWT	SWT		SW	SW		SWT	SWT	SWT	SW	SW	SWT
GOLD	SWT	SWT		SWT	SWT	SWT	SWT	SW	SWT	SW	SW	SWT	SWT	SWT	SWT	SWT	SWT
SILVER	SWT	SWT	SWT		SWT	SWT	SWT	SW	SWT	SW		SWT	SW	SW	SWT	SWT	SWT
ALUMINUM			SWT	SWT		S	SWT										
BRASS	SWT	SWT	SWT	SWT	S				SW	SW	S	SW	SW	SWT	SWT	SWT	
COPPER	SW	SWT	SWT	SWT	SWT				SWT	SW		S	SW	SWT	SWT	SWT	SWT
BERYLLIUM COPPER			SW	SW													
IRON	SW	SW	SWT	SWT		SW	SWT			S	S	S	SW	SW	S		SW
INVAR	SW	SW	SW	SW		SW	SW		S		S	SW	S	S		S	SW
STAINLESS STEEL			SW			S			S	S			S	S			S
PHOS. BRONZE	SWT	SWT	SWT	SWT		SW	S		S	SW			SW	SWT	SWT	SW	SWT
MONEL	SWT	SWT	SWT	SW		SW	SW		SW	S	S	SW				SW	SW
NICKEL	SWT	SWT	SWT	SW		SWT	SWT		SW	S	S	SWT				SW	SW
SILVER SOLDER	SW	SW	SWT	SWT		SWT	SWT		S			SWT	S	S		S	SWT
STEEL SAE 1010	SW	SW	SWT	SWT		SWT	SWT			S		SW	SW	SW	S		SW
NICKEL SILVER	SWT	SWT	SWT	SWT			SWT		SW	SW	S	SWT	SW	SW	SWT	SW	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● Three of the most destructive enemies moving parts have to contend with are vibration, noise and shock. They impair smooth operation, cut down efficiency, produce excess wear and shorten the life of any piece of equipment. Torflex Flexible Bearings were created, engineered and designed to control and eliminate vibration, noise and shock and are manufactured by Harris Products Company, pioneers in the field of engineered vibration control.
Torflex Flexible Bearings consist of a tube or ring of rubber stretched longitudinally between two concentric metal sleeves which prevent the rubber from returning to its original state. The pressure thus exerted by the rubber, on the metal sleeve, insures a high capacity mechanical bond between the rubber and metal under all operating conditions. Torflex Flexible Bearings are free from wear, also have long life, compensate for parallel and angular shaft misalignment, require no lubrication, and are widely used to transmit torque. They come in a wide range of sizes, are simple and easy to incorporate in designs and to install.
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a boat which can carry only a small antenna will have an accurate set. A narrow beam gives higher resolution so that objects close together can be seen separately and irregularities in coast lines can be seen. Shorter wavelengths hug the water better. Narrow beams readily obtainable at from two to five centimeters wavelengths see less of the surface of the sea which, when rough, gives rise to a confusing clutter, due to reflections from the waves, that obscures small objects.

Airplane navigation and traffic control presents a series of complex problems. No one technique or equipment can possibly solve them all. A system of radio and radar needs to be developed and integrated. Some of the elements of which such a system could be built are loran navigation for long flights, especially over water, radar in the airplane for detecting obstacles and for seeing the lay of the land at night or through clouds, and long-range ground radar to give traffic control officers at ground stations complete pictures of the air traffic over a wide area.

Heavy storm clouds are visible on the radar screen, and planes can be given instructions from ground radar stations which will enable them to avoid such storm areas. Short-range, high-precision radar at each airport will give local controllers an accurate picture of traffic in their vicinity and will assist in giving instructions for approach, avoiding collisions, and generally keeping an orderly traffic pattern even in bad weather.

Radar and radio equipment can be used for blind landings. A microwave radio glide path or an accurate ground radar for talking a plane in, or both, may be used.

Static-free and, if desired, directional communication links at microwave frequencies to provide noiseless reliable radio communication from ground to plane can be built from the techniques developed in research on radar. These and other newly developed techniques, used in conjunction with present radio techniques, give promise of providing all-weather reliable flying, the first requirement of really successful air transportation.

THE MODERN METHOD FOR APPLYING

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Decal men like to wager they can find 20 decals in your home or office in half as many minutes. They seldom lose because more trademarks, instructions, patent data, wiring diagrams, etc., are made from Decalcomania than any other material. And Meyercord Decals lead the field. In the home—radios, pianos, washers, heaters, stokers, tools, furniture, even toilet goods. In the office or factory—desks, typewriters, filing cabinets, safes, machinery are but a few of the items your count will include. Meyercord Decals are durable, easily applied at production speeds and can be produced in any colors, size or design for application to all commercial surfaces. Investigate their use on your product. Free technical service. Address inquiries to Department 9-1

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NEWS OF THE INDUSTRY

IRE convention program and data on other meetings to come; counter-mortar radar; guided missiles; f-m channel numbers; many engineers change jobs

Outline of Proposed National Science Foundation

A NATIONAL SCIENCE Foundation having an integral and independent position in the structure of the Government is now definitely assured. It remains for Congress to define the executive setup of the agency, the types of research with which it will concern itself, the nature of its financial support by the Government, and the status of patents obtained by research contractors.

The reasons advanced for formation of a science foundation are: (1) requirements of national security; (2) promotion and maintenance of better national health through medical research; (3) expansion and stimulation of basic scientific research; (4) further development of industrial facilities in the interest of full employment; (5) revival of America's badly depleted scientific talent; (6) publication and dissemination of scientific knowledge including important items held back for security reasons, and the unfettered exchange of basic scientific knowledge, nationally and internationally.

Basic research is to be given top priority in the work of the foundation. This type of research is very

Dr. Vannevar Bush, OSRD director speaking before a Senate committee considering a national science foundation, stated, "A nation which depends upon others for its new basic scientific knowledge will be slow in industrial progress and weak in its competitive position in world trade, regardless of its mechanical skill."

expensive and often slow in bringing results, hence Government will assume part of it as a public responsibility.

Purpose

Basic operating principles now generally accepted are: The function of the foundation will be to implement and energize, and in no sense to supplant any existing agency whether private or governmental. It would not operate laboratories or research facilities of its own. It would work primarily through the colleges, universities and research institutions, rather than through industrial laboratories, except for special products for which these are especially fitted.

Executive Setup

Opinions differ as to top authority for the entire project. One plan would place top power in a board of nine men, chosen by the President on the basis of their demonstrated capacity for the work of the foundation, without regard to political affiliation. They would select the executive head of the foundation and lay down the policies and programs to be followed. Another plan would give the top power to a director, selected by the President with the advice and approval of the Senate, who would set up an advisory board of sixteen, eight of whom would be government officials and the other eight public members.

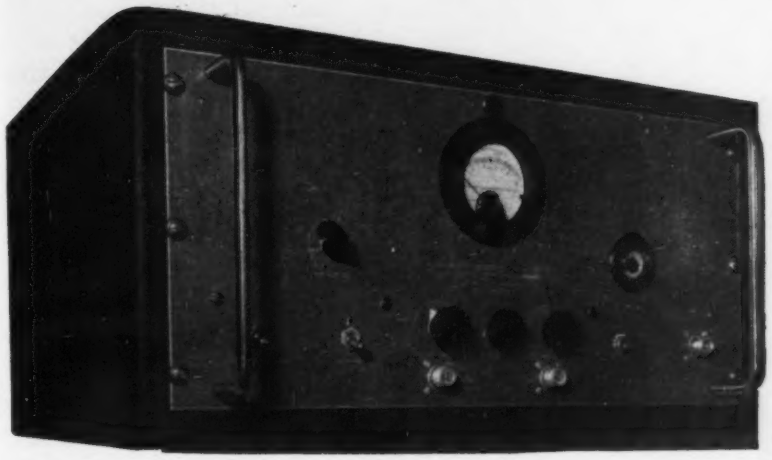
The foundation will be subdivided into various sections, each of which will deal with one of the major areas of the foundation's work, such as: national defense; medical research; physical science; publications and exchange of information; scholarships; possibly others, including a division of social sciences.

Patents

Another highly controversial issue is the control of patents. Existing government agencies work

RADIO PIONEERS GET TOGETHER IN NEW YORK

Over 900 radio pioneers and future pioneers attended the annual Radio Pioneers dinner staged by the New York section of the Institute of Radio Engineers, held Nov. 8, 1945 at the Hotel Commodore. A program featuring reminiscences of the good old days followed the dinner



C-200 Harmonic
Frequency Generator

Lavoie Laboratories

. . . . are *Specialists* in the design, development and manufacture of High Frequency equipment. LAVOIE plant procedure, personnel and equipment are developed especially for this type of production. The LAVOIE trade mark is your guarantee of precision manufacture and dependable performance.

LAVOIE products include:

- **FREQUENCY STANDARDS**
- **FREQUENCY METERS**
- **RECEIVERS**
- **TRANSMITTERS**
- **ANTENNAS and MOUNTS**

Lavoie Laboratories

RADIO ENGINEERS AND MANUFACTURERS
MORGANVILLE, N. J.

Write for detailed information.

Fixed Frequency Receiver

Specialists in The Development of UHF Equipment
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In War or Peacetime..

THE STANDARD SOLDER

for Every Application

• Whatever the production changeover . . . from Radar to FM Receivers or tanks to roadsters . . . Kester Cored Solders remain industry's standard—right for every type of precision and quality manufacturing!

• Kester Cored Solders, a number-one aid to speedy reconversion, are virtually mistake-proof in application—trouble-free in operation. The flux is right in the core, scientifically balanced with superior alloys, ready to be applied in one simple operation. Solder-bonds formed the Kester way are clean, tight, and hold permanently against shock, vibration, bending, and the contraction and expansion of temperature extremes.

• *Kester Rosin-Core Solder,* specially compounded for electrical application, will not harm insulation nor cause corrosion. *Kester Acid-Core Solder,* for general work, is the ideal all-purpose solder. Both are available in a wide range of strand and core sizes, flux and alloy combinations.

• The 47 years of practical Kester experience and research are at your service. Kester engineers will gladly work with you on any solder problem—and at no obligation. Write them fully, anytime.

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KESTER

Cored Solders

STANDARD FOR INDUSTRY

out such patent arrangements with research contractors as are in line with public interest.

Some believe that all inventions or discoveries resulting from projects which have been financed in whole or even in small part by the foundation shall be the sole property of the United States Government.

Others, feeling that industrial laboratories and inventors who have put a considerable investment in a certain field of research would be hesitant to take on government research on such terms, believe that the patent rights which the Government should acquire depend upon the relative degree of the Government's contribution to the particular research project as compared with the contribution of the private organization. Here the Government would get full patent rights to inventions in fields of particular public importance, such as medicine, but otherwise the contractor would hold the rights and grant a royalty-free license in favor of Government.

Finances

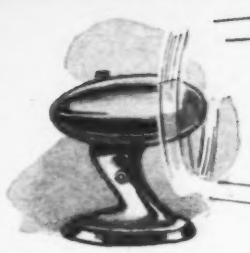
Rough estimates indicate that the cost for the first year of operation would be approximately thirty-three and a half million dollars, rising in the fifth year to about 122.5 million dollars as follows:

Activity	Millions of Dollars: First Year	Fifth Year
Medical Research	$5.0	$20.0
Natural Sciences	10.0	50.0
National Defense	10.0	20.0
Scientific Personnel and Education	7.0	29.0
Publications and Scientific Collaboration	0.5	1.0
Administration	1.0	2.5
	33.5	122.5

IRE Winter Technical Meeting and Radio Engineering Show

THE FIRST POSTWAR Winter Technical Meeting and Radio Engineering Show of the Institute of Radio Engineers at the Hotel Astor, New York, January 23rd through 26th, 1946, gives indications of being one of the largest and most significant gatherings of its type ever held.

The 150 booths originally planned, including three theater booths, were all taken in November by 124



more efficient ...in miniature

The old, slow motion belt driven fan was of questionable value as a breeze maker. Perhaps its best service was that of chasing flies with fluttering streamers. Then came the modern high speed electric fan. Like the miniature electronic tube, it is an outstanding example of the current trend toward increased efficiency in miniature.

TUNG-SOL Miniature Tubes are a part of the trend to smaller component parts. They are a factor in reducing the over-all size of equipment. Shorter leads with low inductance, and low capacity with high mutual conductance make the miniature tube ideal for high frequency circuits. The smaller elements weigh less, tending to reduce the effects of vibration. The smaller size also makes possible a more rigid construction. This reduces the possibilities of element distortion.

ACTUAL SIZE

To aid in the creation of new electronic equipment and in the improvement of old, TUNG-SOL engineers will draw upon their experience and work with manufacturers in the designing of circuits and in the selection of tubes. Of course your plans will be held in strictest confidence.

TUNG-SOL

vibration-tested

ELECTRONIC TUBES

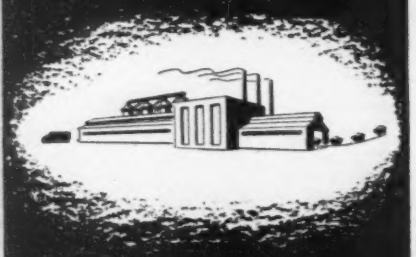
TUNG-SOL LAMP WORKS INC., NEWARK 4, NEW JERSEY

Sales Offices: Atlanta • Chicago • Dallas • Denver • Detroit • Los Angeles • New York

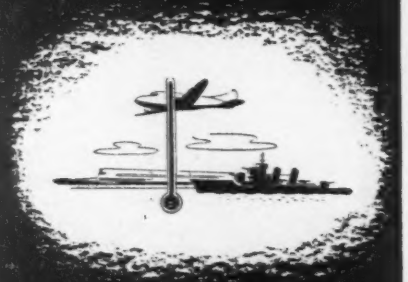
Also Manufacturers of Miniature Incandescent Lamps, All-Glass Sealed Beam Headlight Lamps and Current Intermittors

THERMOTECHNICS

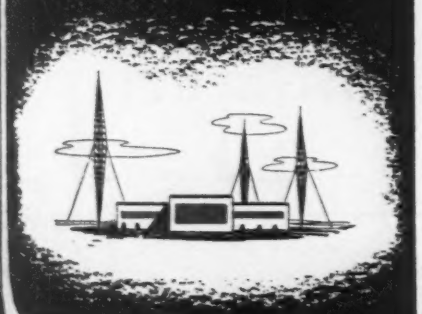
IS IT METALLURGY?



IS IT A THERMAL SCIENCE?



IS IT A METHOD OF COMMUNICATION?



IT IS ALL THREE!

THERMOTECHNICS — a new art in temperature detection . . . based upon known phenomena of the expansion and contraction of metals; it includes regulatory media that harness temperatures to industry's bidding; and, it communicates silently and invisibly critical observations by the action of THERMOTECHNICS' unique instrument, the THERMOSWITCH, and directs appropriate action.

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To learn about THERMOTECHNICS, send for the latest brochure which will be mailed to you upon request... of course, no obligation. Then fill out the information sheet which is enclosed with the brochure and submit the data to our THERMOTECHNIC Analysis Group which will be glad to make a THERMOTECHNIC analysis of your problem and submit the most effective application of THERMOSWITCHES to your specific requirements.

Today, THERMOTECHNICS serves a thousand industries, from Aviation to X-ray equipment.

THERMOTECHNICS can serve you, too . . . economically, efficiently, simply and ruggedly.

Fenwal

exhibitors, for displays of postwar products. The show is scheduled to open at 4:00 p.m., Wednesday, January 23rd, and will close at 4:00 p.m. Saturday, January 26th.

PROGRAM OF EVENTS

WEDNESDAY, JANUARY 23, 1946

9:00 a.m.—5:30 p.m. Promenade—Registration and Sale of Tickets

9:30 a.m.—12:30 p.m. Coral Room—Annual Meeting of Sections' Representatives

12:30 p.m.—2:00 p.m. Rose Room—Luncheon for Sections' Representatives

2:00 p.m.—5:00 p.m. Coral Room—Annual Meeting of Sections' Representatives

4:00 p.m.—8:00 p.m. Eighth and Tenth Floors—Radio Engineering Show

6:00 p.m.—10:00 p.m. Engineering Societies Building—Joint Meeting of A.I.E.E. and I.R.E.

THURSDAY, January 24, 1946

8:30 a.m.—5:30 p.m. Promenade—Registration and Sale of Tickets

9:00 a.m.—7:00 p.m. Eighth and Tenth Floors—Radio Engineering Show

9:45 a.m.—10:30 a.m. Grand Ballroom—Annual Meeting of The Institute of Radio Engineers, Inc.

TECHNICAL SESSIONS
10:30 a.m.—12:30 p.m.

Group A—Grand Ballroom
Military Applications of Electronics

Group B—Rose Room
Frequency Modulation and Standard Broadcasting

Group C—Coral Room
Circuits and Theory

TECHNICAL SESSIONS
2:00 p.m.—5:00 p.m.

Group A—Grand Ballroom
Television

Group B—Rose Room
Radio Navigation Aids

Group C—Coral Room
Vacuum Tubes

Annual I.R.E. Banquet (Dress Optional) 7:30 p.m.—10:30 p.m. Grand Ballroom

Awarding of Medal of Honor, Morris Liebmann Memorial Prize, and Fellowship Awards. Address of Retiring President

Speaker—Dr. Frank B. Jewett, President of the National Academy of Sciences

Toastmaster—Mr. Edgar Kobak, President of the Mutual Broadcasting System, Inc.

FRIDAY, JANUARY 25, 1946

9:00 a.m.—5:00 p.m. Promenade—Registration and Sale of Tickets

9:00 a.m.—10:00 p.m. Eighth and Tenth Floors—Radio Engineering Show

TECHNICAL SESSIONS
9:30 a.m.—12:00 Noon

Group A—Grand Ballroom
Microwave Vacuum Tubes

Group B—Rose Room
Antennas

President's Luncheon Honoring Dr. Frederick B. Llewellyn 12:30 p.m. Grand Ballroom. Speaker—Mr. Paul Porter, Chairman, Federal Communications Commission. Master of Ceremonies—Mr. Lewis M. Clement

"Railroads . . . Like a Giant Conveyor Belt"

"The war has emphasized the importance of American railroads. Like a giant conveyor belt, they link up the industrial, agricultural and mining areas of this country with the many thousands of markets that dot our land. With reconversion a fact, far-sighted railroad management is carefully exploring many technological war developments, and, in particular, radio, with the expectation that radio will help keep American railroads the safe, efficient and modern network of transportation which has so ably served the Nation during the war."

S P Kennedy

President, Detroit, Toledo & Ironton R. R. Co.

RADIO *has* a story to tell the railroads . . . a story that will contribute to their continued safety, efficiency and economy of operation. Through its Mobile Communications Division, the Farnsworth Television & Radio Corporation is now telling its railroad radio story to railroads in all parts of the country.

Farnsworth brings to railroad radio specially designed, thoroughly tested equipment, utilizing either space radiation or induction principles . . . equipment which gives positive, unfailing voice communication between the operating units of a railroad, whether moving or stationary.

Furthermore, Farnsworth brings sound engineering, backed up by adequate facilities and by eight years of research into the problems of railroad radio . . . the experience of the Halstead Traffic Communications Corporation, whose assets and key personnel were recently acquired by Farnsworth. *For instance, the corrosive effects of coal gases on radio equipment is one of a number of problems which have been solved in the design of Farnsworth Mobile equipment.*

Write for the complete story of Farnsworth railroad radio. Address the Farnsworth Television & Radio Corporation, Dept. E-1, Fort Wayne 1, Indiana.

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Farnsworth Radio and Television Receivers and Transmitters • Aircraft Radio Equipment • Farnsworth Television Tubes • Halstead Mobile Communications and Traffic Control Systems for Rail and Highway • the Farnsworth Phonograph-Radio • the Capehart • the Panamuse by Capehart

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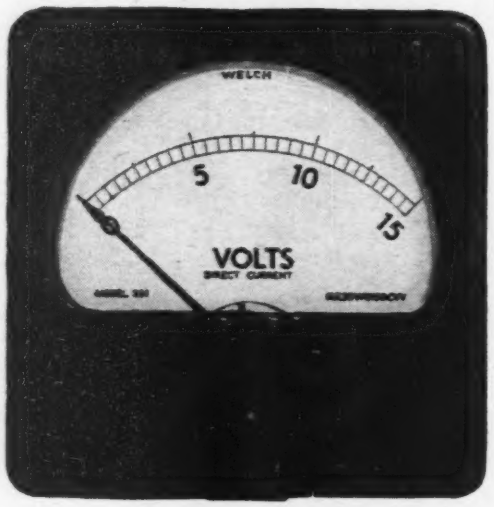

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FEATURES

- Alnico Magnets
- Sapphire Jewel Bearings
- Selected Steel Pivots
- Fume-proof White Metal Scales
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MODELS

350—3½ inch round—2.4 inch scale
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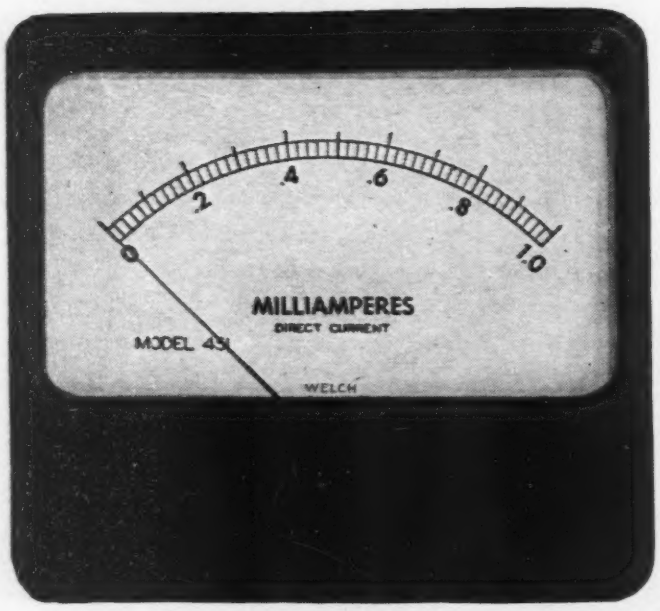

MODEL NO. 451

TYPES

Microammeters 0-20 up to 0-500

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Voltmeters, various ranges, 100 to 50,000 ohms per volt

Zero Center Ammeters

DB Meters

Rectifier Instruments

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Special Sealed Meters

PRICES

$9.00 to $26.00

THE PRODUCT OF MORE THAN 50 YEARS EXPERIENCE IN BUILDING HIGH GRADE INSTRUMENTS

W. M. WELCH SCIENTIFIC COMPANY

Established 1880

1515 Sedgwick Street, Dept. H, Chicago 10, Illinois, U. S. A.

Vice President in charge of Research and Engineering, The Crosley Corporation

TECHNICAL SESSIONS
2:00 p.m.—5:30 p.m.

Group A—Grand Ballroom
Radar

Group B—Rose Room
Microwave Technique

6:30 p.m.—8:00 p.m.—Cocktail Party

SATURDAY, JANUARY 25, 1946

9:00 a.m.—3:00 p.m. Promenade—Registration
9:00 a.m.—2:00 p.m. Eighth and Tenth Floors—Radio Engineering Show

TECHNICAL SESSIONS
9:30 a.m.—12:00 Noon

Group A—Grand Ballroom
Industrial Electronics

Group B—Rose Room
Communication Systems and Relay Lines

Group C—Coral Room
Radio Propagation

TECHNICAL SESSIONS
2:00 p.m.—4:00 p.m.

Group A—Grand Ballroom
Broadcast Receivers

Group B—Rose Room
Quartz Crystals

Group C—Coral Room
Crystal Rectifiers

Final Adjournment—4:00 p.m.

COMMITTEE MEETINGS
(Open to Members of Committees Only)

WEDNESDAY, JANUARY 23, 1946

Morning: Antennas; Radio Receivers; Frequency Modulation; Radio Wave Propagation.

Afternoon: Circuits; Research; Membership; Television; Vacuum Tubes; Railway and Vehicular Communication

THURSDAY, JANUARY 24, 1946

Morning: Standards

Afternoon: Education; Public Relations

WOMEN'S PROGRAM
(Tentative)

THURSDAY, JANUARY 24, 1946

11:00 a.m.—4:00 p.m.

Museum of Costume Art—Sloane's House of Years—Luncheon and Art Exhibition, Town Hall Club—Television Tour of Radio City

FRIDAY, JANUARY 25, 1946
11:00 a.m.—5:00 p.m.

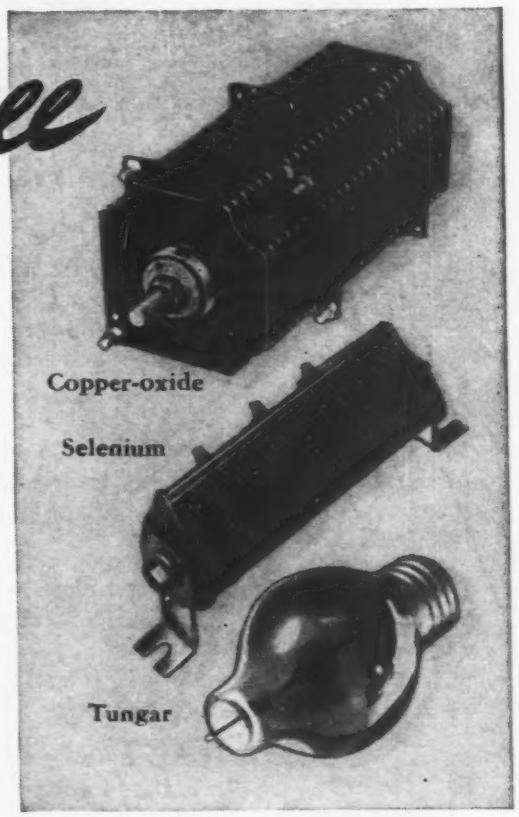
St. John's Cathedral—Luncheon, Stoddards—Riverside Church—Tea, Castleholm

Counter-mortar Radar

TO REDUCE THE NUMBER of casualties among American ground forces from enemy mortar fire, the SCR-584 antiaircraft fire control radar was modified to search at a low angle. It then revealed the rise and fall of enemy mortar shells, giving points on the trajectory from which the position of the mortar could be

THE CORRECT ANSWER TO YOUR RECTIFIER PROBLEM IS HERE

G-E Builds All Three



Selecting the rectifier best suited for a particular d-c application is not a decision that can be made on a "guess" basis. Construction, basic materials, operating characteristics, weight, size, cost and life expectancy are all factors that should be considered.

G.E. and only G.E. builds the three types of low-voltage rectifiers most generally used—copper-oxide, selenium and Tungar. All three are tops in quality and leaders in their field. To say that one type is better than another is as fatuous as saying a bomber is better than a fighter plane. Each performs best when doing the job for which it was specifically designed.

When blueprints call for rectifiers choose the correct size and type from the G-E line. If you're not sure of what is best for your need let G-E engineers help you. Years of experience qualify them to recommend the rectifier which will give you the most economical, most efficient and most reliable performance. Whether they recommend copper-oxide, selenium or Tungar you can be sure their selection is impartial because G.E. offers all three.

For more information write to Section A1613-119, Appliance and Merchandise Dept., General Electric Co., Bridgeport, Conn.

Hear the General Electric radio programs: "The G-E All Girl Orchestra" Sunday 10 p.m. EST, NBC. "The World Today" news every weekday 6:45 p.m. EST, CBS. "The G-E House Party" Monday through Friday 4:00 p.m. EST, CBS.

BUY VICTORY BONDS AND KEEP THEM

GENERAL ELECTRIC

READY NOW!

A MULTI-FEATURE PILOT LIGHT:

"DIALCO" Presents:—
The New PLN-849 PILOT LIGHT

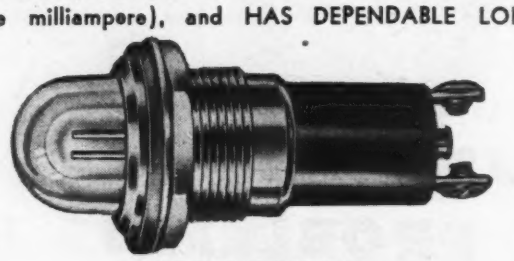
featuring

THE NEW NE-51 NEON BULB

with BUILT-IN RESISTOR

FOR 110 VOLTS (and higher)

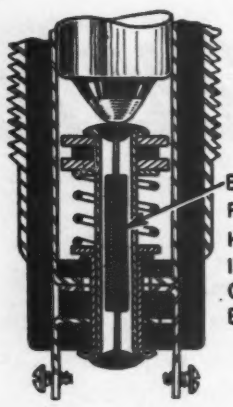
A RUGGED UNIT, CONSUMES A SMALL AMOUNT OF CURRENT (under one milliampere), and HAS DEPENDABLE LONG LIFE.



PATENT PENDING

NOTE THESE IMPORTANT FEATURES OF THE PLN-849 PILOT LIGHT:—

1. RESISTOR INTEGRAL WITH SOCKET ASSEMBLY—VALUE TO SUIT SUPPLY VOLTAGE
2. Moulded Bakelite Socket.
3. Full-view Jewel Plastic Cap for visibility at all angles.
4. Rugged Terminals: Binding Screw or Permanent Soldering Type.
5. High resistance to vibration or shock.



Manufacturers . . . here's the ultimate in Pilot Light design. A compact, rugged unit—housing a BUILT-IN RESISTOR as an integral element of the assembly (not externally attached, or fastened to body or terminal).

The PLN-849 is supplied complete with General Electric Neon NE-51 Bulbs. May also be adapted to accommodate General Electric Radio Panel Bulbs such as 47, 44, etc., for low voltage circuits. Bulbs are removable from front of panel. Plastic Jewel Cap may be had in choice of 7 colors. Units are suitable for all panel thicknesses.

DIALCO

Dialco mass production methods make possible a price so low that you can have the advantage of the PLN-849 110-V. Neon Pilot Light on all of your newly designed products.

Write today for samples and prices. There is no obligation.

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plotted. This unique tactical use of radar led to the design of a new set specifically for directing counter-mortar fire.

Another wartime application of the SCR-584 was in detecting enemy patrols and vehicles in the dark. Its use enabled the Allies to keep Germans off their main supply roads near the front lines at night with minimum expenditure of ammunition. Radar made it possible to fire only when targets actually appeared on the roads, without resorting to costly random interdictory fire. The same radar often alerted American outposts when German patrols and raiding parties approached on dark nights and thereby contributed materially to the Allied collection of German prisoners.

New IRE Officers

THE ELECTION OF Dr. Frederick B. Llewellyn of Summit, New Jersey, as president of the Institute of Radio Engineers for the year 1946, has been announced. He succeeds Dr. William L. Everitt, head of the Department of Electrical Engineering of the University of Illinois.

Dr. Llewellyn, a consulting engineer on the staff of Bell Telephone Laboratories, has specialized in the

Dr. Frederick B. Llewellyn, IRE president for 1946

design of vacuum tubes for communication and electronic control purposes. In 1936 he received the Morris Liebmann Memorial prize for his analysis of reactions within the vacuum tube.

E. M. Deloraine, president of International Telecommunication Laboratories, New York, becomes the new vice president. Three directors were also elected: Dr.

A WAVEMETER with very useful range

NOW AVAILABLE (at the moment from stock) the popular Type 566-A Wavemeter leaves the war effort to return to the civilian laboratory. This meter with its very wide range of 0.5 to 150 Mc (600 to 2 meters) is an extremely handy gadget in any laboratory. Its accuracy is sufficient for a large number of measurements such as determination of coil ranges, oscillator spans, lining up oscillators and transmitters, locating and naming harmonics in either the receiver or the transmitter, and for general experimental work.

All five plug-in coils are stored in a rack on the side of the cabinet. It weighs only three pounds and can be held in the palm of one hand.

Despite its modest price, this wavemeter is built with the same care and is calibrated with the same thoroughness as the most expensive piece of G-R measuring gear.

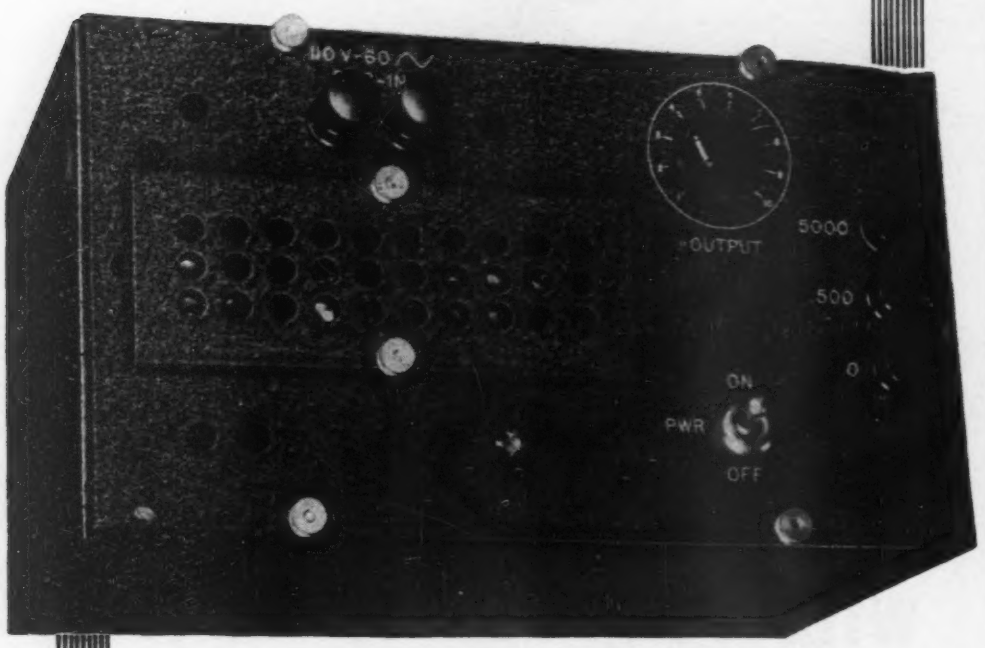
ORDER NOW. Shipment probably can be made from stock.

SPECIFICATIONS
TYPE 566-A WAVEMETER

FREQUENCY RANGE: 0.5 to 150 Mc
COILS: Five plug-in coils furnished
DIAL: Direct reading in frequency to 2% or better
ACCURACY: 2% for 0.5 to 16 Mc; 3% for 16 to 150 Mc
RESONANCE: Indicator is small incandescent lamp. Two spares supplied
DIMENSIONS: 4¾ x 5⅞ x 5¾ inches, overall
NET WEIGHT: 3 pounds
PRICE: $45.00

GENERAL RADIO COMPANY Cambridge 39, Massachusetts
90 West St., New York 6 920 S. Michigan Ave., Chicago 5 1000 N. Seward St., Los Angeles 38

Now... a compact audio signal source for laboratory and production test applications



The Type MR-1000 is a compact source of audio frequency power producing up to 35 MW of signal energy at output impedances of 500 or 5000 ohms.

The distortion at maximum output is less than 2% over all. The built-in attenuator features an arbitrary scale of 1 to 10 units.

The Type MR-1000 operates from a standard source of 110 volts, 60 cycles.

The case, panel and subchassis are 1/4" aluminum. A removable ventilator panel at the top permits easy tube replacement.

Using a new type of oscillating amplifier circuit, this unit maintains an output frequency of 1000 cycles at plus or minus 1 cps over 24 hours and will operate continuously for 720 hours with a variation of less than 4 cps.

OPERATION:—The MR1000 serves to replace tuning forks and offers a source of audio test energy for bridges etc. and can be substituted in existing test positions without wiring changes. The MR-1000 can be furnished at frequencies other than 1000 cps at slight additional cost.

SIZE:—6 1/4" deep x 5 3/4" high x 9 1/4" long.

WEIGHT:—13 1/2 lbs.

PRICE:—$149.00 f.o.b. Chicago—Guaranteed—2 years.

TELEVISO PRODUCTS CO.
7466 IRVING PARK ROAD CHICAGO 34, ILL.

Walter R. G. Baker, vice president of General Electric Co., Syracuse, New York; Dr. Donald B. Sinclair, assistant chief engineer of General Radio Co., Cambridge, Mass.; Virgil M. Graham, plant manager of Sylvania Electric Products, Inc., Williamsport, Pa. Installation of the new officers will take place at the annual meeting in New York on Jan. 23 which is the first day of the Winter Technical Meeting.

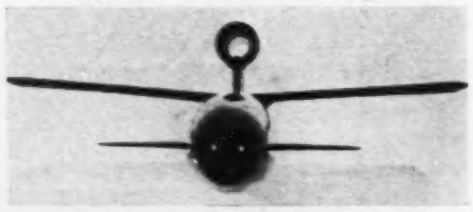
Chicago Engineering Conference

THE CHICAGO SECTION of the Institute of Radio Engineers announces the Chicago Engineering Conference and Banquet, February 9th, 1946, to be held at the Merchants and Manufacturers Club in the Merchandise Mart.

Dr. W. L. Everitt, IRE president in 1945, will deliver the opening address at 9:30 a.m. In one of two concurrent technical sessions the subjects discussed will be F-M Receiver Design Problems, Scanning and High Voltage Supplies for Television, Design of Signal Frequency Circuits for F-M and Television, and Broad Band Antenna Design. In the other session the subjects will be The Vacuum Tube Proximity Fuse, The Loran System, Atomic Energy, and High-Frequency Heating. Nationally known speakers will discuss each subject. The technical sessions will run from 10:00 a.m. to 5:00 p.m. Dinner and dancing will conclude the evening's entertainment.

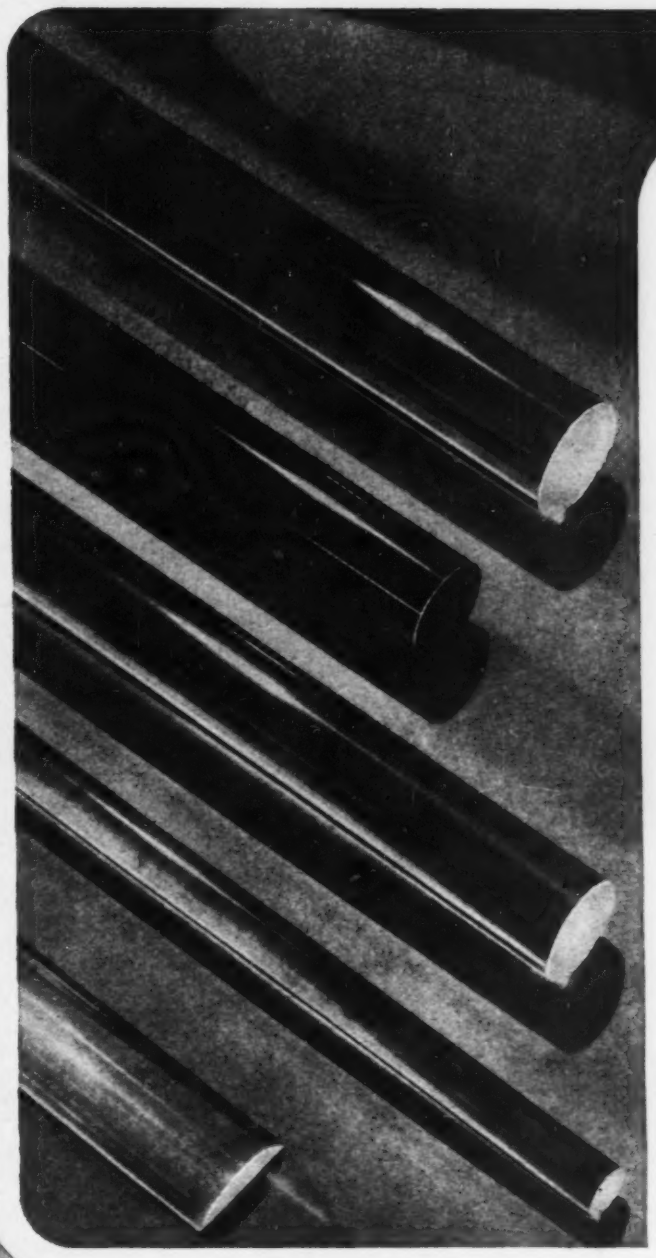
Guided Missiles

AN EXTENSIVE PILOTLESS aircraft program now being carried out by Naval aviation planners includes such successful devices as the



Model of gorgon II-C, a guided missile that carries 1,000 pounds of explosive to its target at 400 miles per hour. The ring at the top is the jet-propulsion engine

POLISHED POLYSTYRENE ROD AVAILABLE NOW



Plax polystyrene rod is now available in a new form: with a highly polished surface lustre. Thus designers may now readily work with crystal-clear rods of this unique material.

In this form, Plax polystyrene rod requires very little fabricating and it is easily polished after fabrication. A cutting-down wheel, using a compound held by a non-petroleum grease, will remove any surface imperfections caused by machining — and the final high lustre is restored by a soft cotton buff, free of compound.

The availability of crystal-clear polystyrene rod will suggest many new uses for this versatile material, which is light, hard, inexpensive, and easily fabricated. A few of the applications which immediately come to mind are push bars and racks, display and decorative fixtures, edge lighting effects, novelties, etc.

Plax polished polystyrene rod is available in standard 4' lengths in all diameters up to 2". Special lengths and colors are available on minimum order. Samples are available for testing purposes.

POLYSTYRENE LITERATURE AVAILABLE:

Bulletins on how to machine, polish and cement polystyrene; on what to tell machinists about polystyrene; and on how to use coolants with polystyrene are available on your request.

While Plax has been the leader in development of uses for polystyrene, we also offer several other plastic materials in unique forms and shapes.

In fact, between the resources of Plax and the Shaw Insulator Company, Irvington 11, N. J., you can obtain help and counsel in the use of most plastic materials and processes. For the literature mentioned above . . . write Plax.

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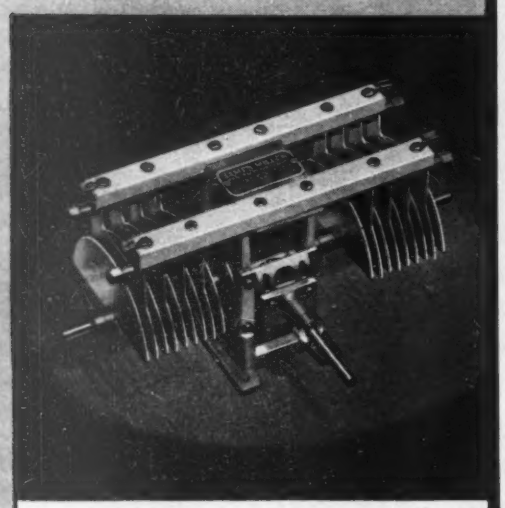

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A new member of the "Designed for Application" series of transmitting variable air capacitors is the 04000 series with peak voltage ratings of 3000, 6000, and 9000 volts. Right angle drive, 1/1 ratio. Adjustable drive shaft angle for either vertical or sloping panels. Sturdy construction, thick, round-edged, polished aluminum plates with 1¾" radius. Constant impedance, heavy current, multiple finger rotor contactor of new design. Available in all normal capacities.

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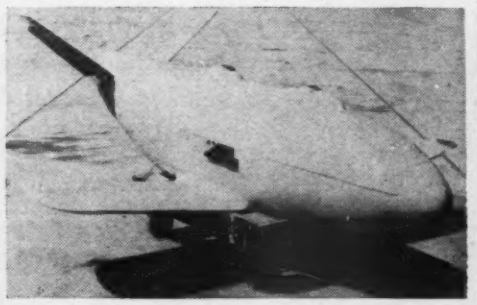
glomb, gorgon, and gargoyle. Each of these is guided to its target after launching either by self-seeking electronic equipment in the missile or by radio and television control.

The glomb is a glider bomb that carries a 4,000-pound bomb. It can be towed by a Navy fighter plane, and when released can be directed into a target through radio control and television.

The gorgon is a jet-propelled missile that can be carried by a bomber and sent into an enemy aircraft either by radio control or by its own automatic target-seeking device.

The gargoyle, also jet-propelled, carries a 1,000-pound bomb and



The gargoyle, a stub-winged pilotless dive bomber having a top speed better than 600 mph and carrying a 1,000-pound special all-purpose bomb

serves as a pilotless dive bomber that automatically seeks and collides with a ship target.

As early as 1940, successful demonstrations of pilotless aircraft were made with a torpedo plane that was radio-controlled and television-directed from a control plane up to ten miles away. From these early experiments several types of guided assault missiles were developed and some were actually used against the Japanese base at Rabaul.

Electronic techniques of guided missiles have also proved important in advancing the design of piloted planes. New planes controlled from the ground can be flown under conditions expected to produce structural failure or that would cause injury to a pilot, and essential test data can be transmitted by radio to the control station on the ground or in a piloted plane overhead.

Man will be too slow for combat of the future, but man's mind can devise both defense and offense. Out of the Navy's research and development programs are expected

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For name plates or any other plastic parts, it will pay you to consult Sillcocks-Miller specialists. This is particularly true if your products demand fabrication to close tolerance. This experienced organization can help you in four ways:

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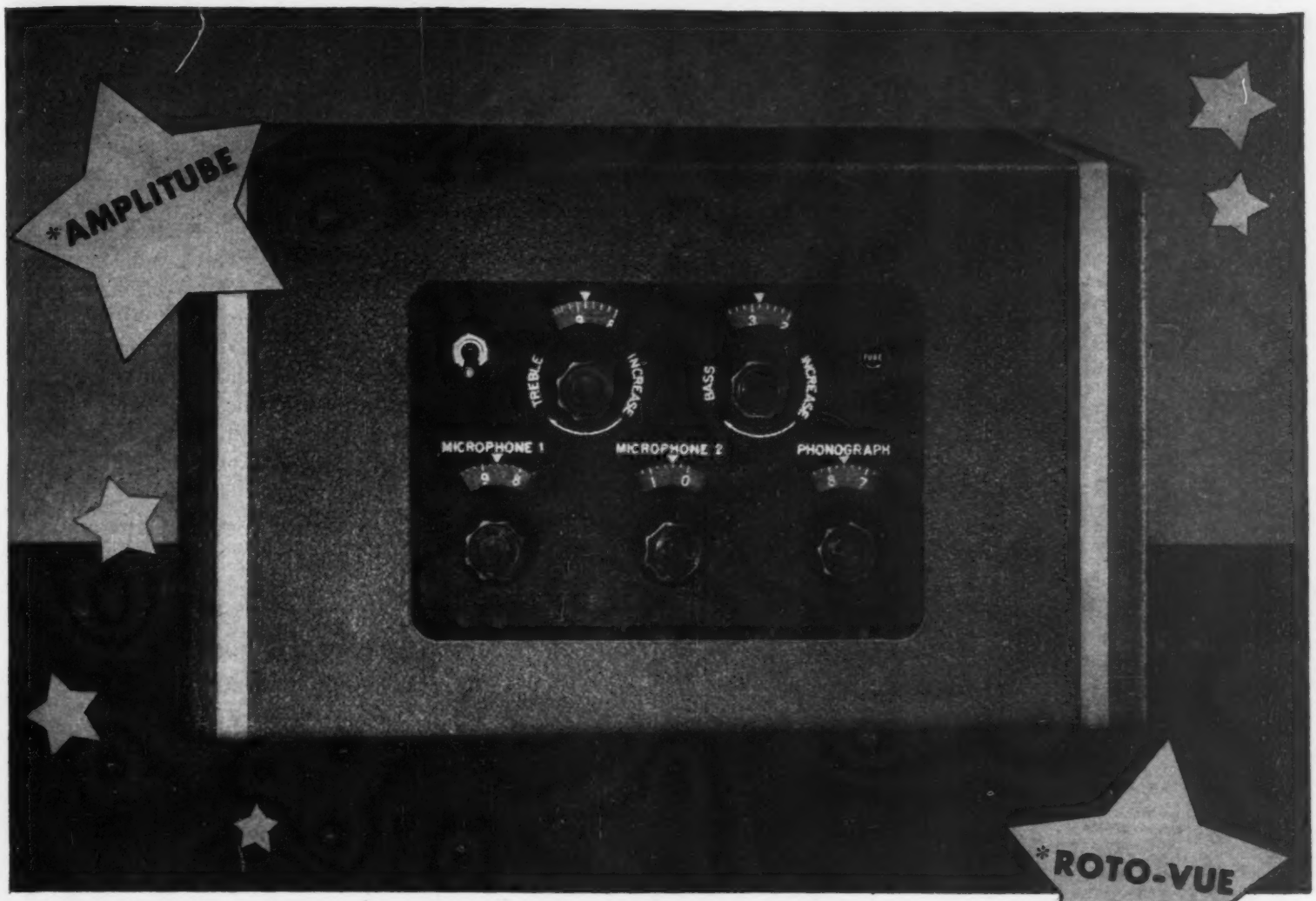

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Ethical engineering at Eastern is the history of many years in the service of sound amplification. The 21 Star Features are the result of intensive experience dating back to the early days of radio—the pioneer 20s! Today this engineering background accounts for the many innovations we have designed for the new 1946 Eastern Amplifiers—the 21 Star Features that produce Eastern's famous *Quality Performance*. No other amplifiers, regardless of price, incorporate so many novel and useful features. . . . For complete information and price list—for the first edition of our 1946 Catalog—write today! . . . Eastern Amplifier Corporation, 794 East 140th Street, New York 54, N. Y. Dept. 1-F.

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That Great Wave of Modern Appliances, Radios, etc., that will use Indicator Lights—

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Ask for catalog of other Gothard Lights.

Gothard Lights are an official requirement in the National All-Amateur Transmitter Contest.

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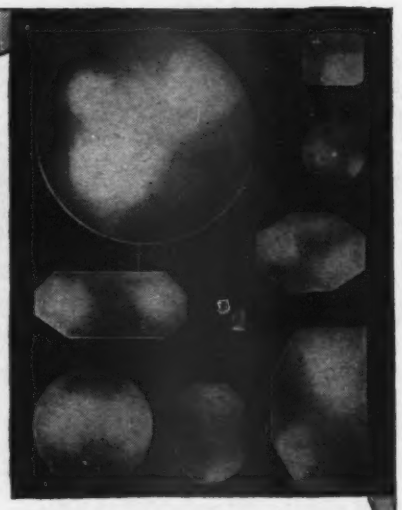

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to come airborne radars that can initiate defense automatically, by triggering automatic control circuits that instantly release the appropriate airborne counter-missile.

The Navy's Bureau of Aeronautics, which has the overall responsibility for new aircraft and equipment, will lean heavily on electronics for these future offensive and defensive weapons. It will have the assistance of the Navy's new Office of Research and Inventions, now engaged in enlarging the airborne electronic facility at Naval Research Laboratory.

F-m Gets Channel Numbers

A SIMPLIFIED SYSTEM of identifying frequencies of stations on f-m receiver tuning dials has been adopted by the FCC. The first channel frequency (88.1 mc) will be numbered 201; the second frequency (88.3 mc) will be numbered 202, and so on up to and including channel number 300 (107.9 mc). This system provides for extension of the f-m band either upstairs or downstairs without disturbing the numbering on existing receivers. The FCC action allows set manufacturers to proceed with production of simplified tuning dials with standardized numbering of dial calibrations.

Proposed RMA Standards

THE FOLLOWING MATTERS have been proposed by various committees of Radio Manufacturers Association for standardization as recommended practice and are being submitted to the RMA membership for comment.

(1) The intermediate frequency for vhf broadcast receivers shall be 10.7 megacycles.

(2) The sound-channel intermediate frequency of television broadcast receivers shall be located in the region 21.25 to 21.9 mc, and the oscillator frequency shall be higher than the signal frequency, thus placing the corresponding upper-frequency limits of the video channel between 26.5 and 27.15 mc.

(3) The antenna-to-set transmission line for television broadcast receivers shall be an unshielded, parallel line of 300 ohms impedance.

(4) The number of tubes counted

ULTRA HIGH SPEED RELAY
OPERATES AT 1,000 TIMES A SECOND

DEPARTING from conventional design to produce a hermetically-sealed relay only slightly larger than a metal receiving tube, Stevens Arnold engineers have developed the MILLISEC relay operating on ½ milliwatt of power, carrying 5 amperes, and functioning in 1 millisecond. We don't claim that all of these features can be incorporated simultaneously in the same relay, but these accomplishments give an indication of the range of operations attainable.

The MILLISEC relay may be used as the basis for a square-wave generator for frequencies up to 1,000 cycles per second. It has possibilities for replacing electron tubes in electronic switches for viewing simultaneously two traces on the screen of a cathode ray oscilloscope.

Because its mechanical resonance can be adjusted over a wide range, the moving element of the ultra-high speed relays provides the basis for the control element of another unconventional Stevens Arnold development — frequency operated switches. Having a band-width of 10 cycles per second, these frequency responsive switches will operate at any selected frequency from 20 to 800 cycles per second. This characteristic makes them admirably suited to remote control applications in radio or wire guided carrier systems.

We don't know all of the potentialities of our relay, but we do know that its unusual characteristics enable it to perform functions not ordinarily expected of relays. We shall be glad to co-operate with your engineers in exploring the possibilities of having MILLISEC relays — or our frequency selective switches — serve your requirements.

In designing the MILLISEC relay our engineers were not handicapped by restrictions which tradition sometimes imposes. In fact, we encourage new ideas and original approaches. We feel that this is one of the reasons why we have been able to build a well-balanced, efficient organization whose engineering achievements we shall report from month to month.

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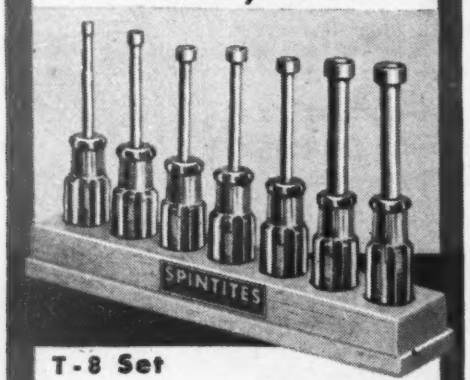
SPINTITE WRENCH
Works Like a Screw Driver

Speeds Production on the assembly line.



T-8 Set
Seven Standard Hex sizes in a Convenient Holder.

No. T-51
CHUCK TYPE SPINTITE
with Seven SPINTITES, three Screw Drivers and Reamawl in Leatherette Roll.

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WALDEN WORCESTER WRENCHES

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in an export radio receiver shall include the total number of evacuated envelopes exclusive of such of these as provide enclosure solely for illuminants.

(5) In specifying the frequency coverage of export broadcast receivers, the assured upper and lower frequency limits of all continuous frequency bands shall be stated in terms of: (a) kilocycles where the lower limit of frequency is less than two megacycles; (b) megacycles where the lower limit of frequency is between one megacycle and one kilomegacycle; (c) kilomegacycles where the lower limit of frequency is between one kilomegacycle and one megamegacycle.

(6) For export receivers, the values or ranges of power supply voltage and frequency on which satisfactory operation of the receiver is assured shall be specified. (Examples: 90-130 volts; 180-260 volts; 50/60 cycles.)

(7) When color coding is used for chassis wiring, it shall be standard to employ the following schedule of solid colors for wire insulation:

No.	Color	Circuit
0	Black	Grounds, grounded elements, and returns
1	Brown	Heaters or filaments, off ground
2	Red	Power supply B plus
3	Orange	Screen grids
4	Yellow	Cathodes
5	Green	Control grids
6	Blue	Plates
7	Violet	Not used
8	Gray	A-c power lines
9	White	Above or below ground returns, avc, etc.

New Call Areas for Amateurs

IN ORDER TO PERMIT assignment of thousands of additional calls without exceeding a limit of five symbols, the Federal Communications Commission has set up a new system of assigning call letters to identify amateur radio stations. The number of U. S. call areas was increased from nine to ten by reassignment of some areas within certain states. Full use of the prefix K will be made in the continental United States also, rather than in outlying areas only. Distinctive 2-letter prefixes will be reserved for outlying areas, such as KG6AA to KG6ZZ for Guam and KV4AA to KV4ZZ for the Virgin Islands. It is expected that in a great majority

Attention GI JOE!

Here's Your Opportunity to be First to
Start Your Own RADIO SERVICE SHOP
Complete Starting-in-Business Package Stocks of
TEST EQUIPMENT
TUBES, PARTS, TOOLS
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Act quickly! Meet the pent up demand for radio service. Turn your special service training into a profitable business of your own. No fuss. No worry. Here's everything you need—$350 up. Details upon request! Write, wire or phone

TRIPLETT
Volt-Ohm-Milliammeter
A. C.-D. C. Volts 0-10-50-250-1000-5000 at 1000 ohms per volt.
D. C. Milliamperes 0-10-100-500
Ohms 0-300-250,000
Case Black molded 3¼"x5⅛"x2⅛".
$16.00 net

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SUPER DEFIANT	SX25	$94.50
SKY CHAMPION	S20R	60.00
SKYRIDER MARINE	S22R	74.50
SKY RANGER	S39	110.00
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ORDERS FILLED AS RECEIVED!!!

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Complete with tubes.

17 Watt	30.30
25 Watt	42.60
35 Watt	54.60
50 Watt	70.50
17 Watt with Phono-top	42.30
25 Watt with Phono-top	52.20
35 Watt with Record-changer	89.10

PHILCO BEAM OF LIGHT
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20% deposit required on all C.O.D. orders. 2% transportation allowance on orders of $25.00 or more accompanied by payment in full.

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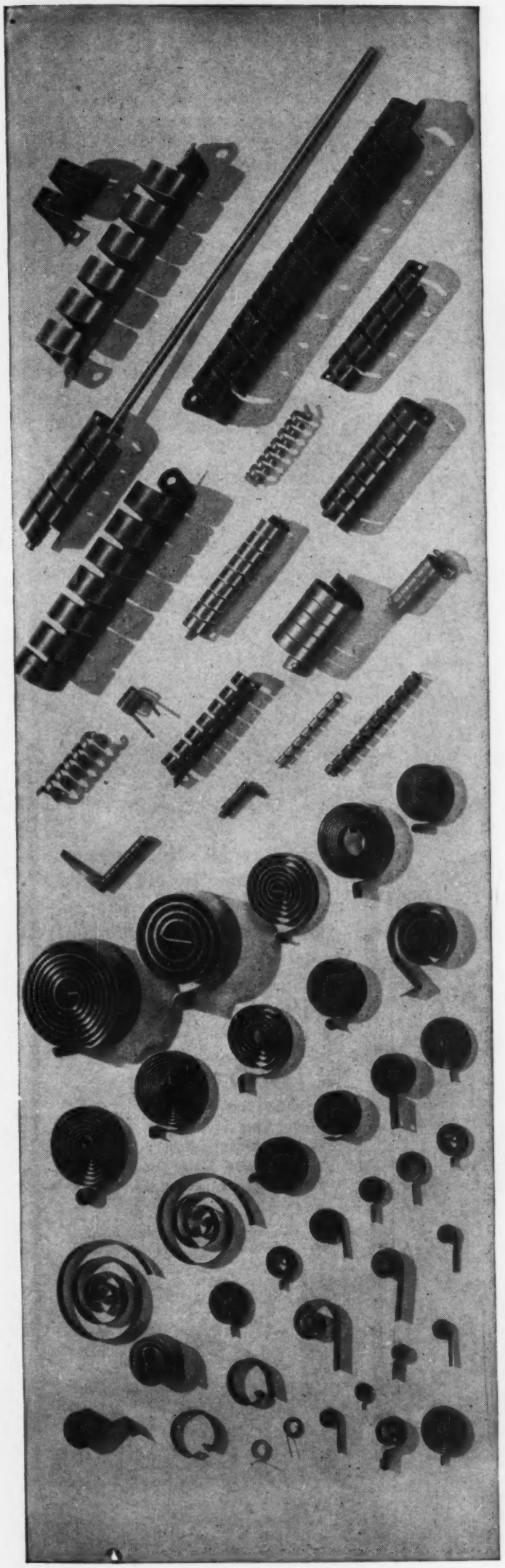

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WILCO THERMOMETAL (Thermostatic Bimetal)

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Higher Deflection Rate • Greater Electrical Resistivity

This new high action WILCO Thermometal broadens the range of Thermostatic Bimetal application.

PROPERTIES AND CHARACTERISTICS—Morflex provides a 40% higher temperature deflection and electrical resistivity for devices or instruments requiring extremely high sensitivity from 50° to 350°F. Whether the desired function is Temperature Indication, Temperature Control, or Temperature Compensation, Morflex operates dependably and uniformly . . . saves space.

WIDE APPLICATION FOR DOMESTIC AND INDUSTRIAL DEVICES—Morflex is supplied in strip or in straight cantilever blades, U-shapes, spirals and helices . . . or as parts of assemblies with shafts, studs, brackets, contacts, braids or springs.

CONSULT OUR ENGINEERING DEPARTMENT—Write the WILCO Engineering Department for help in developing the proper application of WILCO materials to your products. Send for FREE copy of the WILCO Blue Book. It contains charts, formulae, and full descriptions of *all* WILCO Thermometals and other products:

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High and Low Temperature with new high temperature deflection rates.

PRECIOUS METAL COLLECTOR RINGS—
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The same personalized skill and service that helped us produce hundreds of thousands of components and sixteen separate types of test equipment for the Navy's V.T. proximity fuse can now help you produce better, sturdier, more salable products for civilian use. We're equipped to fulfill your Electro-Magnetic Winding, R.F. Coil, Sub-Assembly or special Test Instrument requirements with intelligence and dispatch. Send your specifications for quotes, TODAY!

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Incorporates features and accuracy of expensive custom-built laboratory equipment—yet priced within reach of all wanting the best. Low capacity, high impedance imput. Balanced circuit. Wide Audio, H.F. and V.H.F. response. New small-diameter diode probe reaches difficult places with minimum lead length. Full scale sensitivity of 1 volt. Determines resistances from fraction of 1 ohm to 1000 megohms. Decibel scale for measuring andio gain. Sturdy oak case. Handy carrying handle. Write for details.

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of instances, more than 75 percent, a former call can be assigned without any change.

The old call area boundaries cut through a number of states, tending to cause confusion and delay in processing applications. The new areas will follow state boundaries, eliminating such confusion.

The new call areas are as follows:

No.	Area
1	New England (six states).
2	New York, New Jersey.
3	Pennsylvania, Delaware, Maryland, District of Columbia.
4	Virginia, North and South Carolina, Georgia, Florida, Alabama, Tennessee, Kentucky, Puerto Rico and Virgin Islands.
5	Mississippi, Louisiana, Arkansas, Oklahoma, Texas, New Mexico.
6	California, Hawaii and Pacific possessions except those included in Area 7.
7	Oregon, Washington, Idaho, Montana, Wyoming, Arizona, Nevada, Utah, Alaska and adjacent islands.
8	Michigan, Ohio, West Virginia.
9	Wisconsin, Illinois, Indiana.
0	Colorado, Nebraska, North and South Dakota, Kansas, Minnesota, Iowa, Missouri.

FCC Establishes Laboratory

A LABORATORY DIVISION recently established within the FCC Engineering Department will study the civilian uses of radar as they affect frequency allocations, conduct wave propagation and allocation studies, develop new monitoring equipment, test all types of transmitters for type approval, and test diathermy and industrial heating equipment.

Chief of the Laboratory Division will be Charles A. Ellert, formerly technical supervisor of the Radio

JAPANESE MAGNETRONS

Japanese radar receiving magnetron (held by Col. Marvin Hobbs, consulting engineer for Scott Radio Laboratories, Inc.), transmitting magnetron of cavity resonator type (held by Lorraine Pellegrini), and other Japanese tubes brought back from Japan by Col. Hobbs

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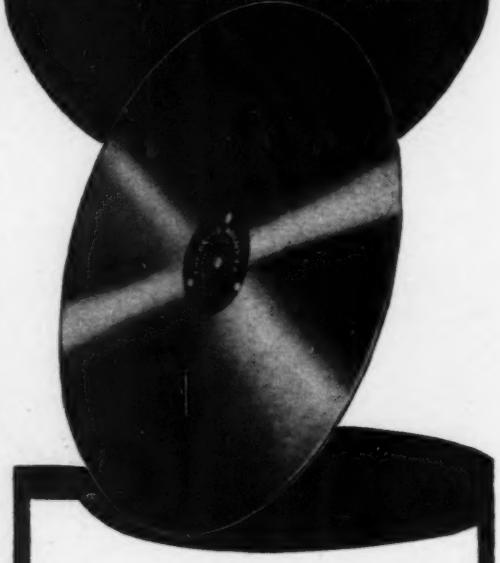

Technological advancements brought about by our wartime assignments provide critical engineers with "Black Seal" blanks of improved cutting and reproduction qualities plus more satisfactory play-back life.

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Natvar Varnished Fiberglas is available in 36" width rolls and sheets, or cut to any desired tape width.

NATVAR Varnished Fiberglas is used *primarily* to insulate equipment which must operate at temperatures above the safe working range of other insulating materials. It is approved Class B insulation.

But there are actually *three* advantages. For, in addition to its superior heat resistance, Natvar Varnished Fiberglas has excellent mechanical and dielectric strength, because it is carefully processed with special varnish to take fullest advantage of the Fiberglas base.

Write, wire, or phone us for quick deliveries, either from nearby wholesaler's stock or from our own.

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- Varnished canvas
- Varnished duck
- Varnished cellulose acetate
- Varnished special rayon
- Varnished Fiberglas cloth
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- Varnished tubings and sleeving
- Varnished identification markers
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IDEAL FOR AMPLIFYING STRINGED INSTRUMENTS
USED WITH ANY AMPLIFIER AND WITH RADIO SETS.
ASK YOUR JOBBER WRITE FOR FOLDER
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Intelligence Division (RID). Willmar K. Roberts will be assistant chief.

Testing of diathermy equipment will be done to prevent such apparatus from interfering with radio communications. Diathermy equipment manufactured since May 25, 1945 is required to operate within the three frequency bands designated for that purpose, with reasonable suppression of harmonic radiations. If the equipment is operated outside these bands, it must be operated in accordance with prescribed engineering standards to prevent interference. Equipment manufactured prior to the above date will be permitted to operate indefinitely as in the past; however, if interference will result from such operation, steps necessary to eliminate the interference must be taken.

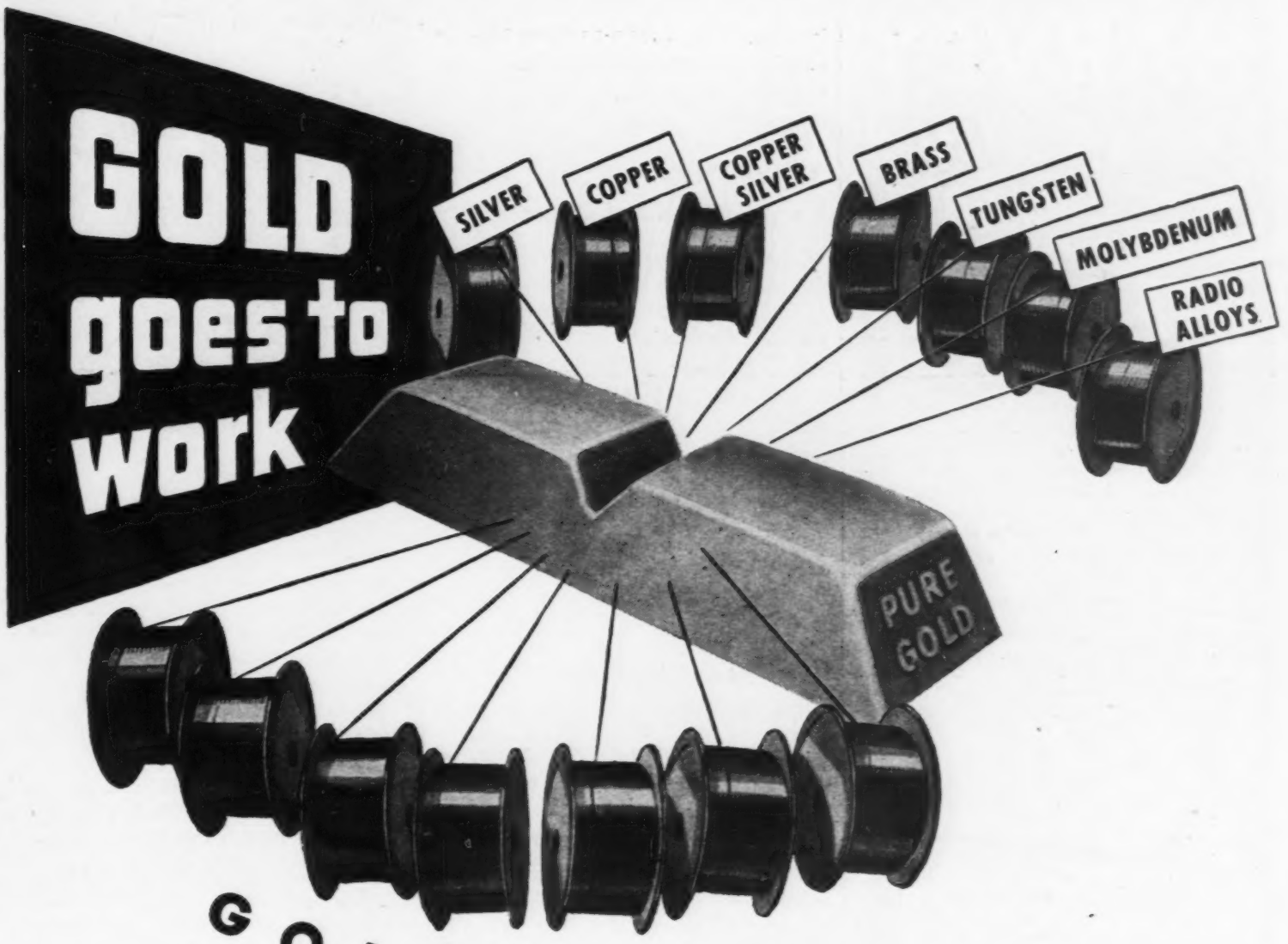
The Commission will test for type-approval diathermy equipment submitted by manufacturers designed to operate within one or more of the frequency bands allocated for such equipment. Equipment found to be capable of operation within one or more of such bands with reasonable suppression of harmonic radiations will be included in a list made available to the public.

Amateurs Go On Air

THREE BANDS of frequencies were released Nov. 15, 1945 by the Federal Communications Commission for use by radio amateurs: (1) the 28,000-kc band which in prewar days carried many an international conversation; (2) the 56 to 60-mc band, due to be shifted to 50 to 54-mc on March 1, 1946 when television stations are moved to a new channel; (3) a 144 to 148-mc band.

New F-m Detector Circuit

A TRUE FREQUENCY-modulation detector circuit employing a new seven-element vacuum tube has been announced by Philco Corporation. The new circuit replaces the discriminator and limiter in conventional f-m circuits, thereby saving either one or two tubes depending on whether a one or two-tube limiter is considered. Radio-frequency signals are used to control a



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From America's *precision* producer of gold clad fine wire . . . a complete line of sizes and materials.

Gold clad wires available include silver, copper silver, copper, brass, molybdenum, tungsten and radio alloys. Sizes .010 to .0005.

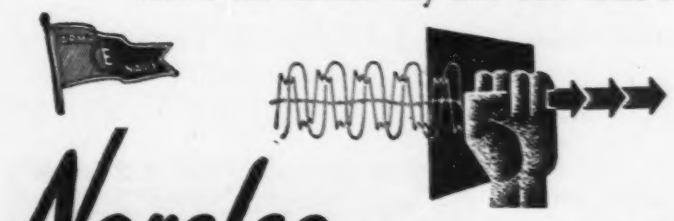
Besides, wire can also be furnished clad with silver, copper and other material on almost any base metal. All with the same high standards of quality.

And not only a wide variety of sizes and materials — but the assurance of a coating which is smooth and even, which will not blister, is non-porous and *sticks* to the wire.

That's the reason why fine wire users rush right to fine wire headquarters when they have a tough problem — they have found North American Philips is *wired for quick action*. Do you have a question about the application of fine wire to your products? Then write, wire or telephone — North American Philips.

In Electronic Tubes: Gold clad wires guard against undesirable grid emission.

In Decorations: The Admiral's braid, women's clothing decorations, religious vestments, decorators braid, fraternal uniforms and jewelry, all use this smart, non-tarnishable and tough gold clad wire.



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OTHER PRODUCTS: Quartz Oscillator Plates; Cathode Ray Tubes; Searchray (Industrial X-ray) Equipment; X-ray Diffraction Apparatus; Medical X-ray Equipment, Tubes and Accessories; Fine Wire; Diamond Dies. • We invite you to visit our office and showroom when in New York City.

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SHALLCROSS MODEL 637
KELVIN-WHEATSTONE BRIDGE

An extremely accurate and versatile bridge for the measurement of resistance over a wide range. Range is from 0.001 ohm to 11.1 megohms. Built-in galvanometer with a sensitivity of 1 microampere per millimeter division is supplied with the instrument. The rheostat arm consists of three decades variable in steps of 10 ohms when used as a Wheatstone bridge, or 10 micro-ohms when used as a Kelvin bridge. Accuracy of component resistors is 0.1%, except 1 ohm resistors, on which the accuracy is 0.25%. Portable; weight approximately 7 lbs. . $80.00

SHALLCROSS MODEL 638-2
KELVIN-WHEATSTONE BRIDGE

Combining both Kelvin and Wheatstone networks in a single, portable unit, the Model 638-2 provides a range of resistance from 0.0001 ohm to 11.11 megohm. Separate keys are provided for the battery and galvanometer circuit. Accuracy of component resistors is 0.1%, except the 1 ohm resistors, on which accuracy is 0.25%. For laboratory, school use, maintenance work, field investigations, and many forms of production line testing, this compact instrument is supplied with built-in galvanometer, sensitivity 0.25 microampres per m.m. Weight approximately 9 lbs .. $120.00

SHALLCROSS MODEL 630
WHEATSTONE BRIDGE

A basic electrical measuring instrument for laboratory and industrial use, providing accurate and direct electrical resistance measurements from 0.1 to 11,100,000 ohms. Accuracy between 10 ohms and 1,000,000 ohms is normally better than 1 percent. Supplied with built-in galvanometer having a sensitivity of 1 microampere per millimeter division. Weight approximately 6 lbs.....$40.00

SHALLCROSS DECADE
RESISTANCE BOXES

Used extensively as laboratory standards, A.C. and D.C. bridge and ratio arms, voltage dividers, etc. The accuracy adjustment of resistors as follows: 0.1 ohms—1%; 1. ohm—0.25%; all others—0.1%.

No. 544—One dial; 1.0 ohm steps; 10 ohms total resistance $13.50

Model 550—One dial; 1,000,000 ohm steps; 10,000,000 ohms total resistance... .$45.00

Telephone: LOngacre 3-1800

HARVEY RADIO COMPANY

103 WEST 43rd ST., NEW YORK 18, N. Y.

quadrature circuit and an oscillator in such a way that a-m noise signals are ignored completely, eliminating the need for limiter action.

20,000,000-Volt Betatron Used in Industrial Radiology

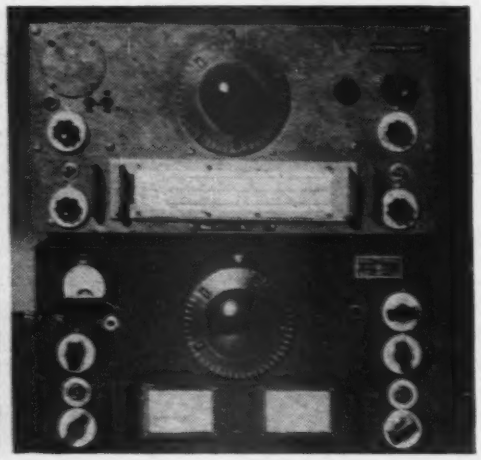
USE OF A 20,000,000-volt betatron during the past two years at a Government arsenal has been disclosed by the University of Illinois and Allis-Chalmers Manufacturing Co., who cooperated in the design and construction of the equipment. The betatron generates x-rays powerful enough to take pictures through 15 inches of solid steel. Laboratory development of the betatron was carried on by Dr. D. W. Kerst of the University of Illinois.

Two New Elements Discovered

DISCOVERY OF TWO new elements, 95 and 96 in the periodic table, was announced by Dr. Glenn T. Seaborg of the University of California at a recent American Chemical Society symposium. Dr. Seaborg, co-discoverer of plutonium, element 94, which was used in the atomic bomb dropped on Nagasaki, said the new elements belong to the heavy type and are of importance from the standpoint of atomic energy.

The new elements were discov-

A COMPLETE STEAL



Nine-tube German communications receiver made by Korting-Radio, shown on top of National Radio Company's type HRO which it imitates in electrical and mechanical design. Both craftsmanship and materials were inferior in the German copy, indicating it was rushed to meet heavy production schedules

In its multiplicity of wiring problems the many new and precious features of Surco Spiralon Keyed Insulation, with the widest range of identification in all sizes and lengths, is proving invaluable to Farnsworth Television & Radio Corp. of Fort Wayne, Ind. The ease with which this new insulated wire can be used in small compact areas or in large or intricate installations found instantaneous favor with this famous concern which is taking full advantage of Spiralon's diverse uses.

Spiralon is non-inflammable, non-fogging, non-corrosive, yet flexible and tough; and highly resistant to oils, dilute acids and alkalies to prove ideal for wiring under any and all conditions. Identification stripes are easily seen even on diameters as small as .025. The absence of all pigment fully preserves every electrical property, increases insulating resistance and allows for greater voltage.

With a Nylon jacket added—resistant to high heat and low temperatures—Spiralon further protects all electrical properties, reduces creepage while soldering terminals, offers a higher rupture point than braids and lacquers, checks deterioration, fungi attack, voids and pin holes.

- SHIELDED WIRE
- HIGH FREQUENCY WIRE and CABLE
- VINYL RESIN SHEETING
- INSULATING TUBING
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RESISTOR DATA

For the convenience of designers of products requiring resistors, Ward Leonard offers this new Resistor Handbook. It describes in detail the full line of wire-wound resistors giving complete information on mountings, enclosures, terminals and resistance values. Write for your copy today.

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RADIO PHONES

MURDOCK Radio Phones are built by sound manufacturing methods to give service—traditionally outlasting ordinary radio phones by years of service.

No other listening experience is like hearing through the "Ears-of-a-Nation" —produced by MURDOCK. Thousands are now listening without strain or discomfort. They get the message right . . . the first time . . . all the time!

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ered as the result of bombardment of uranium 238, an isotope or twin of uranium, and plutonium 239, an isotope of plutonium, with helium ions of 40,000,000 electron volts in the cyclotron at the University of California.

A new hypothesis regarding the relationship of these heavy elements was advanced, suggesting that elements from actinium (number 89) through the newly discovered elements 95 and 96 form a series corresponding to the only previously known series of elements which also have similar properties, the Rare Earths, elements 58 to 71.

Weather-Predicting Radar

A HIGH-ALTITUDE bombing radar set, designed by the Signal Corps to spread destruction in enemy countries, has now been adapted to peaceful meteorological use in detecting the approach of storms.

Developed at the Signal Corps Engineering Laboratories at Bradley Beach, N. J., and originally installed in B-29 bombers, these sets are being dismantled and mounted in observation stations where they foretell the direction, intensity and other characteristics of imminent weather disturbances.

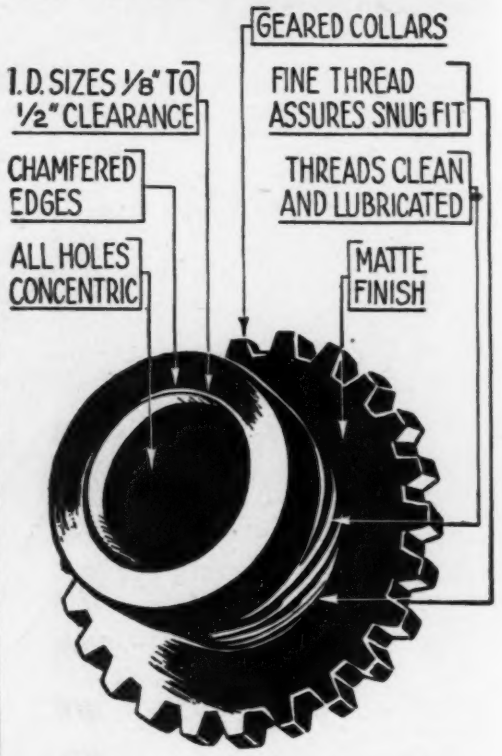
The intensity and distance of an approaching storm are indicated on one oscilloscope, the signals appearing as vertical deflections on a time scale. Another oscilloscope (Plan Position Indicator) records direction by means of glowing points of light that appear on a moving arm. The set is known as the AN/APQ-13 and has located storms as far distant as 200 miles.

Television Schedules for Intercity Coaxial Cable

HERALDING REGULARLY scheduled intercity television to begin early in January over the Bell System coaxial circuit between New York and Washington, the Army-Navy football game in Philadelphia was brought to the NBC television audience in the New York metropolitan area over a portion of this coaxial link. Twenty wide-band amplifiers spaced five miles apart in the coaxial cable made up for line losses.

The Washington-New York cable

NOW
8 SIZES OF CREATIVE GROMMETS



Four new larger sizes of CREATIVE 100% PHENOLIC PLASTIC GROMMETS (up to ½" i.d.) are now available for radio, electronic and electric instruments...*Send for a sample of each of the eight standard stock sizes, mounted on a convenient card.*

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You don't have to build molds to get Plastic Parts with Inserts such as knobs, terminals, etc. Get the facts about this unusual custom service . . . *CALL ON CREATIVE.*

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A REALLY NEW VACUUM GAUGE FOR THE ELECTRONICS INDUSTRY

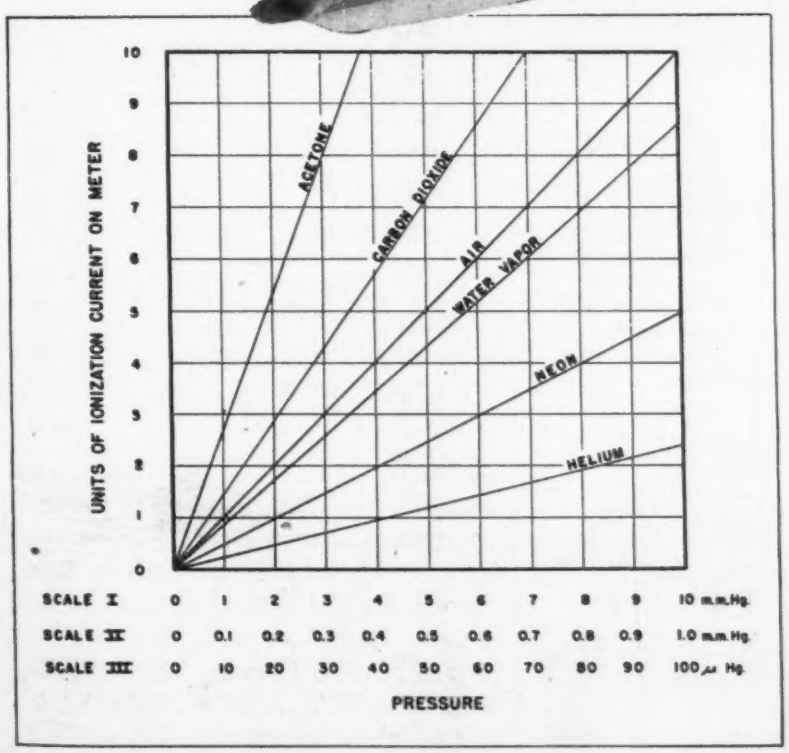
Continuous linear pressure indication for any atmosphere from 0 – 10 millimeters in three ranges: 0 – 0.1 mm., 0 – 1 mm., and 0 – 10 mm.

Simple to operate. Ideal for production gauging in the hands of inexperienced operators.

Radioactive source of ionization is undamaged by exposure to atmospheric pressure.

Contamination originating in gauging is eliminated. Instantaneous response for accurate pressure control and convenient leak detecting.

Reads pressures of gas mixtures directly.



See the Alphatron in operation at the I. R. E. Convention.

For full particulars send for Bulletin G-3

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SQUARE-WAVE GENERATOR

● This unit, generating its own frequency or synchronized from an external source, will be found invaluable in many fields. FM, AM and Television Broadcasting—Telephone and Telegraph Communications—Manufacture of Transmitting and Receiving Equipment and Parts.

● Many additional functions will recommend it for use in school and college research projects and in scientific laboratories.

• • •

For additional information write: *Electronics Department, General Electric Company, Syracuse, New York.*

Electronic Measuring Instruments

GENERAL ELECTRIC

164-E1

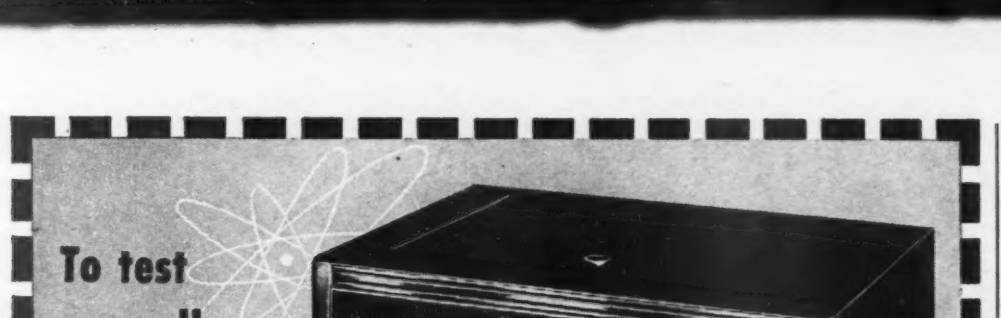


Diagram of experimental intercity television system utilizing coaxial network between Washington and New York

is to be made available to CBS, DuMont, and NBC for two nights per week during an extended experimental period, as well as to other television interests as soon as they have facilities for using this intercity service.

MEETINGS TO COME

JAN. 9; AMERICAN INSTITUTE OF ELECTRICAL ENGINEERS; TRANSIENT Analysis of Linear Servomechanisms, by John R. Ragazzini, professor, Columbia University; Room 301, Pupin Hall, Columbia University; H. E. Farrer, AIEE Headquarters, 33 West 39 St, New York 18, N. Y.

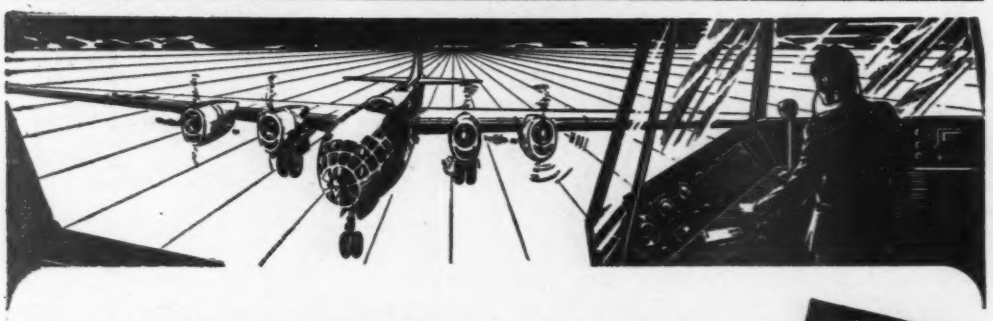
JAN. 23-26, INSTITUTE OF RADIO ENGINEERS, 33d Annual Winter Technical Meeting; Astor Hotel, New York, N. Y.; E. J. Content, chairman of meeting committee, WOR, 1440 Broadway, New York 18, N. Y.

FEB. 6-8; AMERICAN INDUSTRIAL RADIUM & X-RAY SOCIETY, Annual Convention; Hollenden Hotel, Cleveland, Ohio.

FEB. 9; INSTITUTE OF RADIO ENGINEERS, Chicago Section; Chicago Engineering Conference and Banquet; Merchants and Manufacturers Club, Merchandise Mart, Chicago.

FEB. 13; AMERICAN INSTITUTE OF ELECTRICAL ENGINEERS; Frequency Spectrum Theory Applied to Servomechanisms, by E. B. Ferrel, Bell Telephone Laboratories; Room 301, Pupin Hall, Columbia University, 7 p.m.; H. E. Farrer, AIEE Headquarters, 33 West 39 St., New York 18, N. Y.

MARCH 13; AMERICAN INSTITUTE OF ELECTRICAL ENGINEERS; Non-Linearity in Servomechanisms, by



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FOR CONTACTS

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SAMPLES of Silver Graphalloy will be gladly furnished for test on your applications. Silver Graphalloy is usually silver plated to permit easy soldering to leaf springs or holders. Why not WRITE NOW for your test samples?

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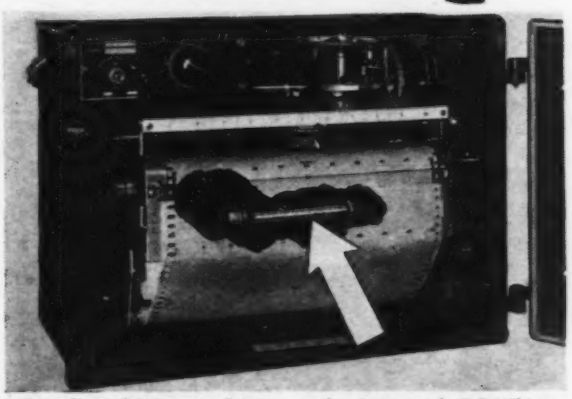
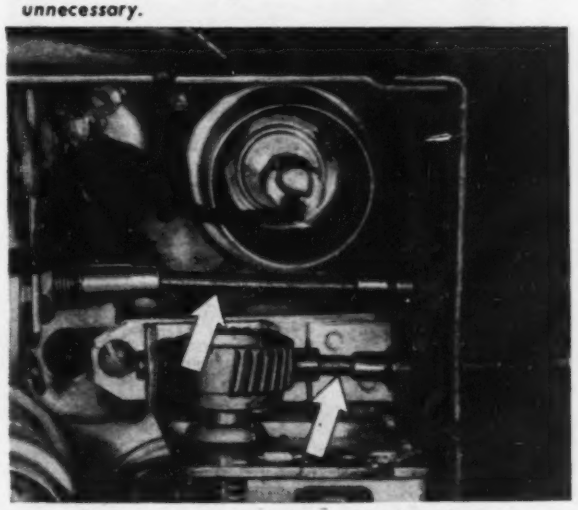
***FLEXY**—the S.S.White Flexible Shaft Man—says . . .*

END YOUR MISALIGNMENT TROUBLES!



Coupling driving to driven mechanisms with S.S.White flexible shafting—as in this recording instrument—provides a smooth-running, vibration-proof drive and makes accurate alignment of connected spindles unnecessary.



Coupling external control shafts to tuning elements—as in this radio receiver—damps noise-causing vibration, eliminates need for precise mounting and alignment of tuning elements—also gives complete freedom in their location.

IN all power drive and control applications within the wide range of flexible shafts, you can readily eliminate misalignment as a factor.

For example, take a small power drive — like the one in the top illustration at the left. By using an S.S.White flexible shaft as a connecting link between driving and driven members you eliminate the need for accurate alignment — and with it the expensive close-tolerance workmanship and assembly needed to obtain accurate alignment. And you eliminate the vibration and uneven operation that any misalignment in a solid shaft drive would cause.

For coupling elements requiring operational adjustments — as in the lower example — S.S.White flexible shafts offer the same advantages.

This is only one of the many ways in which S.S.White shafts can serve and save. To get the complete story —

SEND FOR THIS FLEXIBLE SHAFT HANDBOOK

This 256-page standard handbook size volume covers the subject of flexible shafts from every angle including all essential technical data. A complimentary copy is yours if you will write for it on your business letterhead.

S.S.WHITE INDUSTRIAL DIVISION

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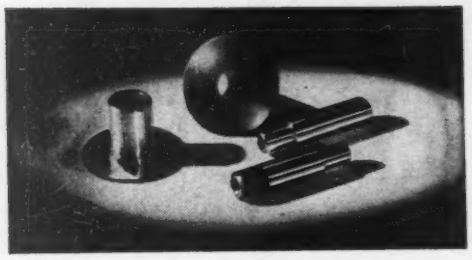
PRECISION PARTS

SMALL PARTS PLAY BIG ROLES



These intricate radio parts are typical of the precision and highly specialized production-capacity that have followed Ace products around the world.

For Ace has acquired the knack of machining and grinding delicate parts to incredible accuracies . . . doing it fast, on a mass-production basis. And this is important to every manufacturer engaged in conversion and production of specialized equipment.

Here at Ace, you'll find the ingenuity and modern machinery to help you design parts for *your* product . . . get them into production, and then turn them out faster, with greater accuracy, and to amazingly close tolerances.

If your production problems involve small parts and assemblies requiring stamping, machining, heat-treating, or grinding, check with Ace now. Send sample, sketch, or blueprint for quotation.

PARTS REQUIRING THREAD GRINDING A SPECIALTY. All types of threads up to 5" in diameter by 8" long on parts up to 20" between centers.

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for Precision Parts

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Dr. L. A. MacColl, Bell Telephone Laboratories; same place as Feb. 13 meeting.

MARCH 18-23; BROADCAST ENGINEERING CONFERENCE; developments since 1942 in broadcast engineering, including f-m and television; directed by Dr. W. L. Everitt, head, Department of Electrical Engineering, University of Illinois, Urbana, Ill., who requests addresses of those interested so they can be kept informed on program details.

BUSINESS NEWS

RESEARCH CORPORATION, New York, N. Y., offers $2,500,000 in grants to college laboratories in order that talented young scientists will be able to undertake important peacetime research in pure science. This nonprofit institution was established in 1912 with the gift by Dr. F. G. Cottrell of patent rights on electrical precipitation, and revenues from this and other patents assigned to it by public-spirited inventors are used to advance research and technology.

MAGUIRE INDUSTRIES, INC., announces its purchase of Radiart Corp., Cleveland.

AIREON MFG. CORP., Kansas City, Kansas, is making available to railroad operators and related groups a one-reel natural-color sound film titled "Railroading by Radio", depicting two-way combined induction and space radiotelephone equipment for train communication. Aireon recently acquired Lewis Electronics, Inc.

GALVIN MFG. CORP., Chicago, Ill. announces use of Motorola two-way radiotelephone equipment by the Greyhound Bus Lines of Chicago on an experimental basis for dispatching and for maintaining contact with buses while enroute. Frequency is in 30-44 mc band; output power of f-m terminal station is 250 watts, with 50 watts output from transmitters in buses.

FARNSWORTH TELEVISION & RADIO CORP. has purchased another plant at Huntington, Ind., in which to manufacture component parts. An east wing will be added to the company's administration and engineering building at Fort Wayne where

RESISTORS
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CONTROLS • WIRE
TRANSFORMERS • CABLE
HEADPHONES
RELAYS • SWITCHES
SOLDERING IRONS
BATTERIES • TUBES
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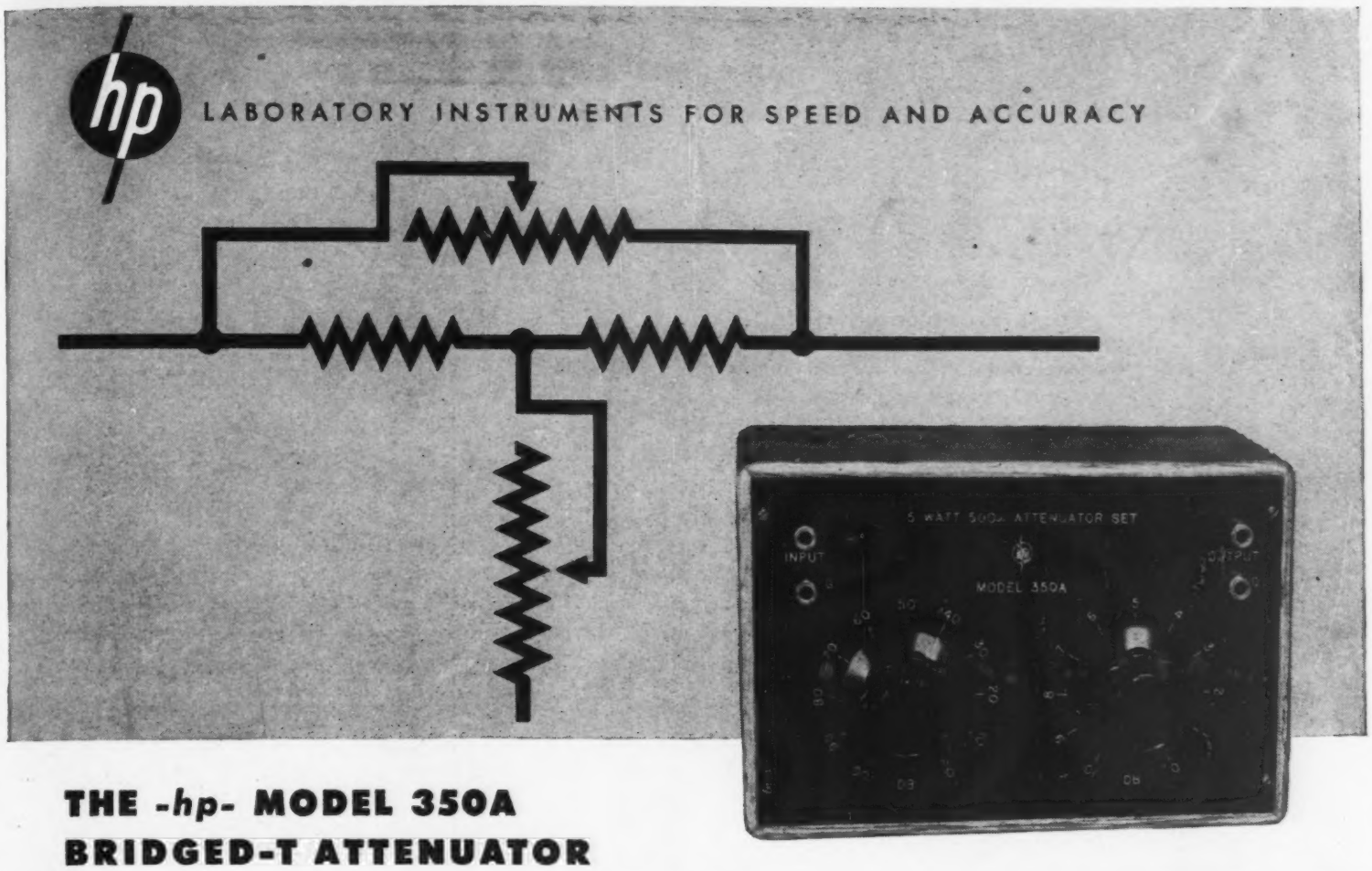


THE -*hp*- MODEL 350A BRIDGED-T ATTENUATOR

A Small Instrument With a Lot of Uses

The schematic diagram above shows the basic bridged-T circuit, two of which make up the -*hp*- 350A attenuator set. One is a 100 db attenuator, calibrated in 10 db steps, and one is a 10 db attenuator, calibrated in 1 db steps. Response is substantially flat at frequencies as high as 100 k.c. See figure 3. Accuracy is assured because the resistors are adjusted to plus or minus ½%.

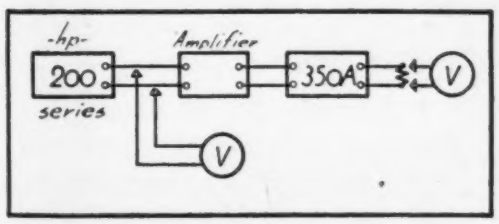


FIG. 1

In conjunction with an -*hp*- Audio Oscillator and two voltmeters, this -*hp*- Model 350A Attenuator may be used to make exact measurements of power gain . . . See figure 1.

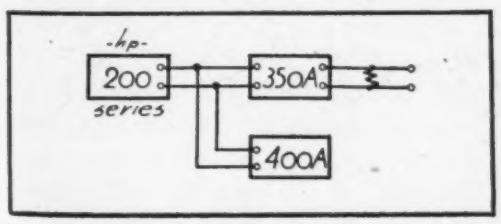


FIG. 2

The 350A may also be used to augment an -*hp*- audio oscillator and a vacuum tube voltmeter (-*hp*- 400A) to form a signal generator. See figure 2.

FOR MEASUREMENT CONTROL

The 350A is built with a large power handling capacity—5 watts continuous duty. It is particularly adapted to work in the supersonic field, and for other measurement work above the range of the conventional AF attenuator. It may also be used down to zero frequency.

The 350A like all -*hp*- instruments is held to a minimum size for convenience in use; actual dimensions are 5" by 8" by 4½". Input and output binding posts are available on the front panel; the unit is completely shielded from moderate fields.

Write today for more information on this and other -*hp*- instruments. 1163

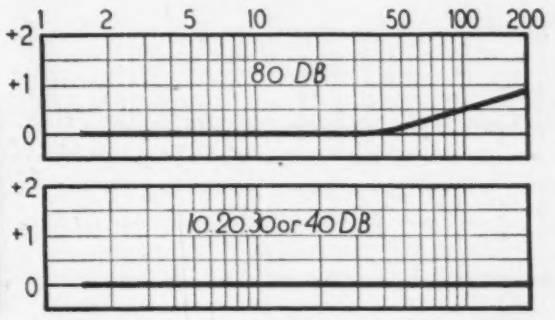


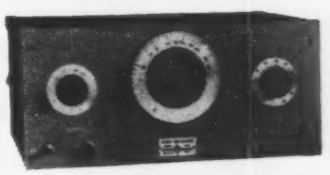
FIG. 3

HEWLETT-PACKARD COMPANY

BOX 1163 • STATION A • PALO ALTO, CALIFORNIA

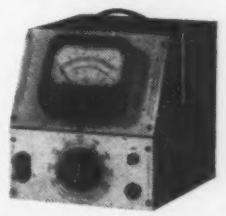
Audio Frequency Oscillators *Signal Generators* *Vacuum Tube Voltmeters*
Noise and Distortion Analyzers *Wave Analyzers* *Frequency Meters*
Square Wave Generators *Frequency Standards* *Attenuators* *Electronic Tachometers*

OTHER -hp- INSTRUMENTS



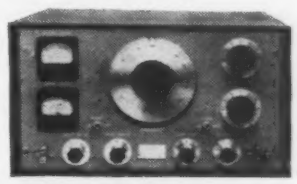
RESISTANCE-TUNED AUDIO OSCILLATORS

Require no zero setting! Several models available in 200 series, covering frequencies from 2 cps. to 200 kc.



VACUUM TUBE VOLTMETERS

For speed and accuracy in making voltage measurements from 1 cycle to 1 megacycle. The 400A covers 9 ranges (.03 to 300 volts) with full scale sensitivity.



AUDIO SIGNAL GENERATOR

The Model 205 AG consists of an -*hp*- resistance-tuned audio oscillator, combined with input and output meters, attenuator, and impedance matching system—all in one compact instrument.

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PRODUCTS

★ Here at Rogan, you are invited to avail yourselves of our complete knowledge and long experience in all phases of plastic molding. Our staff of trained experts will be glad to assist you with your plastic problems, no matter how involved or comprehensive.

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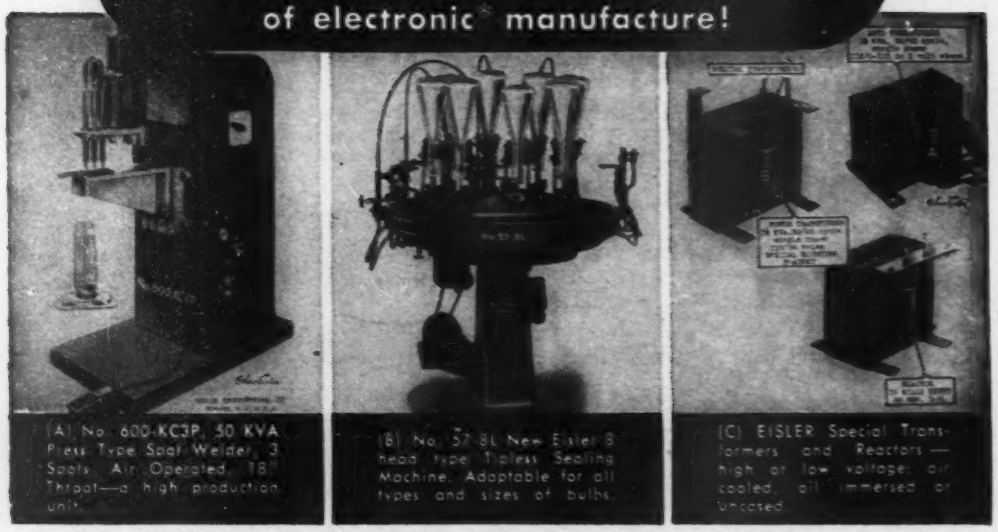
EISLER EQUIPMENT

..complete and diversified for every phase of electronic manufacture!



(A) No. 600-KC3P, 50 KVA Press Type Spot Welder, 3 Spots, Air Operated, 18" Throat—a high production unit.

(B) No. 57 BL New Eisler 8 head type Tipless Sealing Machine. Adaptable for all types and sizes of bulbs.

(C) EISLER Special Transformers and Reactors—high or low voltage; air cooled, oil immersed or encased.

The CHAS. EISLER line of specialized electronic tools, machines and devices is complete and diversified. Included are innumerable types of welders — spot, seam, butt, rocker, arm, pneumatic and special types. Also included are hundreds of devices for vacuum tube manufacture — glass tube cutters, slicers, stem and sealing machines as well as an all-inclusive line of transformers for every industrial and general need.

* EISLER serves 99% of American vacuum tube producers today. Write for completely illustrated catalog.

CHAS. EISLER

EISLER ENGINEERING CO.

751 SO. 13th STREET (Near Avon Ave.) NEWARK, NEW JERSEY

radio and television transmitters, communication, and other special apparatus will be made. Manufacture of phonographs, radios, and television sets continues at their Marion and Bluffton plants.

BELMONT RADIO CORP. has purchased a one-story plant near Waukegan, Ill., to secure additional space for manufacturing operations. When in full production, 250 workers will be employed.

CAMBURN PRODUCTS CO. announces plans for erection of modern manufacturing plants on Long Island, including a laboratory staffed with electronic engineers assigned to development of antennas and other new products for f-m and television entertainment applications as well as communications systems.

COMMUNICATION PRODUCTS CO., INC. has moved from Jersey City to a new location at Route 36 and Palmer Avenue, Keansburg, N. J.

THE GRENBY MFG. CO., manufacturers of precision machine tools and electronic equipment, has acquired full control of the Allen D. Cardwell Mfg. Corp., one of the pioneer manufacturers of radio parts. The manufacturing division has been moved into its new plant at Plainville, Conn., where greatly increased production capacity is available. Ralph H. Soby, vice president and director of Grenby, becomes president of the Allen D. Cardwell Mfg. Corp., following the retirement of Mr. Cardwell.

E. F. JOHNSON CO., Waseca, Minn., has acquired all tools, inventory and manufacturing rights for the cable connectors, pilot and dial light assemblies, tip plugs and tip jacks that were formerly Mallory-Yaxley products.

ELECTRONICS MANUFACTURERS ASSOCIATION, Inc., New York, N. Y., reelected I. W. Wyckoff of Pilot Radio Corp. as president. Newly elected officers for the coming year are A. Freed of Freed Radio Corp. and A. P. Hirsch of Micamold Radio Corp. as vice presidents, I. A. Mitchell of United Transformer Corp. as secretary, and S. J. Novick of Electronic Corp. of America as treasurer.

HALLICRAFTERS CO., Chicago, Ill., has purchased the plant of Shelby

FOR THOSE WHO SEEK

The Finest Cored Solder in the World

Ersin Multicore Solder contains 3 cores of non-corrosive Ersin Flux and provides just that extra rapidity of fluxing action which ensures a precision standard of consistently reliable solder joints. Only 3 independent cores of flux, evenly distributed across the cross-section of the solder wire, can give this extra efficiency. The cost of an individual solder joint in electronic apparatus is so little and each joint so vital that it must pay you to buy the best cored solder—Ersin Multicore.

3 CORES OF ERSIN
THE ACTIVE NON-CORROSIVE FLUX

Ersin, which is contained in the 3 cores of Multicore Solder, is a pure high grade rosin which has been subjected to a complex chemical process to increase its fluxing action to the highest degree without impairing the well known non-corrosive and protective properties of the original rosin. In effect, rosin as a flux suffices only as an agent to avoid oxidation during soldering, whereas Ersin will not only remove surface oxides, but also prevent their formation during the soldering operation. NO extra flux is required. The flux does not tend to run out of cores, so there is always a supply available for the next joint. The utmost economy of flux and solder is achieved.

ALLOYS

Five standard antimony free alloys are available. Ersin Multicore Solder is supplied in bulk quantities in any other tin-lead alloy to special order. Recently 45 tin and 55 lead alloy has been in most demand for electronic equipment. Colour coding of reels and packages makes different alloys instantly recognisable.

GAUGES

Ersin Multicore Solder is made in a wide range of gauges. Standard gauges supplied are from 10 S.W.G. - 22 S.W.G. (·128" - ·028") (3·251 - ·7109 m/ms) 13 S.W.G. (·092", 2·336 m/ms) and 16 S.W.G. (·064", 1.625 m/ms) are the most widely used sizes for the production of electronic equipment.

ERSIN MULTICORE

THE SOLDER WIRE CONTAINING
3 cores **OF NON-CORROSIVE ERSIN FLUX**

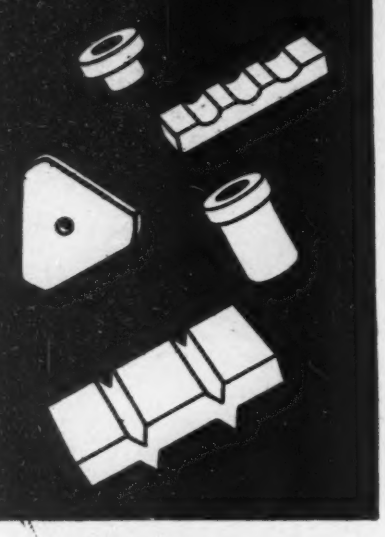
MULTICORE SOLDERS LTD.

MELLIER HOUSE, ALBEMARLE STREET, LONDON, W.1, ENGLAND
Telephone: REGENT 1141 PBX 4 lines Telegrams: DUSTICKON, WESTCENT, LONDON

U.S.A. DISTRIBUTORS:
BRITISH INDUSTRIES SALES CORPORATION
401 BROADWAY, NEW YORK, N.Y.13

CANADIAN DISTRIBUTORS:
SNI-DOR RADIOLECTRIC LTD.,
455 CRAIG ST. WEST, MONTREAL

Specialists in Special Crystals



PRECISION PROCESSING

FUSED QUARTZ

Fused quartz . . . has the lowest coefficient of expansion of any known material . . . high melting point (approximately 1756°C) . . . great resistance to thermal shock, high resistivity . . . low dielectric losses . . .

New skills of processing and metallic coating now allow for more precise engineering. New uses are suggested, new feats of design effected by taking advantage of Crystalab's techniques which allow precision processing of fused quartz to ± .001".

Yankee Ingenuity makes us

SPECIALISTS IN SPECIAL CRYSTALS

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LABORATORIES AND MAIN OFFICE 29 ALLYN STREET, HARTFORD 3, CONN.

New York Office 15 E. 26th Street, New York 10, N. Y. Phone MU 5-2952

ACCURATE INSTRUMENTS for PRECISION TIMING

Table model electric stop clock with a-c clutch and toggle switch

The Stoelting table model electric stop clock is an accurate timer for a wide variety of industrial and laboratory tests . . . such as measuring start-to-stop intervals of relays and instruments, and for checking sequence operations.

Timer with a-c clutch has toggle switch for manually starting the pointer. Timer with d-c clutch has binding posts only for attaching d-c control circuit for starting and stopping the pointer. Both timers have a-c clock motors, and pointers are reset with knob.

The Stoelting electric timer and impulse counter is an accurate, dual-purpose instrument for counting individual electric impulses or for use as a chronoscope.

When used as timer, 11-16 v current is taken from step-down transformer. When used as counter, direct current only is used. Counter capacity—7,200 impulses.

FREE ILLUSTRATED BULLETIN

Send for Stoelting Timer Bulletin No. 1100. Includes illustrations, wiring diagrams, technical data, and complete information on stop clocks, chronoscopes, impulse counters, stop watch controllers, and X-ray timers.

Electric timer and impulse counter

C. H. STOELTING CO.

INDUSTRIAL DIVISION

424-A N. HOMAN AVE. • CHICAGO 24, ILLINOIS

Shops, Inc., Shelbyville, Ill. and will use it to produce part of the cabinet requirements of the firm's Echophone Division. Daily output will be about 250 cabinets.

GENERAL PRECISION LABORATORY INC. has been established to conduct centralized research for General Precision Equipment Corp. and its subsidiaries. The laboratory, to be set up on the Manville estate at Pleasantville, N. Y., will conduct research and development in electronics and supersonics as well as other fields.

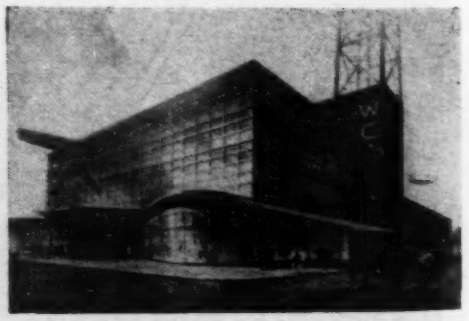
ELECTRONICS CORP. OF AMERICA, New York, N. Y., has acquired a plant in Brooklyn to supplement its two Manhattan factories in the production of radio receivers, electro-medical equipment, and other products.

FRANK RIEVER announces moving of his laboratory to 127 E. 73rd St., New York, N. Y.

MACHLETT LABORATORIES, INC., Springdale, Conn., has approximately doubled the size of its Springdale plant and will eventually consolidate all activities there.

WIRE RECORDER DEVELOPMENT CORP., Chicago, announces licensing of four additional manufacturers to produce Armour magnetic wire sound recorders, bringing the total to 24. Newcomers are Bendix Aviation Corp., Baltimore, Md.; Bang and Olufsen, Copenhagen, Denmark; Pyrox Proprietary, Ltd., Melbourne, Australia; St. George Recording Equipment Co., New York, N. Y.

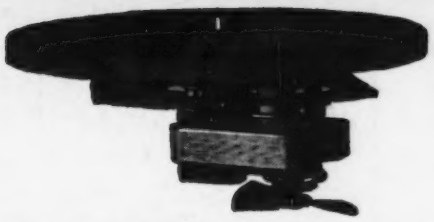
WCAU, CBS affiliate in Philadelphia, plans to erect a $2,000,000 radio and television center occupy-



Proposed WCAU radio and television center in Philadelphia, with helicopter landing field on roof

Eastern Electronics Corp.

- PRECISION WIRE WOUND RESISTORS
- WHEATSTONE BRIDGES
- RADIO & ELECTRONIC TEST EQUIPMENT
- RADAR ASSEMBLIES



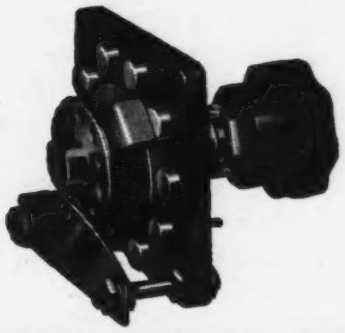
Eastern Electronics Corp.

PHONOGRAPH TURNTABLE UNIT

The need at this time for large quantities of phonograph turntable assemblies has prompted us to quickly design and tool up for the immediate production of this item. Engineers will find this compact turntable meeting all of their requirements for performance. We are prepared to make immediate deliveries and suggest that purchasing agents place their orders at once.

PERFORMANCE: — Correct and uniform speed is secured through the use of a motor of ample capacity, preloaded to operate on the flattest portion of the torque-speed characteristic.

★ ★ ★



ROTARY SELECTOR SWITCH

Designed for use where low contact resistance and mechanical sturdiness is required. Its construction insures long wear with low contact resistance of less than .001 ohm. May be arranged to have several sections to obtain multi-polar switching.

Well suited for precision test instruments; shunt ammeters, thermo-couple types, Wheatstone Bridges, and similar devices.

Korect-Ohm Precision Wire Wound Resistors

General Specifications

KORECT-OHM Resistors are wound on a sectional Ceramic bobbin, the direction of the winding being alternated on each section so that the resistor is non-inductive. Resistors can also be inductively wound when required.

To insure stability Korect-Ohm Resistors are aged and treated to relieve strains due to winding before the final adjustment is made.

Final resistance adjustment to an accuracy of better than .1%.

TEMPERATURE COEFFICIENT. Resistors are wound with selected alloy wire having a resistance change vs. temperature of less than .08% between —55 degree C and plus 55 degree C.

IMPREGNATION. A radically new material, INSO-FLEX, used for impregnating Korect-Ohm resistors has been developed by our engineering laboratory. This new material has several outstanding advantages, being extremely flexible it does not chip nor crack due to expansion or contraction under temperature variations. INSO-FLEX has high insulating qualities and is highly resistant to alkalies and weak acids, is resistant to moisture and mechanical shock. It forms an intimate bond between the ceramic bobbin, winding and lead wires. Being flexible, the lead wires may be bent and formed without disturbing the moisture-proof bond between the lead wires and INSO-FLEX covering.

Korect-Ohm resistors can also be supplied impregnated with our anti-fungi varnish or anti-fungi wax.

Type C
Maximum resistance 500,000 ohms.

Type A
Maximum resistance 1,000,000 ohms.

Type B
Maximum resistance 1,000,000 ohms.

Type XM
Instrument resistance shunt .1 ohms or lower. 25 watts.

We will make special resistors to any value or tolerance.

Our Regular Line of Resistors are ready for delivery.

Eastern Electronics Corp.

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Los Angeles Sales Office
Norman B. Neeley
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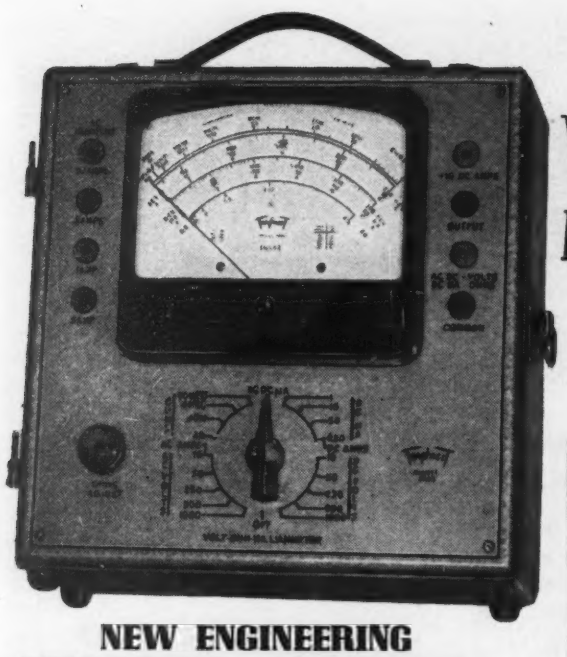
Hollywood, Calif.
Tel. Whitney 1147



MODEL 2405

Volt · Ohm Milliammeter

25,000 OHMS PER VOLT D. C.

STANDARDS ARE SET BY TRIPLETT

SPECIFICATIONS

NEW "SQUARE LINE" metal case, attractive tan "hammered" baked-on enamel, brown trim.

■ PLUG-IN RECTIFIER — replacement in case of overloading is as simple as changing radio tube.

■ READABILITY—the most readable of all Volt-Ohm-Milliammeter scales—5.6 inches long at top arc.

NEW ENGINEERING
NEW DESIGN • NEW RANGES

30 RANGES

Voltage:	5 D.C. 0-10-50-250-500-1000 at 25000 ohms per volt.
	5 A.C. 0-10-50-250-500-1000 at 1000 ohms per volt.
Current:	4 A.C. 0-.5-1-5-10 amp.
	6 D.C. 0-50 microamperes—0-1-10-50-250 milliamperes—0-10 amperes.
4 Resistance	0-4000-40,000 ohms—4-40 megohms
6 Decibel	-10 to +15, +29, +43, +49, +55
Output	Condenser in series with A.C. volt ranges

Model 2400 is similar but has D. C. volts Ranges at 5000 ohms per volt.

Write for complete description

Triplett

ELECTRICAL INSTRUMENT CO.

BLUFFTON ■ OHIO

On special mill shipments we can give prompt delivery. Also complete fabrication service backed by over 20 years of experience.

ELECTRICAL INSULATION CO., INC.

12 Vestry St., New York 13, N. Y.

ing an entire city block on Broad Street, with a television and f-m tower extending 612 feet above ground level. Completion by December 1947 is expected.

PERSONNEL

HOWARD DOOLITTLE, formerly of Radiation Laboratories, NDRC, has joined the engineering staff of Machlett Laboratories, Springdale and Norwalk, Conn., and will be in charge of high-frequency research and development. At Radiation Laboratories, he was in charge of the group responsible for the development of pulse generators for radar purposes, and before then held a professorship of physics at Trinity College.

Dr. H. Doolittle

Dr. R. L. Freeman

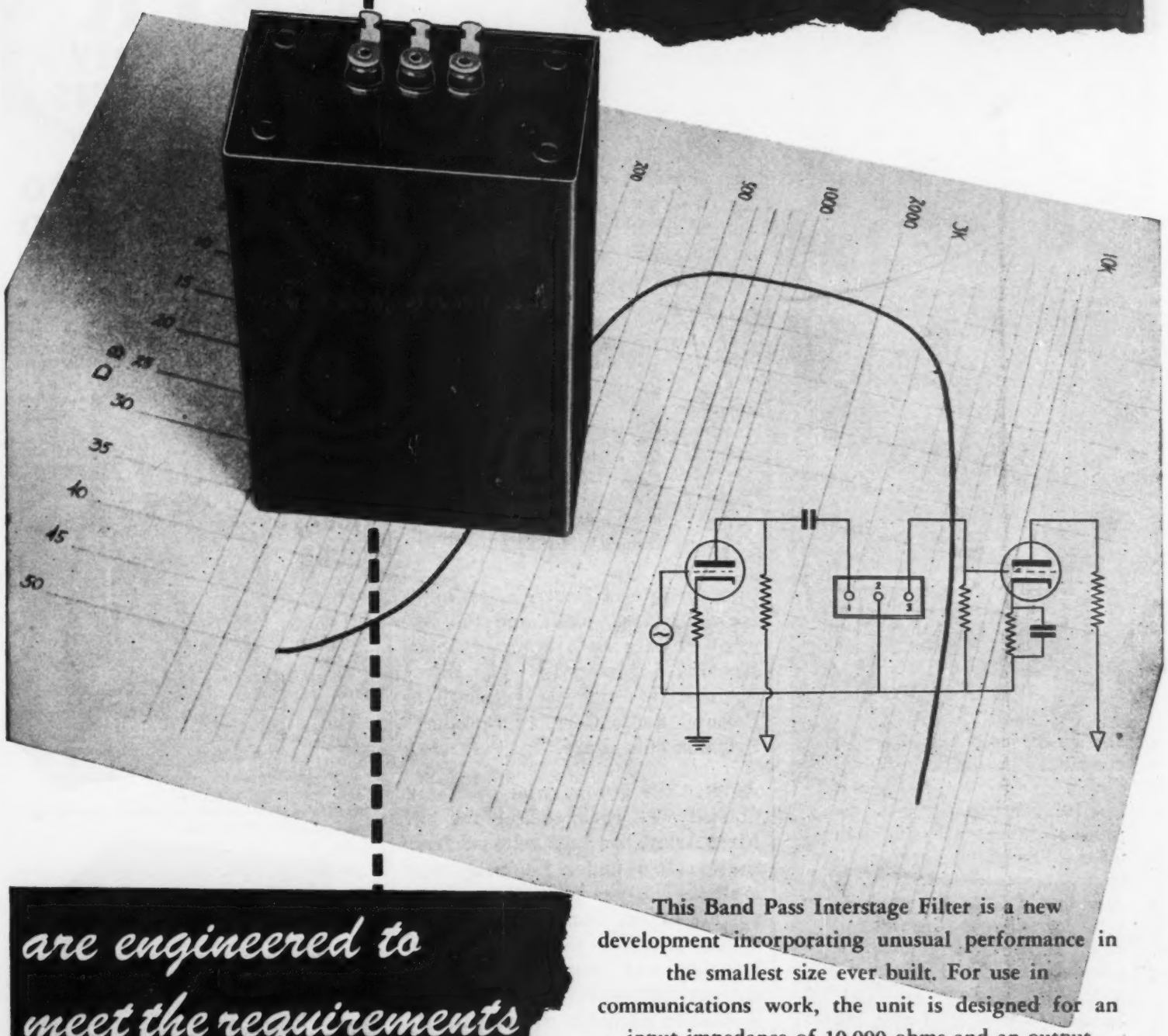
ROBERT LEE FREEMAN is now chief electronics engineer for Lewyt Corp., Brooklyn, N. Y. Dr. Freeman was formerly senior and consulting engineer for Hazeltine Corp., and previously served in various engineering capacities for Farnsworth Television, Inc., and Crosley Radio Corp.

JULES G. SIMMONDS and JOHN J. GUARRERA, until recently staff members of Radiation Laboratory at MIT, are now affiliated with the engineering staff of Bernard Rice's Sons, Inc. in New York City, which is continuing to develop, engineer, and manufacture in the uhf and r-f fields.

RUSSELL A. NIELSEN heads the new Pacific coast high-frequency laboratory opened recently by Westinghouse Electric Corp. He has been a research engineer for Westinghouse at East Pittsburgh, Pa.

H. M. BEVILLE, JR., who recently returned from duty as Lt. Colonel with Headquarters Staff, Intelligence, 1st Army, to assume his ex-

FREED
Transformers



are engineered to meet the requirements of **YOUR** *job!*

This Band Pass Interstage Filter is a new development incorporating unusual performance in the smallest size ever built. For use in communications work, the unit is designed for an input impedance of 10,000 ohms and an output impedance of 40,000 ohms. Provides a step-up ratio of 6 DB. Attenuation at 300 and 3000 CPS is less than 6 DB. Attenuation below 100 CPS and above 7000 CPS is 40 DB. Dimensions in hermetically sealed cans are 1⅝" x 2⅝" x 4".

FREED TRANSFORMER CO.
72 SPRING ST. NEW YORK CITY 12, N. Y.

Full Range

HIGH AND LOW FREQUENCY REPRODUCTION IN ONE ASSEMBLY!

TRU-SONIC
CO-AXIAL SPEAKER

The Tru-Sonic Co-Axial Speaker combines a high frequency metal diaphragm reproducer and a low frequency paper cone reproducer, mounted together with the dividing network in a single, compact assembly, 15" in diameter and 7" in depth giving a horizontal sound distribution of 90 degrees. Outstanding for tonal quality, and excellence before the war, the Tru-Sonic Speaker is finer than ever, but is available at a lower price, because of quantity production. Available now! Write for illustrated brochure.

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STEPHENS
MANUFACTURING CO.
8600 RATIONAL BLVD.
LOS ANGELES 34, CALIF.

ecutive position as research director of National Broadcasting Co., has been appointed a member of the Technical Research Committee of Broadcast Measurement Bureau, New York City. This committee now has equal representation of radio advertisers, advertising agencies, and broadcasters.

FREEMAN A. SPINDELL has been appointed chief engineer of Browning Laboratories, Inc. after working there since the spring of 1941 on design of radar units and radar test equipment.

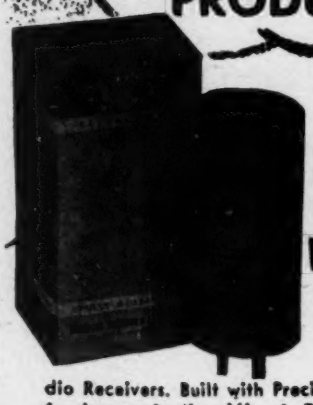
RAYMOND K. MCCLINTOCK is serving as engineering consultant for the international division of Sylvania, after experience since 1936 on tube design and development and circuit development for Sylvania, as

liaison engineer between Sylvania's Ipswich, Mass. plant and the Applied Physics Laboratory of Johns Hopkins University in Silver Spring, Md., and as engineer at Colonial Radio Corp. in Buffalo, a Sylvania subsidiary.

ARTHUR H. JONES, who as Lieut. Colonel was radar officer for the First Army, has been released from the services under the new point system. He rose from the rank of private in less than four years, and during planning for the European invasion was responsible for all anti-aircraft warning systems and utilization of radar equipment. He has served in all major theatres of operation and was at Luzon with the First Army headquarters staff at the time of the Japanese surrender.

CLINTON R. HANNA, associate director of the research laboratories at Westinghouse Electric Corp., was presented with the degree of Doctor of Engineering at Purdue University in recognition of his development of the gyroscopic tank gun stabilizer and other inventions

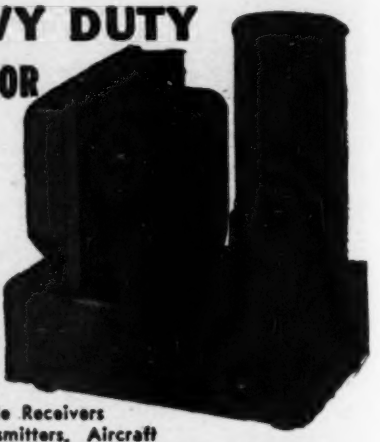
IMMEDIATE DELIVERY ON ALL ATR QUALITY PRODUCTS



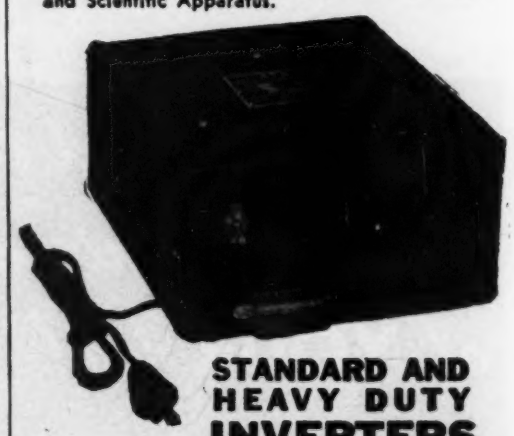
AUTO RADIO Replacement VIBRATORS

Designed for Use in Standard Vibrator - Operated Auto Radio Receivers. Built with Precision Construction for Longer Lasting Life at PRE-WAR PRICES!

HEAVY DUTY VIBRATOR PACKS



For Inverting Low Voltage D.C. to High Voltage D.C. for Operation of Portable Receivers and Transmitters. Aircraft Apparatus, Public Address Systems, Amplifiers, and Scientific Apparatus.

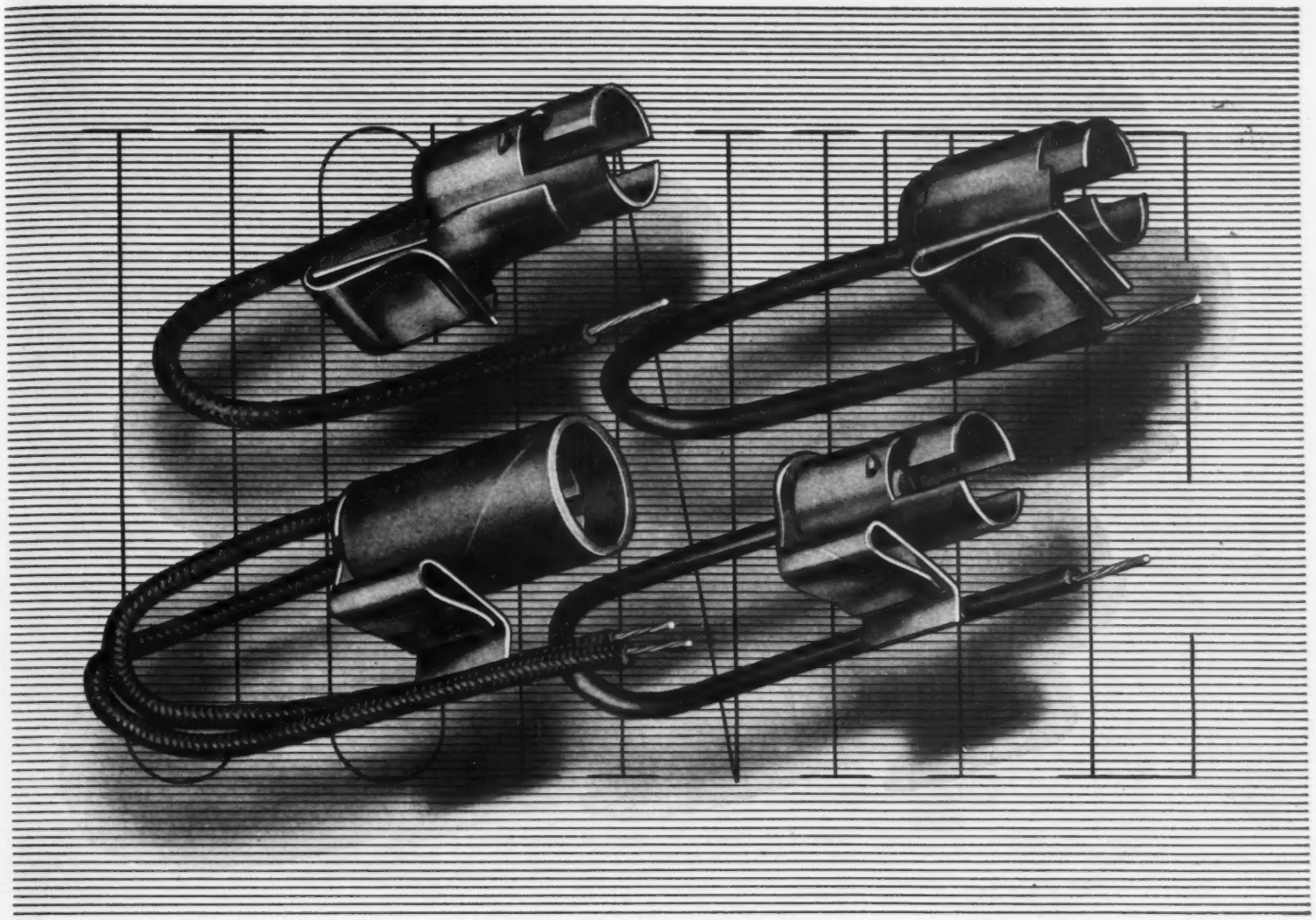


STANDARD AND HEAVY DUTY INVERTERS

Specially Designed for Operating A.C. Radios, Television Sets, Amplifiers, Address Systems, and Radio Test Equipment from D.C. Voltages in Vehicles, Ships, Trains, Planes, and in D.C. Districts.

WRITE for LATEST ATR CATALOG — Just off the press!

AMERICAN TELEVISION & RADIO CO.
Quality Products Since 1931
ST. PAUL 1, MINN. U.S.A.



Where the special is standard

These pilot lamp assemblies illustrate types we now make regularly for postwar radio sets. We hesitate to call them typical, however, because at Ucinite we have the skill, ingenuity and adaptability to manufacture any other style of pilot lamp assembly our customers might require . . . To us the "special" is "standard." We make *what* you want, the *way* you want it, *when* you want it.

The UCINITE CO.

Newtonville 60, Mass.

Division of United-Carr Fastener Corp.

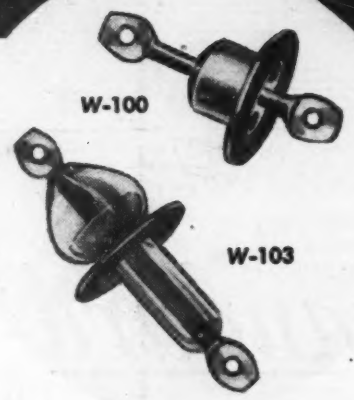
Specialists in **RADIO & ELECTRONICS**

LAMINATED BAKELITE ASSEMBLIES

CERAMIC SOCKETS • BANANA PINS & JACKS • PLUGS • CONNECTORS • ETC.

Build the competitive advantages of longer life better performance into your products . . . with

SPERTI HERMETIC SEALS



BUYERS who have waited through the war years will be looking for big improvements in your products. You'll have to meet civilians' expectations . . . just as you have met military specifications.

You can do it by building longer life, better performance, more trouble-free operation into your products. That calls for Sperti Hermetic Seals, the rugged, dependable, war-proved seals that effectively shut out dust, moisture and deteriorating agents.

Sperti Hermetic Seals are durable, one-piece units, easily soldered-in at less expense. Because of Sperti's advanced manufacturing methods, plus exhaustive tests and inspections, you'll get "true" seals that cut down production delays and costly rejects in the inspection line.

WRITE, TODAY. Get the facts. Find out about the many product applications of Sperti Hermetic Seals and their performance advantages.

Sperti

Incorporated, Dept. E-16, Cincinnati 12, Ohio

RESEARCH • DEVELOPMENT • MANUFACTURING

covered by more than 90 patents. His 23 years at Westinghouse have included research on loudspeakers, power tubes for radio receivers, microphones, and noise-measuring equipment.

NELSON P. CASE has joined the Hallicrafters Co., Chicago, as chief engineer of its receiver division. For the last two years he had been director of engineering design and development for Hamilton Radio Corp., and for 13 years prior was with Hazeltine Electronics Corp. He holds approximately 30 patents on radio receiving circuits.

Nelson P. Case

Paul H. Merriman

PAUL H. MERRIMAN has assumed new duties as head of the electrical and electronics section of the laboratories of The Glenn L. Martin Co., Baltimore, Md. Among his recent achievements is the Martin PBM Mariner radar production installation.

HENRI BUSIGNIES becomes director of the laboratories of Federal Telephone and Radio Corp. in Newark, N. J. He has been active in development of direction finder equipment for naval and marine operations, and has been with the Federal laboratories since 1940.

HAROLD V. NIELSEN has been made chief engineer of United States Television Mfg. Corp., New York, N. Y., in charge of all radio, television, and special product design and production. For 12 years previously he was with Sparks-Withington Co., Jackson, Mich., lately as chief engineer of their radio division.

JOSEPH N. BANKY received a $3,000 graduate fellowship for a year of study at Illinois Institute of Technology in Chicago, awarded by Allis-Chalmers Mfg. Co. of Milwaukee, Wis. He will study electrical

The Home Power Servant also handles many other jobs efficiently, dependably

For quiet operation, dependable performance, long life, maximum power per ounce of weight and per inch of space, use SM Fractional H.P. Motors. Models from 1/10th to 1/200th H.P. Speeds of 3,000 to 20,000 R.P.M. Voltage from 6 to 220 AC-DC Large volume production to your exact specifications.

Small Motors, Inc.

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1308 ELSTON AVE. • CHICAGO 22

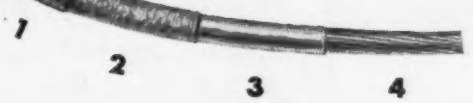
Manufacturers of special small universal, fractional H. P. motors, dynamotors, shaded pole motors, heater motors, generators.

Design, Engineering, Production

For Heat and Flame Resistant Radio and Electronic Hookups, Motor and Transformer Leads . . .

ROCKBESTOS FIREWALL HOOKUP WIRE

REG. U. S. PAT. OFF.



Construction: (1) Lacquer finished, color-coded glass braid. (2) Impregnated felted asbestos firewall. (3) High dielectric synthetic tapes. (4) Stranded tinned copper conductor, perfectly and permanently centered in helically applied insulation. Sizes No. 22 to 4 AWG in 1000 volt rating, and No. 12, 14 and 16 AWG in 3000 volt.

Unusual dielectric strength for its small diameter, high current carrying capacity, and ability to operate *continuously* at rated voltage under temperatures ranging from a maximum of 125°C. to minus 50°C., are a few of the reasons why Rockbestos Firewall Hookup Wire is being used in many varied applications.

Designed with a heatproof, flameproof *firewall* to meet the demands of airborne radio equipment manufacturers for a flame resistant hookup wire . . . widely used since in ground, marine and mobile communications systems . . . it has proven ideal for leads for small motors, transformers, dynamotors, battery chargers and rectifiers; miniature switchboard and relay wiring; and various applications in voice recorders, control devices, alarm systems, blood processing equipment and aircraft heaters, to name a few in a growing list.

This wide degree of adaptability is true of many other wires, cables and cords in the Rockbestos line of 125 *permanently insulated* standard constructions ranging from tiny multi-conductor cable to 5000 volt power cable. And if a "standard" won't fill requirements we are always willing to put Rockbestos Research to work on a "special" that will. For recommendations, samples or wire-engineering assistance, write or phone the nearest district office or:

ROCKBESTOS PRODUCTS CORPORATION
431 Nicoll Street
New Haven 4, Conn.

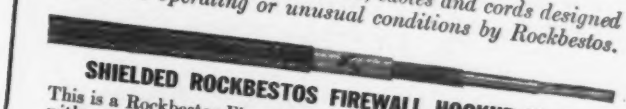
ROCKBESTOS RESEARCH

Solves Difficult Wiring Problems

NEW YORK, BUFFALO, CLEVELAND, CHICAGO, SEATTLE, ST. LOUIS, LOS ANGELES, SAN FRANCISCO, PITTSBURGH, PORTLAND, ORE.

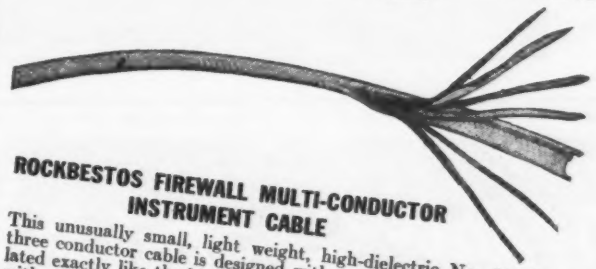
★ *DO YOUR SHARE—BUY VICTORY BONDS* ★

A few of the 125 different wires, cables and cords designed for severe operating or unusual conditions by Rockbestos.



SHIELDED ROCKBESTOS FIREWALL HOOKUP WIRE

This is a Rockbestos Firewall Hookup Wire, insulated and covered with a color-coded braid exactly like the construction detailed to the left, and shielded with a tinned copper braid. It is highly resistant to heat, flame, moisture, oil, grease and gasoline, won't bake out under a maximum operating temperature of 125°C., and won't become brittle at minus 50°C. Sizes No. 22 to 4 AWG in 1000 volt construction, and No. 16 to 12 AWG in 3000 volt. Also available in twisted pair and various multi-conductor constructions.



ROCKBESTOS FIREWALL MULTI-CONDUCTOR INSTRUMENT CABLE

This unusually small, light weight, high-dielectric No. 26 AWG three conductor cable is designed with unbraided individuals insulated exactly like the big illustration of hookup wire above, cabled with asbestos fillers, and braid-covered to make an extremely compact, mechanically strong, multi-conductor cable for use where space is at a premium and dependable performance is essential. It is made to a nominal diameter of .125" and is smaller than a No. 14 AWG Rockbestos Firewall Hookup Wire. Also in 4, 5 and 6 conductor in sizes No. 26 to 20 AWG.

ROCKBESTOS ASBESTOS INSULATED LEAD WIRE

Sizes No. 22 to 4 AWG solid or stranded copper, monel or nickel conductors insulated with .031" or .040" of impregnated felted asbestos in black, white or colors. Heatproof and flame-resistant, this lead wire will not bake brittle and crack under vibration, won't rot, swell or flow when in contact with oil or grease, and has ample moisture resistance for most applications. For high dielectric strength and added moisture resistance specify Type CA Lead Wire with synthetic tape next to the conductor.

ROCKBESTOS ASBESTOS INSULATED MAGNET WIRE

Round, square and rectangular asbestos insulated conductors finished to meet varying winding conditions and coil treatment requirements. Designed for Class B windings and also suitable for use as insulated bus wire where high-dielectric strength is not required. The insulation is non-checking and is unaffected by heat or aging.

Another New Rockbestos Firewall Construction!

War-developed Rockbestos High-Temperature Wire — *with a maximum operating temperature of 400°F.* — designed for jet propelled plane applications and circuits to hot-wing de-icers, fire detectors and extinguishers, and air conditioning and heating units where baking temperatures destroy ordinary insulation. Under continuous operation at rated temperature it retains its original dielectric strength and inherent resistance to heat and flame, and *progressively improves* in its resistance to moisture and abrasion. It is now available for these and other severe applications. Write for complete information and samples.

ANTICIPATE YOUR REQUIREMENTS!

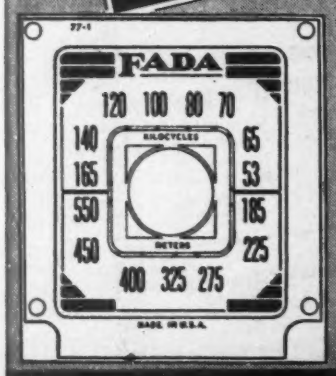
First Come . . . First Served!

Orders for transformers are pouring in so fast that we will soon be booked to capacity for several months to come. If your product requires heavy duty filter reactors, swinging reactors, or plate transformers, anticipate your requirements and place your order NOW!

DONGAN ELECTRIC MFG. CO.
2977 Franklin Detroit 7, Mich.

DONGAN
TRANSFORMERS
The Dongan Line Since Nineteen-Nine

HOPP
Plastic
RADIO DIALS
DIAL WINDOWS, NAME PLATES, SCALES, GAUGES, CHARTS, CALCULATORS, ETC.



PLASTIC RADIO DIALS combine the beauty of radiant color with the utility of perfect light transmission.

The possibilities of design, size, shape and color combination are limitless.

Whether your problem is dials, or any of hundreds of allied applications, let Hopp artists and engineers "sit in" on your designing problem.

Send us your blueprints or samples for quotation.

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460 WEST 34th STREET, NEW YORK 1, N. Y.
ESTABLISHED 1893

Joseph N. Banky (left) and Dr. Louis T. Rader, director of electrical engineering department at Illinois Institute of Technology, stand before a-c network calculator board

problems that can be solved through use of the school's new $90,000 a-c network calculator board.

PETER KAYE has been released by the Canadian Army to resume his former post as associate director of Canadian Research Institute, Toronto. His activities will include supervision of the electronic instrument division.

AWARDS

For excellence in production of military equipment, Army-Navy E burgees were awarded to:

Technifinish Laboratory, Inc.
Rochester, New York

Les Laboratories, L.M.T.
Paris, France (IT&T affiliate)

For meritorious participation in the VT fuze development, the Bureau of Ordnance E was awarded to:

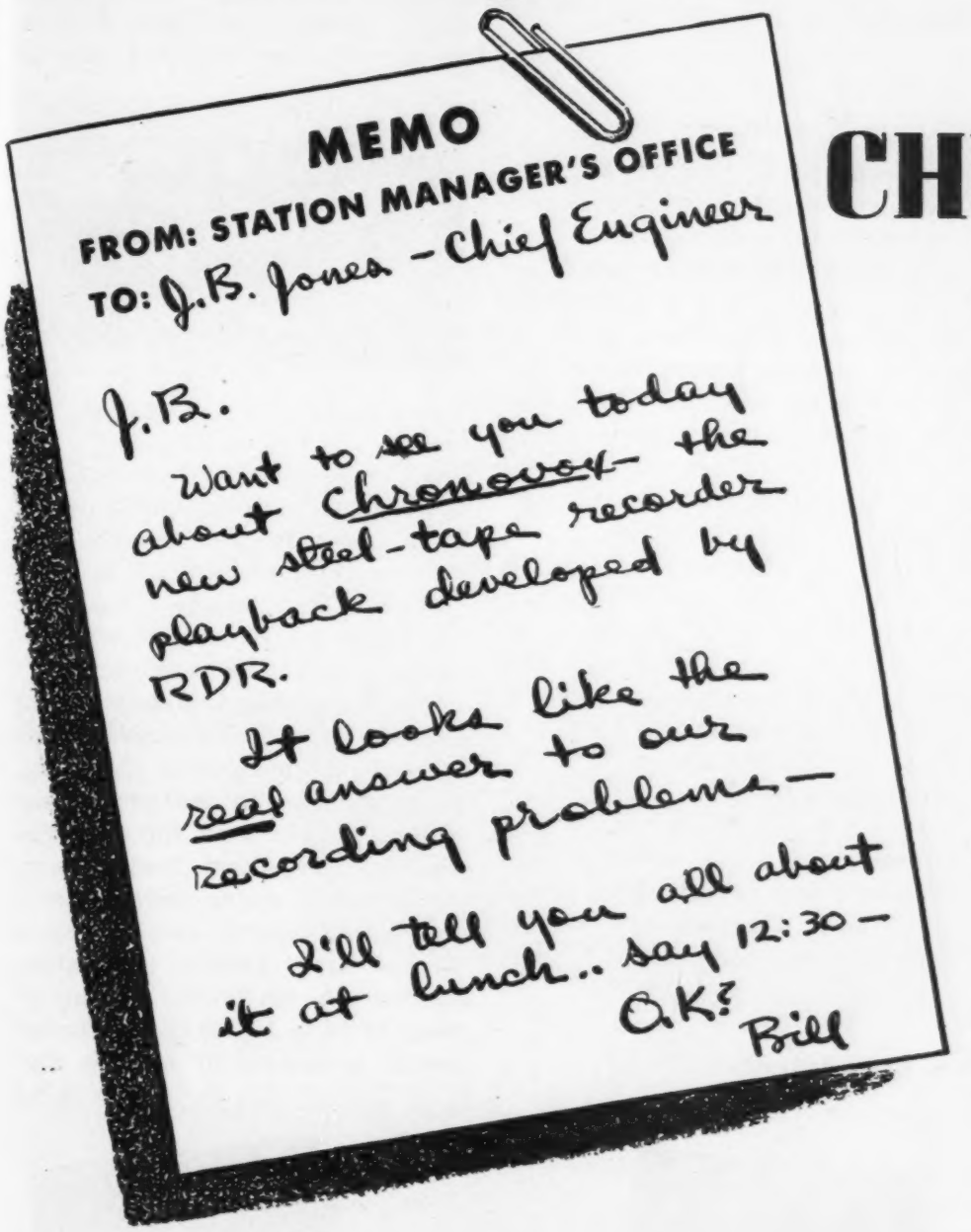
Sprague Electric Company
North Adams, Massachusetts

For exceptionally distinguished service to the United States Navy, the Distinguished Civilian Service Award, highest bestowed on civilians by the Navy, has been presented to:

Dr. Robert M. Page, Naval Research Laboratory, Washington, D. C.—for electronic research, and his outstanding contributions to the development of radar since 1934.

Dr. Herbert Friedman, physicist, Naval Research Laboratory, Washington, D. C.—for development and adaptation of the tube counter to x-ray diffraction methods used in production of quartz crystals.

Making Steel Talk


MEMO
FROM: STATION MANAGER'S OFFICE
TO: J.B. Jones - Chief Engineer
J.B.
Want to see you today about Chronovox - the new steel-tape recorder playback developed by RDR.
It looks like the real answer to our recording problems —
I'll tell you all about it at lunch.. say 12:30 —
O.K?
Bill

with CHRONOVOX

Know these Facts about Chronovox!

- The cost of discs is eliminated!
- The steel tape is permanent . . . indestructible!
- Recordings are erased at will!
- Surface noise lower than any other method of recording!
- Recordings reproduce indefinitely with less than 3DB attenuation!
- It's a complete, self-contained unit!
- Plugs in any 110 volt AC source!
- For a permanent record, "dub" from the final—perfect —Chronovox impression to your disc recorder!

Yes, memos like this one are being written daily. Radio executives KNOW that Chronovox will solve many of their recording problems. The RDR Chronovox is a precision instrument employing an improved method of recording sound on an indestructible steel tape. Recordings are made magnetically not physically—and the Chronovox will repeat the last recording indefinitely or until a new one is made.

Of Course, it's an RDR Achievement!

FOR MORE INFORMATION — CONTACT:

RADIO DEVELOPMENT & RESEARCH CORP.

233 WEST 54TH STREET · NEW YORK 19, N. Y.

AFFILIATE: TRANSFORMER PRODUCTS, INC.
143 W. 51st Street, N Y. C.

MIAMI: SALES—SERVICE
1415 N. E. 2nd Ave., Miami, Fla.

NEW PRODUCTS

New materials, new components, new assemblies; new measuring equipment; new technical bulletins, and new catalogs

1

High Frequency Video Amplifier

COMPACT, RUGGED, high-frequency amplifiers are now available from Sylvania Electric Products Inc., Industrial Electronics Division, Boston, Mass. Supplied for center frequencies between 30 and 70 mc with any bandwidth from 2 to 10 mc, these sets are designed particularly for use as i-f amplifiers in uhf and shf receiver applications.

A typical amplifier has an overall gain of 100 db with a center frequency of 60 mc and a half-power bandwidth of 9.0 mc. An external gain control is easily provided. Unless otherwise specified, a standard 500-ohm input impedance is supplied. The output stages are cathode followers designed to operate into impedances of 75 to 100 ohms with voltages ranging from 0.5 to 2.0 volts, negative or positive.

The video detector may take one of several forms according to the special application of the amplifier. In broad-band circuits the frequency characteristics of the rectified video components will be such that the output at 8 mc will be reduced not more than 3 db from the output at 1 mc. These amplifiers will thus pass a square-top pulse having a duration of 0.15 microsecond or greater without appreciable frequency or phase distortion, making them suitable for television receiver applications.

Either single-ended or balanced-input circuits are supplied. The balanced-input circuits are designed for use with dual-input systems and will distinguish between in-phase and out-of-phase signals from two channels. In one such unit this discrimination is 33 db.

Power supply for a typical amplifier includes 105 d-c at 90 ma, 300 d-c at 20 ma, and 6.3 a-c or d-c at 1.7 amp. External gain control requires 0 to −12.5 v d-c at 1.5 ma.

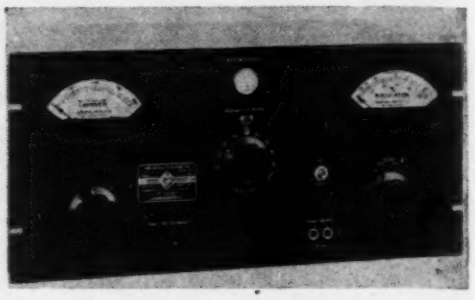
2

Miscellaneous New Instruments

NEW BROADCAST MONITORS; UHF Wavemeter; Light Source for High-Speed Photography General Radio Co., 275 Massachusetts Ave., Cambridge 39, Mass. announces four new instruments, as follows:

Two of these instruments are Type 1931-A, an amplitude-modulation monitor, and Type 1932-A, a distortion and noise meter.

The Type 1931-A amplitude-modulation monitor measures percentage modulation on either positive or negative peaks and gives a continuous indication of modulation peaks in excess of a predetermined percentage set by means of a dial. It can also be used for program-level monitoring and for measuring transmitter audio-frequency response. Two audio output circuits are provided, one at 600 ohms for audible program monitoring, and the other a high-impedance circuit that gives a faithful reproduction of the audio envelope for distortion and noise measurements. The linear rectifier is designed for use at a low power level, so that the problem of coupling to the transmitter is greatly simplified. The required r-f power input is only 0.5



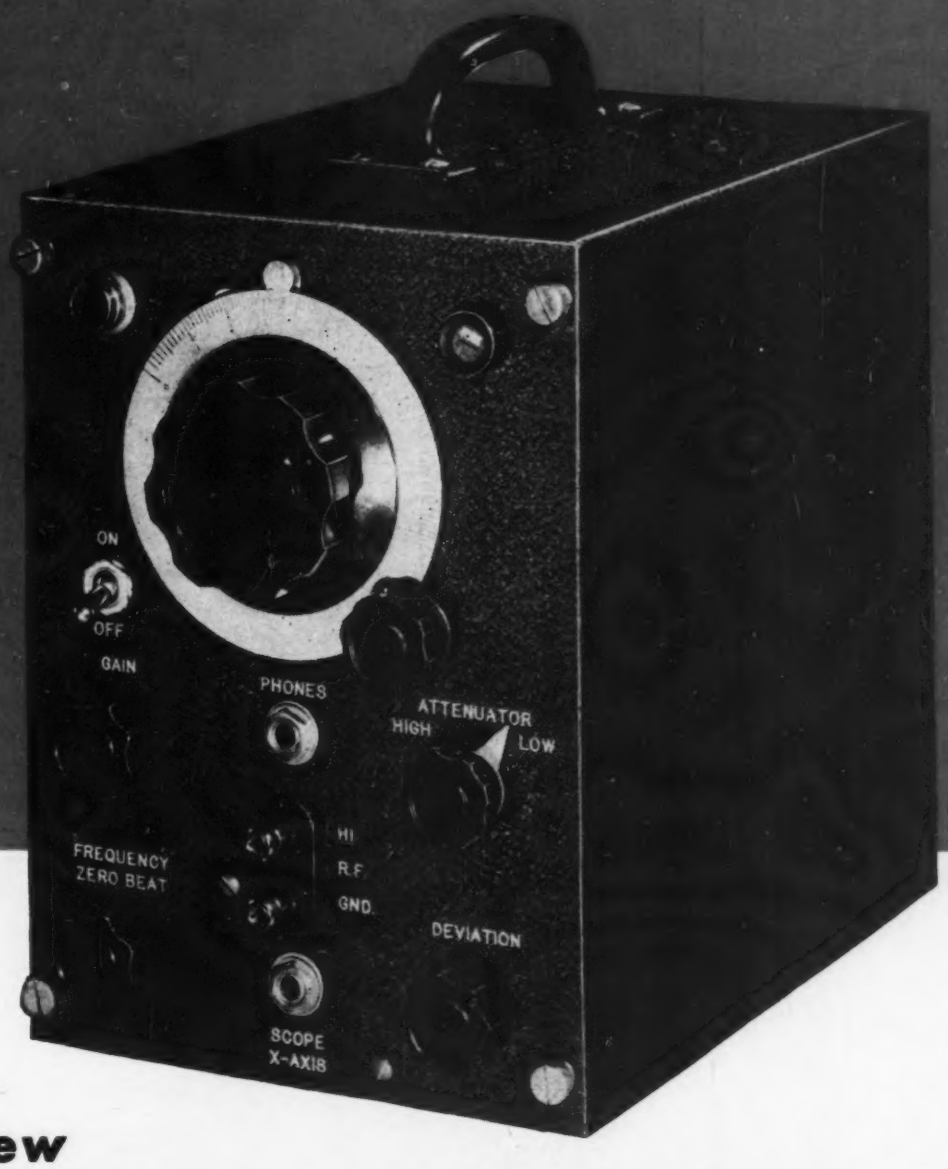
Amplitude-modulation monitor

watt, far less than that necessary to operate pre-war modulation monitors. It operates over a range from zero to 110 percent on positive peaks and from zero to 100 percent on negative peaks; carrier frequency range is 0.5 to 60 mc.

The Type 1932-A distortion and noise meter is a direct-reading instrument for measuring distortion, noise, and hum in audio-frequency systems. When used for measurements on broadcast transmitters, the distortion meter operates from the high-impedance output circuit of the Type 1931-A modulation monitor. A continuous frequency range of 50 to 15,000 cycles, fundamental, is covered by a single dial

Complete equipment for high-speed flash photography

ANNOUNCING



The New
HAR-CAM VISUAL ALIGNMENT SIGNAL GENERATOR

This new HAR-CAM unit provides the most efficient and effective method of aligning the IF circuit of FM receivers. By use of an oscillograph screen, the performance of the IF circuit is shown visually, and rapid, accurate alignment is easily accomplished.

SPECIFICATIONS

1. Frequency range 100kc to 20mc with direct reading dial calibrated in megacycles.

2. Linear frequency sweep deviation adjustable from zero to 900kc peak to peak.

3. Vernier frequency control of 100kc allows zero beat calibration of main tuning dial or for vernier frequency deviations about main dial frequency setting.

4. Stable rf gain control independent of frequency.

5. Five-step attenuator of rf output giving over-all voltage range of 1 microvolt to 1 volt when used in conjunction with the gain control.

6. Output impedance, 1 ohm to 2500 ohms.

7. Phone jack for aural monitoring of zero beat calibration of main tuning dial.

8. Panel jack to feed linear sweep voltage to x-axis amplifier of oscilloscope, thus synchronizing the frequency linear sweep of the generator with the spot trace on the scope screen.

9. Voltage regulated supply for internal oscillators.

10. Careful oscillator design to minimize drift.

11. Stable and proven circuit principles used throughout to insure complete reliability.

12. Size, 7" wide, 9½" high, 10½" deep. Weight, 18 pounds.

For complete information on the HAR-CAM Visual Alignment Signal Generator, write for Bulletin H-40.

HAR-CAM
HARVEY OF CAMBRIDGE

HARVEY RADIO LABORATORIES, INC.

439 CONCORD AVENUE • CAMBRIDGE 38, MASSACHUSETTS

180,000 of these engine thermo-couples were EASY-FLO brazed in quick time with the foot-controlled carbon electro heating set-up below.

EASY-FLO
-JUST RIGHT FOR FABRICATING *thin-gauge* METALS

For joining thin-gauge metals — either for current-carrying or structural purposes — brazing with the low-temperature silver alloy EASY-FLO meets all requirements from both the physical and economic angles.

From the **physical** angle — EASY-FLO brazed assemblies are fully up to solid metal in strength, in ability to stand severe stresses and strains, and in current-carrying efficiency. And EASY-FLO's low working temperature makes it possible to guard against heat damage to thin metals.

From the **economic** angle, EASY-FLO brazing means fast production and big savings in labor and machine-hours.

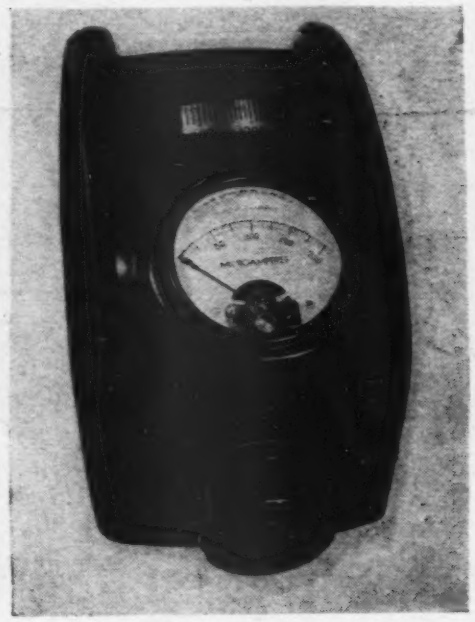
SEND FOR BULLETIN 12-A

This bulletin gives full details about EASY-FLO and tells you how to get the full benefit of its fast, economical production on your metal joining. Write for a copy today.

HANDY & HARMAN
82 FULTON ST., NEW YORK 7, N. Y.
Bridgeport, Conn. • Chicago, Ill. • Los Angeles, Cal.
Providence, R. I. • Toronto, Canada
Agents in Principal Cities

and push-button multiplier. Distortion and noise components up to 45,000 cycles are included in the measurement. The direct-reading distortion meter provides full-scale ranges of 0.3%, 1%, 3%, 10%, and 30% full scale; noise range to 80 db below 100 percent modulation or 80 db below zero v-u.

Type 1140-A wavemeter is useful for rapid measurements of frequency in the range 240 to 1200 mc. This range is covered in a single



direct-reading dial with an accuracy of ±2 percent. The tuning element is a butterfly-type circuit which is coupled to a standard cartridge-type crystal detector. Crystal current, as indicated on a microammeter, gives an indication of resonance. Where the available power is not sufficient to actuate the microammeter, the reaction of the wavemeter upon the current in the circuit under measurement can be used. The entire assembly is housed in a small molded plastic case that can be conveniently held in one hand. Overal dimensions are 3⅞ x 7¼ x 4½ inches; net weight, 3¼ pounds, price: $65.

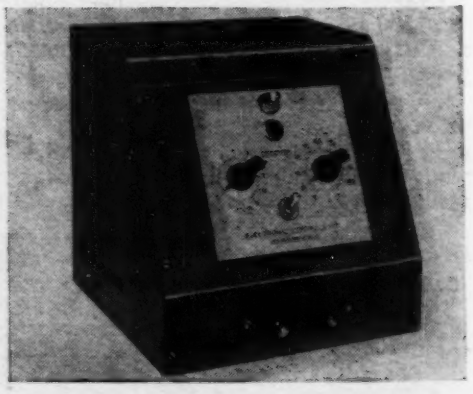
The latest addition to the General Radio line of stroboscopic equipment is the Microflash, a light source for photographic exposures of the order of 2 microseconds. Originally designed for war use in the development and test of new types of ammunition, this instrument provides both knowledge and a record of mechanical phenomena occurring in a very small fraction of a second.

The Microflash consists of a power supply, which charges a capacitor to a high voltage, and means for discharging it through a special gas-filled lamp, designed to dissipate most of the energy in about 2 microseconds, producing an intense, extremely short flash. The flash may be tripped by a make or break contact, by an electrical impulse, or by a microphone which picks up a sound impulse from the phenomenon to be photographed. The microphone is supplied, and an amplifier with built-in gain control.

Ordinary film and camera equipment can be used with the Microflash. Lamp and power supply are mounted in separate metal cases, which lock together in a single unit for transportation. Dimensions (assembled), 24⅛ x 13¼ x 11¾ inches; net weight, 72 pounds; price, $525.00 (plus current Federal tax on photographic equipment.)

Interval Timer

A NEW ELECTRONIC TIMER is suitable for all photographic and industrial applications requiring precise time control. Time range is from 1 to 120 sec in increments of 1 sec. Accuracy is rated better than 5 percent. Two dials provide time selection. One dial is calibrated in single seconds; the other in 10-sec steps. Tap switches are controlled by dials.



Snap-positioning step-switches are used to assure precision control, with exact values of resistance inserted in the circuit at each posi-

REVERE FREE-CUTTING COPPER ROD

. . . INCREASES ELECTRONIC PRODUCTION

SINCE its recent introduction, Revere Free-Cutting Copper has decisively proved its great value for the precision manufacture of copper parts. Uses include certain tube elements requiring both great dimensional precision, and exceptional finish. It is also being used for switch gear, high-capacity plug connectors and in similar applications requiring copper to be machined with great accuracy and smoothness. This copper may also be cold-upset to a considerable deformation, and may be hot forged.

Revere Free-Cutting Copper is oxygen-free, high conductivity, and contains a small amount of tellurium, which, plus special processing in the Revere mills, greatly increases machining speeds, makes possible closer tolerances and much smoother finish. Thus production is increased, costs are cut, rejects lessened. The material's one important limitation is that it does not make a vacuum-tight seal with glass. In all other electronic applications this special-quality material offers great advantages. Write Revere for details.

REVERE

COPPER AND BRASS INCORPORATED

Founded by Paul Revere in 1801

Executive Offices: 230 Park Avenue
New York 17, N. Y.

Mills: Baltimore, Md.; Chicago, Ill.; Detroit, Mich.; New Bedford, Mass.; Rome, N. Y.—Sales Offices in principal cities, distributors everywhere

Listen to Exploring the Unknown *on the Mutual Network every Sunday evening, 9 to 9:30 p.m., EST.*

CUSTOMERS REPORT:

"This material seems to machine much better than our previous hard copper bar; it cuts off smoothly, takes a very nice thread, and does not clog the die." (Electrical parts.)

"Increased feed from 1-1/2" to 6" per minute and do five at one time instead of two." (Switch parts.)

"Spindle speed increased from 924 to 1161 RPM and feed from .0065" to .0105" per spindle revolution. This resulted in a decrease in the time required to produce the part from .0063 hours to .0036 hours. Material was capable of faster machine speeds but machine was turning over at its maximum. Chips cleared tools freely, operator did not have to remove by hand." (Disconnect studs.)

3 OF A KIND – Completely INTERCHANGEABLE

Complete interchangeability is highly important today in units which operate continuously or for long periods under a variety of conditions. If breakdown should occur, a new part may easily be installed and operation resumed in short order.
Original gears for such units, and also replacements must have positive guaranteed accuracy, tooth form and finish, or the mechanism may fail. Our completely modern shop, equipment and methods, can serve you with assurance on such requirements. Write us for quotation on your specifications.

THREAD GRINDING

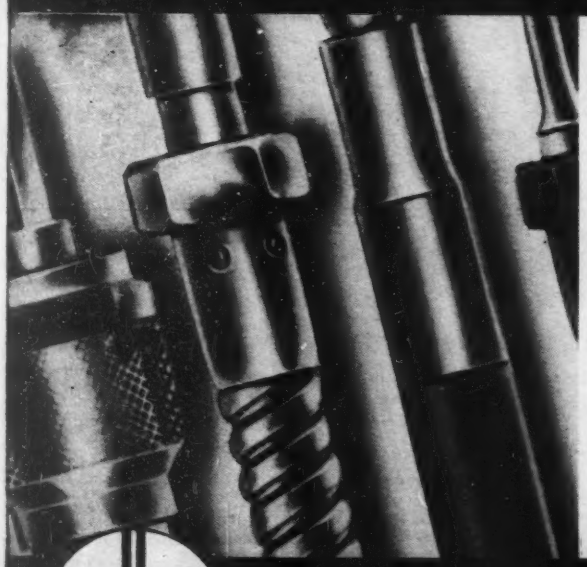
SMALL OR MEDIUM WORMS AND GEARS

SPUR, HELICAL, BEVEL, SPROCKETS AND SPLINES

Beaver Gear Works Inc.

1021 PARMELE STREET, ROCKFORD, ILLINOIS

SIMPLIFY METHODS OF ASSEMBLING AND SERVICING with WALKER-TURNER FLEXIBLE SHAFTING



• You can put controlled elements wherever good design dictates . . . when you use Walker-Turner Flexible Shafting in your products. It goes around and over obstacles—takes up minimum space—centralizes controls for convenient operation, no matter where the condensers, rotary switches, etc., are located.

Scientifically engineered and carefully wound to produce smoother power flow, sensitive control and trouble free operation. Write for further information.

WALKER-TURNER CO., Inc.
PLAINFIELD, NEW JERSEY

walker-turner COMPANY, INC. PLAINFIELD, N.J. U.S.A.

FLEXIBLE SHAFTING
FOR REMOTE CONTROL AND POWER TRANSMISSION

tion. A pilot light, toggle on-off switch and a push-switch are included. A double receptacle permits timing two circuits simultaneously. Contacts are normally wired spdt with double break and are rated at 25 amp, 32 v d-c, 25 amp, 125 v a-c, and 10 amp, 230 v, a-c. Contacts can be wired dpdt at correspondingly lower ratings. The cabinet measure 8 x 8 x 8 in. The tube used in the unit is a 2050 thyratron. High accuracy resistors assure precise timing and timing is independent of the length of time the tripping push button is depressed. A dial switch is provided with a position for continuous circuit closure. The unit is compact, easy to operate and supplied for 115-v a-c operation. Electronic Controls, Inc., 44 Summer Ave., Newark 4, N. J.

4

Vacuum Capacitor

TENTATIVE characteristics of Jennings 1000-$\mu\mu$fd high voltage vacuum capacitor: peak voltage 10 kv (increased voltage ratings may be obtained on request); peak current 100 amp; capacity 0.001 μfd; overall length approximately 7$\frac{7}{8}$ in.;

maximum diameter at center 4$\frac{3}{8}$ in. The capacitor is for use in heavy industrial applications, or for broadcast studios and experimental laboratories where rugged mechanical construction is essential. Jennings Radio Mfg. Co., 1098 E. William St., San Jose 12, Calif.

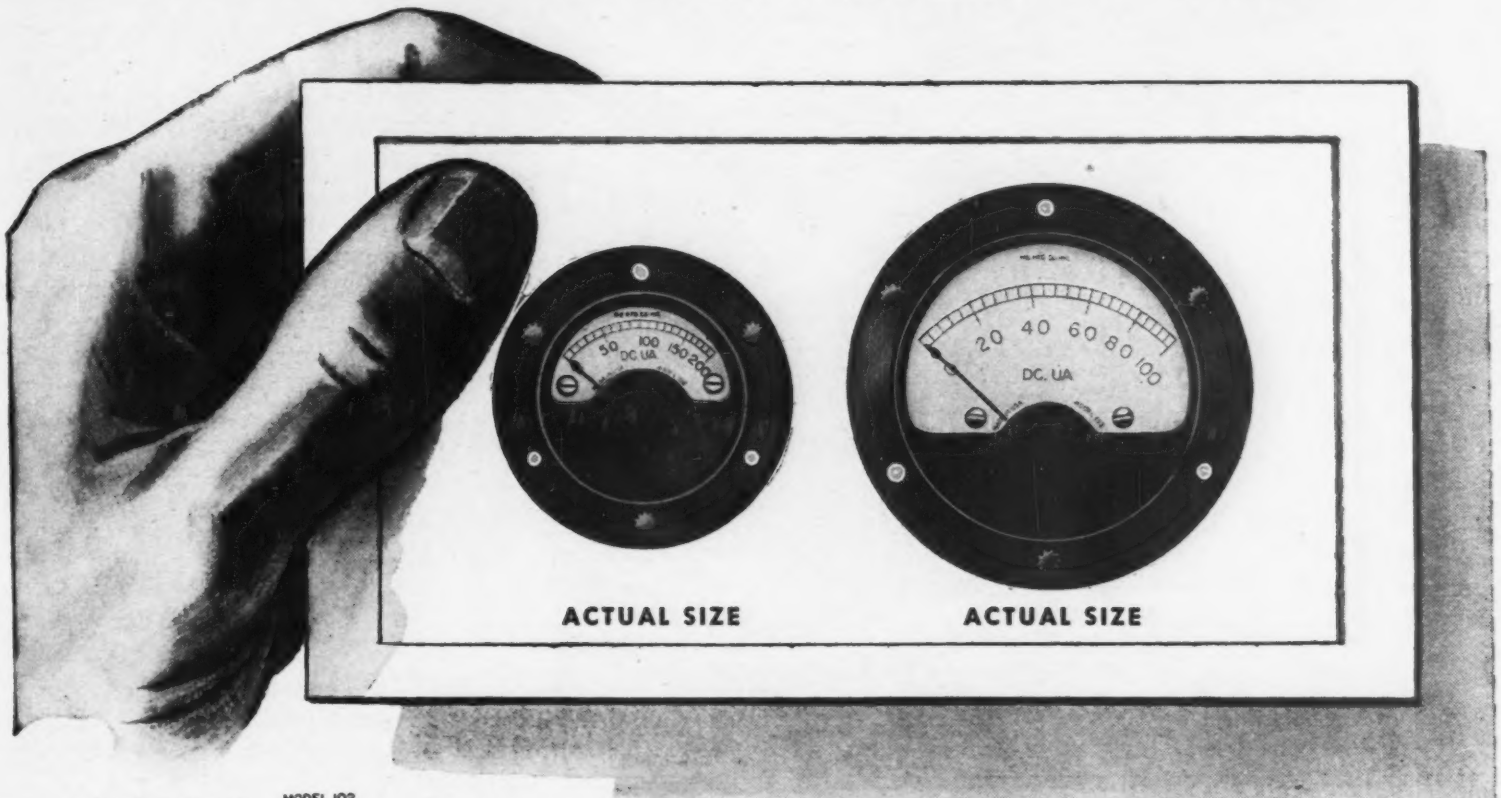
5

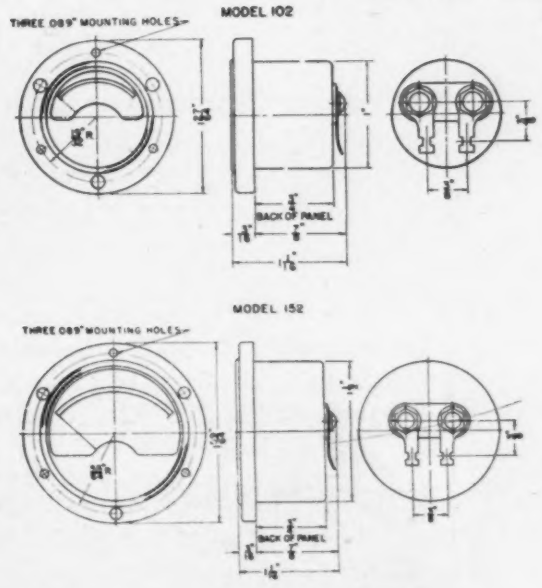
Pulse Generator

DESIGNATED AS Micro-Pulser, a newly developed electronic instrument generating short pulses simplifies production testing of television and radar systems, compon-

Presenting...

NEW MINIATURE METERS BY MB





PRECISION AND PRICE—both have been given the utmost consideration in the design of these two additions to the MB family of quality miniature indicating instruments.

They bring to your portable products or panelboard the economies of simplified manufacturing and production methods, *plus* the advantages of small size and standard accepted accuracy.

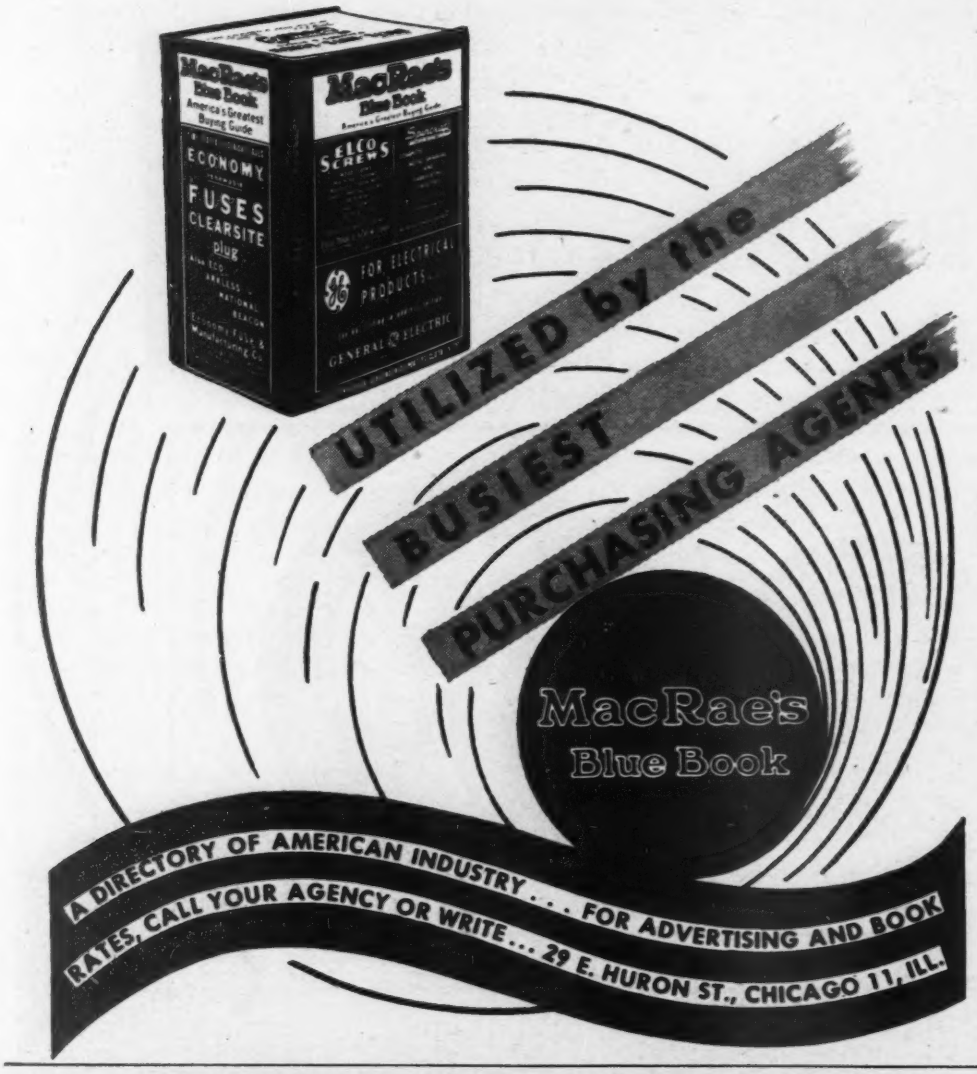
The fine performance of these new lines will fulfill your every expectation . . . for good reasons! Scales are clear, sharply defined. The sensitive, featherweight element moves in the field of an Alnico No. 5 magnet—response is fast. Because they're built with jewel bearings, and are precision pivoted, these rugged instruments will withstand hard usage. The anodized, pressed aluminum cases are securely sealed. Model 152 mounts in a 1½ inch opening . . . model 102 in a *1-inch opening*—it's the most compact construction made!

Write for full details and description. We invite your inquiries for special scales, accessories, or adaptations . . . or qualified engineering assistance.

THE
MB
MANUFACTURING COMPANY, INC.
Instrument Division
331 East St., New Haven 11, Conn.

INSTRUMENT MB DIVISION

ELECTRICAL METERS FOR AIRCRAFT AND PORTABLE EQUIPMENT



. . . 15 CUBIC FEET A MINUTE

The No. 1½* is one of many blowers manufactured by the L-R Mfg. Div. with C.F.M's at 8000 R.P.M. ranging from 15 to 270. They will outperform many larger types and where size and weight are factors, they solve cooling problems presented by electronic tubes or circuit components in airborne communication units as well as in many industrial applications.

*WEIGHT: 2 oz.; CAPACITY: 15 C. F. M. at 8000 R.P.M.; CONSTRUCTION: Housing of high impact phenolic plastic. Wheel is turbotype cadmium-plated steel; SIZE: 2⅝" long x 61/64" wide x 2½" high.

The RIPLEY Company

BLOWER DIVISION

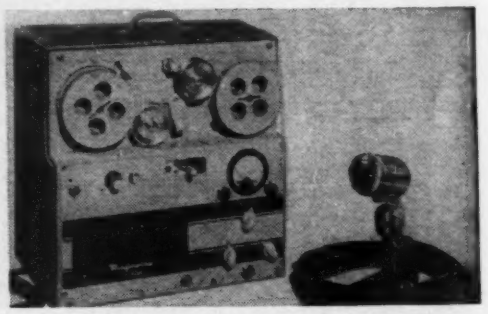
13 NEW LITCHFIELD STREET TORRINGTON, CONNECTICUT

ents, or video amplifiers, or can be used to obtain transient response study of electrical networks, or as a microsecond timing device, or pulse modulating source for high frequencies, or as a laboratory unit for student instruction in radar or in pulse techniques. The generator produces a pulse width of from one-half to 5 microseconds, a pulse repetition rate of 200 to 2000 and from 2000 to 20,000 cps. It will trigger from external positive or negative pulse. Other features include: variable delay in two ranges (from 10 to 100 and from 100 to 1000 microseconds); output pulse is both positive and negative (low impedance positive pulse at approximately 150 ohms, and negative pulse at approximately 400 ohms); amplitude of pulse is continuously variable between zero and 50 v. A positive and negative synchronizing pulse is furnished to trigger external equipment. The generator has a self-contained power supply (117 v, 60-cycle operation). It measures 10 x 14 x 8 in., weighs about 30 lb, and is priced at $150, F.O.B. East Orange. Kay Electric Co., 8 Eaton Place, Newark, N. J.

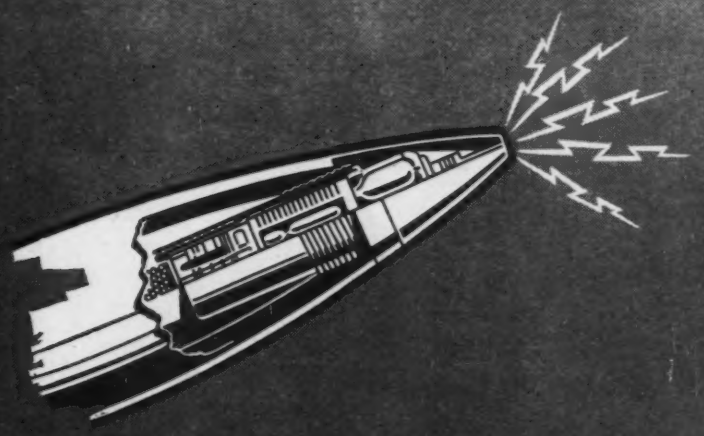
6

Recorder-Reproducer Unit

MANUFACTURED BY Utah Radio Products Co., 812 Orleans St., Chicago 10, Ill., and designated as Magicwire, a portable recording reproducing device records on a moving steel-wire, reproducing sounds immediately. Overall dimensions of the unit are 13¾ x 11½ x 9 in. and the total weight is 37¾ lb. Used re-



cording wire can be preserved indefinitely, or it can be cleared of its record and used again. Recordings can be made regardless of the position of the recorder. Climate, temperature or vibration have no effect on the unit. A timing device enables one to select only a portion



★ *Waterbury Companies, inc.*

MANUFACTURED

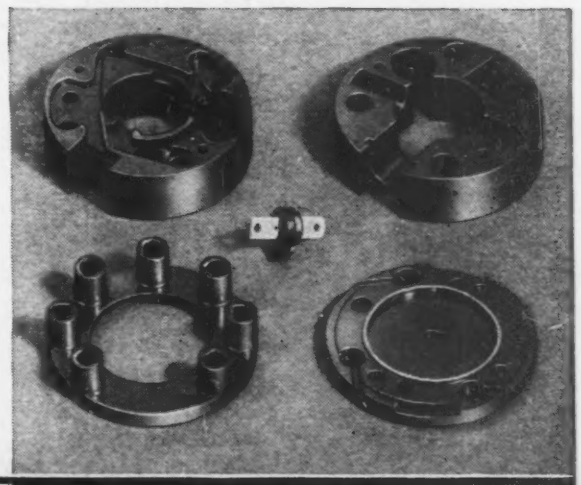
★ 5,132,000 ★

COMPONENT PARTS *for the* VT FUSE

★ NOW IT CAN BE TOLD

Waterbury's technical skill . . . the same skill that produced intricate plastic parts for this secret weapon which affected the course of the war . . . is available to manufacturers who use plastic parts and assemblies in their products.

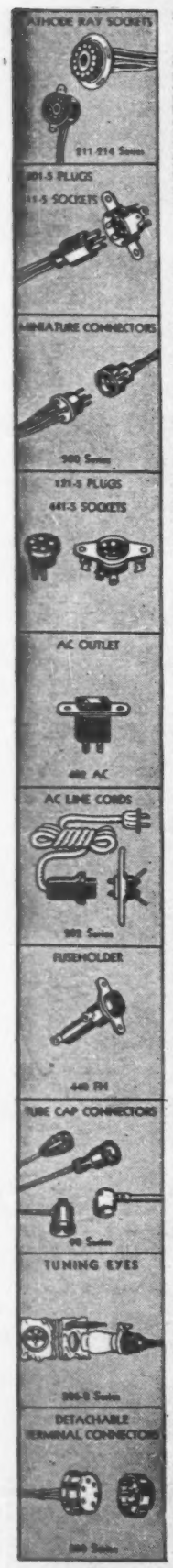
Highly intricate mold making and precision plastic molding by transfer, injection, and compression is a specialty with Waterbury Companies. Let our complete engineering and design staff work with you on the production of your plastic parts.



These small but highly intricate plastic pieces are vital parts of the now famous VT fuse.

WATERBURY COMPANIES, INC.
FORMERLY WATERBURY BUTTON CO., EST. 1812
McGRAW AVE., WATERBURY, CONNECTICUT

ALDEN
RADIO COMPONENTS



211 and 214 SERIES CATHODE RAY TUBE CONNECTOR WITH LEADS
Any requirements in a cathode ray tube connector complete with the proper leads attached are specially engineered as a complete assembly to insure high safety factors in all kinds of service. For octal, magnal, diheptal bases and any new cathode ray tube. Have super-long leakage paths, rounded, "coronaless" clips and individual pocket type insulation and strain relief for each lead. Buying the complete assembly gives you the best obtainable connector plus the economies of standardization.

801-5 SHIELDED PLUGS AND 441-5 METAL SOCKETS
Shielded plug and socket for automobile sets or for any other equipment where leads must be shielded and shield grounded to chassis. One to five circuits. Individual insulation for each lead prevents shorting by vibration. Shield is easy to put on and solder to plug. Supplied with or without shielded cable.

MINIATURE CABLE CONNECTORS 500 SERIES
One to four circuit cable connectors —famous for connecting AC motors in combination sets and all kinds of "through-panel" work. Overall diameter only ¾". Eliminates ponderous "ball and chain" devices. Save labor costs by having our special wire equipment put on leads to your particular needs. Each lead has individual "pocketed" insulation and strain relief. Underwriters approved.

121-5 MINIATURE PLUGS AND 441-5 SOCKETS
Compact plug and metal seal socket of 1 to 5 circuits. Use when you want connector to come directly out of chassis. Plug supplied with leads to your specifications. Socket has easy to solder tabs. "Pocket" type individual insulation on each lead and clip.

AC OUTLET 402AC
Smallest possible outlet that can be eyeletted or rivetted to chassis like other components. Solder tabs designed for easy soldering.

AC LINE CORDS 202 SERIES
Here is a detachable AC line and with socket, that is neat and compact. Socket eyelets or rivets in place like other components. Underwriters approved. Special cords and plugs to your specifications.

FUSEHOLDER 440FH
Here is a fuseholder that rivets or eyelets in place like the other components in your set. Cannot twist or turn, does not require special wrenches for assembly. Has spring to eject fuse if it breaks. Spring makes contact at base of fuse and prevents rattle. Top contact slottted for easy removal of fuse ferrule when glass breaks. Tabs are of special design for ease in atttaching primary leads of ample size for any current carrying requirements.

90 SERIES TUBE CAP CONNECTORS WITH LEADS
Any requirement in tube cap connectors supplied with leads of proper voltage handling characteristics. Many made special, hundreds of moldings, stampings and wire to draw on.

206-8 TUNING EYES WITH LEADS
Socket supplied with tailor-made leads attached. With or without escutcheon and bracket. Individual insulation and strain relief for each lead.

200 SERIES DETACHABLE TERMINAL CONNECTORS
Two to eight circuit detachable connector. Replaces terminal strips. Supplied with leads attached. Each lead has individual insulation and strain relief.

WIRE AND CABLE
Any kind of wire or cable laced, braided, woven or assembled with any of our components or those of other make. Many types of wire in stock and in process.

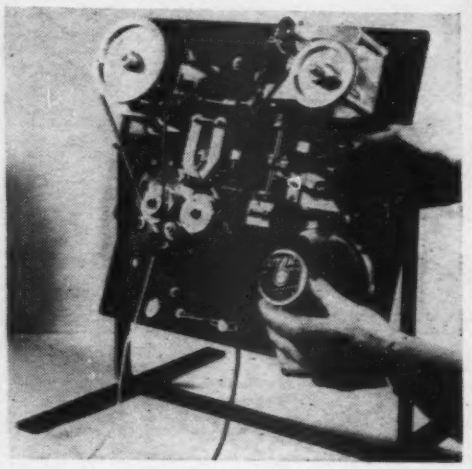
NEW ITEMS
When you want to get into production fast, Alden is a specialist in bringing through special electrical assemblies, new items in small or large quantities. Samples made promptly.

Electrical Recording Instruments

Special instruments to record electrical impulses as they occur with all the minute variations of intensity and duration, free from the lag and inertia of present systems. "Electrographic" recorders we can supply, include a complete line of facsimile recorders, specially engineered recorders for high speed signal analysis, slow speed recorders for day by day events, multi-trace recorders for simultaneous recording of any phenomena that can be reduced to electrical impulses.

ALDEN PRODUCTS COMPANY
BROCKTON 64E3 MASS.

(if desired) of the recording in playback. The unit operates on 115 v, 60-cycles, a-c and is for use in radio stations, home and other noncommercial entertainment. A full-wave rectifier tube (5W4), three-



stage audio amplifier, 30-kc oscillator tube (6V6), record-listening mechanism, drive motor and associated mechanism and necessary accessories comprise the recorder and reproducer. The wire used on the machine is 0.004 inch in diameter, and each spool, which is about 3 in. in diameter, will take 66 minutes of recording.

Hermetic Cans, Terminals

FUSITE "HermetiCans" and Fusite multiple hermetic terminals are available from Cincinnati Electric Products Co., Carthage St. at Hannaford, Cincinnati 12, Ohio. "HermetiCan" is a registered trade name, identifying a new series of metal containers incorporating the hermetic terminals. The terminals are ready for connection to the elec-

trical part. The part is placed in the HermetiCan, and by means of either a manually-operated or motor-driven closing machine, is hermetically sealed at the rate of 150

PLASTIC ENCLOSED

TYPE 1200
ADVANCE ULTRA SENSITIVE* D. C. RELAYS

★ **2½ TO 5 MILLIWATT OPERATION 25% TO 15% DIFFERENTIAL**

COVER: Moulded Plastic, Cellulose Acetate, Clear, Tough Single Screw Attachment • No dust or dirt on contacts • No accidental operation • No short circuits • Instant visual inspection • Low maintenance of contact adjustment

BASE: Moulded black BAKELITE • Good mechanical strength • High dielectric strength and insulation • Negligible water absorption • Compactness and fine appearance

OPERATING POWER: 5 Milliwatts for positive operation • 2½ Milliwatts with careful adjustment and light contact loads

MAGNETIC CIRCUIT: Armature and pole of Nickel-Iron alloy, Hydrogen annealed for high permeability and low retentivity • High overall sensitivity • Small make-break coil current differential—(25% to 15% less current to break than to make)

ARMATURE: Counterbalanced • Prevents action of relay due to moderate vibration • Allows operation in any position

SENSITIVITY ADJUSTMENT: Vernier screw for coil spring tension on armature • Accuracy • Permanent setting, easily changed

CONTACTS: Pure Silver (palladium, platinum or other specified materials at extra cost) • Single pole, double throw • 1 ampere on 110 volt A.C., non-inductive load • Screwdriver adjustment

COIL: Standard resistance from 1 ohm to 10,000 ohms, up to 30,000 ohms at small extra cost • Cellulose acetate insulation • Varnish vacuum impregnation

TERMINALS: Solder lugs and screws, recessed on bottom of base, accessible through panel or through knockouts on side of base

MOUNTING: Surface mounting, any position, fastens with two No. 6 screws

SIZE: 2" x 2-9/16" x 1½" high

WEIGHT: 6¼ ounces

PRICE: Moderate

Write for quotations and catalogs on the Advance Type 1200 Ultra Sensitive D. C. Relay and other Advance Relays

Advance Relays

ADVANCE ELECTRIC & RELAY CO.
1260 West 2nd Street
Los Angeles 26, Calif., U. S. A.
Phone Michigan 9331

ELECTRONIC AND MECHANICAL POWDER METALLURGY

IN the twin fields of electronic and mechanical powder metallurgy, where close attention to dimensional tolerance, porosity, density control, etc., is a production "must", Micro-Ferrocart's reputation for strict adherence to such requirements is well established.

•Typical examples of this attention to "quality in production", are the Radacor Iron Cores shown here. Precision-made, subjected to rigorous tests, they are designed for use at all frequencies in television and FM.

•If you contemplate the manufacture of a product requiring electronic or mechanical powder metallurgy parts, designed to close tolerances and of specific densities, why not outline your problems in a letter to us today. Our mechanical powder metallurgy parts replace those heretofore requiring much expensive milling and machining. Our complete engineering and plant facilities are at your disposal.

The Radacor Iron Core used in this Meissner I. F. Transformer permits higher "Q" with a resultant increase in selectivity and gain.

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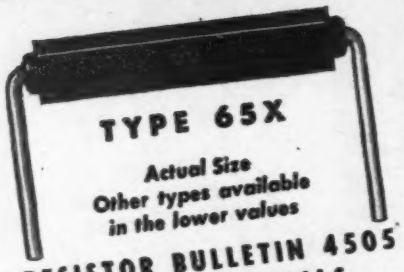
MICRO-FERROCART PRODUCTS DIVISION

MAGUIRE INDUSTRIES, *INCORPORATED*

375 FAIRFIELD AVENUE, STAMFORD, CONN.

The "All-Weather" Resistors



TYPE 65X

Actual Size
Other types available in the lower values

RESISTOR BULLETIN 4505 GIVES FULL DETAILS...

It shows illustrations of the different types of S. S. White Molded Resistors and gives details about construction, dimensions, etc. A copy, with Price List will be mailed on request. Write for it—today.

- Noiseless in operation
- Strong and durable
- Good performance in all climates

STANDARD RANGE
1000 ohms to 10 megohms
• NOISE TESTED •

At slight additional cost, resistors in the Standard Range are supplied with each resistor noise tested to the following standard: "For the complete audio frequency range, resistor shall have less noise than corresponds to a change of resistance of 1 part in 1,000,000."

HIGH VALUES
15 to 1,000,000 megohms

S.S.WHITE INDUSTRIAL DIVISION
THE S. S. WHITE DENTAL MFG. CO.
DEPT. R, 10 EAST 40th ST., NEW YORK 16, N. Y.

FLEXIBLE SHAFTS • FLEXIBLE SHAFT TOOLS • AIRCRAFT ACCESSORIES
SMALL CUTTING AND GRINDING TOOLS • SPECIAL FORMULA RUBBERS
MOLDED RESISTORS • PLASTIC SPECIALTIES • CONTRACT PLASTICS MOLDING

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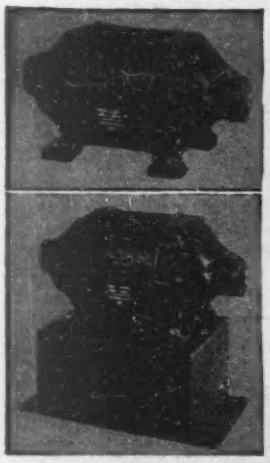
Janette

D.C. to A.C. CONVERTERS

The Janette Rotary Converter is a simple, practical method for obtaining Alternating Current when only D.C. power is available.



.1 to 3.2 K. V. A.

Janette was one of the first manufacturers to build converters especially for use with A.C. electronic tube devices. Since their inception these machines have established a world wide record for reliable, efficient, quiet, trouble free operation, under the most adverse conditions.

TWO TYPES are available; one for commercial applications, the other for marine service. Special filters for suppressing conducted and radiated noise voltage interference can be supplied.

If you want a really dependable converter, guaranteed for one year, we suggest you **TRY A JANETTE!**

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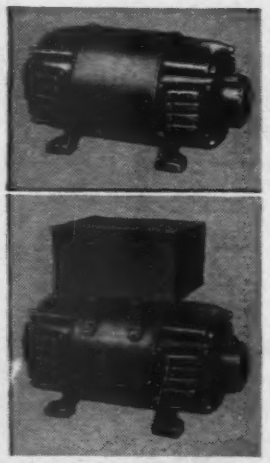
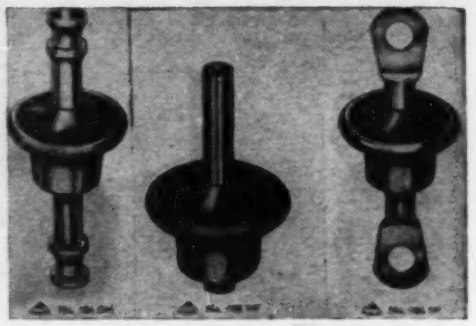


Janette Manufacturing Company
556 W. Monroe St. Chicago 6, Ill.



to 300 per hour, easily and quickly without solder or heat by means of a double seam formed in the automatic closing machine. Containers (of various lengths, range from 1½ to 4 in. in diameter) are sturdy, withstand internal and external pressures satisfactorily. Lids are equipped with either 600 or 800 series hermetic terminals (one of 26 glass-to-metal types). Terminals are available in any number from one to nine or combinations and come in two different styles, hollow tube and flattened and pierced.

General Electric Products

AN ELECTRONIC switch, two new thermoplastic cements and a loudspeaker are all new items available from General Electric Co., 1 River Rd., Schenectady, N. Y.

The electronic switch is type YE-9 designed for special electrical studies of wave form, phase, frequency

relationship, and for the comparison of amplitudes. By using two of these switches in cascade, three independent circuits can be studied simultaneously, using any oscilloscope with a horizontal sweep voltage and available connections to the plate of the cathode-ray tube. Operating on any sweep frequency

Is FLUCTUATING LINE VOLTAGE *interfering with your*

Test Equipment
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A SORENSEN ELECTRONIC A.C. VOLTAGE REGULATOR

will hold your Line Voltage
CONSTANT to 2/10%

NOT AFFECTED *by variations in*

Input Voltage
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Write the factory or your nearest representative (see below) for catalog describing various models available and how they function.

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SUN RADIO

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Where the impossible becomes the usual. 6,000 square feet of space and a considerably enlarged staff, enable us to improve a service that has always been TOP FLIGHT.

With over 10,000 items in stock, of Standard Nationally advertised products, you have a wide selection to fill your every requirement.

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SUN RADIO

& ELECTRONICS CO., Inc.

122 Duane St., New York 7

Tel. BArclay 7-1840

from 10 to 12,000 cps, continuously variable, the YE-9 has an amplifier with frequency response flat within 3 db from 4 cps to 450 kc. The device's input power voltage is 110-125 v, 50-60 cps, and its maximum signal input is 250 rms.

The two thermoplastic cements, Nos. 2142 and 2160, are developed specifically for use in the manufacture of quartz crystals and loudspeakers. Their properties insure a clean, dry and efficient union between porous and non-porous surfaces.

Long-range projection of the human voice was demonstrated by General Electric engineers to officers of the Air Technical Service

Command at Wright Field, Dayton, Ohio, recently. G-E's "Old Loudmouth", a new super loudspeaker was used. The loudspeaker can deliver a whisper a mile. It operates with compressed air on much the same principle as the lungs are used in transmitting human speech. Under average weather conditions, the speaker delivers audible speech two miles. It has been heard at varying distances up to five miles. It is particularly useful in applications where a system is needed to override high noise levels.

New Ceramics

A THERMAL EXPANSION coefficient in the same range as that of Invar, and high resistance to extreme thermal shock are among the properties of a ceramic material recently developed by General Ceramics and Steatite Corp., Keasbey, N. J. The ceramic is designated as Material M-244, and can be used in such applications as in high-temperature furnaces, insulation for high precision instruments in which dimen-

RESISTANCE WIRE

ALLOY "A": Nickel-chromium alloy, resists oxidation at extreme temperatures. Essential for operating temperatures up to 2100° F. Also used for cold resistance. Resists chemical corrosion by many media. Non-magnetic; specific resistance, 650 ohms/C.M.F.

C. O. JELLIFF MFG. CORP.
123 PEQUOT AVE. • SOUTHPORT, CONN.

RESISTANCE WIRE

ALLOY "C": Nominally contains 60% nickel, 15% chromium, and balance iron. High resistance to oxidation and corrosion. Widely used in resistances for radio and electronics, industrial, and domestic equipment. Operating temperature up to 1700° F. Specific resistance 675 ohms/C.M.F.

C. O. JELLIFF MFG. CORP.
123 PEQUOT AVE. • SOUTHPORT, CONN.

RESISTANCE WIRE

ALLOY "180": Nickel-copper alloy with resistivity of 180 ohms/C. M. F. Widely used for resistor elements up to 750° F. (400° C). For radio controls, magnets, rheostats and voltage control relays.

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RESISTANCE WIRE

ALLOY "45": Alloy of 55% copper, 45% nickel with a constant electrical resistance over wide range of temperatures. Specific resistance 294 ohms/C.M.F.; temperature coefficient 0.00002 ohms per degree F; 32 to 212 degrees range. Used in winding of precision resistors.

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RESISTANCE WIRE

KANTHAL: Exclusive manufacturers of KANTHAL, an outstanding achievement in resistance-wire development. Now available—complete data upon request.

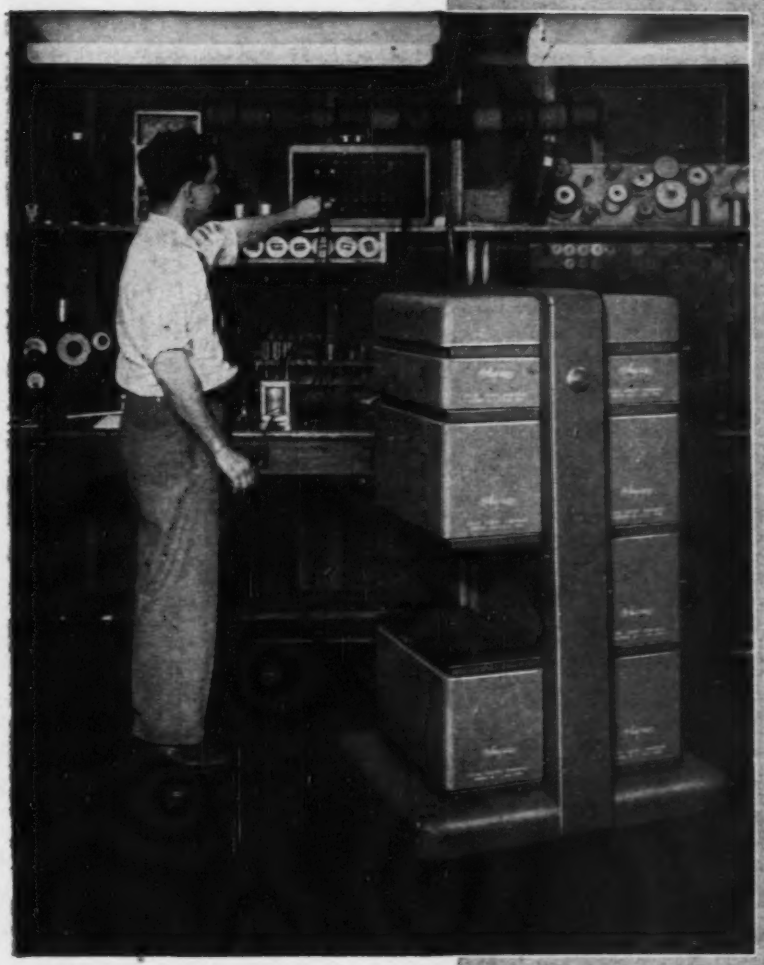
C. O. JELLIFF MFG. CORP.
123 PEQUOT AVE. • SOUTHPORT, CONN.

Note: All alloys are produced in high-frequency type furnaces, and are furnished bright, dull or oxidized finish, also with enamel, silk, or cotton insulation.

At your fingertips -

THE COMPONENTS YOU WANT

when you want them!



Now—at a moment's notice—you have an almost limitless number of circuit components and power sources at your command with the Harvey-Wells Decade Unit.

Entirely new and unique both in design and application, this unit saves time and expense in laboratories, schools or workshops working on DC or low frequency AC electrical circuits. A few of the possible applications are filter circuits, power supply loads, phase shift circuits, phase correction circuits, inductive or capacitive loads for testing relay contacts, resonance circuits, control circuits, voltage divider circuits, etc.

A complete technical data Booklet is now available. Send for it today.

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IN·RES·CO RESISTORS

Close tolerance units for precision performance

IN-RES-CO resistors are engineered to meet all important electronic application needs. They are wire wound for permanent, exact resistance value and both inductive and non-inductive types are standard. Standard tolerances are 1 and 2%—closer tolerances on special order. Rigid quality control assures a uniform standard of excellence and modern manufacturing facilities result in low unit cost and prompt delivery. Manufacturers of electrical and electronic instruments and equipment should investigate the advantages of designating IN-RES-CO as their exclusive wire wound resistor source.

TYPES RN AND SN Tropi-pruf resistors—RN: 3/4" long x 5/8" dia., Max. Res. 1/2 Megohm, 1/2 watt; SN: 1-3/16" long x 5/8" dia., Max. Res. 1 Megohm, 1 watt

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INSTRUMENT RESISTORS CO.

25 AMITY STREET, LITTLE FALLS, N. J.

WEAR RESISTANT
NON-ABRASIVE
NON-CORROSIVE
PERMOPIVOTS
NON-MAGNETIC
SUPER-POLISHED
Yes, Extra Value!
Long after the ordinary pivot has worn out you'll find PERMOPIVOTS in the prime of life—faithfully keeping precision instruments precise! It's all due to the special PERMOMETAL tip—the ideal osmium alloy. . . . Made to your specifications.
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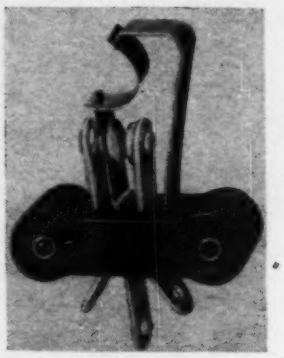
MANUFACTURING METALLURGISTS

sional changes must be absolutely minimized, and for many electrical and electronic purposes where a low thermal expansion coefficient is necessary. Some of the technical data is as follows: Coefficient of thermal expansion up to 600 deg F 1.6×10^{-6}, and up to 1000 deg F 1.9×10^{-6}; dielectric constant (1000 kc) 5.30; power factor 0.540 percent; dielectric strength 90 v/mil; withstands a temperature of 2500 deg F. The ceramic will be available in any shapes that can be formed by pressing, extruding, or casting.

10

Snap Switch

ILLUSTRATED BELOW is a new type smaller size open blade snap switch (designated as Model M) for vertical mounting. It can be mounted singly or in multiples. Overall di-



mensions are approximately $1\frac{13}{32}$ x $1\frac{3}{64}$ x $\frac{7}{16}$ in. Standard operating pressure is 6 to 10 oz. The switch is furnished for sp, normally open, normally closed and dt circuits. Rated at 15 amp, 125 v a-c, and $\frac{1}{3}$ hp 110 a-c. Acro Electric Co., 1316 Superior Ave., Cleveland 14, Ohio.

11

Communications Receiver

TYPE HQ-129-C is a professional or amateur receiver manufactured by The Hammarlund Mfg. Co., Inc., 460 West 34th St., New York 1, N. Y. It has the following characteristics: full range 0.54 to 31 mc, accurately calibrated; band spread on 4 calibrated amateur bands and one arbitrary scale; three i-f amplifier stages; two audio stages; low drift beat oscillator for code; antenna compensator to provide maximum image rejection and high sensitivity; and a compensated oscillator to reduce drift during warm-up. Voltage regulation reduces effects of

A NEW AND DIFFERENT

SWEEP GENERATOR

Needed in

TELEVISION, FM, RADAR, HIGH FREQUENCIES

.1 Volt, Max., 500 Kc to 110 Mcs; 100 Ohms, 10 Mc. Sweep Width

SAVES TIME
—In Production and in Testing

LIGHT
—Only 16 lbs.

SMALL
—14½" x 8" x 8"

$395.00 APPROXIMATE PRICE

For your work in television, FM, radar, high frequencies, etc., you will need one of these newly developed electronic frequency modulated signal generators covering a continuous range between 500 Kc and 110 Mc. Designed primarily for use for field, laboratory, or production alignment of wide band r.f., i.f., or video amplifiers used in radar, direction finders, television, or other wide-band systems.

Sweep range is adjustable from 10 Mc. down to 5,000 cycles at any frequency within the above range for alignment of narrow-band receivers or amplifiers. Self-contained power supply. Input 110 V., 50-60 cycles. A.C. 60 watts. Two internal "markers" are provided, one at intervals of 10 Mc., the other at intervals of 1 Mc. for band-width measurement. The amplitude of these markers is adjustable from the panel. The main dial is marked in megacycles/sec. and when set at any frequency the sweep is plus and minus 5 Mc. from this setting.

An attenuator is provided which reduces the output signal of .1 V. to about 30 microvolts, which is well below the gain control region of most receiver or amplifier systems.

Also Available Immediately In Sample Quantities

HIGH VOLTAGE RF POWER SUPPLIES
(10 KV, 24 KV, or 30 KV) for 7, 10, 12, or 14 inch direct-viewing Kinescopes and for projection sets

ICONOSCOPE YOKES

CATHODE RAY RECEIVING TUBE YOKES

DEFLECTION TRANSFORMERS

Write for preliminary technical data. A limited number of orders placed now for the Sweep Generator can be filled immediately.

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ELECTRODE

V these tube characteristics to meet the most exacting requirements of fine instrumentation.

They are intended primarily for service where ordinary commercial tubes are not suitable.

Actual Size

Series VW-41 Characteristics

Filament Current	0.015 amperes
Filament Voltage	1.5 volts
MU G-1 •	18
Transconductance	65 micromhos
Plate Resistance	275000 ohms

Especially suitable for measuring very small currents or voltages in very high resistance circuits particularly where input resistance may be of the order of 10^{12} ohms or greater.

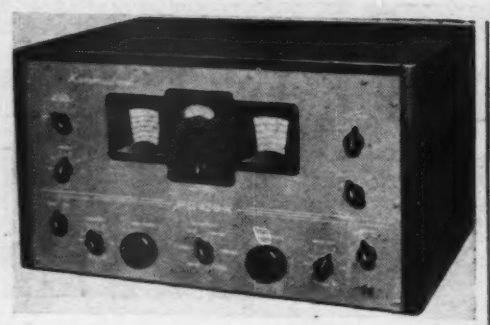
Also available as . . .

Electrometers Triodes
Pentodes Diodes

Hi-Meg Resistors

Designed for high precision instrumentation where ranges of 10 millivolts to 10 volts are used. The same physical size is maintained for all values from 1 megohm to 1,000,000 megohms. Vacuum sealed in special treated glass—size of envelope 1⅞ inches long 3/16 inches in diameter.

THE VICTOREEN INSTRUMENT CO.
5806 HOUGH AVENUE
CLEVELAND 3, OHIO



line voltage fluctuation. Earphone jack cuts out speaker when phones are used. The set has 11 tubes, including voltage regulator and rectifier. A 10-in. permanent magnet dynamic speaker with cabinet (to match) is priced at $10.50 net, the unit itself sells for $129. Overall dimensions are 13-in. deep, 19½-in. wide, and 11-in. high. Speaker cabinet dimensions are 7¼-in. deep, 12½-in. wide, and 12¼-in. high.

12

Pilot Light Assemblies

A BUILT-IN RESISTOR is an integral element of Series PL-849 pilot light assemblies, manufactured by Dial Light Co. of America, Inc., 900 Broadway, New York 3, N. Y., to enable direct connection to 115-v circuits. Ratings for the resistor are available as follows: 100,000 ohms for bright glow, 200,000 for dimmer flow; 270,000 ohms for 220-v circuits. Jewel caps are available in 7 different colors.

13

Battery Eliminator

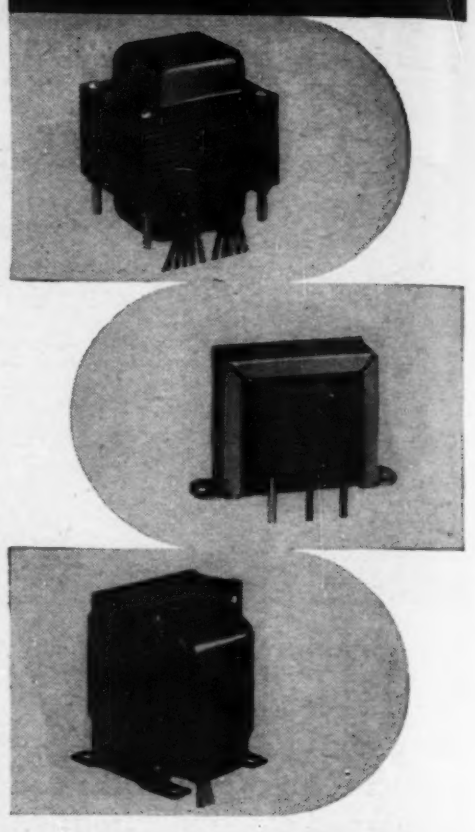
DESIGNED TO REPLACE two dry cells, this new Electrox battery eliminator (type A3V) is for use where 3 volts of direct current is required. It operates on 110 v, 60 cycles, and delivers a smooth, noiseless d-c output of 3 v, 150 ma. It is enclosed in a metal can housing which measures 3 x 2½-in. Schauer Machine Co., Electrox Div., 2048 Reading Rd., Cincinnati 2, Ohio.

14

High Voltage Capacitors

THESE UNITS are especially adaptable for radar and similar applications, and incorporate special high dielectric-strength refractory film and an exclusive fabrication technique. Capacitors are made to individual specifications only (an example of the types available is a 15

HUDSON AMERICAN TRANSFORMERS



FOR GENERAL ELECTRONIC AND RADIO APPLICATIONS

HUDSON AMERICAN'S 20 years engineering experience has developed 6 *outstanding points* of perfection in the manufacture of high quality Transformers:

1. **Most modern coil producing machinery available.**
2. **Vacuum impregnating equipment for wax or varnish.**
3. **Completely automatic production test equipment.**
4. **Precise winding and meticulous assembly.**
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Specify HUDSON AMERICAN TRANSFORMERS for your radio and electronic requirements.

Write for catalog 1045

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Velocity Microphones

Flat Frequency Response
Insures True Reproduction

Dead at Sides, Top, Bottom—
To Reduce Unwanted Sounds

**Wide-Angle Pick-up—
Greater Pick-up Range**

**Usable Front and Back
at the Same Time**

Model V-I

Years of special E-V "know how" make possible the full utilization of the ribbon velocity's high fidelity characteristic.

In Model V-1, you get true reproduction without imposed peaks and dips. Pole pieces are designed to insure adequate high frequency reproduction and provide ample output. One-piece frame and internal mounting structure is exclusively E-V—jars and mechanical shocks cannot shift parts.

Bi-directional polar pattern provides equal front and back pick-up, zero pick-up at sides, top, and bottom. Proper tilting reduces feedback and unwanted sounds.

Excellent for solo or orchestra—for individual or chorus—for single speaker or groups . . . in public address, broadcasting, and recording.

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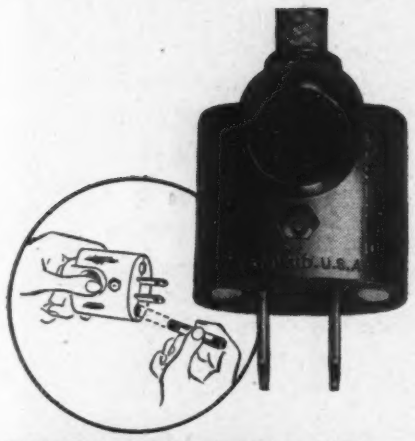

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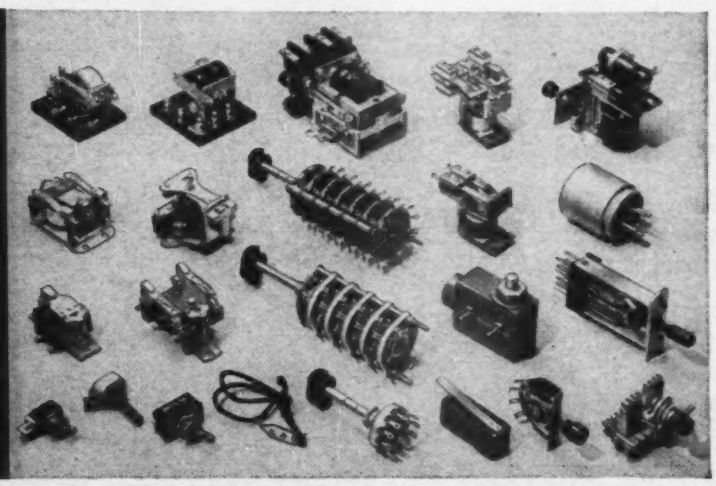

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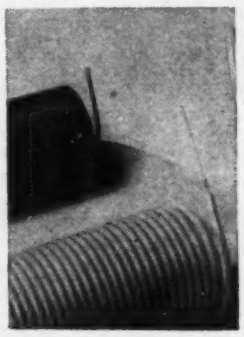
15

Thermoplastic Insulated Wire

THIS EXTRUDED thermoplastic insulated wire (designated Turbo-Therm) is resistant to high temperatures, oils, organic solvents, acids, etc. Because of its poly-



vinyl insulation it has breakdown strength in excess of 1000 v per thousandth inch wall thickness. For the time being the wire is available in guages from 24 to 30. William Brand & Co., 276 Fourth Ave., New York 10, N. Y.

16

Insulator Test Set

THE NEW INSULATION TEST set type RD-50; manufactured by Radio Development Labs., 362 Atlantic Ave., Brooklyn 2, N. Y., is designed as a safe, simple and quick means for testing the breakdown voltage of materials and components. It has a continuous duty rating of 50 kv d-c at 10 ma. It can also deliver up to 50 kv peak, a-c. Output voltages are continuously variable and are divided into three ranges, 0-10 kv, 0-25 kv, and 0-50 kv. Full 315 degree rotation of a variae is employed for each range of output. The output voltage is measured directly in the output circuit with a 4 in. long instrument scale, accurate to 2 percent.

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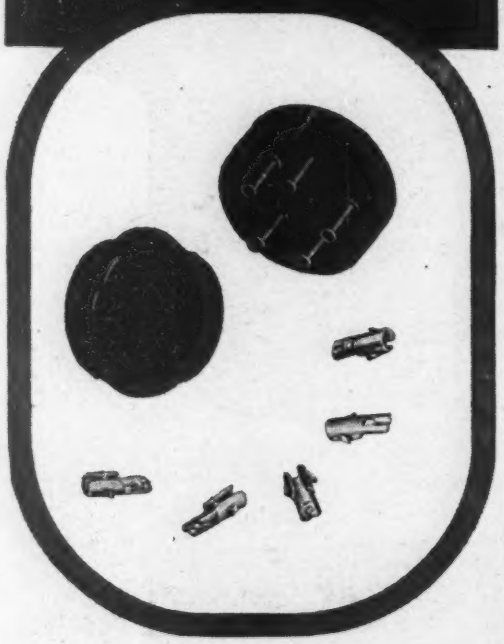

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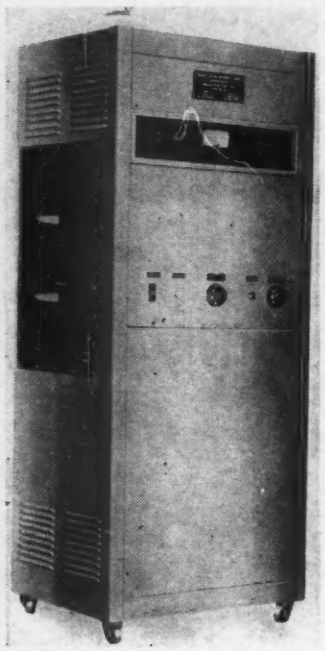
switching interlock, oil immersed 50 kv filter-capacitor-shorting switch interlocked with the rear door, electrically operated power cutoff switch also interlocked with the rear door.

A repeating chime gives continuous audible warning that high voltage is being generated and a red



indicator light simultaneously gives visual indication.

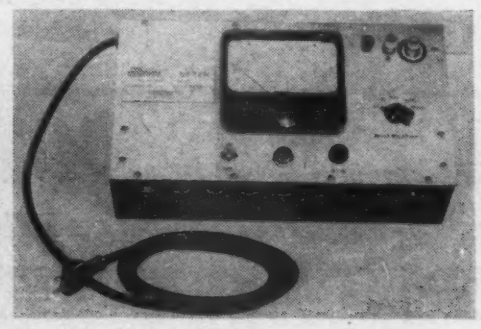
To insure adequate protection for the components, current limiting devices are provided in both a-c and d-c circuits, while an adjustable overload relay is provided in the output circuit.

The cabinet is 80 x 32 x 24 in. deep, gray crackle finish, and has a total weight of 700 lb. Operation is from 115 v 60 cps; power consumption is 1 kw.

17

Megohm Meter

UNUSUAL STABILITY, portability, and a range from 400,000 ohms to 100,000 megohms in five ranges on single-scale four-inch meter, are among the outstanding features of the new model 1500 megohm meter,



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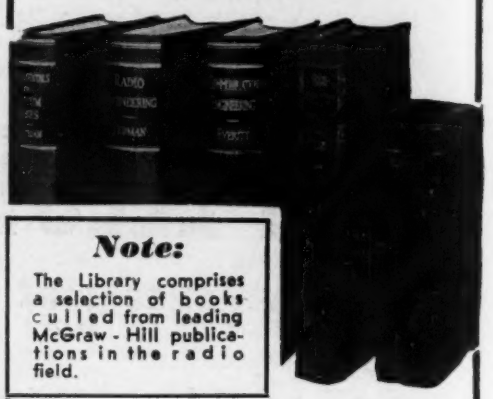

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18

Midget Vibrator

The Radiart Corporation, 3571 W. 62nd St., Cleveland 2, Ohio, has available a new midget vibrator type VR-2, it measures 2⅛ inches high by 1⅛ inches in diameter. A war development designed for operation from a small 6-volt storage battery to furnish power for certain communications equipment, the entire power supply including the storage battery had to be made for a space 6½ x 3½ x 1¾ inches.

Specifications of the VR-2 are as follows:

Vibrator frequency, 185 cps ±10 percent; input voltage nominal, 6.0 v; input voltage range, 4.5 v to 7.5 v; input current, 1.5 amps max at 6.0 v; output voltage, 200 v d-c max; potential difference between primary reed and secondary reed, 25 v Max.

The manufacturer offers blue prints and engineering cooperation.

19

Special Terminal for Heavy Duty Switches

DONALD P. MOSSMAN, INC., 612 N. Michigan Ave., Chicago, Ill., has announced an optional special terminal arrangement for its Series 4100 and 4500 heavy duty lever switches, and its Series 6300 and 061T heavy duty turn Switches.

The addition of the 41ET and 41ETM terminals does not change the operation of the switch in any manner, and the terminals corre-

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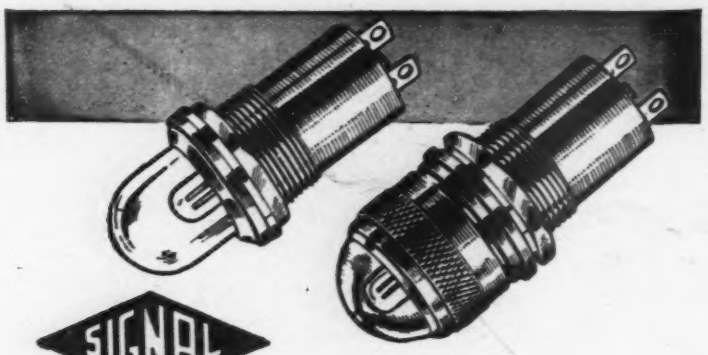

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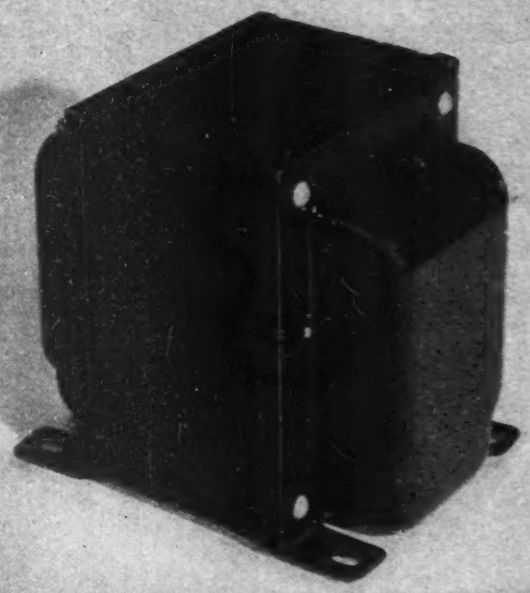

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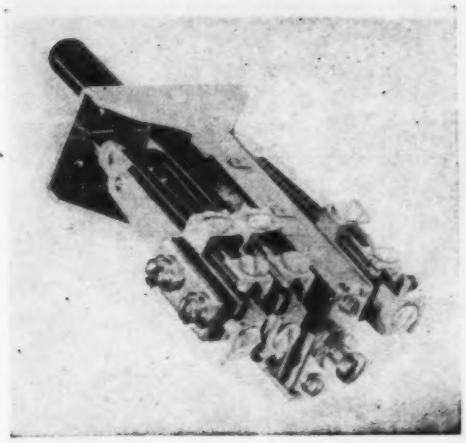


spond with the spring pile-ups. This new type terminal, however, is more effective for specific applications, because by elimination of all soldering, the switch may be installed and serviced more easily.

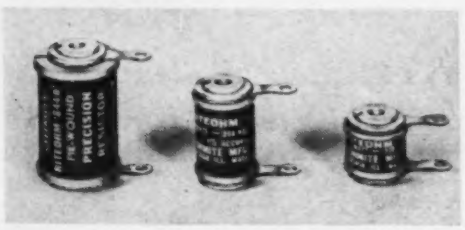
The terminals are heavy, silver-plated brass with 6-32 binding head machine screws and a No. 6 plain washer. Up to No. 10 wire may be used.

The switch is particularly adapted to heavy loads or in installations which, of necessity, are not serviced for long periods of time.

Small Pie-wound Resistors

A NEW SERIES of precision resistors, mounted by means of a through-bolt and equipped with a radial lug at each end, types 842A, 844A, 844B, is being produced by Ohmite Manufacturing Company, 4835 Flournoy St., Chicago 44, Ill.

The Riteohm 84 is pie-wound to 1 percent accuracy and is available in 3 sizes—$\frac{5}{16}$ in. diam by $\frac{1}{8}$ in.



long, $\frac{5}{16}$ in diam by $\frac{1}{8}$ in. long, and $\frac{3}{4}$ in. diam by $1\frac{3}{16}$ in. long. The smallest is a 2 pie while the other two are 4 pie units. The minimum resistance is 1.0 ohm for the 2 pie unit and small 4 pie unit, and 0.10 ohm for the large 4 pie unit. The maximum resistance is 200,000 ohms for the 2 pie, 400,000 ohms for the small 4 pie, and 1.5 meg for the large 4 pie unit.

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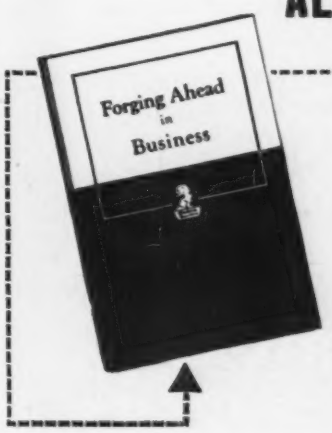

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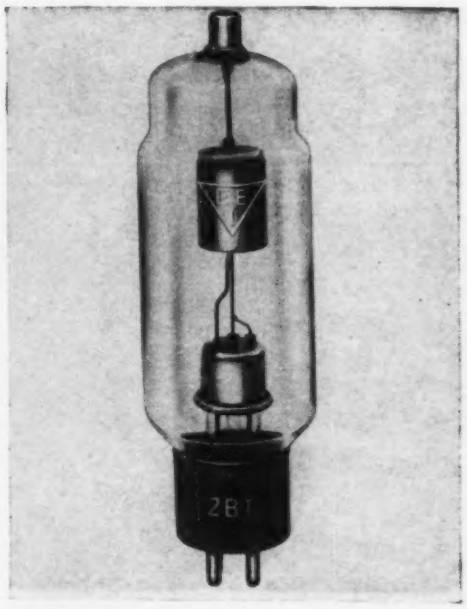
21

High Vacuum Television Rectifier

TYPE 2BT RECTIFIER manufactured by Electronic Enterprises, Inc., 67 Seventh Ave., Newark 4, N. J. closes the gap in available types of rectifiers for television applications. It is a high vacuum type having



high peak inverse rating and expressly designed for plate supply in video receivers requiring potentials to 12,700 volts on the projector tube. Flashback is eliminated in television design without specifying a rectifier with ratings in excess of actual requirements.

22

Remote Antenna Ammeter

OPERATING ON a new principle without the usual thermocouples, an electronic remote antenna ammeter is described by the Andrew Company, 363 E. 75 St., Chicago 19, in

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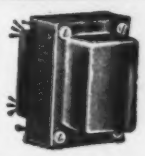

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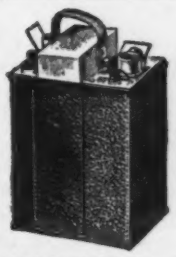

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Hermetically sealed. Six studs. 1, 2, and 3 are pri. 4, 5, and 6 the sec. Pri. ind. at 5 V. 1000 cy.; .20 H. Ratio sec. to pri. 3.02:1. size: 3¼ x 2 41/64"
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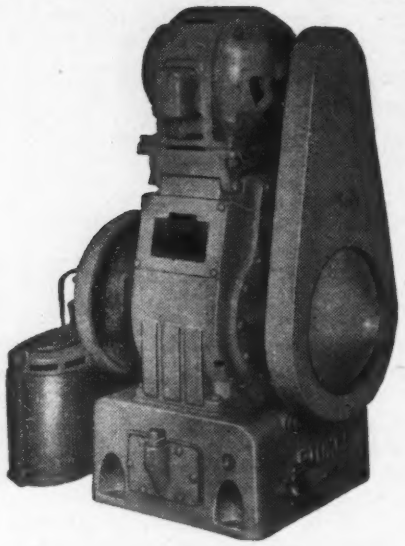

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Bulletin 28A. The remotely-located d-c microammeter is actuated by a current transformer feeding a diode-rectifier tube located at the antenna. Since the regular thermocouple antenna ammeters can be disconnected most of the time, the station using this unit is spared the frequent cost of meter replacement. Likewise, station shutdowns due to thermocouple failure in lightning storms are eliminated.

23

Supersonic Inspection for Quality Control

RAPID AND NON-DESTRUCTIVE testing of materials for internal defects is accomplished with the Supersonic Reflectoscope manufactured by Sperry Products, Inc., 15th and Willow Ave., Hoboken, N. J.

The apparatus sends supersonic vibrations through material under test and measures the length of time it takes the vibrations to penetrate, reflect from the opposite side, or an internal defect, and return to the sending point.

An oscilloscope pattern is viewed by the operator who can be trained in a short time to evaluate the pattern. Access to only one side of material as thick as 10 feet and ability to inspect assembled components make quality control easy.

24

Hand-Held Microphone and Retrax Cord

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How much do you know about U.H.F. Circuits, Cavity Resonators, Wave Guides, Klystrons, Magnetrons and other tubes? All these revolve largely around U.H.F. applications. And here is where CREI training can help you. In our proved home-study course, you learn not only *how* . . . but *why!* Easy-to-read-and-understand lessons are provided well in advance, and each student has his personal instructor who corrects, criticizes and offers suggestions on each lesson examination.

Let CREI train you now to trade that "screwdriver" for a slide rule. Do something about increasing your technical ability and advance to the better-paying radio jobs that offer security and opportunity. The facts are in the free booklet. Send for it today.

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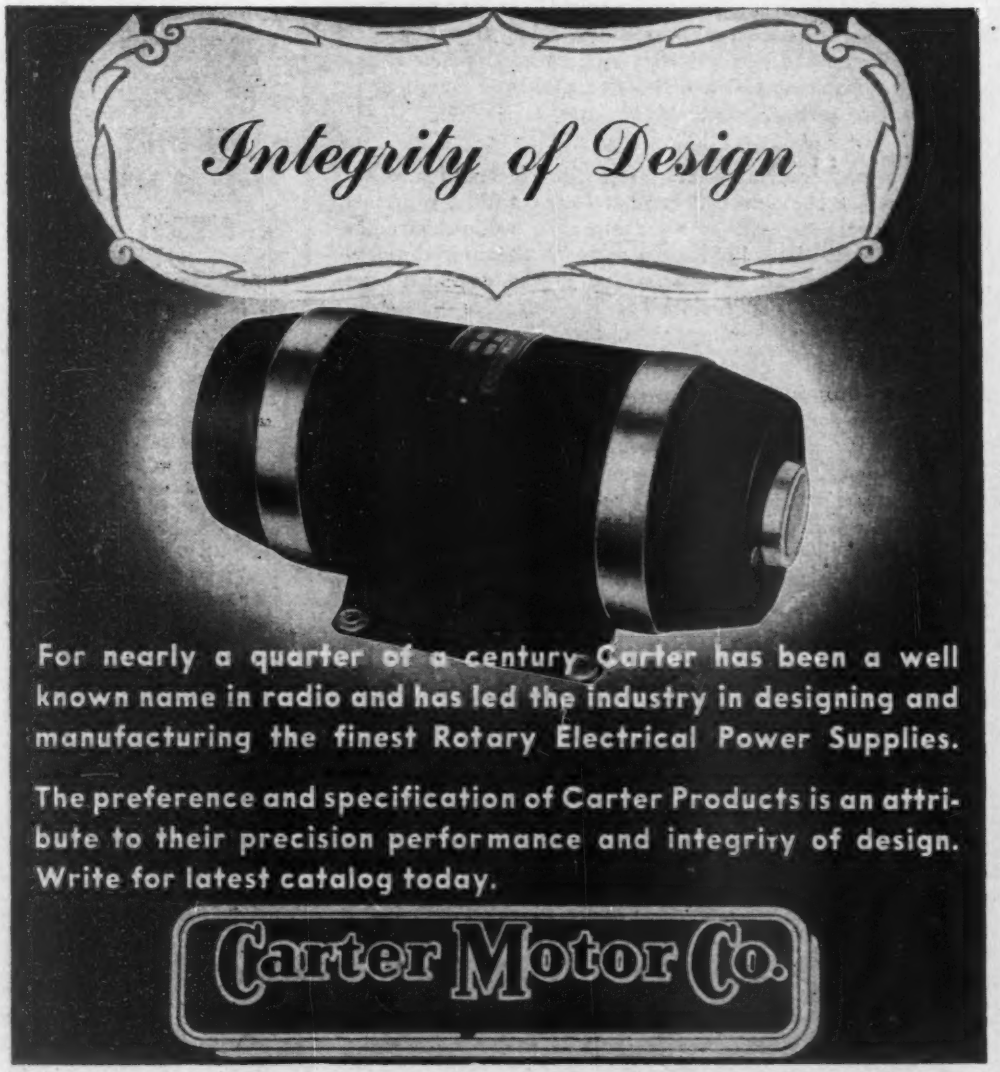
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acoustic low-frequency cutoff and response to 5000 cycles guarantees performance under extreme noise conditions.

The plunger, with recessed finger grip, moves against a strong detent spring, giving clear disconnect warning. The unit is spray-proof and equipped with flexible rubber Retrax cord which extends 40 inches on less than a 2-pound pull.

25

Dielectric Filament Coaxial Cable

A NEW LOW-LOSS coaxial line called CO-X using dielectric filaments between outer and inner conductors to maintain concentric alignment is a new product of the Boston Insulated Wire & Cable Co., Uphams Corner, Boston, Mass.

The lines, suitable for high-power, high-frequency transmission, are furnished in either a flexible or rigid sheath, in both of which gas pressure can be maintained. The dielectric filaments which may be varied to obtain a range of line properties, allow operation up to 250° C without breakdown.

Fabrication of coaxial harness in a systems network and other difficult mechanical processes which involve bending the line, can be accomplished more easily with this than with some conventional air core lines.

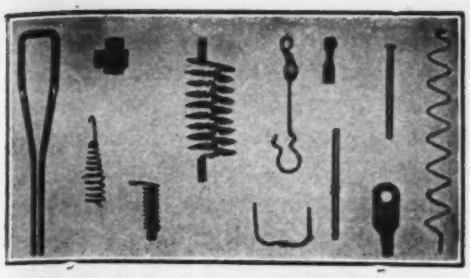
26

Plywood Tubing Masts

TWO NEW LIGHTWEIGHT masts are in production by the Plymold Corporation, Lawrence, Mass., for civilian, f-m, television and amateur use.

Mast, antenna system, and all fittings suitable for erecting atop a roof or side of a building are designed to retail for about $20.

The Ham-Mast is designed for amateur radio uhf and vhf opera-



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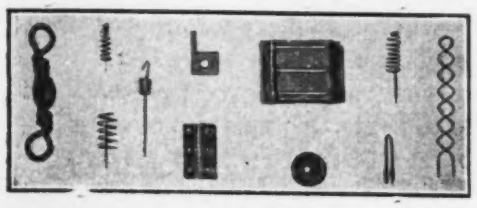

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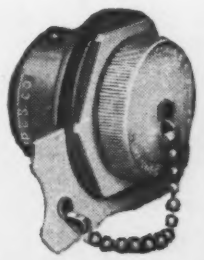

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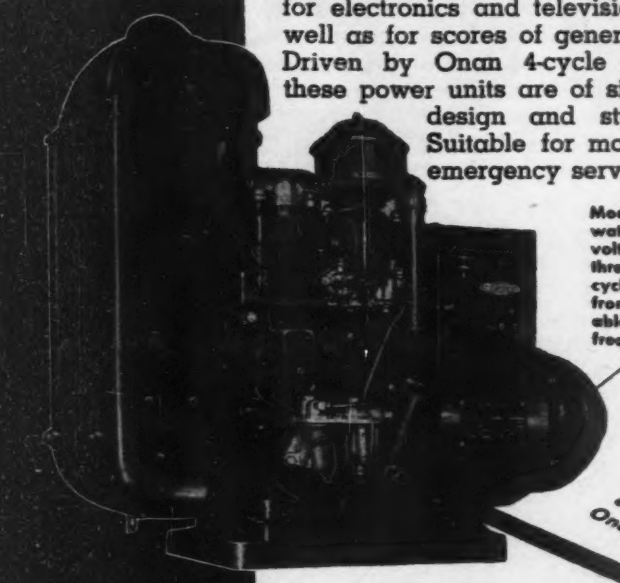

Models range from 350 to 35,000 watts, A. C. types from 115 to 660 volts, 50, 60, 180 cycles, single or three-phase and 400, 500 and 800 cycles, single phase. D. C. types from 6 to 4000 volts. Also available in dual voltage and special frequency types.

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Model shown is from W2C series 2000 to 3500 watts powered by Onan two-cylinder water-cooled engine

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tions, but can be modified for f-m and television reception. It is made up of four telescoping sections of plywood, guyed at two levels, and can be speedily erected.

Mast, fittings and an erection kit are priced at $98.50.

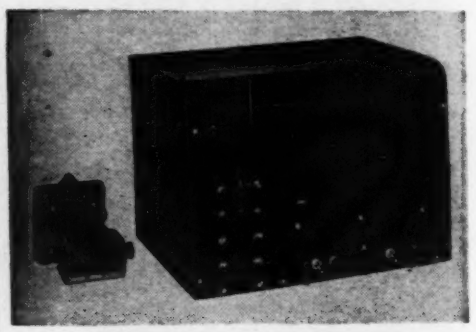
Each section is 14 feet 3 in. long and four sections together weigh 29 lb. Fittings weigh 41 lb. and the erection kit, 15 lb.

27

Two-Decade High-Speed Counter

A COUNTER UNIT particularly applicable for counts exceeding 10 cycles per second, a rate too fast for conventional counters is being built by Potter Instrument Company of 136-56 Roosevelt Avenue, Flushing, N. Y. It is also recommended for installations where mechanical counters would wear out prematurely because of the high-speed continuous operation.

Used alone as a two-decade instrument, the maximum count capacity of the electronic counter is 100. A tube-operated relay is provided for cases where the quantity to be counted exceeds 100. The relay has a spdt contact which is brought out to terminals on the front panel of the unit, and operates once for each 100 counts. An electro-mechanical counter may be connected in series with these terminals and an appropriate external power source, such as the a-c line. Each 100 counts of the electronic counter will then cause one operation of the electro-mechanical counter. When operation of the relay and an external mechanical counter are not involved in the application of the two-decade electronic counter, it may be used alone, at counting rates up to 20,000 per second.



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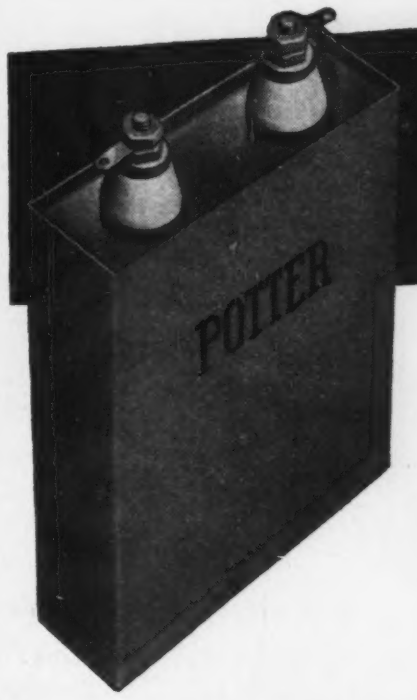

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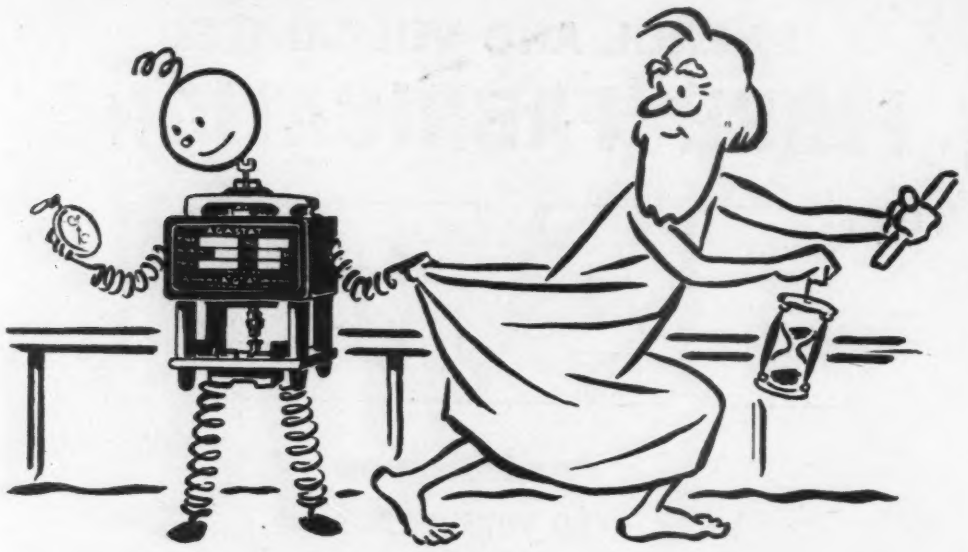

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sine-wave signals, and square-wave signals can all be used to actuate the units decade and the tens decade counters.

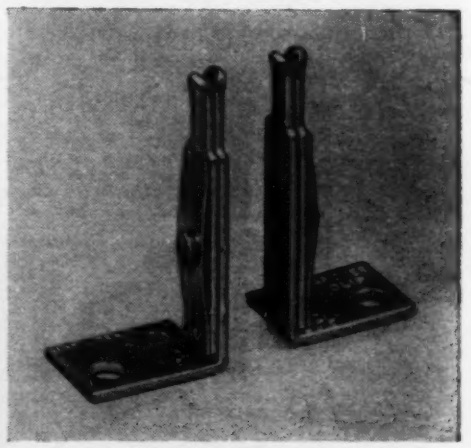
The two-decade electronic counter is complete with output-relay stage, power supply, and is ready for operation from a 115-v, 60-cps, a-c power line. Special selection of vacuum tubes is not required for the counter's 12-tube complement.

Outside dimensions of the cased instrument are: width 13½ in., height 8¾ in., depth 10 in., weight is 26 lb.

28

Testing Clips

A NEW TYPE OF CONNECTOR designed and manufactured by the Rapid Specialties Company, 327-9 West Huron Street, Chicago 10, Illinois, promotes quick test wiring on test panels and bridges. The clips hold



wire leads firmly without injury. Each clip has a base, ⅝ in. x ½ in., and is 1½ in. high made of 0.051 polished brass with a 0.016 phosphor bronze spring.

29

Aircraft Transceiver

A LIGHTWEIGHT, compact, and highly sensitive transmitter-receiver unit, the ATR-3, which serves as a means of radio communications from personal plane to ground and ground to plane, as a direction finder to insure navigational safety, and for broadcast entertainment is now available from Harvey-Wells Electronics, Inc., Department N, Southbridge, Massachusetts.

Functional in design, the transceiver weighs only 12¾ lb, measures 4½ in. high x 5½ in. wide x 8 in. deep, and will operate on a 6- or

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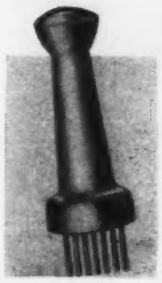

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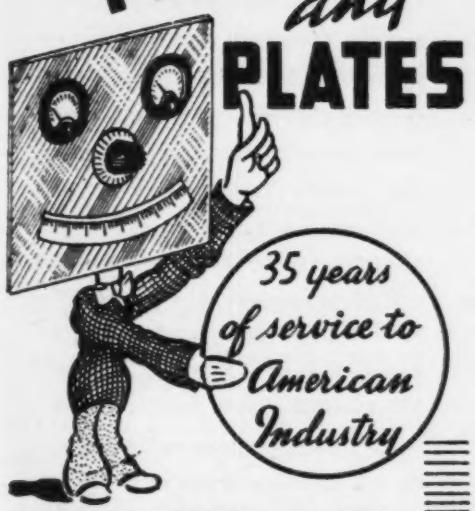

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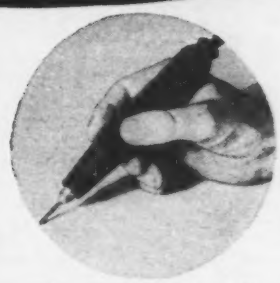

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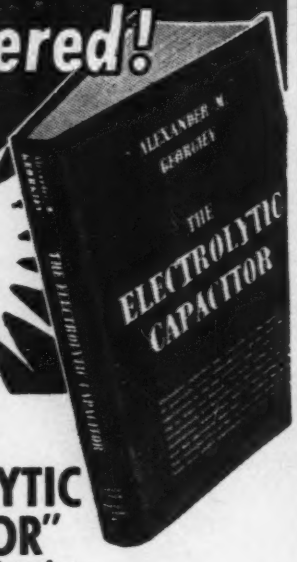

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30

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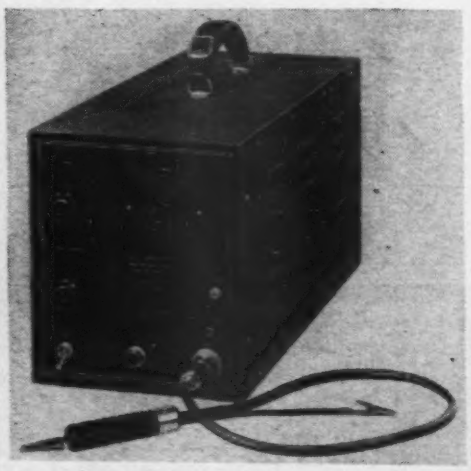
31

Video Amplifier for CRO

A NEW VIDEO AMPLIFIER designed primarily for amplifying complex waves to be viewed on an oscilloscope is announced by the United Cinephone Corporation, Torrington, Connecticut. The amplifier is also useful in laboratory work as an audio amplifier, for tracing and



measuring small r-f voltages in the early stages of radio receivers and in similar applications.

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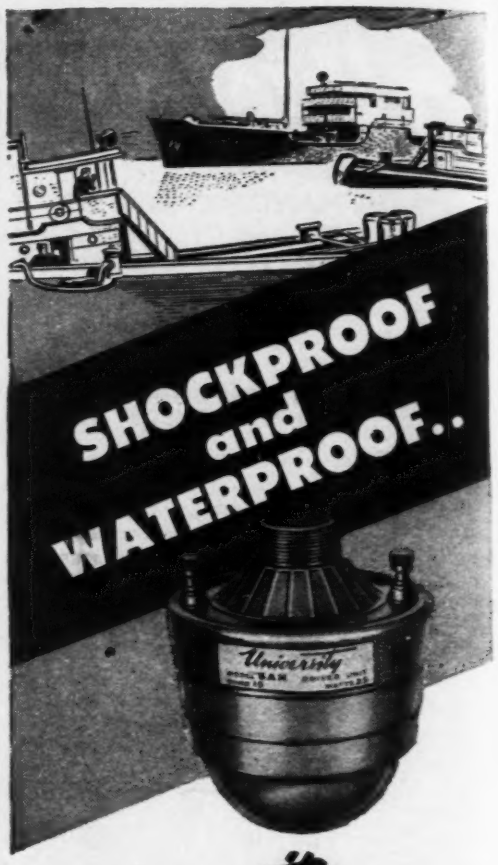

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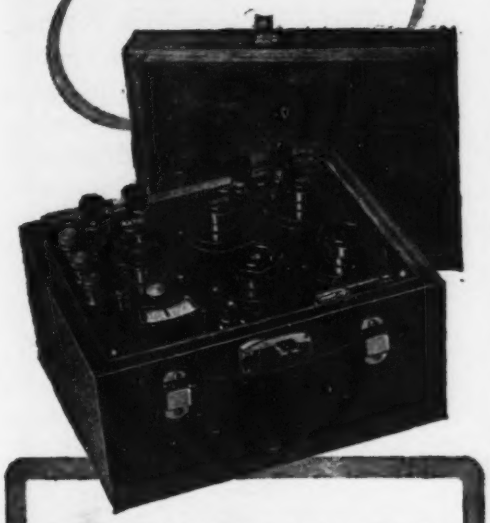

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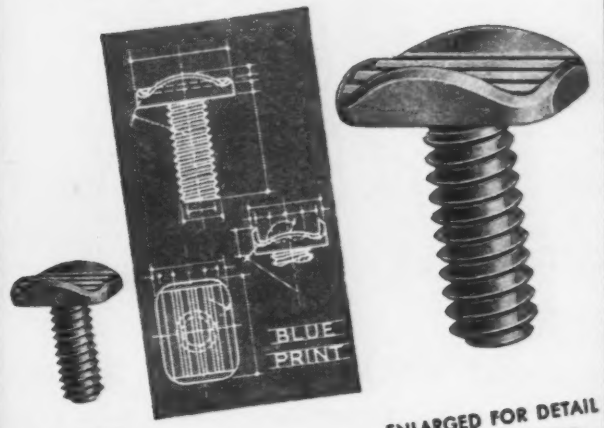

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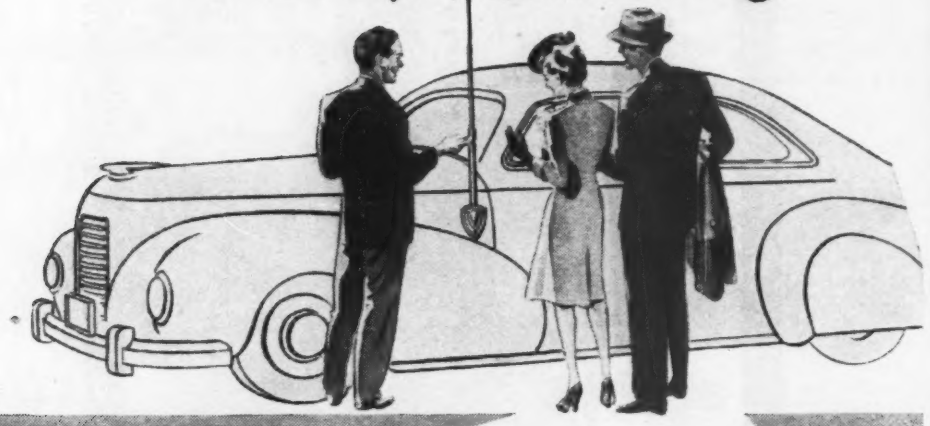

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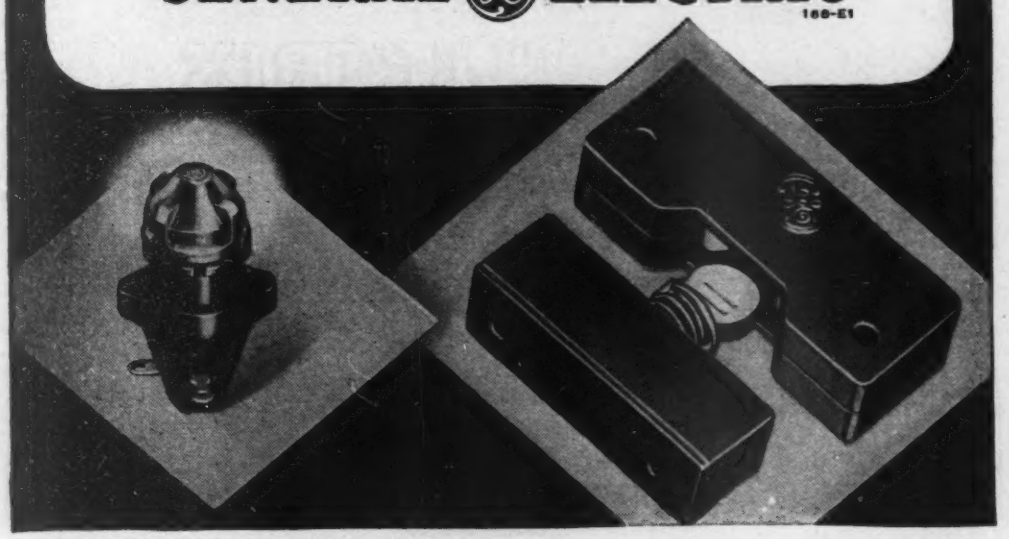

from 15 cps to 4 mc, and 3 db from 10 cps to 4.5 mc. The phase shift is reduced to a minimum to provide satisfactory reproduction of pulses of the order of one microsecond, and square waves at repetition rates as low as 100 a second.

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The equipment operates on 110 to 120 v, 60 cps with power consumption 100 w. Complete with tubes and probe, it measures 7¾x9x 20¾ in. and weighs 35 lbs.

32

Electronic Motor Control

THE ELECTRON Equipment Corp., 917 Meridian Ave., South Pasadena, Calif., offers among its new products a series of variable-speed drives called Varitronic Motor Controls. These devices provide automatic starting, fast, stepless, acceleration and dynamic braking. All makes or sizes of standard d-c motors can be operated, with the aid of the control, from a-c lines. Preselected speeds may be set at any point required and single lever control of starting, stopping, reversing and automatic braking may be provided.

33

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A NEW HEAT dissipating unit for use in television, radar, short-wave radio communication, high pressure mercury lamps, x-ray tubes, induction heating unit is available from The Eastern Engineering Co., New Haven, Connecticut.

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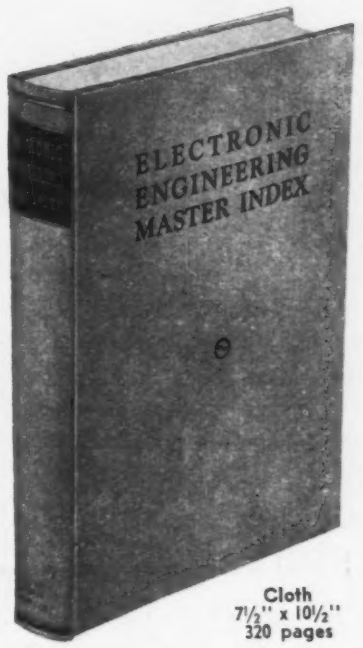

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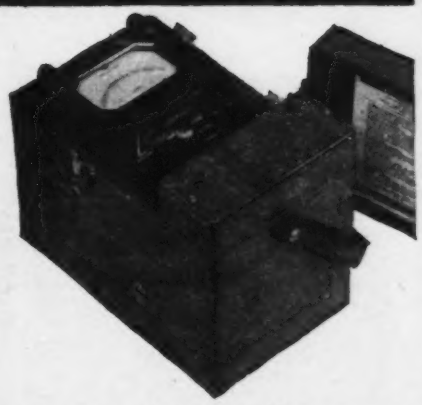

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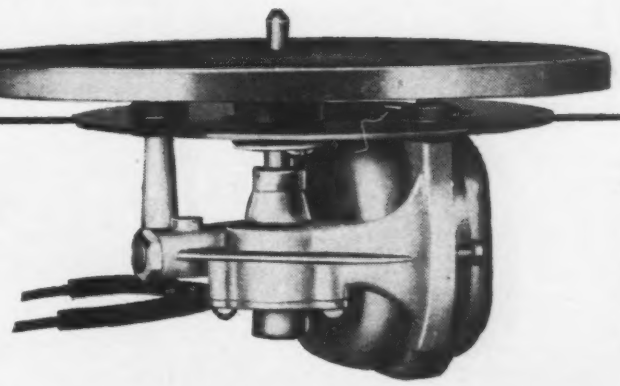

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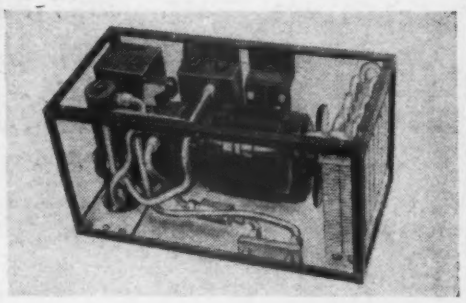

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Literature

34

High Fidelity PA Amplifier. Form No. 130 describes in detail the first in a new series of high-fidelity P-A Amplifiers available from Clark Radio Equipment Corp., 4313 Lincoln Ave., Chicago 18, Ill. Response and distortion curves, as well as specifications, are included.

35

Aircraft and Electrical Circuit Plugs. A revised edition of electric connectors (type K and RK plugs) are described in a 64-page catalog from Cannon Electric Development Co., 3209 Humboldt St., Los Angeles 31, Calif.

36

Training Material. Bulletin No. GES-3303A illustrates and describes a complete set of training material on industrial electronics. The material described in the bulletin includes sound slide films designed to present industrial electronic instruction in a practical and easily understood manner. General Electric Co., 1 River Road, Schenectady, N. Y.

37

Micro Switch House Organ. "Uses Unlimited" designates the title of a bulletin published every now and then by Micro Switch, Freeport, Ill. to describe uses of their products. Vol. 1, No. 4 contains an ar-

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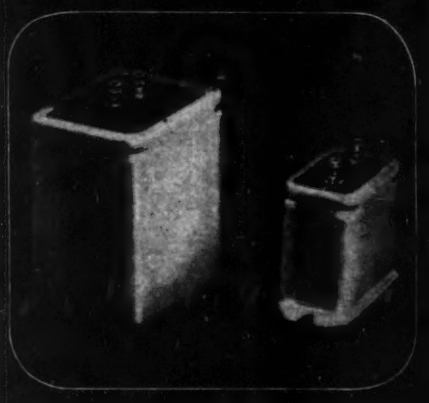

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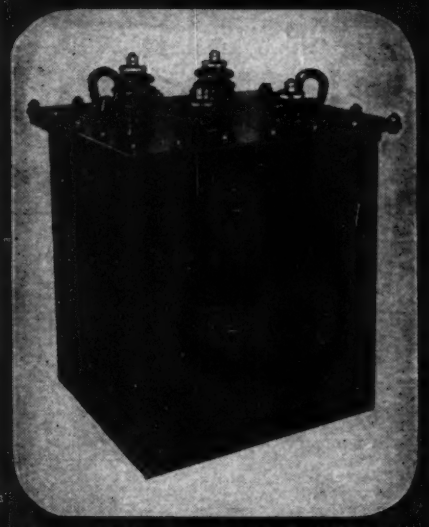

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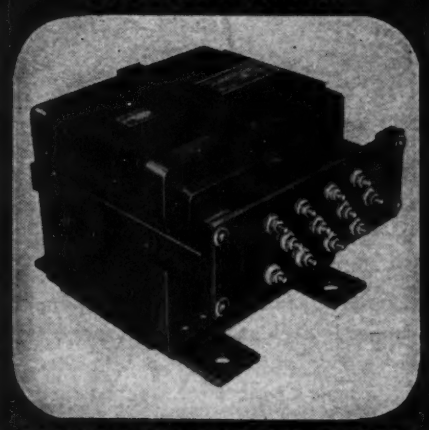

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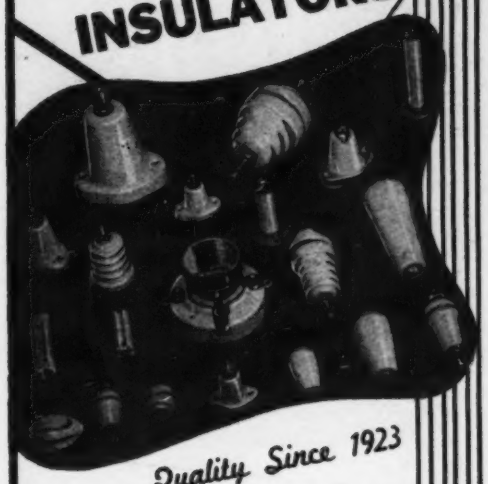

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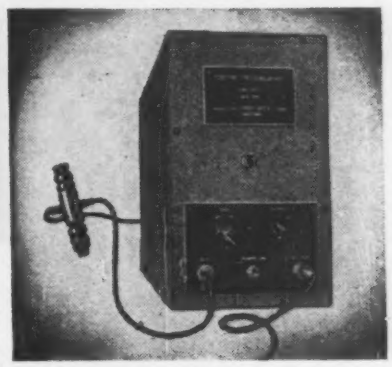

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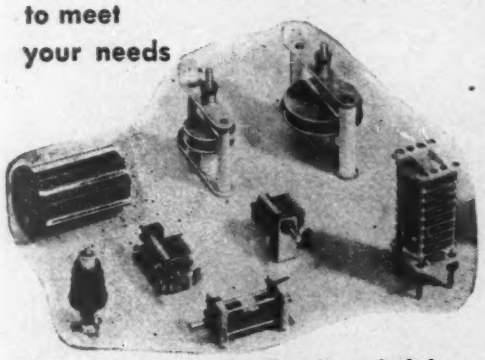

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ticle on "The Micro Switch in Temperature Controls" as well as short article on "The Make Before Break Switch."

38

Guide to Electronics. "The Business Man's Guide to Electronics" (Form A4726) booklet of 28 pages from Westinghouse Electric Corp., Electronic Tube Sales Dept., Bloomfield, N. J. attempts to help one determine how electronic equipment can be profitably applied to business by giving explanations of the six fundamental functions of electronic tubes and describing the applications of electronic equipment that have found widespread use. Most of the applications given can be handled by standardized electronic equipment.

39

Variable Voltage Transformer. Bulletin No. 149-A contains 12 pages of data (including motor-drives, capacities and applications) on a variable transformer designated Powerstat which is designed to obtain continuously on variable output voltage from a-c power lines. Superior Electric Co., Bristol, Conn.

40

Rotary Solenoids. Solenoids made by George H. Leland (development engineers), 123 Webster St., Dayton 2, Ohio, are illustrated and described in an 8-page booklet.

41

Plastics Primer. This designates the title of a booklet which features a concise chart that points out the physical and chemical properties of a representative group of Durex phenolic molding materials. Included also, are two pages devoted to uses of resins. Durez Plastics & Chemicals, Inc., North Tonawanda, N. Y.

42

C-R Tubes for Television. This bulletin illustrates and lists several 5, 7, 10, 12 and 20-in. tubes of both the electrostatic and the magnetic deflection categories, together with the more significant characteristics. It also announces a 15-in. magnetic-deflection tube now in the development stage. Allen B. DuMont Laboratories, Inc., 2 Main Ave., Passaic, N. J.

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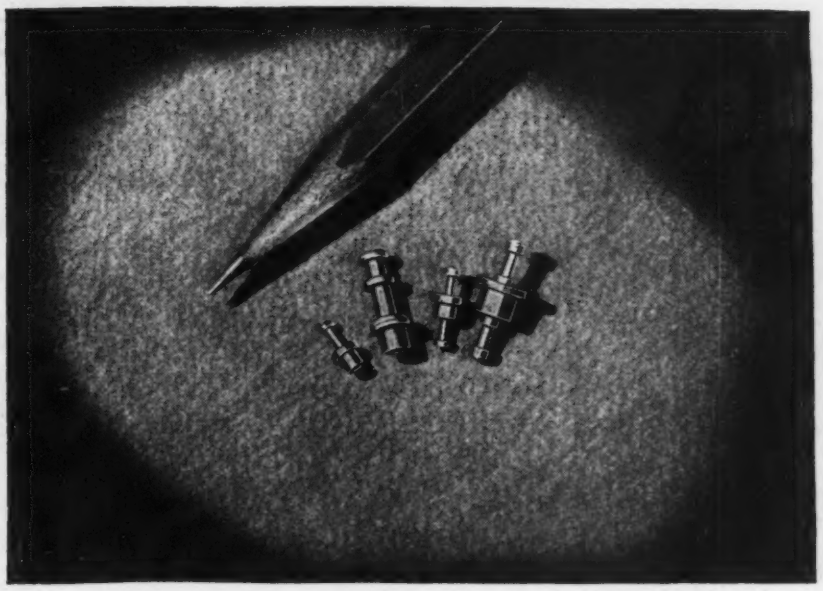

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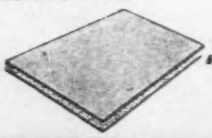

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Bibliography on Industrial Radiology

By H. R. ISENBURGER. *St. John X-Ray Service, Inc., Long Island City 1, N. Y., 16 mimeographed pages, $1.00.*

AUTHORS, TITLES, and locations of about 400 articles on industrial radiology published in the period 1942-1945. This list supplements the compilation of over 1,300 references in the book "Industrial Radiology", Second Edition, by Ancel St. John and H. R. Isenburger, which was published in 1943 by John Wiley & Sons, New York. Together these works are an impressive testimonial to the increasing use of x-ray equipment in industry for visual inspection of materials.—J.M.

• • •

Electronics Dictionary

By NELSON M. COOKE, *Lieut. Comdr., USN, Executive Officer, Radio Materiel School, Naval Research Laboratory, and* JOHN MARKUS, *Associate Editor,* ELECTRONICS. *McGraw-Hill Book Co., New York 18, 433 pages, $5.00.*

TO OTHERS, like this reviewer, who have spent hours in Engineering Society committee meetings formulating a few dozen definitions, the task of setting down thousands of them would indeed seem a colossal undertaking. It must have been.

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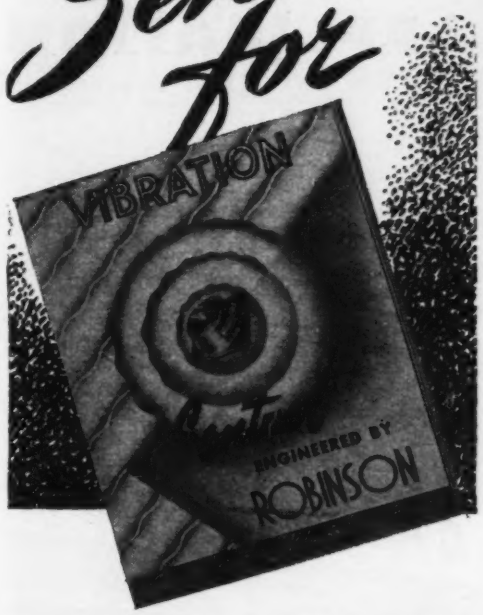

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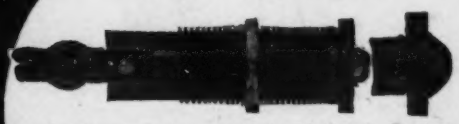

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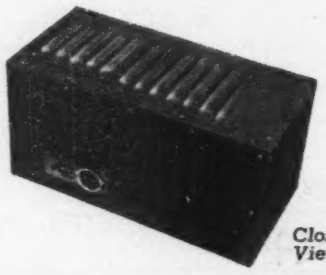

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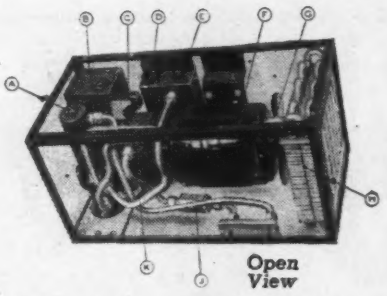

Open View

A. Thermo flow-regulator
B. Reservoir
C. Flow Switch
D. Thermostat
E. Auxiliary Heater
F. Motor
G. Fan
H. Radiator
I. Filter
J. Pump

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with waveguides. Here the authors' policy of including diagrams is particularly helpful.

The authors have made a bold and broad approach to the troublesome problems of abbreviations and compound words. All too few abbreviations have been adopted by ASA but this book continues the general policies that have been standardized. In the case of compound words, the system followed is consistent, even if at times somewhat unorthodox. Engineers in general will devoutly hope that such a uniform system "takes", but reformers of the English language have had hard going over the years. Some of the results of this standardization look a little strange at first glance, as for example antinode and leadin.

A casual glance at a few pages might give the impression that it is an elementary text, but this is far from the case. What would be your score if confronted with the problem of defining these ten words: allochromatic; anelectrotonus; aniseikon; Barnett effect; barytron; Bronson resistance; episcotister; Matteucci effect; trautonium; daraf.

A considerable number of words in the electromedical field are included and may be useful to workers in electronics as medical dictionaries are conveniently available only to very few engineers. Included also are some colloquialisms in the field of sound recording and studio operation.

By frequent repetition, the newly adopted abbreviations vlf, vhf, uhf, etc, are emphasized and, although this system for designating frequency bands is not universally liked, it will be most helpful to have these terms mean the same whenever used.

Engineers are by nature individualists and many of them examining this dictionary would list a considerable number of definitions with which they would not agree or would feel that the wording was not sufficiently qualifying. These would be questions of engineering opinion in general, as the book seems to be remarkably free from errors. This reviewer is sending to the authors, as suggested in the preface, a detailed list of his comments and, if a considerable num-

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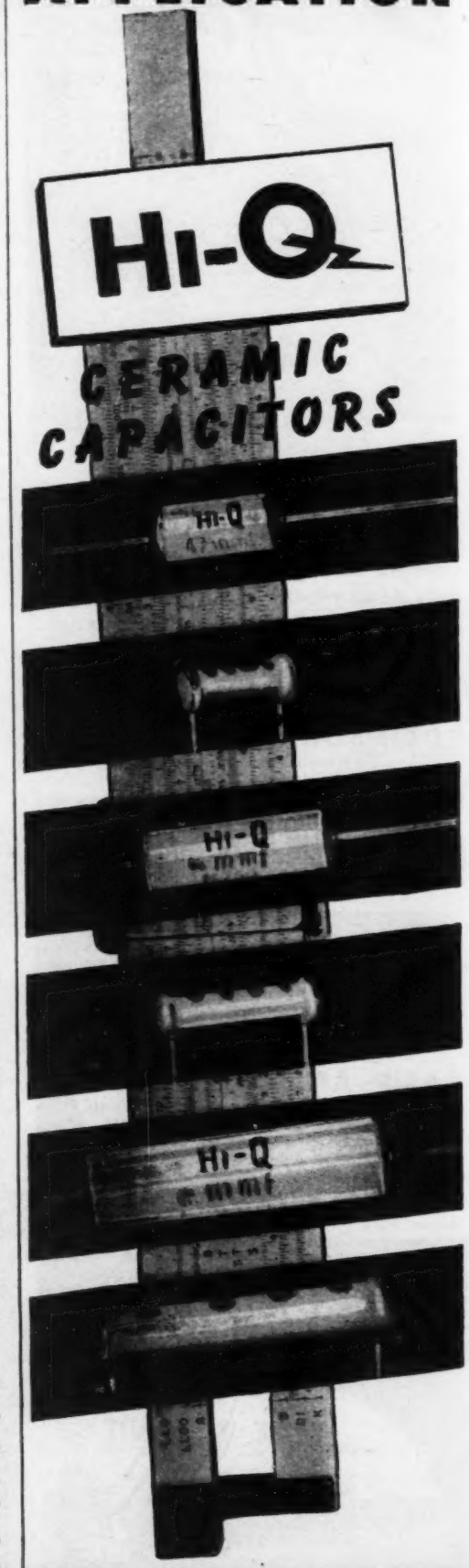

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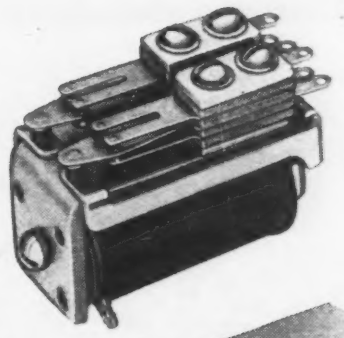

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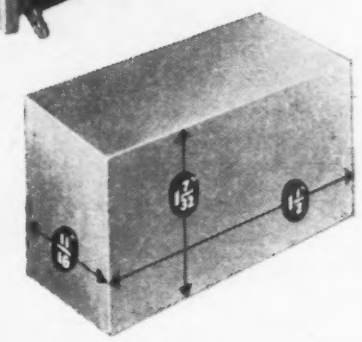

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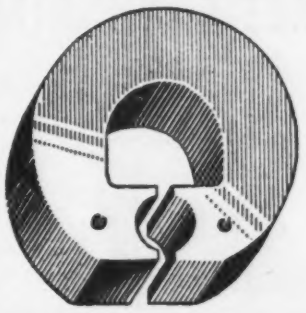

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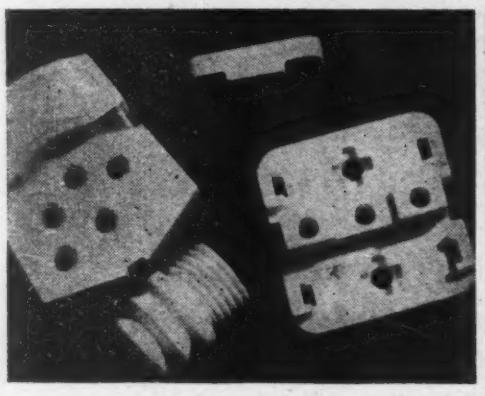

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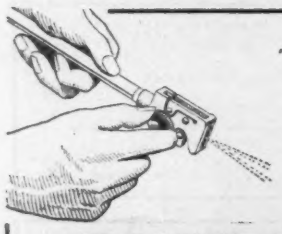

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collecting all like terms, thus giving apparent resistance, inductance and capacitance. The variety of differential equations that must be solved is thus kept to a minimum. Furthermore, the author points out that in a practical circuit under most conditions, stray impedances and other impedances of extreme values can be omitted, thereby simplifying still further the mathematics.

In his introduction the author calls attention to the weakness of the classic approach used throughout the book, and justifies it on the basis of its familiarity to most engineers and the necessity of obtaining a solution. In the problems, curves and oscillographs indicate the response of circuits and the effects of varying circuit parameters, thus illustrating the design knowledge that can be obtained from the analysis, cumbersome as it is in some instances.

The reader should read the text with a critical mind; there are some misleading statements. On page 236 the author states, "Since the period of the applied pulses, 200 microseconds, is *small* compared with the discharging time constant, 700 microseconds, the transit due to a given pulse is very large when the next pulse arrives." What he means is that because the period of the applied pulses is *comparable to* the circuit discharging time, the circuit has appreciably discharged by the time the next pulse arrives.

In connection with a resonant circuit, the text on page 239 explains that ". . . the transit due to one pulse does not have time to diminish to a negligible value before the subsequent pulse arrives, unless the pulse period is extremely large compared with the *natural or freely oscillating* period of the plate circuit." What is meant is that the pulse period is extremely long compared to the *decay* period of the resonant circuit. Had the technical editor been familiar with transient behavior he could have caught such inaccuracies; as it is, the book will have to be read with care. Another fault of the editing is the describing of pulse periods as large and small, which does not give as readily visualized a concept of the time functions as would the conventional designation of long and short.—F.R.

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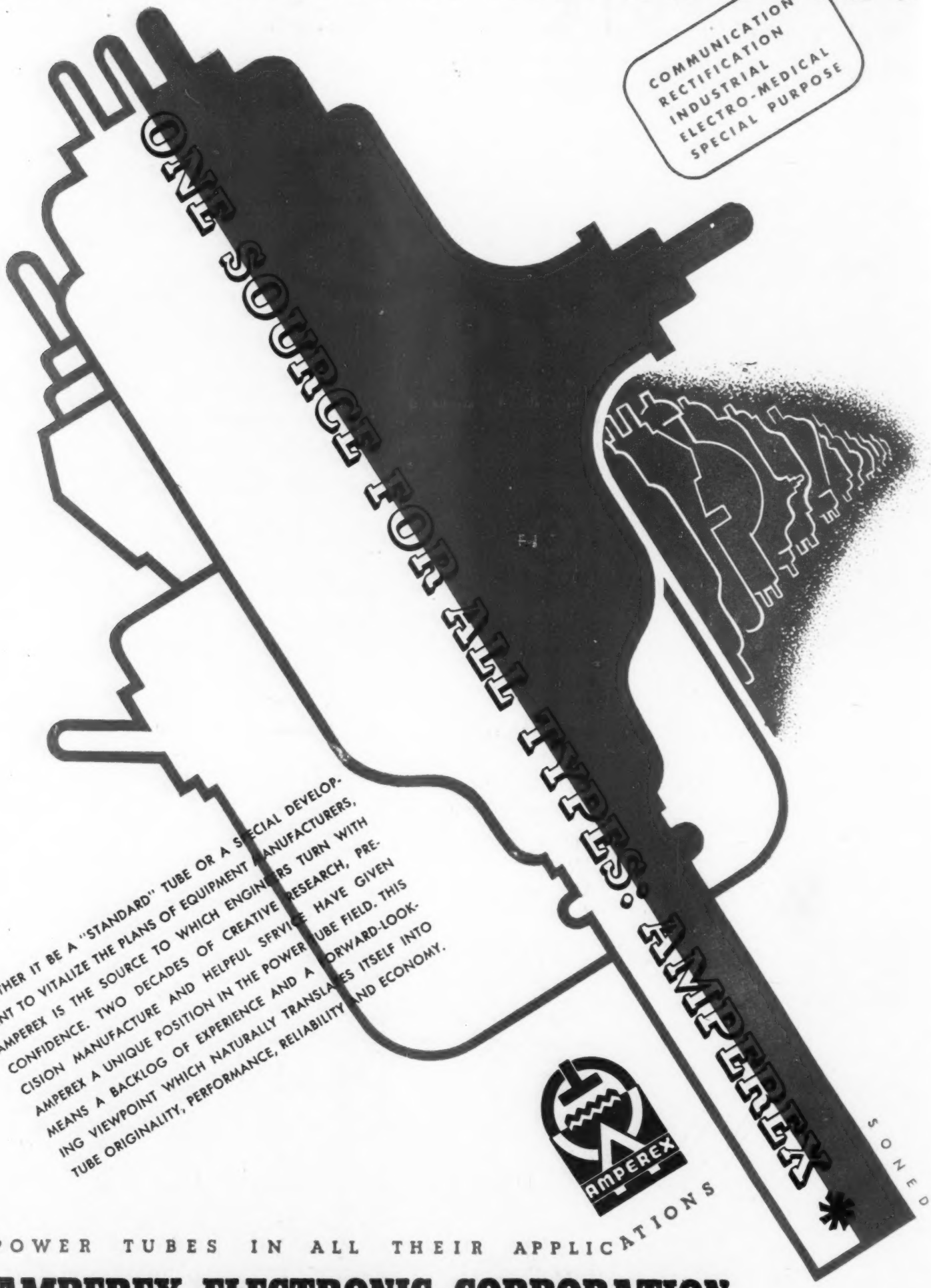

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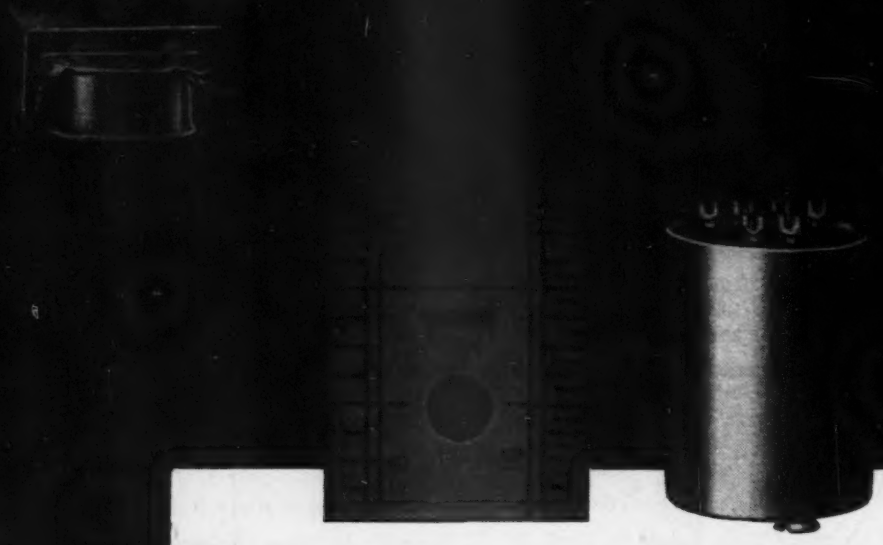

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